## Data Sheet

## FEATURES

Worldwide NTSC/PAL/SECAM color demodulation support One 10-bit analog-to-digital converter (ADC), $4 \times$ oversampling per channel for CVBS, Y/C mode, and YPrPb
Four analog video input channels with on-chip antialiasing filter CVBS (composite), Y/C (S-Video), and YPrPb (component) video input support
Fully differential, pseudo differential, and single-ended CVBS video input support
Up to 4 V common-mode input range solution
Excellent common-mode rejection capabilities
Five-line adaptive comb filters and CTI/DNR video enhancement
TBC functionality provided by adaptive digital line length tracking (ADLLT), signal processing, and enhanced first in, first out (FIFO) management
Integrated automatic gain control (AGC) with adaptive peak white mode
Video fast switch capability
Adaptive contrast enhancement (ACE)
Down dither ( 8 bits to 6 bits)

Rovi ${ }^{\text {TM }}$ (Macrovision) copy protection detection NTSC/PAL/SECAM autodetection
8-bit ITU-R BT. 656 YCrCb 4:2:2 output and HS, VS, or FIELD
Full-featured VBI data slicer with teletext support (WST)
Power-down mode and ultralow sleep mode current Two-wire serial MPU interface ( $I^{2} \mathrm{C}$ compatible)
Single 1.8 V supply possible
Automotive qualified models available
$-40^{\circ} \mathrm{C}$ to $+105^{\circ} \mathrm{C}$ automotive temperature grade
$-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ industrial qualified temperature grade
32-lead, $5 \mathrm{~mm} \times 5 \mathrm{~mm}$, RoHS-compliant LFCSP

## APPLICATIONS

Automotive infotainment DVRs for video security Media players

FUNCTIONAL BLOCK DIAGRAM


Figure 1.

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## GENERAL DESCRIPTION

The ADV7182 automatically detects and converts standard analog baseband video signals compatible with worldwide NTSC, PAL, and SECAM standards into a 4:2:2 component video data stream. This video data stream is compatible with the 8 -bit ITU-R BT. 656 interface standard.
External HS, VS, and FIELD signals can provide timing references for LCD controllers and other video ASICs. The accurate 10-bit analog-to-digital conversion provides professional quality video performance for consumer applications with true 8-bit data resolution. The analog video inputs accept both single-ended, pseudo-differential, and fully differential composite video signals as well as S -Video and Y PbPr video signals, supporting a wide range of consumer and automotive video sources.
The ADV7182 along with an external resistor divider provide a common-mode input range of 4 V , enabling the removal of large signal, common-mode transients present on the video lines. Common-mode rejection (CMR) values of up to 80 dB can be achieved without the need for external amplifier circuitry.

The AGC and clamp restore circuitry allow an input video signal peak-to-peak range to 1.0 V at the analog video input pin of the ADV7182. Alternatively, these can be bypassed for manual settings.
The ADV7182 can be protected from short-to-battery (STB) events with standard ac coupling capacitors.
The ADV7182 is programmed via a two-wire, serial bidirectional port ( $\mathrm{I}^{2} \mathrm{C}^{\star}$ compatible) and is fabricated in a 1.8 V CMOS process. Its monolithic CMOS construction ensures greater functionality with lower power dissipation.
The ADV7182 is provided in a space-saving LFCSP surface-mount, RoHS compliant package. The ADV7182 is available in an automotive grade that is rated over the $-40^{\circ} \mathrm{C}$ to $+105^{\circ} \mathrm{C}$ temperature range. This makes the ADV7182 ideal for automive applications. The ADV7182 is also available in a $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ temperature range, making it ideal for industrial applications.
The ADV7182 is a versatile one-chip multiformat video decoder that automatically detects PAL, NTSC, and SECAM standards in the form of composite, S-Video, and component video. The ADV7182 can receive composite signals in either single-ended or differential modes. This makes the ADV7182 ideal for automotive applications.

The ADV7182 converts these analog video formats into a digital 8-bit ITU-R BT. 656 video stream.
The digital video output stream of the ADV7182 interfaces easily to a wide range of MPEG encoders, codecs, mobile video processors, and Analog Devices, Inc., digital video encoders, such as the ADV7391. External HS, VS, and FIELD signals provide timing references for LCD controllers and other video ASICs.

## OVERVIEW OF THE ANALOG FRONT END

The ADV7182 analog front end (AFE) comprises a single high speed, 10-bit ADC that digitizes the analog video signal before applying it to the standard definition processor. The AFE employs differential channels to the ADC to ensure high performance in mixed-signal applications and to enable differential CVBS to be connected directly to the ADV7182.
The front end also includes a 4-channel input mux that enables multiple composite video signals to be applied to the ADV7182. Current clamps are positioned in front of the ADC to ensure that the video signal remains within the range of the converter. A resistor divider network is required before each analog input channel to ensure that the input signal is kept within the range of the ADC (see Figure 23). The choice of this resistor divider ratio provides a common-mode range of up to 4 V . Fine clamping of the video signal is performed downstream by digital fine clamping within the ADV7182.
Table 1 shows the three ADC clocking rates that are determined by the video input format to be processed. These clock rates ensure $4 \times$ oversampling per channel for CVBS, $\mathrm{Y} / \mathrm{C}$, and YPrPb modes.

The ADV7182 has a fully differential AFE. This allows for inherent small and large signal noise rejection, improved electromagnetic interference (EMI), and the ability to absorb ground bounce. Support is offered for both true differential and pseudodifferential signals.

Table 1. ADC Clock Rates

| Input Format | ADC Clock Rate (MHz) | Oversampling <br> Rate per Channel |
| :--- | :--- | :--- |
| CVBS | 57.27 | $4 \times$ |
| Y/C (S-Video) | 114 | $4 \times$ |
| YPrPb $^{2}$ | 172 | $4 \times$ |

${ }^{1}$ Based on a 28.63636 MHz clock input to the ADV7182.
${ }^{2}$ See INSEL[4:0] in Table 95 for writes needed to set Y/C (S-Video) and YPrPb modes.

## OVERVIEW OF THE STANDARD DEFINITION PROCESSOR

The ADV7182 is capable of decoding a large selection of baseband video signals in composite (both single-ended and differential), S-Video, and component formats. The video standards supported by the video processor include PAL B/PAL D/PAL I/PAL G/PAL H, PAL 60, PAL M, PAL N, PAL Nc, NTSC M/NTSC J, NTSC 4.43, and SECAM B/SECAM D/SECAM G/SECAM K/SECAM L. The ADV7182 can automatically detect the video standard and process it accordingly.
The ADV7182 has a five-line, superadaptive, 2D comb filter that gives superior chrominance and luminance separation when decoding a composite video signal. This highly adaptive filter automatically adjusts its processing mode according to the video standard and signal quality without requiring user intervention. Video user controls such as brightness, contrast, saturation, and hue are also available with the ADV7182.

The ADV7182 implements a patented ADLLT ${ }^{\text {pm }}$ algorithm to track varying video line lengths from sources such as a VCR. ADLLT enables the ADV7182 to track and decode poor quality video sources such as VCRs and noisy sources from tuner outputs,

VCD players, and camcorders. The ADV7182 contains a chroma transient improvement (CTI) processor that sharpens the edge rate of chroma transitions, resulting in sharper vertical transitions.
The ACE offers improved visual detail using an algorithm that automatically varies contrast levels to enhance picture detail. This enables the contrast in dark areas of an image to be increased without saturating the bright areas of an image. This is particularly useful in automotive applications, where it can be important to be able to discern objects in shaded areas.
Down dithering from eight bits to six bits enables ease of design for standard LCD panels.
The video processor can process a variety of VBI data services, such as closed captioning (CCAP), wide screen signaling (WSS), copy generation management system (CGMS), and teletext data slicing for world standard teletext (WST). Data is transmitted via the 8 -bit video output port as ancillary data packets (ANC). The ADV7182 is fully Macrovision ${ }^{*}$ certified; detection circuitry enables Type I, Type II, and Type III protection levels to be identified and reported to the user. The decoder is also fully robust to all Macrovision signal inputs.

## SPECIFICATIONS

## ELECTRICAL CHARACTERISTICS

$A_{\mathrm{VDD}}, \mathrm{D}_{\mathrm{VDD}}$, and $\mathrm{P}_{\mathrm{VDD}}=1.71 \mathrm{~V}$ to 1.89 V ; $\mathrm{D}_{\mathrm{VDDIO}}=1.62 \mathrm{~V}$ to 3.63 V , specified at operating temperature range, unless otherwise noted.
Table 2.

| Parameter | Symbol | Test Conditions/Comments | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| STATIC PERFORMANCE <br> Resolution (Each ADC) <br> Integral Nonlinearity Differential Nonlinearity | N <br> INL <br> DNL | CVBS mode CVBS mode |  | $\begin{aligned} & 2 \\ & -0.6 /+0.6 \end{aligned}$ | 10 | $\begin{aligned} & \text { Bits } \\ & \text { LSB } \\ & \text { LSB } \end{aligned}$ |
| DIGITAL INPUTS <br> Input High Voltage (DVDDIO = 3.3 V) <br> Input High Voltage ( DVDDIO $=1.8 \mathrm{~V}$ ) <br> Input Low Voltage (DVDDIO $=3.3 \mathrm{~V}$ ) <br> Input Low Voltage (DVDDIO $=1.8 \mathrm{~V}$ ) <br> Crystal Inputs <br> Input Leakage Current <br> Input Leakage Current (SDATA, SCLK) <br> Input Leakage Current ( $\overline{\text { PWRDWN, ALSB) }}$ Input Capacitance | $\mathrm{V}_{\mathrm{H}}$ <br> $\mathrm{V}_{\mathrm{IH}}$ <br> VIL <br> VII <br> $\mathrm{V}_{\mathrm{H}}$ <br> VII <br> In <br> In <br> In <br> CIN |  | $\begin{aligned} & 2 \\ & 1.2 \\ & 1.2 \\ & -10 \\ & -10 \\ & -10 \end{aligned}$ |  | $\begin{aligned} & 0.8 \\ & 0.4 \\ & \\ & 0.4 \\ & +10 \\ & +15 \\ & +48 \\ & 10 \end{aligned}$ | $\begin{aligned} & \mathrm{V} \\ & \mathrm{~V} \\ & \mathrm{~V} \\ & \mathrm{~V} \\ & \mathrm{~V} \\ & \mathrm{~V} \\ & \mu \mathrm{~A} \\ & \mu \mathrm{~A} \\ & \mu \mathrm{~A} \\ & \mathrm{pF} \end{aligned}$ |
| DIGITAL OUTPUTS <br> Output High Voltage (DVDDIO $=3.3 \mathrm{~V}$ ) <br> Output High Voltage (DVDDIO $=1.8 \mathrm{~V}$ ) <br> Output Low Voltage (DVDDIO $=3.3 \mathrm{~V}$ ) <br> Output Low Voltage (DVDDIO $=1.8 \mathrm{~V}$ ) <br> High Impedance Leakage Current <br> Output Capacitance | Voн <br> Voн <br> Voı <br> Vol <br> leak <br> Cout | $\begin{aligned} & \mathrm{I}_{\text {SOURCE }}=0.4 \mathrm{~mA} \\ & \mathrm{I}_{\text {SOURCE }}=0.4 \mathrm{~mA} \\ & \mathrm{I}_{\text {SINK }}=3.2 \mathrm{~mA} \\ & \mathrm{I}_{\text {SINK }}=1.6 \mathrm{~mA} \end{aligned}$ | $\begin{aligned} & 2.4 \\ & 1.4 \end{aligned}$ |  | $\begin{aligned} & 0.4 \\ & 0.2 \\ & 10 \\ & 20 \end{aligned}$ | $\begin{aligned} & \mathrm{V} \\ & \mathrm{~V} \\ & \mathrm{~V} \\ & \mathrm{~V} \\ & \mu \mathrm{~A} \\ & \mathrm{pF} \end{aligned}$ |
| POWER REQUIREMENTS ${ }^{1,2}$ <br> Digital I/O Power Supply PLL Power Supply Analog Power Supply Digital Power Supply Digital I/O Supply Current PLL Supply Current Analog Supply Current <br> Digital Supply Current | Dvido <br> Pvid <br> Avdd <br> Dvod <br> Idvodio <br> Ipvod <br> IAvDD <br> IDvDD | Single-ended CVBS input <br> Differential CVBS input <br> Single-ended CVBS fast switch <br> Differential CVBS fast switch <br> Y/C input <br> YPrPb input <br> Single-ended CVBS input <br> Differential CVBS input <br> Single-ended CVBS fast switch <br> Differential CVBS fast switch <br> Y/C input <br> YPrPb input | $\begin{aligned} & 1.62 \\ & 1.71 \\ & 1.71 \\ & 1.71 \end{aligned}$ | $\begin{aligned} & 3.3 \\ & 1.8 \\ & 1.8 \\ & 1.8 \\ & 3 \\ & 12 \\ & 47 \\ & 69 \\ & 47 \\ & 69 \\ & 60 \\ & 75 \\ & 60 \\ & 66 \\ & 60 \\ & 66 \\ & 60 \\ & 60 \end{aligned}$ | $\begin{aligned} & 3.63 \\ & 1.89 \\ & 1.89 \\ & 1.89 \end{aligned}$ | V <br> V <br> V <br> V <br> mA <br> mA <br> mA <br> mA <br> mA <br> mA <br> mA <br> mA <br> mA <br> mA <br> mA <br> mA <br> mA <br> mA |


| Parameter | Symbol | Test Conditions/Comments | Min | Typ | Max |
| :--- | :--- | :--- | :--- | :--- | :--- | Unit

${ }^{1}$ Guaranteed by characterization.
${ }^{2}$ Typical current consumption values are recorded with nominal voltage supply levels and an SMPTEBAR test pattern.

## VIDEO SPECIFICATIONS

Guaranteed by characterization. $\mathrm{A}_{\mathrm{VDD}}, \mathrm{D}_{\mathrm{VDD}}$, and $\mathrm{P}_{\mathrm{VDD}}=1.71 \mathrm{~V}$ to $1.89 \mathrm{~V} ; \mathrm{D}_{\mathrm{VDDIO}}=1.62 \mathrm{~V}$ to 3.63 V , specified at operating temperature range, unless otherwise noted.

Table 3.

| Parameter | Symbol | Test Conditions/Comments | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| NONLINEAR SPECIFICATIONS ${ }^{1}$ <br> Differential Phase <br> Differential Gain <br> Luma Nonlinearity | DP <br> DG <br> LNL | CVBS input, modulate five-step CVBS input, modulate five-step CVBS input, five-step |  | $\begin{aligned} & 0.9 \\ & 0.5 \\ & 2.0 \end{aligned}$ |  | Degrees <br> \% <br> \% |
| NOISE SPECIFICATIONS <br> SNR Unweighted <br> Analog Front-End Crosstalk Common-Mode Rejection ${ }^{2}$ | CMR | Luma ramp Luma flat field |  | $\begin{aligned} & 57 \\ & 58 \\ & 60 \\ & 75 \end{aligned}$ |  | $\begin{aligned} & \mathrm{dB} \\ & \mathrm{~dB} \\ & \mathrm{~dB} \\ & \mathrm{~dB} \end{aligned}$ |
| LOCK TIME SPECIFICATIONS <br> Horizontal Lock Range Vertical Lock Range fsc Subcarrier Lock Range Color Lock-In Time Sync Depth Range Color Burst Range Vertical Lock Time Autodetection Switch Speed ${ }^{3}$ Fast Switch Speed ${ }^{4}$ |  |  | $\begin{aligned} & -5 \\ & 40 \\ & 20 \\ & 5 \end{aligned}$ | $\pm 1.3$ <br> 60 $2$ $100$ $100$ | $\begin{aligned} & +5 \\ & 70 \end{aligned}$ $200$ $200$ | \% <br> Hz <br> kHz <br> Lines <br> \% <br> \% <br> Fields <br> Lines <br> ms |
| LUMA SPECIFICATIONS Luma Brightness Accuracy Luma Contrast Accuracy |  | CVBS, 1 V input CVBS, 1 V input |  | $1$ |  |  |

[^0]
## ADV7182

## TIMING SPECIFICATIONS

Guaranteed by characterization. $\mathrm{A}_{\mathrm{VDD}}, \mathrm{D}_{\mathrm{VDD}}$, and $\mathrm{P}_{\mathrm{VDD}}=1.71 \mathrm{~V}$ to 1.89 V ; $\mathrm{D}_{\mathrm{VDDIO}}=1.62 \mathrm{~V}$ to 3.63 V , specified at operating temperature range, unless otherwise noted.

