LDO Linear Regulators -Micropower, ENABLE, **DELAY, RESET**

400 mA

The NCV8505 is a family of precision micropower voltage regulators. Their output current capability is 400 mA. The family has output voltage options for Adjustable, 2.5 V, 3.3 V and 5.0 V.

The output voltage is accurate within $\pm 2.0\%$ with a maximum dropout voltage of 0.6 V at 400 mA. Low quiescent current is a feature drawing less than 1.0 μ A with ENABLE = 0 V. With ENABLE = 5.0 V, the part only draws 200 µA with 100 µA load. This part is ideal for any and all battery operated microprocessor equipment.

Microprocessor control logic includes an active RESET (with DELAY).

The active RESET circuit operates correctly at an output voltage as low as 1.0 V. The \overline{RESET} function is activated during the power up sequence or during normal operation if the output voltage drops below the regulation limits.

The regulator is protected against reverse battery, short circuit, and thermal overload conditions. The device can withstand load dump transients making it suitable for use in automotive environments. The device has also been optimized for EMC conditions.

Features

- Output Voltage Options: Adjustable, 2.5 V, 3.3 V, 5.0 V
- ± 2.0% Output
- Low < 1.0 μA Sleep Current
- Low 200 μA Quiescent Current
- Fixed or Adjustable Output Voltage
- Active RESET
- ENABLE
- 400 mA Output Current Capability
- Fault Protection
 - ♦ +60 V Peak Transient Voltage
 - → -15 V Reverse Voltage
 - ♦ Short Circuit
 - Thermal Overload
- NCV Prefix for Automotive and Other Applications Requiring Site and Change Control
- AEC Qualified
- PPAP Capable
- These are Pb-Free Devices



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MARKING DIAGRAM



D²PAK-7 **DPS SUFFIX** CASE 936AB



= Voltage Ratings as Indicated Below:

A = Adjustable

2 = 2.5 V

3 = 3.3 V5 = 5.0 V

= Assembly Location

= Wafer Lot

= Year

WW = Work Week

= Pb-Free Package

ORDERING INFORMATION

See detailed ordering and shipping information in the package dimensions section on page 11 of this data sheet.

PIN CONNECTIONS

ADJUSTABLE OUTPUT FIXED OUTPUT Tab = GND Tab = GND Lead 1. DELAY Lead 1. DELAY 0 0 2. ENABLE 2. ENABLE 3. RESET 3. RESET 4. GND 4. GND 5. V_{ADJ} 6. V_{OUT} 5. SENSE 6. V_{OUT} 7. V_{IN} 7. V_{IN}

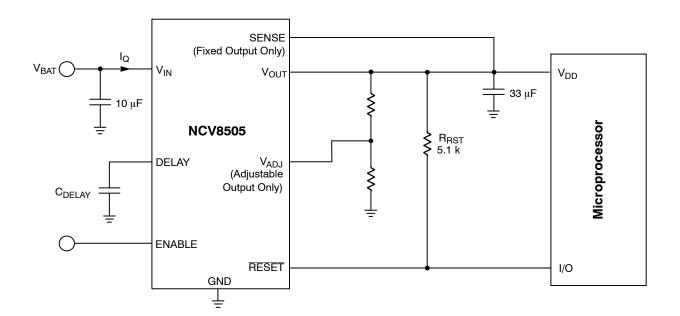


Figure 1. Application Diagram

MAXIMUM RATINGS

Rating	Value	Unit
V _{IN} (DC)	-15 to 45	V
Peak Transient Voltage (46 V Load Dump @ V _{IN} = 14 V)	60	V
Operating Voltage	45	V
V _{OUT} (DC)	-0.3 to 16	V
Voltage Range (RESET, DELAY)	-0.3 to 10	V
Input Voltage Range: V _{ADJ}	-0.3 to 16	V
Input Voltage Range (ENABLE)	-0.3 to 10*	V
ESD Susceptibility (Human Body Model) (Machine Model)	4.0 200	kV V
Junction Temperature, T _J	-40 to +150	°C
Storage Temperature, T _S	-55 to 150	°C
Package Thermal Resistance, 7 Lead D ² PAK Junction-to-Case, R _{θJC} Junction-to-Ambient, R _{θJA}	2.0 10–50**	°C/W °C/W
Lead Temperature Soldering: Reflow: (SMD styles only) (Note	1) 240 peak (Note 2)	°C

Stresses exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stresses above the Recommended Operating Conditions may affect device reliability.