Table 4.

| Parameter | Symbol | Test Conditions/Comments | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SYSTEM CLOCK AND CRYSTAL <br> Nominal Frequency Frequency Stability |  |  |  | $28.63636$ | $\pm 50$ | MHz <br> ppm |
| $1^{2} \mathrm{C}$ PORT <br> SCLK Frequency <br> SCLK Minimum Pulse Width High SCLK Minimum Pulse Width Low Hold Time (Start Condition) Setup Time (Start Condition) SDATA Setup Time SCLK and SDATA Rise Times SCLK and SDATA Fall Times Setup Time for Stop Condition | $\begin{aligned} & \mathrm{t}_{1} \\ & \mathrm{t}_{2} \\ & \mathrm{t}_{3} \\ & \mathrm{t}_{4} \\ & \mathrm{t}_{5} \\ & \mathrm{t}_{6} \\ & \mathrm{t}_{7} \\ & \mathrm{t}_{8} \\ & \hline \end{aligned}$ |  | $\begin{aligned} & 0.6 \\ & 1.3 \\ & 0.6 \\ & 0.6 \\ & 100 \end{aligned}$ | $0.6$ | $400$ $\begin{aligned} & 300 \\ & 300 \end{aligned}$ | kHz <br> $\mu \mathrm{s}$ <br> $\mu \mathrm{s}$ <br> $\mu \mathrm{s}$ <br> $\mu \mathrm{s}$ <br> ns <br> ns <br> ns <br> $\mu \mathrm{s}$ |
| $\overline{\text { RESET FEATURE }}$ $\overline{\text { RESET Pulse Width }}$ |  |  | 5 |  |  | ms |
| CLOCK OUTPUTS <br> LLC Mark Space Ratio | t9: $\mathrm{t}_{10}$ |  | 45:55 |  | 55:45 | \% duty cycle |
| DATA AND CONTROL OUTPUTS Data Output Transitional Time | $t_{11}$ $t_{12}$ | Negative clock edge to start of valid data ( $\mathrm{t}_{\text {ACCESS }}=\mathrm{t}_{10}-\mathrm{t}_{11}$ ) <br> End of valid data to negative clock edge $\left(\mathrm{t}_{\text {HoLD }}=\mathrm{t}_{9}+\mathrm{t}_{12}\right)$ |  |  | 3.8 6.9 | ns |

## Timing Diagrams



Figure 3. Pixel Port and Control Output Timing

## Data Sheet

## ANALOG SPECIFICATIONS

Guaranteed by characterization. $\mathrm{A}_{\mathrm{VDD}}, \mathrm{D}_{\mathrm{VDD}}$, and $\mathrm{P}_{\mathrm{VDD}}=1.71 \mathrm{~V}$ to $1.89 \mathrm{~V} ; \mathrm{D}_{\mathrm{VDDIO}}=1.62 \mathrm{~V}$ to 3.63 V , specified at operating temperature range, unless otherwise noted.

Table 5.

| Parameter | Test Conditions/Comments | Min | Typ |
| :--- | :--- | :--- | :--- |
| CLAMP CIRCUITRY |  |  | Max |
| External Clamp Capacitor |  | 0.1 |  |
| Input Impedance | Clamps switched off | 10 | $\mu \mathrm{~F}$ |
| Large Clamp Source Current |  | 0.32 | $\mathrm{M} \Omega$ |
| Large Clamp Sink Current |  | 0.32 | mA |
| Fine Clamp Source Current |  | 7 | mA |
| Fine Clamp Sink Current |  | 7 | $\mu \mathrm{~A}$ |

## THERMAL SPECIFICATIONS

Table 6.

| Parameter | Symbol | Test Conditions/Comments | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| THERMAL CHARACTERISTICS |  |  |  |  |  |  |
| Junction-to-Ambient Thermal Resistance (Still Air) | $\theta_{\mathrm{JA}}$ | 4-layer printed circuit board (PCB) with solid ground plane, 32-lead LFCSP |  | 32.5 |  | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| Junction-to-Case Thermal Resistance | $\theta \mathrm{fc}$ | 4-layer PCB with solid ground plane, 32-lead LFCSP |  | 2.3 |  | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |

## ABSOLUTE MAXIMUM RATINGS

Table 7.

| Parameter $^{1}$ | Rating |
| :--- | :--- |
| AVDD to GND | 2.2 V |
| DVDD to GND | 2.2 V |
| PVDD to GND | 2.2 V |
| DVDDIO to GND | 4 V |
| PVDD to DVDD | -0.9 V to +0.9 V |
| AVDD to DVDD | -0.9 V to +0.9 V |
| Digital Inputs Voltage | $\mathrm{GND}-0.3 \mathrm{~V}$ to DVDDIO +0.3 V |
| Digital Outputs Voltage | $\mathrm{GND}-0.3 \mathrm{~V}$ to DVDDIO +0.3 V |
| Analog Inputs to Ground | $\mathrm{GND}-0.3 \mathrm{~V}$ to AVDD +0.3 V |
| Maximum Junction <br> $\quad$ Temperature (Tر max) | $140^{\circ} \mathrm{C}$ |
| Storage Temperature Range <br> Infrared Reflow Soldering <br> $(20$ sec) | $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$ |

This device is a high performance integrated circuit with an ESD rating of $<2 \mathrm{kV}$, and it is ESD sensitive. Proper precautions must be taken for handling and assembly.

## ESD CAUTION

ESD (electrostatic discharge) sensitive device.
Charged devices and circuit boards can discharge
without detection. Although this product features
patented or proprietary protection circuitry, damage
may occur on devices subjected to high energy ESD.
Therefore, proper ESD precautions should be taken to
avoid performance degradation or loss of functionality.
${ }^{1}$ The Absolute Maximum Ratings assume that DGND pins and the exposed pad of the ADV7182 are connected together to a common ground plane (GND). This is part of the recommended layout scheme. See the PCB Layout Recommendations section for more information. The Absolute Maximum Ratings are stated in relation to this common ground plane.
Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## PIN CONFIGURATION AND FUNCTION DESCRIPTIONS



Figure 4. Pin Configuration
Table 8. Pin Function Descriptions

| Pin No. | Mnemonic | Type | Description |
| :---: | :---: | :---: | :---: |
| 1,4 | DGND | G | Ground for Digital Supply. |
| 2 | DVDDIO | P | Digital I/O Supply Voltage (1.8 V to 3.3 V ). |
| 3,13 | DVDD | P | Digital Supply Voltage (1.8V). |
| 5 to 12 | P7 to P0 | 0 | Video Pixel Output Port. |
| 14 | XTALP | 0 | This pin should be connected to the 28.6363 MHz crystal or not connected if an external $1.8 \mathrm{~V}, 28.6363 \mathrm{MHz}$ clock oscillator source is used to clock the ADV7182. In crystal mode, the crystal must be a fundamental crystal. |
| 15 | XTALN | I | Input Pin for the 28.6363 MHz Crystal. This pin can be overdriven by an external 1.8 V , 28.6363 MHz clock oscillator source. In crystal mode, the crystal must be a fundamental crystal. |
| 16 | PVDD | P | PLL Supply Voltage (1.8V). |
| 17, 18, 22, 23 | $A_{\text {In }} 1$ to $\mathrm{A}_{\text {IN }} 4$ | 1 | Analog Video Input Channels. |
| 19 | VREFP | 0 | Internal Voltage Reference Output. |
| 20 | VREFN | 0 | Internal Voltage Reference Output. |
| 21 | AVDD | P | Analog Supply Voltage (1.8V). |
| 24 | $\overline{\text { INTRQ }}$ | 0 | Interrupt Request Output. Interrupt occurs when certain signals are detected on the input video. |
| 25 | $\overline{\text { RESET }}$ | 1 | System Reset Input. Active low. A minimum low reset pulse width of 5 ms is required to reset the ADV7182 circuitry. |
| 26 | ALSB | 1 | This pin selects the $I^{2} \mathrm{C}$ address for the ADV7182. For ALSB set to Logic 0 , the address selected for a write is $0 \times 40$; for ALSB set to Logic 1 , the address selected is $0 \times 42$. |
| 27 | SDATA | I/O | $1^{2} \mathrm{C}$ Port Serial Data Input/Output Pin. |
| 28 | SCLK | 1 | $1^{2} \mathrm{C}$ Port Serial Clock Input. The maximum clock rate is 400 kHz . |
| 29 | VS/FIELD/SFL | 0 | Vertical Synchronization Output Signal/Field Synchronization Output Signal/Subcarrier Frequency Lock. This pin contains a serial output stream that can be used to lock the subcarrier frequency when this decoder is connected to any Analog Devices digital video encoder. |
| 30 | HS | 0 | Horizontal Synchronization Output Signal. |
| 31 | $\overline{\text { PWRDWN }}$ | 1 | A logic low on this pin places the ADV7182 is in power-down mode. |
| 32 | LLC EPAD (EP) | 0 | Line-Locked Output Clock for Output Pixel Data. Nominally 27 MHz but varies up or down according to the video line length. <br> The exposed pad must be connected to DGND. |

## POWER SUPPLY SEQUENCING

OPTIMAL POWER-UP SEQUENCE
The optimal power-up sequence for the ADV7182 is to first power up the $3.3 \mathrm{~V} \mathrm{D}_{\text {VDDIo }}$ supply, followed by the 1.8 V supplies (Dvid, Pvid, and Avdd).
When powering up the ADV7182, follow these steps. During power-up, all supplies must adhere to the specifications listed in the Absolute Maximum Ratings section.

1. Assert the $\overline{\text { PWRDWN }}$ pin and the $\overline{\text { RESET }}$ pins (pull the pins low).
2. Power up the $\mathrm{D}_{\text {vddio }}$ supply.
3. After $\mathrm{D}_{\text {vDDio }}$ is fully asserted, power up the 1.8 V supplies.
4. After the 1.8 V supplies are fully asserted, pull the $\overline{\text { PWRDWN }}$ pin high.
5. Wait 5 ms and then pull the $\overline{\mathrm{RESET}}$ pin high.
6. After all power supplies, the $\overline{\text { PWRDWN }}$ pin, and the $\overline{\text { RESET }}$ pin are powerd up and stable, wait an additional 5 ms before initiating $\mathrm{I}^{2} \mathrm{C}$ communication with the ADV7182.

## SIMPLIFIED POWER-UP SEQUENCE

Alternatively, the ADV7182 can be powered up by asserting all supplies, the PWRDWN pin, and the RESET pins simultaneously. After this operation, perform a software reset, then wait 10 ms before initiating $\mathrm{I}^{2} \mathrm{C}$ communication with the ADV7182.

While the supplies are being established, take care to ensure that a lower rated supply does not go above a higher rated supply level. During power-up, all supplies must adhere to the specifications listed in the Absolute Maximum Ratings section.

## POWER-DOWN SEQUENCE

The ADV7182 supplies can be deasserted simultaneously as long as DvdDio does not go below a lower rated supply.

## UNIVERSAL POWER SUPPLY

It is possible to power all the supplies ( $\mathrm{D}_{\mathrm{VDD}}, \mathrm{P}_{\mathrm{VDD}}, \mathrm{A}_{\mathrm{VDD}}$, and $\mathrm{D}_{\text {vDiof }}$ ) to 1.8 V . In this case, apply the power-up sequences as described in the Optimal Power-Up Sequence section and the Simplified Power-Up Sequence section. The only change is that Dvddo is powered up to 1.8 V instead of 3.3 V .
In this setup, note the following:

- Power up the $\overline{\text { PWRDWN }}$ pin and the $\overline{\text { RESET }}$ pin to 1.8 V instead of 3.3 V .
- Set the drive strengths of the digital outputs of the ADV7182 to their maximum setting. See the Global Pin Control section.
- Connect any pull-up resistors connected to pins on the ADV7182 (such as the SCLK pin and the SDATA pin) to 1.8 V and not 3.3 V .


Figure 5. Recommended Power-Up Sequence

## INPUT NETWORKS

This section describes the input networks (external resistor and capacitor circuits) that should be placed on the analog video input ( $\mathrm{A}_{\text {IN }}$ ) pins of the ADV7182. Different input networks are required for different analog input video formats.

## SINGLE-ENDED INPUT NETWORK

Use the input network described in Figure 6 on each AIN input pin of the ADV7182 when any of the following video input formats are used: single-ended CVBS, YC (S-Video), or YPrPb.
It is recommended that the input network circuit shown in Figure 6 be placed as close as possible to the $\mathrm{A}_{\text {IN }}$ pins of the ADV7182.


Figure 6. Input Single-Ended Network
The $24 \Omega$ and $51 \Omega$ resistors supply the $75 \Omega$ end termination required for the analog video input. In addition, these resistors create a resistor divider with a 0.68 gain that attenuates the amplitude of the inputted analog video and scales the input to the ADC range of the ADV7182. This allows the ADV7182 to have an input range of up to 1.47 V p-p.
Note that amplifiers within the ADV7182 restore the amplitude of the input signal so that signal-to-noise (SNR) performance is maintained.

The 100 nF ac coupling capacitor removes the dc bias of the analog input video before it is fed into the analog input pins of the ADV7182.

The clamping circuitry within the ADV7182 restores the dc bias of the input signal to the optimal level before it is fed into the ADC of the ADV7182. See the Clamp Operation section for more information.

## DIFFERENTIAL INPUT NETWORK

Use the input network described in Figure 7 when differential CVBS video is input on the $\mathrm{A}_{\mathrm{IN}}$ input pins of the ADV7182.
It is recommended that the input network circuit shown in Figure 7 be placed as close as possible to the $\mathrm{A}_{\text {IN }}$ pins of the ADV7182.


Figure 7. Input Differential Network

Differential video transmission involves transmitting two complementary CVBS signals. It has several key advantages over single-ended transmission, some of which include the following:

- Inherent small signal and large signal noise rejection
- Improved EMI performance
- Ability to absorb ground bounce

Resistor R1 provides the RF end termination for the differential CVBS input lines. For a pseudo differential CVBS input, a value of $75 \Omega$ is recommended for R1. For a fully differential CVBS input, a value of $150 \Omega$ is recommended for R1.

The $1.3 \mathrm{k} \Omega$ and $430 \Omega$ resistors provide a resistor divider with a 0.25 gain. This results in an attenuation of the inputted analog video but also an increase in the input common-mode range of the ADV7182 of up to 4 V p-p.
Note that amplifiers within the ADV7182 restore the amplitude of the input signal so that SNR performance is maintained.

The 100 nF ac coupling capacitor removes the dc bias of the analog input video before it is fed into the analog video input pins of the ADV7182.
The clamping circuitry within the ADV7182 restores the dc bias of the optimized level before it is fed into the ADC of the ADV7182. See the Clamp Operation section for further information.

The combination of the $1.3 \mathrm{k} \Omega$ and $430 \Omega$ resistors and the 100 nF ac coupling capacitor limits current flow into the ADV7182 during short-to-battery (STB) events. See Short-to-Battery Protection section.

To achieve optimal performance, closely match the $1.3 \mathrm{k} \Omega$ and $430 \Omega$ resistors; that is, all the $1.3 \mathrm{k} \Omega$ and $430 \Omega$ resistors should have the same resistance tolerance, and this tolerance should be as low as possible.

## SHORT-TO-BATTERY PROTECTION

In differential mode, the ADV7182 is protected against short-tobattery (STB) events by the external 100 nF ac coupling capacitors (see Figure 7). The external input network resistors are sized to be large enough to reduce the current flow during a STB event, but to be small enough not to effect the operation of the ADV7182.

Choose the power rating of the input network resistors to withstand the high voltages of STB events. Similarly, choose the breakdown voltage of the AC coupling capacitors to be robust to STB events.

The R1 resistor is protected because no current or limited current flows through it during an STB event.

ANALOG FRONT END


Figure 8. Manual Muxing

## INPUT CONFIGURATION

The following two steps are key for configuring the ADV7182 to correctly decode the input video.

1. Use INSEL[4:0] to configure the routing and format decoding (CVBS, Y/C, or YPrPb).
2. If the input requirements are not met using the INSEL[4:0] options, the analog input muxing section must be configured manually to correctly route the video from the analog input pins to the ADC. The standard definition processor block, which decodes the digital data, should be configured to process the CVBS, $\mathrm{Y} / \mathrm{C}$, or YPrPb format. This is performed by INSEL[4:0] selection.

## INSEL[4:0], Input Control, Address 0x00[4:0]

The INSEL bits allow the user to select the input format. They also configure the standard definition processor core to process CVBS, differential CVBS, S-Video (Y/C), or component (YPrPb) format.
INSEL[4:0] has predefined analog input routing schemes that do not require manual mux programming (see Table 9). This allows the user to route the various video signal types to the decoder and select them using INSEL[4:0] only. The added benefit is that if, for example, the CVBS input is selected, the remaining channels are powered down.

Table 9. INSEL[4:0]

| INSEL[4:0] | Video Format | Analog Input |
| :---: | :---: | :---: |
| 00000 | CVBS | CVBS input on AIN1 |
| 00001 | CVBS | CVBS input on $\mathrm{AlN}^{2}$ |
| 00010 | CVBS | CVBS input on Als 3 |
| 00011 | CVBS | CVBS input on AIn4 |
| 01000 | Y/C (S-Video) | Y input on $\mathrm{Als}^{1}$ |
|  |  | C input on $\mathrm{AlN}^{\text {2 }}$ |
| 01001 | Y/C (S-Video) | Y input on $\mathrm{A}_{\text {IN }} 3$ |
|  |  | C input on Ain 4 |
| 01100 | YPrPb | Y input on $\mathrm{A}_{\text {IN }} 1$ |
|  |  | Pb input on $\mathrm{AlN}^{2}$ |
|  |  | Pr input on Ain3 |
| 01110 | Differential CVBS | Positive on $\mathrm{Aln}^{1}$ |
|  |  | Negative on $\mathrm{AlN}^{2}$ |
| 01111 | Differential CVBS | Positive on $\mathrm{A}_{\text {In }} 3$ |
|  |  | Negative on Ais4 |

## ANALOG INPUT MUXING

The ADV7182 has an integrated analog muxing section that allows more than one source of video signal to be connected to the decoder.
A maximum of four CVBS inputs can be connected to and decoded by the ADV7182. As shown in the Pin Configuration and Function Description section, these analog input pins lie in close proximity to one another, which requires careful design of the PCB layout. For example, route ground shielding between all signals through tracks that are physically close together. It is strongly recommended that any unused analog input pins be connected to AGND to act as a shield.

## MAN_MUX_EN, Manual Input Muxing Enable, Address 0xC4[7]

To configure the ADV7182 analog muxing section, the user must select the analog input $\mathrm{A}_{\mathbb{I N}} 1$ to $\mathrm{A}_{\mathbb{N}} 8$ that is to be processed by the ADC. MAN_MUX_EN must be set to 1 to enable the following muxing blocks:

- MUX0[2:0], ADC mux configuration, Address 0xC3[2:0]
- MUX1[2:0], ADC mux configuration, Address 0xC3[6:4]
- MUX2[2:0], ADC mux configuration, Address 0xC4[2:0]
- MUX3[2:0], ADC mux configuration, Address 0x60[2:0]

The four mux sections are controlled by the signal buses, MUX0/MUX1/MUX2/MUX3[2:0]. Table 10 explains the control words used.

The input signal that contains the timing information (HS and VS) must be processed by MUX0. For example, in a Y/C input configuration, connect MUX0 to the Y channel and MUX1 to the C channel. When one or more muxes are not used to process video, such as the CVBS input, the idle mux and associated channel clamps and buffers should be powered down (see the description of Register 0x3A in Table 95).

Table 10. Manual Mux Settings for ADC (MAN_MUX_EN Must be Set to 1)

| MUXO[2:0] | ADC Connected To | MUX1[2:0] | ADC Connected To | MUX2[2:0] | ADC Connected To | MUX3[2:0] | ADC Connected To |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 000 | No connect | 000 | No connect | 000 | No connect | 000 | No connect |
| 001 | $\mathrm{AlN}^{1}$ | 001 | No connect | 001 | No connect | 001 | No connect |
| 010 | Ain2 | 010 | Ain2 | 010 | Ain2 | 010 | Ain2 |
| 011 | $\mathrm{AlN}^{3}$ | 011 | No connect | 011 | $\mathrm{AlN}^{3}$ | 011 | No connect |
| 100 | AlN 4 | 100 | AIN 4 | 100 | No connect | 100 | AIN 4 |

## Notes:

- CVBS can only be processed by MUX0.
- Differential CVBS can only be processed by MUX0 (positive channel) and MUX3 (negative channel).
- Y/C can only be processed by MUX0 and MUX1.
- YPrPb can only be processed by MUX0, MUX1, and MUX2.


## ANTIALIASING FILTERS

The ADV7182 has optional on-chip antialiasing (AA) filters on each of the four channels that are multiplexed to the ADC (see Figure 9). The filters are designed for standard definition video up to 10 MHz bandwidth. Figure 10 and Figure 11 show the filter magnitude and phase characteristics.
The antialiasing filters are enabled by default and the selection of INSEL[4:0] determines which filters are powered up at any given time. For example, if CVBS mode is selected, the filter circuits for the remaining input channels are powered down to conserve power. However, the antialiasing filters can be disabled or bypassed using the AA_FILT_MAN_OVR control.


AA_FILT_MAN_OVR, Antialiasing Filter Override, Address 0xF3[4]

This feature allows the user to override the antialiasing filters on/off settings, which are automatically selected by INSEL[4:0].

## AA_FILT_EN[3:0], Antialiasing Filter Enable, Address 0xF3[3:0]

These bits allow the user to enable or disable the antialiasing filters on each of the four input channels multiplexed to the ADC. When disabled, the analog signal bypasses the AA filters and is routed directly to the ADC.

## AA_FILT_EN[0], Antialiasing Filter Enable, Address 0xF3[0]

When AA_FILT_EN[0] is 0 , AA Filter 1 is disabled.
When AA_FILT_EN[0] is 1, AA Filter 1 is enabled.

## AA_FILT_EN[1], Antialiasing Filter Enable, Address 0xF3[1]

When AA_FILT_EN[1] is 0, AA Filter 2 is disabled.
When AA_FILT_EN[1] is 1, AA Filter 2 is enabled.

## AA_FILT_EN[2], Antialiasing Filter Enable, Address 0xF3[2]

When AA_FILT_EN[2] is 0 , AA Filter 3 is disabled.
When AA_FILT_EN[2] is 1, AA Filter 3 is enabled.

## AA_FILT_EN[3], Antialiasing Filter Enable, Address 0xF3[3]

When AA_FILT_EN[3] is 0 , AA Filter 4 is disabled.
When AA_FILT_EN[3] is 1, AA Filter 4 is enabled.


Figure 10. Antialiasing Filter Magnitude Response


Figure 11. Antialiasing Filter Phase Response

## GLOBAL CONTROL REGISTERS

Register control bits listed in this section affect the whole chip.

## POWER-SAVING MODES

## Power-Down

## PWRDWN, Address 0x0F[5]

The ADV7182 can be placed into a chip-wide, power-down mode by setting the PWRDWN bit or by using the PWRDWN pin. The power-down mode stops the clock from entering the digital section of the chip, thereby freezing its operation. No $\mathrm{I}^{2} \mathrm{C}$ bits are lost during power-down mode. The PWRDWN bit also affects the analog blocks and switches them into low current modes. The $I^{2} \mathrm{C}$ interface is unaffected and remains operational in power-down mode.
When PWRDWN is 0 , the chip is operational. When PWRDWN is 1 (default), the ADV7182 is in a chip-wide, power-down mode.

## RESET CONTROL

## Reset, Chip Reset, Address 0x0F[7]

Setting this bit, which is equivalent to controlling the $\overline{\text { RESET }}$ pin on the ADV7182, issues a full chip reset. All I ${ }^{2} \mathrm{C}$ registers are reset to their default/power-up values. Note that some register bits do not have a reset value specified. They keep their last written value. Those bits are marked as having a reset value of $x$ in the register tables (see Table 95 and Table 97). After the reset sequence, the part immediately starts to acquire the incoming video signal.
After setting the reset bit (or initiating a reset via the $\overline{\operatorname{RESET}}$ pin), the part returns to the default for its primary mode of operation. All $\mathrm{I}^{2} \mathrm{C}$ bits are loaded with their default values, making this bit self-clearing. Executing a software reset takes approximately 2 ms . However, it is recommended to wait 5 ms before any further $\mathrm{I}^{2} \mathrm{C}$ writes are performed.
The $\mathrm{I}^{2} \mathrm{C}$ master controller receives a no acknowledge condition on the ninth clock cycle when chip reset is implemented (see the MPU Port Description section).
When the reset bit is 0 (default), operation is normal.
When the reset bit is 1 , the reset sequence starts.

## GLOBAL PIN CONTROL

## Tristate Output Drivers

TOD, Address 0x03[6]
This bit allows the user to tristate the output drivers of the ADV7182.
Upon setting the TOD bit, the P7 to P0, HS, and VS/FIELD/SFL pins are tristated.
The timing pins (HS and VS/FIELD/SFL pins) can be forced active via the TIM_OE bit. For more information on tristate control, see the Tristate LLC Driver and the Timing Signals Output Enable sections.
Individual drive strength controls are provided via the DR_STR_x bits.

When TOD is 0 , the output drivers are enabled.
When TOD is 1 (default), the output drivers are tristated.

## Tristate LLC Driver

TRI_LLC, Address 0x1D[7]
This bit allows the output drivers for the LLC pin of the ADV7182 to be tristated. For more information on tristate control, refer to the Tristate Output Drivers and Timing Signals Output Enable sections.

Individual drive strength controls are provided via the DR_STR_x bits.
When TRI_LLC is 0 , the LLC pin drivers work according to the DR_STR_C[1:0] setting (pin enabled).
When TRI_LLC is 1 (default), the LLC pin drivers are tristated.

## Timing Signals Output Enable

TIM_OE, Address 0x04[3]
The TIM_OE bit should be regarded as an addition to the TOD bit. Setting it high forces the output drivers for HS, VS/FIELD/SFL into the active state (that is, driving state) even if the TOD bit is set. If TIM_OE is set to low, the HS and VS/FIELD/SFL pins are tristated depending on the TOD bit. This functionality is beneficial if the decoder is used only as a timing generator. This may be the case if only the timing signals are extracted from an incoming signal or if the part is in free-run mode, where a separate chip can output a company logo, for example.

For more information on tristate control, see the Tristate Output Drivers and Tristate LLC Driver sections.
Individual drive strength controls are provided via the
DR_STR_x bits.
When TIM_OE is 0 (default), HS and VS/FIELD/SFL are tristated according to the TOD bit.

When TIM_OE is 1, HS and VS/FIELD/SFL are forced active all the time.

## VS/FIELD/SFL Sync Mux Selection

## FLD_OUT_SEL[2:0], Address 0x6B[2:0]

The FLD_OUT_SEL[2:0] bits select whether the VS/FIELD/SFL pin outputs vertical sync, horizontal sync, field sync, data enable (DE), or subcarrier frequency lock (SFL) signals.
Note that the VS/FIELD/SFL pin must be active for this selection to occur. See the Tristate Output Drivers and Tristate LLC Driver sections.

Table 11. FLD_OUT_SEL Function

| FLD_OUT_SEL[2:0] | Description |
| :--- | :--- |
| 000 | The VS/FIELD/SFL pin outputs horizontal <br> sync information. |
| 001 | The VS/FIELD/SFL pin outputs vertical sync <br> information. |
| 010 (default) | The VS/FIELD/SFL pin outputs field sync <br> information. <br> The VS/FIELD/SFL pin outputs data enable <br> (DE) information. <br> The VS/FIELD/SFL pin outputs subcarrier <br> frequency lock information. |
| 100 |  |

## HS Sync Mux Selection

HS_OUT_SEL[2:0], Address 0x6A[2:0]
The HS_OUT_SEL[2:0] bits allow the user to change the operation of the HS pin. The HS pin is set to output horizontal sync signals as the default. The user can also set the HS pin to output vertical sync, field sync, data enable (DE), or subcarrier frequency lock (SFL) information.
Note that the HS pin must be active for this selection to occur. See the Tristate Output Drivers and Tristate LLC Driver sections.

Table 12. HS_OUT_SEL Function

| HS_OUT_SEL[2:0] | Description |
| :--- | :--- |
| 000 (default) | The HS pin output horizontal sync <br> information. |
| 001 | The HS pin outputs vertical sync information. <br> 010 |
| 011 | The HS pin outputs field sync information. <br> The HS pin outputs data enable (DE) <br> information. |
| 100 | The HS pin outputs subcarrier frequency <br> lock (SFL) information. |

## Drive Strength Selection (Data)

DR_STR[1:0], Address 0xF4[5:4]
For EMC and crosstalk reasons, it may be desirable to strengthen or weaken the drive strength of the output drivers. The DR_STR[1:0] bits affect the drive strength for the pixel output pins ( $\mathrm{P}[7: 0]$ ) and the timing pins (HS and VS/FIELD/SFL).

For more information on tristate control, see the Tristate Output Drivers and Tristate LLC Driver sections.

Table 13. DR_STR Function

| DR_STR[1:0] | Description |
| :--- | :--- |
| 00 | Low drive strength $(1 \times)$ |
| 01 (default) | Medium low drive strength $(2 \times)$ |
| 10 | Medium high drive strength $(3 \times)$ |
| 11 | High drive strength $(4 \times)$ |

## Drive Strength Selection (Clock)

DR_STR_C[1:0], Address 0xF4[3:2]
The DR_STR_C[1:0] bits can be used to select the strength of the clock signal output driver (LLC pin). For more information, see the Drive Strength Selection (Data) Section.

Table 14. DR_STR_C Function

| DR_STR_C[1:0] | Description |
| :--- | :--- |
| 00 | Low drive strength $(1 \times)$ |
| 01 (default) | Medium low drive strength $(2 \times)$ |
| 10 | Medium high drive strength $(3 \times)$ |
| 11 | High drive strength $(4 \times)$ |

Drive Strength Selection ( $I^{2} C$ )
DR_STR_S[1:0], Address 0xF4[1:0]
The DR_STR_S[1:0] bits allow the user to select the strength of the $I^{2} \mathrm{C}$ signal output drivers. This affects the drive strength for the SDA and SCL pins.

Table 15. DR_STR_S Function

| DR_STR_S[1:0] | Description |
| :--- | :--- |
| 00 | Low drive strength $(1 \times)$ |
| 01 (default) | Medium low drive strength $(2 \times)$ |
| 10 | Medium high drive strength $(3 \times)$ |
| 11 | High drive strength $(4 \times)$ |

## Enable Subcarrier Frequency Lock Pin

EN_SFL_PIN, Address 0x04[1]
The EN_SFL_PIN bit enables the output of subcarrier lock information (also known as genlock) from the ADV7182 core to an encoder in a decoder/encoder back-to-back arrangement.
When the EN_SFL_PIN is 0 (default), the subcarrier frequency lock output is disabled.
When EN_SFL_PIN is 1, the subcarrier frequency lock information is presented on the SFL pin.

## GLOBAL STATUS REGISTER

Four registers provide summary information about the video decoder. The IDENT register allows the user to identify the revision code of the ADV7182. The other three registers (Address $0 \times 10$, Address $0 \times 12$, and Address $0 \times 13$ ) contain status bits from the ADV7182.

## IDENTIFICATION

## IDENT[7:0], Address 0x11[7:0]

This is the register identification of the ADV7182 revision. Table 16 describes the various versions of the ADV7182.

Table 16. IDENT CODE

| IDENT[7:0] | Description |
| :--- | :--- |
| $0 \times 40$ | Prerelease silicon |
| $0 \times 41$ | Released silicon |

STATUS 1

## Status 1[7:0], Address 0x10[7:0]

This read-only register provides information about the internal status of the ADV7182.
See the CIL[2:0], Count into Lock, Address 0x51[2:0] section and the COL[2:0], Count out of Lock, Address 0x51[5:3] section for details on timing.