1. 60 second maximum above 183°C.

 $\textbf{ELECTRICAL CHARACTERISTICS} \ (I_{OUT} = 1.0 \ \text{mA}, \ \text{ENABLE} = 5.0 \ \text{V}, \ -40 \ \text{°C} \le T_{J} \le 150 \ \text{°C}; \ V_{IN} = \text{dependent on voltage option (Note 3)}; \ (I_{OUT} = 1.0 \ \text{mA}, \ ENABLE = 5.0 \ \text{V}, \ -40 \ \text{°C} \le T_{J} \le 150 \ \text{°C}; \ V_{IN} = \text{dependent on voltage option (Note 3)}; \ (I_{OUT} = 1.0 \ \text{mA}, \ ENABLE = 5.0 \ \text{V}, \ -40 \ \text{°C} \le T_{J} \le 150 \ \text{°C}; \ V_{IN} = \text{dependent on voltage option (Note 3)}; \ (I_{OUT} = 1.0 \ \text{mA}, \ ENABLE = 5.0 \ \text{V}, \ -40 \ \text{°C} \le T_{J} \le 150 \ \text{°C}; \ V_{IN} = \text{dependent on voltage option (Note 3)}; \ (I_{OUT} = 1.0 \ \text{mA}, \ ENABLE = 5.0 \ \text{V}, \ -40 \ \text{°C} \le T_{J} \le 150 \ \text{°C}; \ V_{IN} = \text{dependent on voltage option (Note 3)}; \ (I_{OUT} = 1.0 \ \text{mA}, \ ENABLE = 5.0 \ \text{V}, \ -40 \ \text{°C} \le T_{J} \le 150 \ \text{°C}; \ V_{IN} = \text{dependent on voltage option (Note 3)}; \ (I_{OUT} = 1.0 \ \text{mA}, \ ENABLE = 5.0 \ \text{V}, \ -40 \ \text{°C} \le T_{J} \le 150 \ \text{°C}; \ V_{IN} = \text{dependent on voltage option (Note 3)}; \ (I_{OUT} = 1.0 \ \text{mA}, \ ENABLE = 5.0 \ \text{V}, \ -40 \ \text{°C} \le T_{J} \le 150 \ \text{°C}; \ V_{IN} = \text{dependent on voltage option (Note 3)}; \ (I_{OUT} = 1.0 \ \text{mA}, \ ENABLE = 5.0 \ \text{V}, \ -40 \ \text{°C} \le T_{J} \le 150 \ \text{°C}; \ V_{IN} = \text{dependent on voltage option (Note 3)}; \ (I_{OUT} = 1.0 \ \text{mA}, \ I_{OUT} = 1.0 \$ unless otherwise specified.)