Depending on the setting of the FSCLE bit, the status registers are based solely on horizontal timing information or on the horizontal timing and lock status of the color subcarrier. See the FSCLE, fSC Lock Enable, Address 0x51[7] section.
Table 17. Status 1 Function

| Status 1[7:0] | Bit Name | Description |
| :--- | :--- | :--- |
| 0 | IN_LOCK | In lock (now) |
| 1 | LOST_LOCK | Lost lock (since last read) |
| 2 | FSC_LOCK | fsc locked (now) |
| 3 | FOLLOW_PW | AGC follows peak white algorithm |
| 4 | AD_RESULT[0] | Result of autodetection |
| 5 | AD_RESULT[1] | Result of autodetection |
| 6 | AD_RESULT[2] | Result of autodetection |
| 7 | COL_KILL | Color kill active |

## STATUS 2

Status 2[7:0], Address 0x12[7:0]
Table 18. Status 2 Function

| Status 2[7:0] | Bit Name | Description |
| :--- | :--- | :--- |
| 0 | MVCS DET | Detected Macrovision color <br> striping |
| 1 | MVCS T3 | Macrovision color striping <br> protection; conforms to Type 3 <br> if high, Type 2 if low |
| 2 | MV PS DET | Detected Macrovision pseudo- <br> sync pulses |
| 3 | MV AGC DET | Detected Macrovision AGC pulses |
| 4 | LL NSTD | Line length is nonstandard |
| 5 | FSC NSTD | fsc frequency is nonstandard |
| 6 | Reserved |  |
| 7 | Reserved |  |

## STATUS 3

Status 3[7:0], Address 0x13[7:0]
Table 19. Status 3 Function

| Status 3[7:0] | Bit Name | Description |
| :--- | :--- | :--- |
| 0 | INST_HLOCK | Horizontal lock indicator <br> (instantaneous) |
| 1 | Reserved <br> SD_OP_50Hz | Flags whether 50 Hz or 60 Hz is <br> present at output <br> Reserved |
| 3 | Reserved | FREE_RUN_ACT |
| $\mathbf{4}$ | Flags if ADV7182 has entered <br> free-run mode (see Free-Run <br> Operation section) |  |
| $\mathbf{5}$ | Interlaced | Field length is correct for <br> currently selected video <br> standard <br> Interlaced video detected <br> (field sequence found) <br> Reliable sequence of <br> swinging bursts detected |
| 7 |  |  |

## AUTODETECTION RESULT

## AD_RESULT[2:0], Address 0x10[6:4]

The AD_RESULT[2:0] bits report back on the findings from the ADV7182 autodetection block. See the General Setup section for more information on enabling the autodetection block and the Autodetection of SD Modes section for more information on how to configure it.
Table 20. AD_RESULT Function

| AD_RESULT[2:0] | Description |
| :--- | :--- |
| 000 | NTSC M/NTSC J |
| 001 | NTSC 4.43 |
| 010 | PAL M |
| 011 | PAL 60 |
| 100 | PAL B/PAL G/PAL H/PAL I/PAL D |
| 101 | SECAM |
| 110 | PAL Combination N |
| 111 | SECAM 525 |

## VIDEO PROCESSOR



Figure 12. Block Diagram of Video Processor

Figure 12 shows a block diagram of the ADV7182 video processor. The ADV7182 can handle standard definition video in CVBS, $\mathrm{Y} / \mathrm{C}$, and YPrPb formats. It can be divided into a luminance and chrominance path. If the input video is of a composite type (CVBS), both processing paths are fed with the CVBS input.

## SD LUMA PATH

The input signal is processed by the following blocks:

- Luma digital fine clamp. This block uses a high precision algorithm to clamp the video signal.
- Luma filter. This block contains a luma decimation filter (YAA) with a fixed response and some shaping filters (YSH) that have selectable responses.
- Luma gain control. The AGC can operate on a variety of different modes, including gain based on the depth of the horizontal sync pulse, peak white mode, and fixed manual gain.
- Luma resample. To correct for line length errors as well as dynamic line length changes, the data is digitally resampled.
- Luma 2D comb. The 2D comb filter provides Y/C separation.
- AV code insertion. At this point, the decoded luma (Y) signal is merged with the retrieved chroma values. AV codes can be inserted (as per ITU-R BT.656).


## SD CHROMA PATH

The input signal is processed by the following blocks:

- Chroma digital fine clamp. This block uses a high precision algorithm to clamp the video signal.
- Chroma demodulation. This block employs a color subcarrier ( $\mathrm{fsc}_{\mathrm{sc}}$ ) recovery unit to regenerate the color subcarrier for any modulated chroma scheme. The demodulation block then performs an AM demodulation for PAL and NTSC, and an FM demodulation for SECAM.
- Chroma filter. This block contains a chroma decimation filter (CAA) with a fixed response and some shaping filters (CSH) that have selectable responses.
- Chroma gain control. AGC can operate on several different modes, including gain based on the color subcarrier amplitude, gain based on the depth of the horizontal sync pulse on the luma channel, or fixed manual gain.
- Chroma resample. The chroma data is digitally resampled to keep it perfectly aligned with the luma data. The resampling is done to correct for static and dynamic line length errors of the incoming video signal.
- Chroma 2D comb. The 2D, five line, superadaptive comb filter provides high quality Y/C separation in case the input signal is CVBS.
- AV code insertion. At this point, the demodulated chroma ( Cr and Cb ) signal is merged with the retrieved luma values. AV codes can be inserted (as per ITU-R BT.656).


## ACE AND DITHER PROCESSING BLOCKS

- ACE. This block offers improved visual detail by using an algorithm to automatically vary the contrast levels to enhance picture detail. See the Adaptive Contrast Enhancement section.
- Dither. When enabled, this block converts the digital output of the ADV7182 from 8-bit pixel data down to 6-bit pixel data. This function makes it easier for the ADV7182 to communicate with some LCD panels. See the Dither Function section.


## SYNC PROCESSING

The ADV7182 extracts syncs embedded in the analog input video signal. There is currently no support for external HS/VS inputs. The sync extraction is optimized to support imperfect video sources, such as VCRs with head switches. The actual algorithm used employs a coarse detection based on a threshold crossing, followed by a more detailed detection using an adaptive interpolation algorithm. The raw sync information is sent to a line length measurement and prediction block. The output of this is then used to drive the digital resampling section to ensure that the ADV7182 outputs 720 active pixels per line.

The sync processing on the ADV7182 also includes the following specialized postprocessing blocks that filter and condition the raw sync information retrieved from the digitized analog video:

- VSync processor. This block provides extra filtering of the detected VSyncs to improve vertical lock.
- HSync processor. The HSync processor is designed to filter incoming HSyncs that were corrupted by noise, providing much improved performance for video signals with a stable time base but poor SNR.

VBI DATA RECOVERY
The ADV7182 can retrieve the following information from the input video:

- Wide screen signaling (WSS)
- Copy generation management system (CGMS)
- Closed captioning (CCAP)
- Macrovision protection presence
- Teletext

The ADV7182 is also capable of automatically detecting the incoming video standard with respect to the following:

- Color subcarrier frequency
- Field rate
- Line rate

The ADV7182 can configure itself to support PAL B/PAL D/ PAL I/PAL G/PAL H, PAL M, PAL N, PAL Combination N, NTSC M/NTSC J, SECAM $50 \mathrm{~Hz} / 60 \mathrm{~Hz}$, NTSC 4.43 , and PAL 60.

## GENERAL SETUP

## Video Standard Selection

The VID_SEL[3:0] bits (Address 0x02[7:4]) allow the user to force the digital core into a specific video standard. This is not necessary under normal circumstances. The VID_SEL[3:0] bits default to an autodetection mode that supports PAL, NTSC, SECAM, and variants thereof.

## Autodetection of SD Modes

To guide the autodetect system of the ADV7182, individual enable bits are provided for each of the supported video standards. Setting the relevant bit to 0 inhibits the standard from being detected automatically. Instead, the system chooses the closest of the remaining enabled standards. The results of the autodetection block can be read back via the status registers (see the Global Status Register section for more information).
VID_SEL[3:0], Address 0x02[7:4]
Table 21. VID_SEL Function

| VID_SEL[3:0] | Description |
| :--- | :--- |
| 0000 (default) | Autodetect PAL B/PAL G/PAL H/PAL I/PAL D, <br> NTSC J (no pedestal), SECAM <br> Autodetect PAL B/PAL G/PAL H/PAL I/PAL D, <br> NTSC M (pedestal), SECAM |
| 0001 | Autodetect PAL N (pedestal), NTSC J <br> (no pedestal), SECAM <br> Autodetect PAL N (pedestal), NTSC M <br> (pedestal), SECAM |
| 0010 | NTSC J |
| 0011 | NTSC M |
| 0100 | PAL 60 |
| 0101 | NTSC 4.43 |
| 0110 | PAL B/PAL G/PAL H/PAL I/PAL D |
| 0111 | PAL N = PAL B/PAL G/PAL H/PAL I/PAL D (with |
| 1000 | pedestal) |
| 1001 | PAL M (without pedestal) |
| 1010 | PAL M |
| 1011 | PAL Combination N |
| 1100 | PAL Combination N (with pedestal) |
| 1101 | SECAM |
| 1110 | SECAM |
| 1111 |  |

## AD_SEC525_EN, SECAM 525 Autodetect Enable, Address 0x07[7]

Setting AD_SEC525_EN to 0 (default) disables the autodetection of a 525-line system with a SECAM style, FM-modulated color component.

Setting AD_SEC525_EN to 1 enables the detection of a SECAM style, FM-modulated color component.

## AD_SECAM_EN, SECAM Autodetect Enable, Address 0x07[6]

Setting AD_SECAM_EN to 0 (default) disables the autodetection of SECAM.

Setting AD_SECAM_EN to 1 enables the detection of SECAM.

## AD_N443_EN, NTSC 4.43 Autodetect Enable, Address 0x07[5]

Setting AD_N443_EN to 0 disables the autodetection of NTSC style systems with a 4.43 MHz color subcarrier.

Setting AD_N443_EN to 1 (default) enables the detection of NTSC style systems with a 4.43 MHz color subcarrier.
AD_P60_EN, PAL 60 Autodetect Enable, Address 0x07[4]
Setting AD_P60_EN to 0 disables the autodetection of PAL systems with a 60 Hz field rate.

Setting AD_P60_EN to 1 (default) enables the detection of PAL systems with a 60 Hz field rate.

## AD_PALN_EN, PAL N Autodetect Enable, Address 0x07[3]

Setting AD_PALN_EN to 0 (default) disables the detection of the PAL N standard.
Setting AD_PALN_EN to 1 enables the detection of the PAL N standard.

## AD_PALM_EN, PAL M Autodetect Enable, Address 0x07[2]

Setting AD_PALM_EN to 0 (default) disables the autodetection of PAL M.

Setting AD_PALM_EN to 1 enables the detection of PAL M.

## AD_NTSC_EN, NTSC Autodetect Enable,

 Address 0x07[1]Setting AD_NTSC_EN to 0 (default) disables the detection of standard NTSC.

Setting AD_NTSC_EN to 1 enables the detection of standard NTSC.

## AD_PAL_EN, PAL B/PAL D/PAL I/PAL G/PAL H Autodetect Enable, Address 0x07[0]

Setting AD_PAL_EN to 0 (default) disables the detection of standard PAL.

Setting AD_PAL_EN to 1 enables the detection of standard PAL.

## SFL_INV, Subcarrier Frequency Lock Inversion, Address 0x41[6]

This bit controls the behavior of the PAL switch bit in the SFL (genlock telegram) data stream. It was implemented to solve some compatibility issues with video encoders. It solves two problems.
First, the PAL switch bit is meaningful only in PAL. Some encoders (including Analog Devices encoders) also look at the state of this bit in NTSC.

Second, there was a design change in Analog Devices encoders from ADV717x to ADV719x. The older versions used the SFL (genlock telegram) bit directly, whereas the newer ones invert the bit prior to using it. The reason for this is that the inversion compensated for the one line delay of an SFL (genlock telegram) transmission.

As a result, for the ADV717x and ADV73xx encoders, the PAL switch bit in the SFL (genlock telegram) must be set to 0 for NTSC to work. For the older video encoders, the PAL switch bit in the SFL must be set to 1 to work in NTSC. If the state of the PAL switch bit is wrong, a $180^{\circ}$ phase shift occurs.
In a decoder/encoder back-to-back system in which SFL is used, this bit must be set up properly for the specific encoder used.

Setting SFL_INV to 0 (default) makes the part SFL compatible with the ADV717x and ADV73xx video encoders.
Setting SFL_INV to 1 makes the part SFL compatible with the older video encoders.

## Lock Related Controls

Lock information is presented to the user through Bits[2:0] of the Status 1 register (see the Status 1[7:0], Address 0x10[7:0] section). Figure 13 outlines the signal flow and the controls that are available to influence the way the lock status information is generated.


Figure 13. Lock Related Signal Path

## SRLS, Select Raw Lock Signal, Address 0x51[6]

Using the SRLS bit, the user can choose between two sources for determining the lock status (per Bits[1:0] in the Status 1 register). See Figure 13.

- The TIME_WIN signal is based on a line-to-line evaluation of the horizontal synchronization pulse of the incoming video. It reacts quite quickly.
- The FREE_RUN signal evaluates the properties of the incoming video over several fields, taking vertical synchronization information into account.

Setting SRLS to 0 (default) selects the FREE_RUN signal (that is, evaluate over several fields).
Setting SRLS to 1 selects the TIME_WIN signal (that is, evaluate on a line-to-line basis).

## FSCLE, $\mathrm{f}_{\mathrm{sC}}$ Lock Enable, Address 0x51[7]

The FSCLE bit allows the user to choose whether the status of the color subcarrier loop is taken into account when the overall lock status is determined and presented via Bits[1:0] in the Status 1 register. This bit must be set to 0 when operating the ADV7182 in YPrPb component mode to generate a reliable HLOCK status bit.

When FSCLE is 0 (default), the overall lock status is dependent only on horizontal sync lock.
When FSCLE is 1 , the overall lock status is dependent on horizontal sync lock and fsc lock.

## CIL[2:0], Count into Lock, Address 0x51[2:0]

CIL[2:0] determines the number of consecutive lines for which the lock condition must be true before the system switches into the locked state and reports this via Status 1[1:0]. The bit counts the value in lines of video.

Table 22. CIL Function

| CIL[2:0] | Number of Video Lines |
| :--- | :--- |
| 000 | 1 |
| 001 | 2 |
| 010 | 5 |
| 011 | 10 |
| 100 (default) | 100 |
| 101 | 500 |
| 110 | 1000 |
| 111 | 100,000 |

## COL[2:0], Count out of Lock, Address 0x51[5:3]

COL[2:0] determines the number of consecutive lines for which the out-of-lock condition must be true before the system switches into the unlocked state and reports this via Status 1[1:0]. It counts the value in lines of video.

Table 23. COL Function

| COL[2:0] | Number of Video Lines |
| :--- | :--- |
| 000 | 1 |
| 001 | 2 |
| 010 | 5 |
| 011 | 10 |
| 100 (default) | 100 |
| 101 | 500 |
| 110 | 1000 |
| 111 | 100,000 |

## COLOR CONTROLS

These registers allow the user to control picture appearance, including control of active data in the event of video being lost. These controls are independent of any other controls. For instance, brightness control is independent of picture clamping, although both controls affect the dc level of the signal.
CON[7:0], Contrast Adjust, Address 0x08[7:0]
This register allows the user to control contrast adjustment of the picture.

Table 24. CON Function

| CON[7:0] | Description |
| :--- | :--- |
| $0 \times 80$ (default) | Gain on luma channel $=1$ |
| $0 \times 00$ | Gain on luma channel $=0$ |
| $0 \times F F$ | Gain on luma channel $=2$ |

## SD_SAT_Cb[7:0], SD Saturation Cb Channel, Address 0xE3[7:0]

This register allows the user to control the gain of the Cb channel only, which in turn adjusts the saturation of the picture.

Table 25. SD_SAT_Cb Function

| SD_SAT_Cb[7:0] | Description |
| :--- | :--- |
| $0 \times 80$ (default) | Gain on Cb channel $=0 \mathrm{~dB}$ |
| $0 \times 00$ | Gain on Cb channel $=-42 \mathrm{~dB}$ |
| $0 \times F F$ | Gain on Cb channel $=+6 \mathrm{~dB}$ |

## SD_SAT_Cr[7:0], SD Saturation Cr Channel,

Address 0xE4[7:0]
This register allows the user to control the gain of the Cr channel only, which in turn adjusts the saturation of the picture.

Table 26. SD_SAT_Cr Function

| SD_SAT_Cr[7:0] | Description |
| :--- | :--- |
| $0 \times 80$ (default) | Gain on Cr channel $=0 \mathrm{~dB}$ |
| $0 \times 00$ | Gain on Cr channel $=-42 \mathrm{~dB}$ |
| 0xFF | Gain on Cr channel $=+6 \mathrm{~dB}$ |

SD_OFF_Cb[7:0], SD Offset Cb Channel, Address 0xE1[7:0]
This register allows the user to select an offset for the Cb channel only and to adjust the hue of the picture. There is a functional overlap with the HUE[7:0] register (Address 0x0B).

Table 27. SD_OFF_Cb Function

| SD_OFF_Cb[7:0] | Description |
| :--- | :--- |
| 0x80 (default) | 0 mV offset applied to the Cb channel |
| 0x00 | -312 mV offset applied to the Cb channel |
| 0xFF | +312 mV offset applied to the Cb channel |

## SD_OFF_Cr[7:0], SD Offset Cr Channel, Address 0xE2[7:0]

This register allows the user to select an offset for the Cr channel only and to adjust the hue of the picture. There is a functional overlap with the HUE[7:0] register.

Table 28. SD_OFF_Cr Function

| SD_OFF_Cr[7:0] | Description |
| :--- | :--- |
| $0 \times 80$ (default) | 0 mV offset applied to the Cr channel |
| $0 \times 00$ | -312 mV offset applied to the Cr channel |
| 0xFF | +312 mV offset applied to the Cr channel |

## BRI[7:0], Brightness Adjust, Address 0x0A[7:0]

This register controls the brightness of the video signal. It allows the user to adjust the brightness of the picture.

Table 29. BRI Function

| BRI[7:0] | Description |
| :--- | :--- |
| 0x00 (default) | Offset of the luma channel $=0$ IRE |
| $0 \times 7$ F | Offset of the luma channel $=+30$ IRE |
| $0 \times 80$ | Offset of the luma channel $=-30$ IRE |

## HUE[7:0], Hue Adjust, Address 0x0B[7:0]

This register contains the value for the color hue adjustment. It allows the user to adjust the hue of the picture.

HUE[7:0] has a range of $\pm 90^{\circ}$, with $0 \times 00$ equivalent to an adjustment of $0^{\circ}$. The resolution of $\operatorname{HUE}[7: 0]$ is $1 \mathrm{bit}=0.7^{\circ}$.
The hue adjustment value is fed into the AM color demodulation block. Therefore, it applies only to video signals that contain chroma information in the form of an AM-modulated carrier (CVBS or Y/C in PAL or NTSC). It does not affect SECAM and does not work on component video inputs ( YPrPb ).

Table 30. HUE Function

| HUE[7:0] | Description (Adjust Hue of the Picture) |
| :--- | :--- |
| $0 \times 00$ (default) | Phase of the chroma signal $=0^{\circ}$ |
| $0 \times 7 \mathrm{~F}$ | Phase of the chroma signal $=-90^{\circ}$ |
| $0 \times 80$ | Phase of the chroma signal $=+90^{\circ}$ |

## DEF_Y[5:0], Default Value Y, Address 0x0C[7:2]

When the ADV7182 loses lock on the incoming video signal or when there is no input signal, the DEF_Y[5:0] register allows the user to specify a default luma value to be output. This value is used under the following conditions:

- If the DEF_VAL_AUTO_EN bit is 1 and the ADV7182 has lost lock to the input video signal, this is the intended mode of operation (automatic mode).
- If the DEF_VAL_EN bit is 1, regardless of the lock status of the video decoder, this is a forced mode that may be useful during configuration.

The DEF_Y[5:0] values define the six MSBs of the output video. The remaining LSBs are padded with 0 s. For example, in 8 -bit mode, the output is $\mathrm{Y}[7: 0]=\left\{\mathrm{DEF}_{-} \mathrm{Y}[5: 0], 0,0\right\}$.

For DEF_Y[5:0], 0x0D (blue) is the default value for Y.
Register 0x0C has a default value of 0x36.
DEF_C[7:0], Default Value C, Address 0x0D[7:0]
The DEF_C[7:0] register complements the DEF_Y[5:0] value. It defines the four MSBs of Cr and Cb values to be output if:

- The DEF_VAL_AUTO_EN bit is set to high and the ADV7182 cannot lock to the input video (automatic mode).
- The DEF_VAL_EN bit is set to high (forced output).

The data that is finally output from the ADV7182 for the chroma side is $\mathrm{Cr}[4: 0]=\left\{\mathrm{DEF} \_\mathrm{C}[7: 4]\right\}$ and $\mathrm{Cb}[4: 0]=\left\{\mathrm{DEF} \_\mathrm{C}[3: 0]\right\}$.

For DEF_C[7:0], 0x7C (blue) is the default value for Cr and Cb .

## FREE-RUN OPERATION

Free-run mode provides the user with a stable clock and predictable data if the input signal cannot be decoded; for example, if input video is not present.

The ADV7182 automatically enters free-run mode if the input signal cannot be decoded. The user can prevent this operation by setting the DEF_VAL_AUTO_EN to 0 . When the DEF_VAL_ AUTO_EN bit is 0 , the ADV7182 outputs noise if it cannot decode the input video. It is recommended that the user keep DEF_VAL_AUTO_EN set to 1 .
The user can force free-run mode by setting the DEF_VAL_EN bit to 1 . This can be a useful tool in debugging system level issues.
The VID_SEL[3:0] bits can be used to force the video standard output in free-run mode (see the Video Standard Selection section).

The user can also specify which data is output in free-run mode with the FREE_RUN_PAT_SEL bits. The following test patterns can be set using this function:

- Single color
- Color bars
- Luma ramp
- Boundary box


## Single Color Test Pattern

In this mode, the ADV7182 device can be set to output the default luma and chroma data stored in DEF_Y and DEF_C (see the Color Controls section).

## Color Bars Test Pattern

In this mode, the ADV7182 device outputs the $100 \%$ color bars pattern.

## Luma Ramp Test Pattern

In this mode, the ADV7182 device outputs a series of vertical bars. Each vertical bar is progressively brighter than the vertical bar to its left.

## Boundary Box Test Pattern

In this mode, the ADV7182 device outputs a black screen with a 1-pixel depth white border (see Figure 14).


Figure 14. Boundary Box Free-Run Test Pattern

DEF_VAL_AUTO_EN, Default Value Automatic Enable, Address 0x0C[1]
This bit enables the ADV7182 to enter free-run mode if it cannot decode the video signal that has been input.

Table 31. DEF_VAL_AUTO_EN Function

| DEF_VAL_AUTO_EN | Description |
| :--- | :--- |
| 0 | The ADV7182 outputs noise if it loses <br> lock with the inputted video signal. <br> 1 (default) <br> The ADV7182 enters free-run mode if it <br> loses lock with the inputted video signal. |

## DEF_VAL_EN, Default Value Enable, Address 0x0C[0]

This bit forces free-run mode.
Table 32. DEF_VAL_EN Function

| DEF_VAL_EN | Description |
| :--- | :--- |
| 0 (default) | Do not force free-run mode (that is, free-run <br> mode dependent on DEF_VAL_AUTO_EN) |
| 1 | Force free-run mode |

FREE_RUN_PAT_SEL[2:0], Free Run Pattern Select, Address 0x14[2:0]
This function selects what data is output in free-run mode.
Table 33. FREE_FUN_PAT_SEL Function

| FREE_RUN_PAT_SEL | Description |
| :--- | :--- |
| 000 (default) | Single color set by DEF_C and DEF_Y <br> controls; see Color Controls section |
| 001 | $100 \%$ color bars <br> 010 |
| Luma ramp. Note that in order to display <br> properly, the DEF_C register should be <br> set to 0x88. See Color Controls section <br> Boundary box |  |
| 101 |  |

## VS_COAST_MODE[1:0], Address 0xF9[3:2]

If no video source is connected, then this function can set the video output standard during free-run mode.
If a valid input video source is connected to the ADV7182 and free-run mode is forced, the VS_COAST_MODE bits are ignored. The free-run standard will be the same as the valid inputted video standard.

Table 34. VS_COAST_MODE Function

| VS_COAST_MODE | Description |
| :--- | :--- |
| 00 (default) | The ADV7182 outputs in the same <br> standard as it did before it entered free-run <br> mode. If no valid standard was output <br> before entering free-run mode, the |
|  | ADV7182 outputs a 576i 50 Hz signal in <br> free-run mode. |
| 01 | Outputs a 576i 50 Hz signal in free-run mode. |
| 10 | Outputs a 480i 60 Hz signal in free-run mode. |
| 11 | Reserved. |

## CLAMP OPERATION

The input video is ac-coupled into the ADV7182. This has the advantage of protecting the ADV7182 from short-to-battery events. However, the dc value of the input video needs to be restored. This process is referred to as clamping the video. This section explains the general process of clamping on the ADV7182 in both singleended and differential modes. This section also shows the different ways in which a user can configure clamp operation behavior.

## Single-Ended CVBS Clamp Operation

The ADV7182 uses a combination of current sources and a digital processing block for clamping, as shown in Figure 15. The analog processing channel shown in Figure 15 is replicated four times inside the IC. Whereas only a single channel is needed for a single-ended CVBS signal, two independent channels are needed for Y/C (SVHS) type signals, and three independent channels are needed to allow component signals ( YPrPb ) to be processed.
The clamping can be divided into two sections:

- Clamping before the ADC (analog domain): current sources and voltage sources.
- Clamping after the ADC (digital domain): digital processing block.

The primary task of the analog clamping circuits is to ensure that the video signal stays within the valid 1.0 V ADC input window so that the analog-to-digital conversion can take place. The current sources in Figure 15 correct the dc level of the ac-coupled input video signal before it is fed into the ADC. The digitized data from the ADC is then fed into the video processor. The digital fine clamp block within the video processor corrects for any remaining variation in the dc level. The video processor also sends clamp control signals to the current sources. This feedback loop fine tunes the current clamp operation and compensates for any noise on the inputted video signal. This maintains the dc level of the video signal during normal operation.

## Differential CVBS Clamping Operation

The differential clamping operation works in a similar manner to the single-ended clamping operation (see Single-Ended CVBS Clamp Operation section). In differential mode, a coarse clamp pulls the positive and negative video input to a common-mode voltage $\mathrm{V}_{\text {Cml }}$ (see Figure 16). The feedback loop between the current clamps and the video processor fine tune this coarse dc offset and make the clamping robust to noise on the video input.
Note that the current clamps are controlled within a feedback loop between the AFE and the video processor, and the coarse clamps are not.


Figure 15. Single-Ended Clamping Overview


Figure 16. Differential Clamping Overview

## Clamp Operation Controls

The following sections describe the $\mathrm{I}^{2} \mathrm{C}$ signals that can be used to influence the behavior of the clamping block.

## CCLEN, Current Clamp Enable, Address 0x14[4]

The current clamp enable bit allows the user to switch off all the current sources in the AFE simultaneously. This may be useful if the incoming analog video signal is clamped externally.

When CCLEN is 0 , the current sources are switched off.
When CCLEN is 1 (default), the current sources are enabled.

## DCT[1:0], Digital Clamp Timing, Address 0x15[6:5]

The clamp timing register determines the time constant of the digital fine clamp circuitry. Note that the digital fine clamp reacts quickly because it immediately corrects any residual dc level error for the active line. The time constant from the digital fine clamp must be much quicker than the one from the analog blocks.

By default, the time constant of the digital fine clamp is adjusted dynamically to suit the currently connected input signal.

Table 35. DCT Function

| DCT[1:0] | Description |
| :--- | :--- |
| 00 (default) | Slow $(T C=1 \mathrm{sec})$ |
| 01 | Medium $(\mathrm{TC}=0.5 \mathrm{sec})$ |
| 10 | Fast $(\mathrm{TC}=0.1 \mathrm{sec})$ |
| 11 | Determined by ADV7182, depending on the |
| input video parameters |  |

DCFE, Digital Clamp Freeze Enable, Address 0x15[4]
This register bit allows users to freeze the digital clamp loop at any time (do their own clamping). Users can disable the current sources for analog clamping via the appropriate register bits, wait until the digital clamp loop settles, and then freeze it via the DCFE bit.

When DCFE is set to 0 (default), the digital clamp is operational. When DCFE is 1 , the digital clamp loop is frozen.

## LUMA FILTER

Data from the digital fine clamp block is processed by the three sets of filters that follow. The data format at this point is CVBS for CVBS input or luma only for $\mathrm{Y} / \mathrm{C}$ and YPrPb input formats.

- Luma antialias filter (YAA). The ADV7182 receives video at a rate of 28.6363 MHz . (In the case of $4 \times$ oversampled video, the ADC samples at 57.27 MHz , and the first decimation is performed inside the DPP filters. Therefore, the data rate into the ADV7182 is always 28.6363 MHz .) The ITU-R BT. 601 recommends a sampling frequency of 13.5 MHz . The luma antialias filter decimates the oversampled video using a high quality linear phase, low-pass filter that preserves the luma signal while, at the same time, attenuating out-ofband components. The luma antialias filter (YAA) has a fixed response.
- Luma shaping filters (YSH). The shaping filter block is a programmable low-pass filter with a wide variety of responses. It can be used to reduce selectively the luma video signal bandwidth (needed prior to scaling, for example). For some video sources that contain high frequency noise, reducing the bandwidth of the luma signal improves visual picture quality. If the video is low-pass filtered, a follow-on video compression stage can work more efficiently. The ADV7182 has two responses for the shaping filter: one that is used for good quality composite, component, and SVHS type sources; and a second for nonstandard CVBS signals. The YSH filter responses also include a set of notches for PAL and NTSC. However, using the comb filters for Y/C separation is recommended.
- Digital resampling filter. This block allows dynamic resampling of the video signal to alter parameters such as the time base of a line of video. Fundamentally, the resampler is a set of low-pass filters. The actual response is chosen by the system with no requirement for user intervention.

Figure 18 through Figure 21 show the overall response of all filters together. Unless otherwise noted, the filters are set into a typical wideband mode.

## Y Shaping Filter

For input signals in CVBS format, the luma shaping filters play an essential role in removing the chroma component from a composite signal. Y/C separation must aim for best possible crosstalk reduction while still retaining as much bandwidth (especially on the luma component) as possible. High quality Y/C separation can be achieved by using the internal comb filters of the ADV7182. Comb filtering, however, relies on the frequency relationship of the luma component (multiples of the video line rate) and the color subcarrier ( $\mathrm{f}_{\mathrm{sc}}$ ). For good quality CVBS signals, this relationship is known; the comb filter algorithms can be used to separate luma and chroma with high accuracy.
In the case of nonstandard video signals, the frequency relationship may be disturbed, and the comb filters may not be able to remove all crosstalk artifacts in the best fashion without the assistance of the shaping filter block.

An automatic mode is provided that allows the ADV7182 to evaluate the quality of the incoming video signal and select the filter responses in accordance with the signal quality and video standard. YFSM, WYSFMOVR, and WYSFM allow the user to manually override the automatic decisions in part or in full.

The luma shaping filter has the following control bits.

- YSFM[4:0] allows the user to manually select a shaping filter mode (applied to all video signals) or to enable an automatic selection (depending on video quality and video standard).
- WYSFMOVR allows the user to manually override the WYSFM decision.
- WYSFM[4:0] allows the user to select a different shaping filter mode for good quality composite (CVBS), component ( YPrPb ), and SVHS ( $\mathrm{Y} / \mathrm{C}$ ) input signals.

In automatic mode, the system preserves the maximum possible bandwidth for good CVBS sources (because they can be successfully combed) as well as for luma components of YPrPb and $\mathrm{Y} / \mathrm{C}$ sources (because they need not be combed). For poor quality signals, the system selects from a set of proprietary shaping filter responses that complements comb filter operation to reduce visual artifacts.