Characteristic	Test Conditions	Min	Тур	Max	Unit
Output Stage					
Output Voltage for 2.5 V Option (V _O)	$6.5 \text{ V} < V_{\text{IN}} < 16 \text{ V}, \ 1.0 \text{ mA} \leq I_{\text{OUT}} \leq 400 \text{ mA} \\ \textbf{4.5 V} < \textbf{V}_{\text{IN}} < \textbf{26 V}, \ 1.0 \text{ mA} \leq I_{\text{OUT}} \leq 400 \text{ mA}$	2.450 2.425	2.5 2.5	2.550 2.575	V V
Output Voltage for 3.3 V Option (V _O)	7.3 V < V_{IN} < 16 V, 1.0 mA \leq I_{OUT} \leq 400 mA 4.5 V < V_{IN} < 26 V, 1.0 mA \leq I_{OUT} \leq 400 mA	3.234 3.201	3.3 3.3	3.366 3.399	V V
Output Voltage for 5.0 V Option (V _O)	$9.0~V < V_{IN} < 16~V,~1.0~mA \le I_{OUT} \le 400~mA \\ \textbf{6.0}~V < \textbf{V}_{IN} < \textbf{26}~V,~1.0~mA \le I_{OUT} \le 400~mA$	4.90 4.85	5.0 5.0	5.10 5.15	V V
Output Voltage for Adjustable Option (V _O)	V _{OUT} = V _{ADJ} (Unity Gain) 6.5 V < V _{IN} < 16 V, 1.0 mA < I _{OUT} < 400 mA 4.5 V < V_{IN} < 26 V , 1.0 mA < I _{OUT} < 400 mA	1.274 1.261	1.300 1.300	1.326 1.339	V V
Dropout Voltage (V _{IN} – V _{OUT}) (5.0 V and Adj. > 5.0 V Options Only)	$I_{OUT} = 400 \text{ mA}$ $I_{OUT} = 1.0 \text{ mA}$	-	400 30	600 150	mV mV
Load Regulation	V _{IN} = 14 V, 5.0 mA ≤ I _{OUT} ≤ 400 mA	-30	5.0	30	mV
Line Regulation (2.5 V, 3.3 V, and Adjustable Options)	4.5 V < V _{IN} < 26 V, I _{OUT} = 1.0 mA	-	5.0	25	mV
Line Regulation (5.0 V Option)	6.0 V < V _{IN} < 26 V, I _{OUT} = 1.0 mA	-	5.0	25	mV
Quiescent Current, (I _Q) Active Mode	$I_{OUT} = 100 \mu A, V_{IN} = 12 V$ $I_{OUT} = 75 \text{ mA}, V_{IN} = 14 V$ $I_{OUT} \le 400 \text{ mA}, V_{IN} = 14 V$	- - -	200 2.5 25	350 5.0 45	μΑ mA mA
Quiescent Current, (I _Q) Sleep Mode	ENABLE = 0 V, V_{IN} = 12 V, -40° C \leq T _J \leq 125 $^{\circ}$ C	-	-	1.0	μΑ
Current Limit	-	425	800	-	mA
Short Circuit Output Current	V _{OUT} = 0 V	100	500	-	mA
Thermal Shutdown	(Guaranteed by Design)	150	180	-	°C

^{3.} Voltage range specified in the Output Stage of the Electrical Characteristics in boldface type.

^{2. -5°}C/+0°C allowable conditions.

^{*}Reference Figure 16 for switched–battery ENABLE application. **Depending on thermal properties of substrate, $R_{\theta JA} = R_{\theta JC} + R_{\theta CA}$.

 $\textbf{ELECTRICAL CHARACTERISTICS (continued)} \; (I_{OUT} = 1.0 \; \text{mA, ENABLE} = 5.0 \; \text{V, } -40 ^{\circ}\text{C} \leq T_{J} \leq 150 ^{\circ}\text{C}; \; V_{IN} = \text{dependent on } 1.0 \; \text{m} \; \text{C} \;$ voltage option (Note 4); unless otherwise specified.)

Characteristic	Characteristic Test Conditions		Тур	Max	Unit
Reset Function (RESET)			•	•	
RESET Threshold for 2.5 V Option HIGH (V _{RH}) LOW (V _{RL}) Hysteresis	V _{IN} = 4.5 V (Note 5) (Note 6) V _{OUT} Increasing V _{OUT} Decreasing		- - -	1.0 × V _O	V V mV
RESET Threshold for 3.3 V Option HIGH (V _{RH}) LOW (V _{RL}) Hysteresis	V _{IN} = 4.5 V (Note 5) (Note 6) V _{OUT} Increasing V _{OUT} Decreasing	3.10 3.00 35	- - -	1.0 × V _O	V V mV
RESET Threshold for 5.0 V Option HIGH (V _{RH}) LOW (V _{RL}) Hysteresis	V _{IN} = 6.0 V (Note 6) V _{OUT} Increasing V _{OUT} Decreasing	4.70 4.60 50	- - -	1.0×V _O	V V mV
$\label{eq:RESET} \hline \textbf{RESET} \ \textbf{Threshold} \ \textbf{for Adjustable Option} \\ \textbf{HIGH (V_{RH})} \\ \textbf{LOW (V_{RL})} \\ \textbf{Hysteresis} \\ \hline$	V _{IN} = 4.5 V (Note 5) (Note 6) V _{OUT} Increasing V _{OUT} Decreasing	1.22 1.19 10	- - -	1.0 × V _O	> > mV
RESET Output Voltage Low (V _{RLO})	V_{IN} = Minimum (Note 6) (Note 7) 1.0 V \leq V _{OUT} \leq V _{RL} , R _{RESET} = 5.1 k	-	0.1	0.4	٧
DELAY Switching Threshold (V _{DT}) (2.5 V, 3.3 V, and 5.0 V Options)	V _{IN} = Minimum (Note 6) (Note 7)	1.4	1.8	2.2	V
DELAY Switching Threshold (V _{DT}) (Adjustable Option)	V _{IN} = Minimum (Note 6) (Note 7)		1.3	1.6	V
DELAY Low Voltage	V _{IN} = Minimum (Note 6) (Note 7) V _{OUT} < RESET Threshold Low(min)		-	0.2	V
DELAY Charge Current	V _{IN} = Minimum (Note 6) (Note 7) DELAY = 1.0 V, V _{OUT} > V _{RH}		4.0	5.5	μΑ
DELAY Discharge Current	V _{IN} = Minimum (Note 6) (Note 7) DELAY = 1.0 V, V _{OUT} < V _{RL}	5.0	-	-	mA
Voltage Adjust (Adjustable Output only)			I.		
Input Current	V _{ADJ} = 1.25 V, V _{IN} = Minimum (Note 6) (Note 7)	-0.5	_	0.5	μΑ
ENABLE					
Input Threshold Low, V_{IN} = 14 V (Note 6) High, V_{IN} = 14 V (Note 6)		_ 2.0		1.0	V V
Input Current	ut Current ENABLE = 5.0 V, V _{IN} = 14 V (Note 6)			75	μΑ