The decisions of the control logic are shown in Figure 17.

## YSFM[4:0], Y Shaping Filter Mode, Address 0x17[4:0]

The Y shaping filter mode bits allow the user to select from a wide range of low-pass and notch filters. When switched in automatic mode, the filter selection is based on other register selections, such as detected video standard, as well as properties extracted from the incoming video itself, such as quality and time base stability. The automatic selection always selects the widest possible bandwidth for the video input encountered. (See Table 36.)
The Y-shaping filter mode operates as follows:

- If the YSFM settings specify a filter (that is, YSFM is set to values other than 00000 or 00001), the chosen filter is applied to all video, regardless of its quality.
- In automatic selection mode, the notch filters are only used for bad quality video signals. For all other video signals, wideband filters are used.


## WYSFMOVR, Wideband Y Shaping Filter Override, Address 0x18[7]

Setting the WYSFMOVR bit enables the use of the WYSFM[4:0] settings for good quality video signals. For more information on luma shaping filters, see the Y Shaping Filter section and the flowchart shown in Figure 17.
When WYSFMOVR is set to 0 , the shaping filter for good quality video signals is selected automatically.
Setting WYSFMOVR is set to 1 (default), it enables manual override via WYSFM[4:0].


Figure 17. YSFM and WYSFM Control Flowchart
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Table 36. YSFM Function

| YSFM[4:0] | Description |
| :--- | :--- |
| 00000 | Automatic selection including a wide notch <br> response (PAL/NTSC/SECAM) <br> 00001 (default) <br> Automatic selection including a narrow notch <br> response (PAL/NTSC/SECAM) |
| 00010 | SVHS 1 |
| 00011 | SVHS 2 |
| 00100 | SVHS 3 |
| 00101 | SVHS 4 |
| 00110 | SVHS 5 |
| 00111 | SVHS 6 |
| 01000 | SVHS 7 |
| 01001 | SVHS 8 |
| 01010 | SVHS 9 |
| 01011 | SVHS 10 |
| 01100 | SVHS 11 |
| 01101 | SVHS 12 |
| 01110 | SVHS 13 |
| 01111 | SVHS 14 |
| 10000 | SVHS 15 |
| 10001 | SVHS 16 |
| 10010 | SVHS 17 |
| 10011 | SVHS 18 (CCIR 601) |
| 10100 | PAL NN1 |
| 10101 | PAL NN2 |
| 10110 | PAL NN3 |
| 10111 | PAL WN1 |
| 11000 | PAL WN2 |
| 11001 | NTSC NN1 |
| 11010 | NTSC NN2 |
| 11011 | NTSC NN3 |
| 11100 | NTSC WN1 |
| 11101 | NTSC WN2 |
| 11110 | NTSC WN3 |
| 11111 | Reserved |

## WYSFM[4:0], Wideband Y Shaping Filter Mode, Address 0x18[4:0]

The WYSFM[4:0] bits allow the user to manually select a shaping filter for good quality video signals, for example, CVBS with stable time base, luma component of YPrPb , and luma component of Y/C. The WYSFM bits are active only if the WYSFMOVR bit is set to 1 . See the general discussion of the shaping filter settings in the Y Shaping Filter section.

Table 37. WYSFM Function

| WYSFM[4:0] | Description |
| :--- | :--- |
| 00000 | Reserved, do not use |
| 00001 | Reserved, do not use |
| 00010 | SVHS 1 |
| 00011 | SVHS 2 |
| 00100 | SVHS 3 |
| 00101 | SVHS 4 |
| 00110 | SVHS 5 |
| 00111 | SVHS 6 |
| 01000 | SVHS 7 |
| 01001 | SVHS 8 |
| 01010 | SVHS 9 |
| 01011 | SVHS 10 |
| 01100 | SVHS 11 |
| 01101 | SVHS 12 |
| 01110 | SVHS 13 |
| 01111 | SVHS 14 |
| 10000 | SVHS 15 |
| 10001 | SVHS 16 |
| 10010 | SVHS 17 |
| 10011 (default) | SVHS 18 (CCIR 601) |
| 10100 to 11111 | Reserved, do not use |

The filter plots in Figure 18 show the SVHS 1 (narrowest) to SVHS 18 (widest) shaping filter settings. Figure 20 shows the PAL notch filter responses. The NTSC notch filter responses are shown in Figure 21.


Figure 18. Y SVHS Combined Responses

## CHROMA FILTER

Data from the digital fine clamp block is processed by the three sets of filters that follow. The data format at this point is CVBS for CVBS inputs, chroma only for $\mathrm{Y} / \mathrm{C}$, or U/V interleaved for YPrPb input formats.

- Chroma antialias filter (CAA). The ADV7182 oversamples the CVBS by a factor of 4 . A decimating filter (CAA) is used to preserve the active video band and to remove any out-ofband components. The CAA filter has a fixed response.
- Chroma shaping filters (CSH). The shaping filter block (CSH) can be programmed to perform a variety of low-pass responses. It can be used to selectively reduce the bandwidth of the chroma signal for scaling or compression.
- Digital resampling filter. This block allows dynamic resampling of the video signal to alter parameters such as the time base of a line of video. Fundamentally, the resampler is a set of low-pass filters. The actual response is chosen by the system without user intervention.

Figure 22 shows the overall response of all filters together.


Figure 19. Combined Y Antialias, CCIR Mode Shaping Filter COMBINED Y ANTIALIAS, PAL NOTCH FILTERS,


Figure 20. Combined Y Antialias, PAL Notch Filters


Figure 21. Combined Y Antialias Filter, NTSC Notch Filters
COMBINED C ANTIALIAS, C SHAPING FILTER, C RESAMPLER


Figure 22. Chroma Shaping Filter Responses
CSFM[2:0], C Shaping Filter Mode, Address 0x17[7:5]
The C shaping filter mode bits allow the user to select from a range of low-pass filters for the chrominance signal. When switched in automatic mode, the widest filter is selected based on the video standard/format and user choice (see the 000 and 001 settings in Table 38).

Table 38. CSFM Function

| CSFM[2:0] | Description |
| :--- | :--- |
| 000 (default) | Autoselection 1.5 MHz bandwidth |
| 001 | Autoselection 2.17 MHz bandwidth |
| 010 | SH1 |
| 011 | SH2 |
| 100 | SH3 |
| 101 | SH4 |
| 110 | SH5 |
| 111 | Wideband mode |

Figure 22 shows the responses of SH1 (narrowest) to SH5 (widest) in addition to the wideband mode (shown in red).

## GAIN OPERATION

The gain control within the ADV7182 is done on a purely digital basis. The input ADC supports a 10 -bit range mapped into a 1.0 V analog voltage range. Gain correction takes place after the digitization in the form of a digital multiplier.

Advantages of this architecture over the commonly used programmable gain amplifier (PGA) before the ADC include the fact that the gain is now completely independent of supply, temperature, and process variations.
As shown in Figure 25, the ADV7182 can decode a video signal as long as it fits into the ADC window. The components for this are the amplitude of the input signal and the dc level it resides on. The dc level is set by the clamping circuitry (see the Clamp Operation section).
If the amplitude of the analog video signal is too high, clipping may occur, resulting in visual artifacts. The analog input range of the ADC, together with the clamp level, determines the maximum supported amplitude of the video signal.
Figure 23 and Figure 24 show the typical voltage divider networks required to keep the input video signal within the allowed range of the $\mathrm{ADC}, 0 \mathrm{~V}$ to 1 V . Place the circuit in Figure 23 before all the single-ended analog inputs to the ADV7182, and place the circuit in Figure 24 before all the differential inputs to the ADV7182.


The minimum supported amplitude of the input video is determined by the ability of the ADV7182 to retrieve horizontal and vertical timing and to lock to the color burst, if present.
There are separate gain control units for luma and chroma data. Both can operate independently of each other. The chroma unit, however, can also take its gain value from the luma path.

The possible AGC modes are shown in Table 39.
Table 39. AGC Modes

| Input Video Type | Luma Gain | Chroma Gain |
| :--- | :--- | :--- |
| Any | Manual gain luma | Manual gain chroma |
| CVBS | Dependent on <br> horizontal sync depth | Dependent on color- <br> burst amplitude <br> taken from luma path <br> Pependent on color- <br> burst amplitude <br> taken from luma path |
| Y/C white | Dependent on <br> horizontal sync depth | Dependent on color- <br> burst amplitude <br> taken from luma path <br> Peak white <br> Dependent on color- |
| YPrPb | Dependent on <br> horizontal sync depth | Taken from luma path |

It is possible to freeze the automatic gain control loops. This causes the loops to stop updating and the AGC determined gain at the time of the freeze to stay active until the loop is either unfrozen or the gain mode of operation is changed.
The currently active gain from any of the modes can be read back. Refer to the description of the dual-function manual gain registers, $\operatorname{LG}[11: 0]$ luma gain and CG[11:0] chroma gain, in the Luma Gain and Chroma Gain sections.

Figure 24. Differential Input Voltage Divider Network


Figure 25. Gain Control Overview

## Luma Gain

## LAGC[2:0], Luma Automatic Gain Control, Address 0x2C[6:4]

The luma automatic gain control mode bits select the operating mode for the gain control in the luma path.

Table 40. LAGC Function

| LAGC[2:0] | Description |
| :--- | :--- |
| 000 | Manual fixed gain (use LMG[11:0]) |
| 001 | AGC (blank level to sync tip), peak white algorithm off |
| 010 (default) | AGC (blank level to sync tip), peak white algorithm on |
| 011 | Reserved |
| 100 | Reserved |
| 101 | Reserved |
| 110 | Reserved |
| 111 | Freeze gain |

## LAGT[1:0], Luma Automatic Gain Timing, Address 0x2F[7:6]

The luma automatic gain timing register allows the user to influence the tracking speed of the luminance automatic gain control. This register has an effect only if the LAGC[2:0] register is set to 001 or 010 (automatic gain control modes).

If peak white AGC is enabled and active (see the Status 1[7:0], Address $0 \times 10[7: 0]$ section), the actual gain update speed is dictated by the peak white AGC loop and, as a result, the LAGT settings have no effect. As soon as the part leaves peak white AGC, LAGT becomes relevant again.

Table 41. LAGT Function

| LAGT[1:0] | Description |
| :--- | :--- |
| 00 | Slow $(\mathrm{TC}=2 \mathrm{sec})$ |
| 01 | Medium $(\mathrm{TC}=1 \mathrm{sec})$ |
| 10 | Fast $(\mathrm{TC}=0.2 \mathrm{sec})$ |
| 11 (default) | Adaptive |

LG[11:0], Luma Gain, Address 0x2F[3:0], Address 0x30[7:0] LMG[11:0], Luma Manual Gain, Address 0x2F[3:0], Address 0x30[7:0]

Luma gain[11:0] is a dual-function register. If all these bits are written to, a desired manual luma gain can be programmed. This gain becomes active if the LAGC[2:0] mode is switched to manual fixed gain. Equation 1 shows how to calculate a desired gain.
If read back, this register returns the current gain value. Depending on the setting in the LAGC[2:0] bits, the value is one of the following:

- Luma manual gain value (LAGC[2:0] set to luma manual gain mode)
- Luma automatic gain value (LAGC[2:0] set to either of the automatic modes)

Table 42. LG/LMG Function

| LG[11:0]/LMG[11:0] | Read/Write | Description |
| :--- | :--- | :--- |
| LMG[11:0] $=\mathrm{x}$ | Write | Manual gain for luma path |
| LG[11:0] $=\mathrm{x}$ | Read | Actual used gain |
| Luma Gain $\cong \frac{\text { LMG[11:0] }}{\text { Luma Calibration Factor }}$ |  |  |

where $L M G[11: 0]$ is a decimal value between 1024 and 4095.

## Calculation of the Luma Calibration Factor

1. Using a video source, set the content to a gray field and apply as a standard CVBS signal to the CVBS input of the board.
2. Using an oscilloscope, measure the signal at the CVBS input to ensure that its sync depth, color burst, and luma are at the standard levels.
3. Connect the output parallel pixel bus of the ADV7182 to a backend system that has unity gain and monitor the output voltage.
4. Measure the luma level correctly from the black level. Turn off the luma AGC and manually change the value of the luma manual gain control register, LMG[11:0], until the output luma level matches the input measured in Step 2.

This value, in decimal, is the luma calibration factor.

## BETACAM, Enable Betacam Levels, Address 0x01[5]

If YPrPb data is routed through the ADV7182, the automatic gain control modes can target different video input levels, as outlined in Table 46. The BETACAM bit is valid only if the input mode is YPrPb (component). The BETACAM bit sets the target value for AGC operation.

A review of the following sections is useful:

- The MAN_MUX_EN, Manual Input Muxing Enable, Address 0xC4[7] section for how component video $(\mathrm{YPrPb})$ can be routed through the ADV7182.
- The Video Standard Selection section to select the various standards, for example, with and without pedestal.

The AGC algorithms adjust the levels based on the setting of the BETACAM bit (see Table 45).

## PW_UPD, Peak White Update, Address 0x2B[0]

The peak white and average video algorithms determine the gain based on measurements taken from the active video. The PW_UPD bit determines the rate of gain change. LAGC[2:0] must be set to the appropriate mode to enable the peak white or average video mode in the first place. For more information, see the LAGC[2:0], Luma Automatic Gain Control, Address 0x2C[6:4] section.

Setting PW_UPD to 0 updates the gain once per video line.
Setting PW_UPD to 1 (default) updates the gain once per field.

## Chroma Gain

CAGC[1:0], Chroma Automatic Gain Control, Address 0x2C[1:0]

The two bits of the color automatic gain control mode select the basic mode of operation for the automatic gain control in the chroma path.
Table 43. CAGC Function

| CAGC[1:0] | Description |
| :--- | :--- |
| 00 | Manual fixed gain (use CMG[11:0]) |
| 01 | Use Luma gain for chroma |
| 10 (default) | Automatic gain (based on color burst) |
| 11 | Freeze chroma gain |

CAGT[1:0], Chroma Automatic Gain Timing, Address 0x2D[7:6]

The chroma automatic gain timing register allows the user to influence the tracking speed of the chroma automatic gain control. This register has an effect only if the CAGC[1:0] bits are set to 10 (automatic gain).

Table 44. CAGT Function

| CAGT[1:0] | Description |
| :--- | :--- |
| 00 | Slow $(\mathrm{TC}=2 \mathrm{sec})$ |
| 01 | Medium $(\mathrm{TC}=1 \mathrm{sec})$ |
| 10 | Reserved |
| 11 (default) | Adaptive |

Table 45. BETACAM Function

| BETACAM | Description |
| :--- | :--- |
| 0 (default) | Assuming YPrPb is selected as input format: <br>  <br>  <br>  <br>  <br>  <br>  <br>  <br> Selecting PAL with pedestal selects MII. <br> Selecting PAL without pedestal selects SMPTE. <br> Selecting NTSC with pedestal selects MII. <br> Selecting NTSC without pedestal selects SMPTE. <br> 1 |
| Assuming YPrPb is selected as input format: <br> Selecting PAL with pedestal selects BETACAM. <br>  <br>  <br> Selecting PAL without pedestal selects BETACAM variant. <br> Selecting NTSC with pedestal selects BETACAM. <br> Selecting NTSC without pedestal selects BETACAM variant. |  |

Table 46. BETACAM Levels

| Name | BETACAM (mV) | BETACAM Variant (mV) | SMPTE (mV) | MII (mV) |
| :--- | :--- | :--- | :--- | :--- |
| Y | 0 to +714 (including 7.5\% pedestal) | 0 to +714 | 0 to +700 | 0 to +700 (including 7.5\% pedestal) |
| Pb and Pr | -467 to +467 | -505 to +505 | -350 to +350 | -324 to +324 |
| Sync Depth | +286 | +286 | +300 | +300 |

## CG[11:0], Chroma Gain, Address 0x2D[3:0],

 Address 0x2E[7:0]; CMG[11:0], Chroma Manual Gain, Address 0x2D[3:0], Address 0x2E[7:0]Chroma gain[11:0] is a dual-function register. If written to, a desired manual chroma gain can be programmed. This gain becomes active if the CAGC[1:0] function is switched to manual fixed gain. See Equation 2 for calculating a desired gain.