Voltage range specified in the Output Stage of the Electrical Characteristics in boldface type.
 For V_{IN} ≤ 4.5 V, a RESET = Low may occur with the output in regulation.
 Part is guaranteed by design to meet specification over the entire V_{IN} voltage range, but is production tested only at the specified V_{IN} voltage.
 Minimum V_{IN} = 4.5 V for 2.5 V, 3.3 V, and Adjustable options. Minimum V_{IN} = 6.0 V for 5.0 V option.

PACKAGE PIN DESCRIPTION, ADJUSTABLE OUTPUT

Pin Number	Pin Symbol	Function
1	DELAY	Timing capacitor for RESET function.
2	ENABLE	ENABLE control for the IC. A high powers the device up.
3	RESET	Active reset (accurate to V _{OUT} ≥ 1.0 V)
4	GND	Ground. All GND leads must be connected to Ground
5	V_{ADJ}	Voltage Adjust. A resistor divider from V _{OUT} to this lead sets the output voltage.
6	V _{OUT}	±2.0%, 400 mA output.
7	V _{IN}	Input Voltage.

PACKAGE PIN DESCRIPTION, FIXED OUTPUT

Pin Number	Pin Symbol	Function
1	DELAY	Timing capacitor for RESET function.
2	ENABLE	ENABLE control for the IC. A high powers the device up.
3	RESET	Active reset (accurate to V _{OUT} ≥ 1.0 V)
4	GND	Ground. All GND leads must be connected to Ground
5	SENSE	Kelvin connection which allows remote sensing of output voltage for improved regulation. If remote sensing is not desired, connect to V_{OUT} .
6	V _{OUT}	±2.0%, 400 mA output.
7	V _{IN}	Input Voltage.

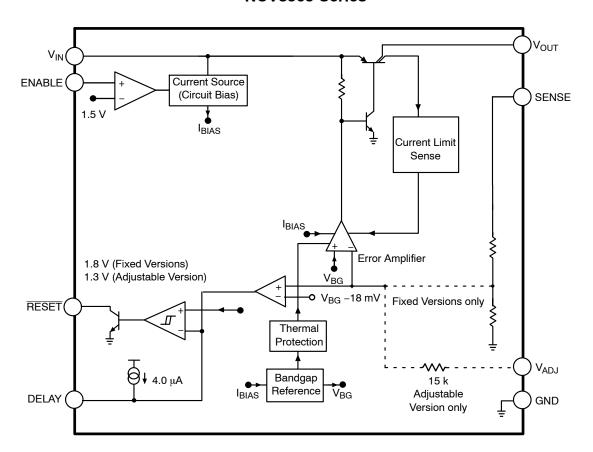


Figure 2. Block Diagram

TYPICAL PERFORMANCE CHARACTERISTICS

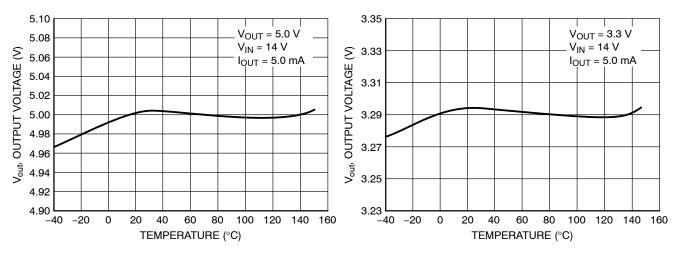


Figure 3. 5 V Output Voltage vs. Temperature

Figure 4. 3.3 V Output Voltage vs. Temperature

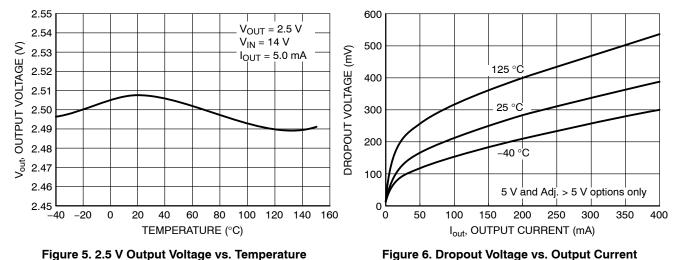


Figure 5. 2.5 V Output Voltage vs. Temperature

OUTPUT VOLTAGE (V)

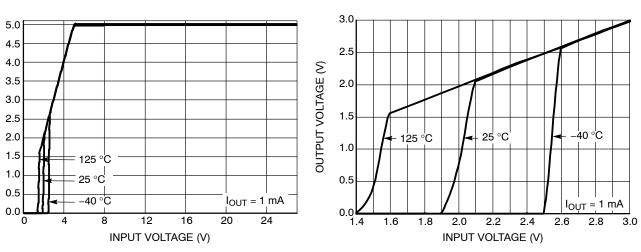


Figure 7. Output Voltage vs. Input Voltage

Figure 8. Output Voltage vs. Input Voltage

TYPICAL PERFORMANCE CHARACTERISTICS

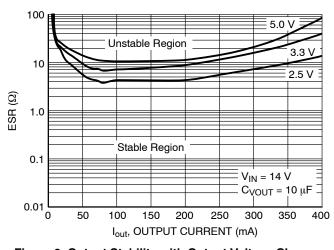


Figure 9. Output Stability with Output Voltage Change

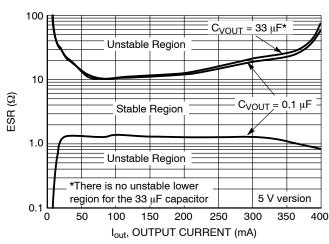


Figure 10. Output Stability with Output Capacitor
Change

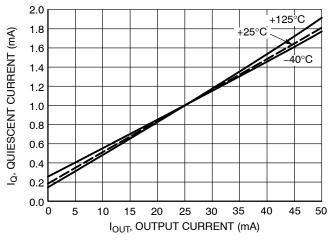


Figure 11. Quiescent Current vs. Output Current

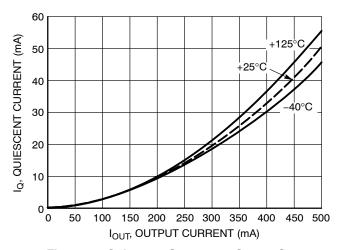


Figure 12. Quiescent Current vs. Output Current

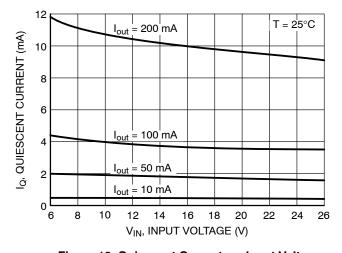


Figure 13. Quiescent Current vs. Input Voltage

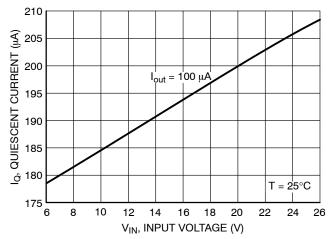


Figure 14. Quiescent Current vs. Input Voltage

CIRCUIT DESCRIPTION

REGULATOR CONTROL FUNCTIONS

The NCV8505 contains the microprocessor compatible control function \overline{RESET} (Figure 15).

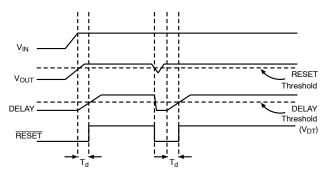


Figure 15. Reset and Delay Circuit Wave Forms

RESET Function

A \overline{RESET} signal (low voltage) is generated as the IC powers up until V_{OUT} is within 1.5% of the regulated output voltage, or when V_{OUT} drops out of regulation,and is lower than 4.0% below the regulated output voltage. Hysteresis is included in the function to minimize oscillations.