If read back, this register returns the current gain value. Depending on the setting in the CAGC[1:0] bits, this is either:

- The chroma manual gain value (CAGC[1:0] set to chroma manual gain mode).
- The chroma automatic gain value (CAGC[1:0] set to either of the automatic modes).

Table 47. CG/CMG Function

| CG[11:0]/CMG[11:0] | Read/Write | Description |
| :--- | :--- | :--- |
| CMG[11:0] | Write | Manual gain for chroma path |
| CG[11:0] | Read | Currently active gain |

$$
\begin{equation*}
\text { Chroma_Gain } \cong \frac{C M G[11: 0]_{\text {decimal }}}{\text { ChromaCalibrationFactor }} \tag{2}
\end{equation*}
$$

where ChromaCalibrationFactor is a decimal value between 0 and 4095.

## Calculation of Chroma Calibration Factor

Take the following steps to calculate the chroma calibration factor:

1. Apply a CVBS signal with the color bars/SMPTE bars test pattern content directly to the measurement equipment.
2. Ensure correct termination of $75 \Omega$ on the measurement equipment. Measure chroma output levels.
3. Reconnect the source to the CVBS input of the ADV7182 system that has a back end gain of 1 . Repeat the measurement of chroma levels.
4. Turn off the chroma AGC and manually change the chroma gain control register, CMG[11:0], until the chroma level matches that measured directly from the source.
This value, in decimal, is the chroma calibration factor.

## CKE, Color Kill Enable, Address 0x2B[6]

The color kill enable bit allows the optional color kill function to be switched on or off.

For QAM-based video standards (PAL and NTSC), as well as FM-based systems (SECAM), the threshold for the color kill decision is selectable via the CKILLTHR[2:0] bits.
If color kill is enabled and the color carrier of the incoming video signal is less than the threshold for 128 consecutive video lines, color processing is switched off (black and white output). To switch the color processing back on, another 128 consecutive lines with a color burst greater than the threshold are required.

The color kill option works only for input signals with a modulated chroma part. For component input ( YPrPb ), there is no color kill.
Set CKE to 0 to disable color kill.
Set CKE to 1 (default) to enable color kill.

## CKILLTHR[2:0], Color Kill Threshold, Address 0x3D[6:4]

The CKILLTHR[2:0] bits allow the user to select a threshold for the color kill function. The threshold applies only to QAM-based (NTSC and PAL) or FM-modulated (SECAM) video standards.

To enable the color kill function, the CKE bit must be set. For the $000,001,010$, and 011 settings, chroma demodulation inside the ADV7182 may not work satisfactorily for poor input video signals.

Table 48. CKILLTHR Function

| CKILLTHR[2:0] | Description |  |
| :--- | :--- | :--- |
|  | NTSC, PAL | SECAM |
| 000 | Kill at $<0.5 \%$ | No color kill |
| 001 | Kill at $<1.5 \%$ | Kill at $<5 \%$ |
| 010 (default) | Kill at $<2.5 \%$ | Kill at $<7 \%$ |
| 011 | Kill at $<4 \%$ | Kill at $<8 \%$ |
| 100 | Kill at $<8.5 \%$ | Kill at $<9.5 \%$ |
| 101 | Kill at $<16 \%$ | Kill at $<15 \%$ |
| 110 | Kill at $<32 \%$ | Kill at $<32 \%$ |
| 111 | Reserved for Analog Devices internal use only; <br> do not select |  |

## CHROMA TRANSIENT IMPROVEMENT (CTI)

The signal bandwidth allocated for chroma is typically much smaller than that for luminance. In the past, this was a valid way to fit a color video signal into a given overall bandwidth because the human eye is less sensitive to chrominance than to luminance.

The uneven bandwidth, however, may lead to visual artifacts in sharp color transitions. At the border of two bars of color, both components (luma and chroma) change at the same time (see Figure 26). Due to the higher bandwidth, the signal transition of the luma component is usually much sharper than that of the chroma component. The color edge is not sharp, and in the worst case, it can be blurred over several pixels.


Figure 26. CTI Luma/Chroma Transition
The chroma transient improvement block examines the input video data. It detects transitions of chroma and can be programmed to create steeper chroma edges in an attempt to artificially restore lost color bandwidth. The CTI block, however, operates only on edges above a certain threshold to ensure that noise is not emphasized. Care was taken to ensure that edge ringing and undesirable saturation or hue distortion are avoided.

Chroma transient improvements are needed primarily for signals that have severe chroma bandwidth limitations. For those types of signals, it is strongly recommended to enable the CTI block via CTI_EN.

## CTI_EN, Chroma Transient Improvement (CTI) Enable, Address 0x4D[0] <br> Set CTI_EN to 0 to disable the CTI block.

Set CTI_EN to 1 (default) to enable the CTI block.

## CTI_AB_EN, Chroma Transient Improvement Alpha Blend Enable, Address 0x4D[1]

The CTI_AB_EN bit enables an alpha blend function within the CTI block. If set to 1 , the alpha blender mixes the transient improved chroma with the original signal. The sharpness of the alpha blending can be configured via the CTI_AB[1:0] bits.

For the alpha blender to be active, the CTI block must be enabled via the CTI_EN bit.
Set CTI_AB_EN to 0 to disable the CTI alpha blender.
Set CTI_AB_EN to 1 (default) to enable the CTI alpha-blend mixing function.

## CTI_AB[1:0], Chroma Transient Improvement Alpha Blend, Address 0x4D[3:2]

The CTI_AB[1:0] controls the behavior of alpha-blend circuitry that mixes the sharpened chroma signal with the original one. It thereby controls the visual impact of CTI on the output data.

For CTI_AB[1:0] to become active, the CTI block must be enabled via the CTI_EN bit, and the alpha blender must be switched on via CTI_AB_EN.

Sharp blending maximizes the effect of CTI on the picture; however, it may also increase the visual impact of small amplitude, high frequency chroma noise.

Table 49. CTI_AB Function

| CTI_AB[1:0] | Description |
| :--- | :--- |
| 00 | Sharpest mixing between sharpened and <br> original chroma signal <br> Sharp mixing |
| 01 | Smooth mixing <br> 10 |
| Smoothest mixing between sharpened and <br> original chroma signal |  |

## CTI_C_TH[7:0], CTI Chroma Threshold, Address 0x4E[7:0]

The CTI_C_TH[7:0] value is an unsigned, 8-bit number specifying how big the amplitude step in a chroma transition must be if it is going to be steepened by the CTI block. Programming a small value into this register causes even smaller edges to be steepened by the CTI block. Making CTI_C_TH[7:0] a large value causes the block to improve large transitions only.
The default value for CTI_C_TH[7:0] is 0x08.

## DIGITAL NOISE REDUCTION (DNR) AND LUMA PEAKING FILTER

Digital noise reduction is based on the assumption that high frequency signals with low amplitude are probably noise and that their removal, therefore, improves picture quality. The two DNR blocks in the ADV7182 are the DNR1 block before the luma peaking filter and the DNR2 block after the luma peaking filter, as shown in Figure 27.


Figure 27. DNR and Peaking Block Diagram
DNR_EN, Digital Noise Reduction Enable, Address 0x4D[5]
The DNR_EN bit enables the DNR block or bypasses it.
Table 50. DNR_EN Function

| Setting | Description |
| :--- | :--- |
| 0 | Bypasses the DNR block (disable) |
| 1 (default) | Enables the DNR block |

DNR_TH[7:0], DNR Noise Threshold 1, Address 0x50[7:0]
The DNR1 block is positioned before the luma peaking block.
The DNR_TH[7:0] value is an unsigned, 8 -bit number used to determine the maximum edge that is interpreted as noise and, therefore, blanked from the luma data. Programming a large value into DNR_TH[7:0] causes the DNR block to interpret even large transients as noise and remove them. As a result, the effect on the video data is more visible. Programming a small value causes only small transients to be seen as noise and to be removed.

Table 51. DNR_TH[7:0] Function

| Setting | Description |
| :--- | :--- |
| $0 x 08$ (default) | Threshold for maximum luma edges to be <br> interpreted as noise |

## PEAKING_GAIN[7:0], Luma Peaking Gain, Address 0xFB[7:0]

This filter can be manually enabled. The user can select to boost or to attenuate the midregion of the $Y$ spectrum around 3 MHz . The peaking filter can visually improve the picture by showing more definition on the picture details that contain frequency components around 3 MHz . The default value on this register passes through the luma data unaltered. A lower value attenuates the signal, and a higher value gains the luma signal. A plot of the responses of the filter is shown in Figure 28.

Table 52. PEAKING_GAIN[7:0] Function

| Setting | Description |
| :--- | :--- |
| $0 \times 40$ (default) | 0 dB response |



Figure 28. Peaking Filter Responses

## DNR_TH2[7:0], DNR Noise Threshold 2, Address 0xFC[7:0]

The DNR2 block is positioned after the luma peaking block and, therefore, affects the gained luma signal. It operates in the same way as the DNR1 block; however, there is an independent threshold control, DNR_TH2[7:0], for this block. This value is an unsigned, 8 -bit number used to determine the maximum edge that is interpreted as noise and, therefore, blanked from the luma data. Programming a large value into DNR_TH2[7:0] causes the DNR block to interpret even large transients as noise and remove them. As a result, the effect on the video data is more visible. Programming a small value causes only small transients to be seen as noise and to be removed.

Table 53. DNR_TH2[7:0] Function

| Setting | Description |
| :--- | :--- |
| 0x04 (default) | Threshold for maximum luma edges to be <br> interpreted as noise |

## COMB FILTERS

The comb filters of the ADV7182 can automatically handle video of all types, standards, and levels of quality. The NTSC and PAL configuration registers allow the user to customize the comb filter operation depending on which video standard is detected (by autodetection) or selected (by manual programming).

## NTSC Comb Filter Settings

These settings are used for NTSC M/NTSC J CVBS inputs.
NSFSEL[1:0], Split Filter Selection, NTSC, Address 0x19[3:2]
The NSFSEL[1:0] control selects how much of the overall signal bandwidth is fed to the combs. A narrow split filter selection results in better performance on diagonal lines but more dot crawl in the final output image. The opposite is true for selecting a wide bandwidth split filter.

Table 54. NSFSEL Function

| NSFSEL[1:0] | Description |
| :--- | :--- |
| 00 (default) | Narrow |
| 01 | Medium |
| 10 | Medium |
| 11 | Wide |

CTAPSN[1:0], Chroma Comb Taps, NTSC, Address 0x38[7:6]

Table 55. CTAPSN Function

| CTAPSN[1:0] | Description |
| :--- | :--- |
| 00 | Do not use |
| 01 | NTSC chroma comb adapts three lines to two lines |
| 10 (default) | NTSC chroma comb adapts five lines to three lines |
| 11 | NTSC chroma comb adapts five lines to four lines |

CCMN[2:0], Chroma Comb Mode, NTSC, Address 0x38[5:3]
Table 56. CCMN Function

| CCMN[2:0] | Description | Configuration |
| :---: | :---: | :---: |
| $\begin{aligned} & \hline 000 \\ & \text { (default) } \end{aligned}$ | Adaptive comb mode | Three-line adaptive chroma comb for CTAPSN = 01 <br> Four-line adaptive chroma comb for CTAPSN $=10$ <br> Five-line adaptive chroma comb for CTAPSN $=11$ |
| 100 | Disable chroma comb |  |
| 101 | Fixed chroma comb (top lines of line memory) | Fixed two-line chroma comb for CTAPSN $=01$ <br> Fixed three-line chroma comb for CTAPSN = 10 Fixed four-line chroma comb for CTAPSN = 11 |
| 110 | Fixed chroma comb (all lines of line memory) | Fixed three-line chroma comb for CTAPSN = 01 <br> Fixed four-line chroma comb for CTAPSN = 10 Fixed five-line chroma comb for CTAPSN = 11 |
| 111 | Fixed chroma comb (bottom lines of line memory) | Fixed two-line chroma comb for CTAPSN $=01$ <br> Fixed three-line chroma comb for CTAPSN $=10$ <br> Fixed four-line chroma comb for CTAPSN = 11 |

YCMN[2:0], Luma Comb Mode, NTSC, Address 0x38[2:0]
Table 57. YCMN Function

| YCMN[2:0] | Description | Configuration |
| :--- | :--- | :--- |
| 000 (default) | Adaptive comb mode | Adaptive three-lines <br> (three taps) luma comb |
| 100 | Disable luma comb | Use low-pass/notch <br> filter; see the Y Shaping <br> Filter section |
| 101 | Fixed luma comb (top <br> lines of line memory) | Fixed luma comb two- <br> line (two taps) |
| 110 | Fixed luma comb (all <br> lines of line memory) | Fixed luma comb three- <br> line (three taps) |
| 111 | Fixed luma comb <br> (bottom lines of line <br> memory) | Fixed luma comb two- <br> line (two taps) |

## PAL Comb Filter Settings

These settings are used for PAL B/PAL G/PAL H/PAL I/PAL D, PAL M, PAL Combinational N, PAL 60, and NTSC 4.43 CVBS inputs.
PSFSEL[1:0], Split Filter Selection, PAL, Address 0x19[1:0]
The PSFSEL[1:0] control selects how much of the overall signal bandwidth is fed to the combs. A wide split filter selection eliminates dot crawl but shows imperfections on diagonal lines. The opposite is true for selecting a narrow bandwidth split filter.

Table 58. PSFSEL Function

| PSFSEL[1:0] | Description |
| :--- | :--- |
| 00 | Narrow |
| 01 (default) | Medium |
| 10 | Wide |
| 11 | Widest |

CTAPSP[1:0], Chroma Comb Taps, PAL, Address 0x39[7:6]
Table 59. CTAPSP Function
\(\left.$$
\begin{array}{l|l}\hline \text { CTAPSP[1:0] } & \text { Description } \\
\hline 00 & \begin{array}{l}\text { Do not use } \\
\text { PAL chroma comb adapts five lines (three taps) } \\
\text { to three lines (two taps); cancels cross luma only }\end{array} \\
10 & \begin{array}{l}\text { PAL chroma comb adapts five lines (five taps) to } \\
\text { three lines (three taps); cancels cross luma and }\end{array}
$$ <br>

hue error less well\end{array}\right\}\) (default) | PAL chroma comb adapts five lines (five taps) to |
| :--- |
| four lines (four taps); cancels cross luma and hue |
| error well |

CCMP[2:0], Chroma Comb Mode, PAL, Address 0x39[5:3]
Table 60. CCMP Function

| CCMP[2:0] | Description | Configuration |
| :---: | :---: | :---: |
| 000 (default) | Adaptive comb mode | Adaptive three-line chroma comb for CTAPSN $=01$ <br> Adaptive four-line chroma comb for CTAPSN = 10 <br> Adaptive five-line chroma comb for CTAPSN $=11$ |
| 100 | Disable chroma comb |  |
| 101 | Fixed chroma comb (top lines of line memory) | Fixed two-line chroma comb for CTAPSN = 01 <br> Fixed three-line chroma comb for CTAPSN = 10 Fixed four-line chroma comb for CTAPSN = 11 |
| 110 | Fixed chroma comb (all lines of line memory) | Fixed three-line chroma comb for CTAPSN = 01 Fixed four-line chroma comb for CTAPSN = 10 Fixed five-line chroma comb for CTAPSN = 11 |
| 111 | Fixed chroma comb (bottom lines of line memory) | Fixed two-line chroma comb for CTAPSN = 01 <br> Fixed three-line chroma comb for CTAPSN = 10 Fixed four-line chroma comb for CTAPSN = 11 |

YCMP[2:0], Luma Comb Mode, PAL, Address 0x39[2:0]
Table 61. YCMP Function

| YCMP[2:0] | Description | Configuration |
| :--- | :--- | :--- |
| 000 (default) | Adaptive comb mode | Adaptive five lines (three <br> taps) luma comb |
| 100 | Disable luma comb | Use low-pass/notch filter; <br> see the Y Shaping Filter <br> section <br> 101 <br> 110 <br> 111 |
| Fixed three lines (two luma comb (top <br> lines of line memory) <br> taps) luma comb |  |  |
| Fixed luma comb (all <br> lines of line memory) | Fixed five lines (three taps) <br> luma comb <br> Fixed luma comb <br> (bottom lines of line <br> memory) | Fixed three lines (two <br> taps) luma comb |

## IF FILTER COMPENSATION

IFFILTSEL[2:0], IF Filter Select, Address 0xF8[2:0]
The IFFILTSEL[2:0] register allows the user to compensate for SAW filter characteristics on a composite input, as would be observed on tuner outputs. Figure 29 and Figure 30 show IF filter compensation for NTSC and PAL, respectively.
The options for this feature are as follows:

- Bypass mode
- NTSC, consisting of three filter characteristics
- PAL, consisting of three filter characteristics

See Table 95 for programming details.