The \overline{RESET} output is an open collector NPN transistor, controlled by a low voltage detection circuit. The circuit is functionally independent of the rest of the IC thereby guaranteeing that the \overline{RESET} signal is valid for V_{OUT} as low as 1.0 V.

ENABLE Function

The part stays in a low I_Q sleep mode when the ENABLE pin is held low. The part has an internal pull down if the pin is left floating.

The integrity of the ENABLE pin allows it to be tied to the battery line through an external resistor. It will withstand load dump potentials in this configuration.

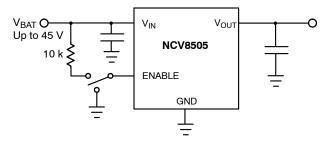


Figure 16. ENABLE Function

DELAY Function

The reset delay circuit provides a programmable (by external capacitor) delay on the \overline{RESET} output lead.

The DELAY lead provides source current (typically $4.0 \,\mu A$) to the external DELAY capacitor during the following proceedings:

- 1. During Power Up (once the regulation threshold has been verified).
- 2. After a reset event has occurred and the device is back in regulation. The DELAY capacitor is discharged when the regulation (RESET threshold) has been violated. This is a latched incident. The capacitor will fully discharge and wait for the device to regulate before going through the delay time event again.

Voltage Adjust

Figure 17 shows the device setup for a user configurable output voltage. The feedback to the V_{ADJ} pin is taken from a voltage divider referenced to the output voltage. The loop is balanced around the Unity Gain threshold (1.30 V typical).

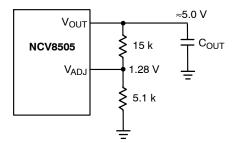


Figure 17. Adjustable Output Voltage

APPLICATION NOTES

SETTING THE DELAY TIME

The delay time is controlled by the Reset Delay Low Voltage, Delay Switching Threshold, and the Delay Charge Current. The delay follows the equation:

$$t_{DELAY} = \frac{\left[C_{DELAY}(V_{dt} - \text{Reset Delay Low Voltage})\right]}{\text{Delay Charge Current}}$$

Example:

Using $C_{DELAY} = 33 \text{ nF}.$

Assume reset Delay Low Voltage = 0.

Use the typical value for $V_{dt} = 1.8 \text{ V}$ (2.5 V, 3.3 V, and 5.0 V options).

Use the typical value for Delay Charge Current = $4.2 \mu A$.

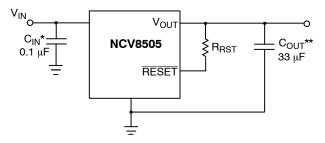
$$t_{DELAY} = \frac{\left[33 \text{ nF}(1.8 - 0)\right]}{4.2 \,\mu\text{A}} = 14 \text{ ms}$$

STABILITY CONSIDERATIONS

The output or compensation capacitor helps determine three main characteristics of a linear regulator: start-up delay, load transient response and loop stability.

The capacitor value and type should be based on cost, availability, size and temperature constraints. A tantalum or aluminum electrolytic capacitor is best, since a film or ceramic capacitor with almost zero ESR can cause instability. The aluminum electrolytic capacitor is the least expensive solution, but, if the circuit operates at low temperatures (-25° C to -40° C), both the value and ESR of the capacitor will vary considerably. The capacitor manufacturers data sheet usually provides this information.

The value for the output capacitor C_{OUT} shown in Figure 18 should work for most applications, however it is not necessarily the optimized solution.



 $^{^{\}star}C_{IN}$ required if regulator is located far from the power supply filter.

Figure 18. Test and Application Circuit Showing Output Compensation

CALCULATING POWER DISSIPATION IN A SINGLE OUTPUT LINEAR REGULATOR

The maximum power dissipation for a single output regulator (Figure 19) is:

$$P_{D(max)} = [V_{IN(max)} - V_{OUT(min)}]I_{OUT(max)}$$

$$+ V_{IN(max)}I_{Q}$$
(1)

where:

V_{IN(max)} is the maximum input voltage,

V_{OUT(min)} is the minimum output voltage,

 $I_{OUT(max)}$ is the maximum output current for the application, and

 I_Q is the quiescent current the regulator consumes at $I_{OUT(max)}$.

Once the value of $P_{D(max)}$ is known, the maximum permissible value of $R_{\theta JA}$ can be calculated:

$$R_{\theta JA} = \frac{150^{\circ}C - T_{A}}{P_{D}}$$
 (2)

The value of $R_{\theta JA}$ can then be compared with those in the package section of the data sheet. Those packages with $R_{\theta JA}$'s less than the calculated value in equation 2 will keep the die temperature below 150°C.

In some cases, none of the packages will be sufficient to dissipate the heat generated by the IC, and an external heatsink will be required.

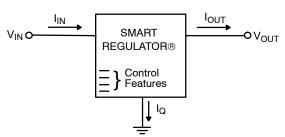


Figure 19. Single Output Regulator with Key Performance Parameters Labeled

HEAT SINKS

A heat sink effectively increases the surface area of the package to improve the flow of heat away from the IC and into the surrounding air.

Each material in the heat flow path between the IC and the outside environment will have a thermal resistance. Like series electrical resistances, these resistances are summed to determine the value of $R_{\rm BIA}$:

$$R_{\theta}JA = R_{\theta}JC + R_{\theta}CS + R_{\theta}SA \tag{3}$$

where:

 $R_{\theta IC}$ = the junction-to-case thermal resistance,

 $R_{\theta CS}$ = the case-to-heatsink thermal resistance, and

 $R_{\theta SA}$ = the heatsink-to-ambient thermal resistance.

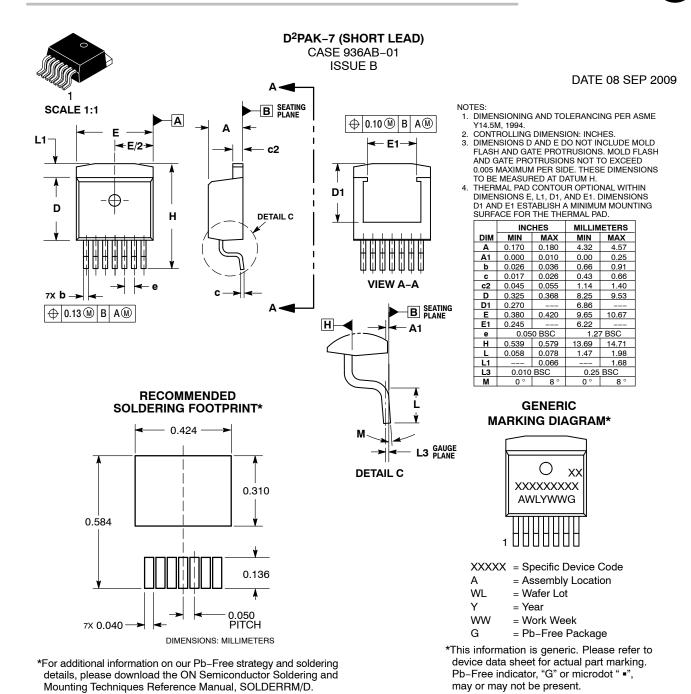
 $R_{\theta JC}$ appears in the package section of the data sheet. Like $R_{\theta JA},$ it too is a function of package type. $R_{\theta CS}$ and $R_{\theta SA}$ are functions of the package type, heatsink and the interface between them. These values appear in heat sink data sheets of heat sink manufacturers.

^{**}C_{OUT} required for stability. Capacitor must operate at minimum temperature expected.

ORDERING INFORMATION

Device	Output Voltage	Package	Shipping [†]
NCV8505D2TADJG		D ² PAK-7 (Pb-Free)	50 Units / Rail
NCV8505D2TADJR4G	Adjustable	D ² PAK-7 (Pb-Free)	750 Tape & Reel
NCV8505D2T25G	0.51/	D ² PAK-7 (Pb-Free)	50 Units / Rail
NCV8505D2T25R4G	2.5 V	D ² PAK-7 (Pb-Free)	750 Tape & Reel
NCV8505D2T33G	0.01/	D ² PAK-7 (Pb-Free)	50 Units / Rail
NCV8505D2T33R4G	3.3 V	D ² PAK-7 (Pb-Free)	750 Tape & Reel
NCV8505D2T50G	501/	D ² PAK-7 (Pb-Free)	50 Units / Rail
NCV8505D2T50R4G	5.0 V	D ² PAK-7 (Pb-Free)	750 Tape & Reel

[†]For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.



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