Figure 29. NTSC IF Filter Compensation


Figure 30. PAL IF Filter Compensation

## ADAPTIVE CONTRAST ENHANCEMENT (ACE)

ACE allows for the contrast of an image to be increased, depending on the content of the picture. Normally, this allows bright areas to be made brighter and dark areas to be made darker. However, the ADV7182 ACE feature also allows for the contrast within dark areas to be increased without significantly affecting bright areas. This is particularly useful in automotive applications where it can be important to be able to discern objects in shaded areas.
The ACE feature works by sampling the chroma and luma levels in the inputted image. This information is then histogrammed, and the resulting correction is applied to the entire image. This correction is done in a nonlinear fashion so that more correction can be applied to dark areas if required.

For normal use, the luma and chroma gain controls can be used; however, in automotive applications, where dark areas may need to be further enhanced, also use the gamma gain controls.
The reaction time of the ACE function can be set using the ACE_RESPONSE_SPEED[7:4] bits (see Table 96). The corrected image is faded over the original image using alpha blending, giving a gradual change in contrast with scene changes. The ACE_ RESPONSE_SPEED[7:4] bits determine the duration of the transition from the original to the corrected image. A larger value for these bits results in a faster transition time; however, a smaller value gives more stability to rapid scene changes.
The ACE_CHROMA_MAX[7:4] bits are used to set a maximum value that clips the chroma gain regardless of the ACE_CHROMA_GAIN[3:0] settings.
The ACE_GAMMA_GAIN[3:0] bits are particularly useful in automotive applications because they allow dramatic image enhancement in dark regions by stretching the contrast of pixels at the low (dark) values of the image histogram. The luma and chroma gain controls are normally used; however, the ACE_GAMMA_GAIN[3:0] bits should be used when further stretching of the contrast in the dark areas of an image is needed.
ACE_ENABLE, User Sub Map 2, Address 0x80[7]
Enables ACE.
Table 62. ACE_ENABLE Function

| ACE_ENABLE | Description |
| :--- | :--- |
| 0 (default) | Disable ACE |
| 1 | Enable ACE |

ACE_LUMA_GAIN[4:0], User Sub Map 2, Address 0x83[4:0]
A control to set the auto-contrast level for the luma channel when ACE_ENABLE is 1 .

Table 63. ACE_LUMA_GAIN Function

| ACE_LUMA_GAIN[4:0] | Description |
| :--- | :--- |
| 00000 | Set ACE luma auto-contrast level to <br> minimum value |
| 01101 (default) | Set ACE luma auto-contrast level to <br> default value <br> Set ACE luma auto-contrast level to <br> maximum value |
| 11111 |  |

ACE_RESPONSE_SPEED[3:0], User Sub Map 2, Address 0x85[7:4]

Sets the reaction time of the ACE function.
Table 64. ACE_RESPONSE_SPEED Function

| ACE_ <br> RESPONSE_SPEED[3:0] | Description |
| :--- | :--- |
| 0000 | Set speed of ACE response to <br> slowest value <br> Set speed of ACE response to <br> fastest value |
| 1111 (default) |  |

ACE_CHROMA_GAIN[3:0], User Sub Map 2, Address 0x84[3:0]

This control sets the color-saturation level for the color channels when ACE_ENABLE is 1 .

Table 65. ACE_CHROMA_GAIN Function

| ACE_CHROMA_GAIN[3:0] | Description |
| :--- | :--- |
| 0000 | Set ACE color auto-saturation level <br> to minimum value |
| 1000 (default) | Set ACE color auto-saturation level <br> to default value |
| 1111 | Set ACE color auto-saturation level <br> to maximum value |

ACE_CHROMA_MAX[3:0], User Sub Map 2, Address 0x84[7:4]
This control sets a maximum threshold value that clips the chroma gain regardless of the ACE_CHROMA_GAIN[3:0] settings.

Table 66. ACE_CHROMA_MAX Function

| ACE_CHROMA_MAX[3:0] | Description |
| :--- | :--- |
| 0000 | Set maximum threshold for ACE <br> color auto-saturation level to <br> minimum value |
| 1000 (default) | Set maximum threshold for ACE <br> color auto-saturation level to <br> default value |
| 1111 | Set maximum threshold for ACE <br> color auto-saturation level to <br> maximum value |

ACE_GAMMA_GAIN[3:0], User Sub Map 2, Address 0x85[3:0]

This control provides further contrast enhancement to the luma and chroma gain controls and is particularly effective in the darker areas of an image.

Table 67. ACE_GAMMA_GAIN[3:0] Function

| ACE_GAMMA_GAIN[3:0] | Description |
| :--- | :--- |
| 0000 | Set further contrast enhancement <br> to minimum value |
| 1000 (default) | Set further contrast enhancements <br> to default value <br> Set further contrast enhancement <br> to maximum value |
| 1111 |  |

## DITHER FUNCTION

The dither function converts the digital output of the ADV7182 from 8-bit pixel data down to 6-bit pixel data. This function makes it easier for the ADV7182 to communicate with some LCD panels. The dither function is turned off by default. It is activated by the BR_DITHER_MODE bit.

BR_DITHER_MODE, User Sub Map 2, Address 0x92[0]
Table 68. BR_DITHER_MODE Function

| BR_DITHER_MODE | Description |
| :--- | :--- |
| 0 (default) | 8 -bit to 6-bit down dither disabled |
| 1 | 8 -bit to 6-bit down dither enabled |

## AV CODE INSERTION AND CONTROLS

This section describes the $\mathrm{I}^{2} \mathrm{C}$-based controls that affect the following:

- Insertion of $A V$ codes into the data stream
- Data blanking during the vertical blank interval (VBI)
- The range of data values permitted in the output data stream
- The relative delay of luma vs. chroma signals


## BT.656-4, ITU-R BT.656-3/ITU-R BT.656-4 Enable, Address 0x04[7]

Between Revision 3 and Revision 4 of the ITU-R BT. 656 standards, the ITU has changed the toggling position for the V bit within the SAV EAV codes for NTSC. The ITU-R BT.656-4 standard bit allows the user to select an output mode that is compliant with either the ITU-R BT.656-3 standard or ITU-R BT.656-4 standard. For further information, visit the International Telecommunication Union website.
Note that the standard change affects only NTSC and has no bearing on PAL.
When ITU-R BT.656-4 is set to 0 (default), the ITU-R BT.656-3 specification is used. The V bit goes low at EAV of Line 10 and Line 273.

When ITU-R BT.656-4 is 1 , the ITU-R BT.656-4 specification is used. The V bit goes low at EAV of Line 20 and Line 283.

## VBI_EN, Vertical Blanking Interval Data Enable, Address 0x03[7]

The VBI enable bit allows data such as intercast and closed caption data to be passed through the luma channel of the decoder with a minimal amount of filtering. All data for Line 1 to Line 21 is passed through and available at the output port. The ADV7182 does not blank the luma data and automatically switches all filters along the luma data path into their widest bandwidth. For active video, the filter settings for YSH and YPK are restored.
See the BL_C_VBI, Blank Chroma During VBI, Address $0 \times 04$ [2] section for information on the chroma path.
When VBI_EN is set to 0 (default), all video lines are filtered/scaled.

When VBI_EN is 1 , only the active video region is filtered/scaled.
BL_C_VBI, Blank Chroma During VBI, Address 0x04[2]
Setting BL_C_VBI to 1 blanks the Cr and Cb values of all VBI lines. This is done so any data that may arrive during VBI is not decoded as color and is output through Cr and Cb . As a result, it is possible to send VBI lines into the decoder and then output them through an encoder again, undistorted. Without this blanking, any color that is incorrectly decoded is encoded by the video encoder, thus distorting the VBI lines.
Setting BL_C_VBI to 0 decodes and outputs color during VBI.
Setting BL_C_VBI to 1 (default) blanks Cr and Cb values during VBI.

## Range, Range Selection, Address 0x04[0]

AV codes (as per ITU-R BT.656, formerly known as CCIR-656) consist of a fixed header made up of 0 xFF and 0 x 00 values. These two values are reserved and, therefore, are not to be used for active video. Additionally, the ITU specifies that the nominal range for video should be restricted to values between 16 and 235 for luma and 16 and 240 for chroma.

The range bit allows the user to limit the range of values output by the ADV7182 to the recommended value range. In any case, it ensures that the reserved values of $255 \mathrm{~d}(0 \mathrm{xFF})$ and $00 \mathrm{~d}(0 \mathrm{x} 00)$ are not presented on the output pins unless they are part of an AV code header.

Table 69. Range Function

| Range | Description |
| :--- | :--- |
| 0 | $16 \leq \mathrm{Y} \leq 235,16 \leq \mathrm{C} / \mathrm{P} \leq 240$ |
| 1 (default) | $1 \leq \mathrm{Y} \leq 254,1 \leq \mathrm{C} / \mathrm{P} \leq 254$ |

## AUTO_PDC_EN, Automatic Programmed Delay Control, Address 0x27[6]

Enabling AUTO_PDC_EN activates a function within the ADV7182 that automatically programs the LTA[1:0] and CTA[2:0] registers to have the chroma and luma data match delays for all modes of operation. If AUTO_PDC_EN is 1, the LTA[1:0] and CTA[2:0] manual registers are not used. If the automatic mode is disabled (by setting the AUTO_PDC_EN bit to 0 ), the values programmed into the LTA[1:0] and CTA[2:0] registers become active.
When AUTO_PDC_EN is set to 0, the ADV7182 uses the LTA[1:0] and CTA[2:0] values for delaying luma and chroma samples. See the LTA[1:0], Luma Timing Adjust, Address $0 \times 27[1: 0]$ section and the CTA[2:0], Chroma Timing Adjust, Address 0x27[5:3] section.
When AUTO_PDC_EN is 1 (default), the ADV7182 automatically determines the LTA and CTA values to have luma and chroma aligned at the output.

## LTA [1:0], Luma Timing Adjust, Address 0x27[1:0]

The luma timing adjust bits allow the user to specify a timing difference between chroma and luma samples.

There is a functionality overlap with the CTA[2:0] register. For manual programming, use the following defaults:

- CVBS input LTA [1:0] $=00$
- $\mathrm{Y} / \mathrm{C}$ input LTA $[1: 0]=01$
- YPrPb input LTA [1:0] $=01$

Table 70. LTA Function

| LTA[1:0] | Description |
| :--- | :--- |
| 00 (default) | No delay |
| 01 | Luma one clock (37 ns) late |
| 10 | Luma two clock (74 ns) early |
| 11 | Luma one clock (37 ns) early |

## CTA[2:0], Chroma Timing Adjust, Address 0x27[5:3]

The chroma timing adjust register allows the user to specify a timing difference between chroma and luma samples. This can be used to compensate for external filter group delay differences in the luma vs. chroma path and to allow a different number of pipeline delays while processing the video downstream. Review this functionality together with the LTA[1:0] bits.
The chroma can be delayed or advanced only in chroma pixel steps. One chroma pixel step is equal to two luma pixels. The programmable delay occurs after demodulation, where delay cannot be made by luma pixel steps.
For manual programming, use the following defaults:

- CVBS input CTA[2:0] $=011$
- Y/C input CTA[2:0] = 101
- YPrPb input $\mathrm{CTA}[2: 0]=110$

Table 71. CTA Function

| CTA[2:0] | Description |
| :--- | :--- |
| 000 | Reserved |
| 001 | Chroma + two pixels (early) |
| 010 | Chroma + one pixel (early) |
| 011 (default) | No delay |
| 100 | Chroma - one pixel (late) |
| 101 | Chroma - two pixels (late) |
| 110 | Chroma - three pixels (late) |
| 111 | Reserved |

## SYNCHRONIZATION OUTPUT SIGNALS

## HSYNC Configuration

The following controls allow the user to configure the behavior of the HSYNC output signal only:

- Beginning of HSYNC signal via HSB[10:0]
- End of HSYNC signal via HSE[10:0]
- Polarity of HSYNC using PHS

The HSYNC signal can be output on the VS/FIELD/SFL pin or the HS pin (see the Global Pin Control section.)
The HSYNC begin (HSB) and HSYNC end (HSE) registers allow the user to freely position the HSYNC signal within the video line. The values in $\operatorname{HSB}$ [10:0] and HSE[10:0] are measured in
pixel units from the falling edge of HSYNC. Using both values, the user can program both the position and length of the HSYNC output signal.

## HSB[10:0], HSYNC Begin, Address 0x34[6:4], Address 0x35[7:0]

The position of this edge is controlled by placing a binary number into HSB[10:0]. The number applied offsets the edge with respect to an internal counter that is reset to 0 immediately after EAV Code FF, 00, 00, XY (see Figure 31). HSB is set to 00000000010 b , which is two LLC clock cycles from count [0]. The default value of $\operatorname{HSB}[10: 0]$ is $0 \times 02$, indicating that the HSYNC pulse starts two pixels after the falling edge of HSYNC.
HSE[10:0], HSYNC End, Address 0x34[2:0], Address 0x36[7:0]
The position of this edge is controlled by placing a binary number into HSE[10:0]. The number applied offsets the edge with respect to an internal counter that is reset to 0 immediately after EAV Code FF, 00, 00, XY (see Figure 31). HSE is set to 00000000000 b , which is 0 LLC clock cycles from count [0].
The default value of $\operatorname{HSE}$ [10:0] is 00 , indicating that the HSYNC pulse ends 0 pixels after the falling edge of HSYNC.
For example,

- To shift the HSYNC toward active video by 20 LLCs, add 20 LLCs to both HSB and HSE, that is, HSB [10:0] = [00000010110], HSE[10:0] $=$ [000000010100].
- To shift the HSYNC away from active video by 20 LLCs, add 1696 LLCs to both HSB and HSE (for NTSC), that is, $\operatorname{HSB}[10: 0]=[11010100010], \operatorname{HSE}[10: 0]=[11010100000]$. Therefore, 1696 is derived from the NTSC total number of pixels, 1716.
- To move 20 LLCs away from active video, subtract 20 from 1716 and add the result in binary to both HSB [10:0] and HSE[10:0].


## PHS, HSYNC Polarity, Address 0x37[7]

The polarity of the HSYNC signal can be inverted using the PHS bit.
When PHS is 0 (default), HSYNC is active low.
When PHS is 1 , HSYNC is active high.

Table 72. HS Timing Parameters

| Standard | Characteristic |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | HS Begin Adjust, HSB[10:0] (Default) | HS End Adjust, HSE[10:0] (Default) | HS to Active Video, LLC Clock Cycles, C in Figure 31 (Default) | Active Video Samples/ Line, D in Figure 31 | Total LLC Clock Cycles, E in Figure 31 |
| NTSC | 00000000010b | 00000000000b | $\begin{aligned} & 272 \\ & \hline, 88 \end{aligned}$ | $\begin{aligned} & 720 Y+720 C=1440 \\ & 720 Y+720 C=1440 \end{aligned}$ | $1716$ |
| Figure 31. HSYNC Timing <br> Rev. C\| Page 42 of 96 |  |  |  |  |  |

## VSYNC and FIELD Configuration

The following controls allow the user to configure the behavior of the VSYNC and FIELD output signals, as well as the generation of embedded AV codes. Note that the VSYNC and FIELD signals can be output on the VS/FIELD/SFL pin or the HS pin (see the Global Pin Control section).

## NEWAVMODE, New AV Mode, Address 0x31[4]

When NEWAVMODE is 0 , EAV/SAV codes are generated to suit Analog Devices encoders. No adjustments are possible.

When NEWAVMODE is 1 (default), it enables the manual position of the VSYNC, FIELD, and AV codes using Register 0x32 to Register 0x33 and Register 0xE5 to Register 0xEA. Default register settings are CCIR656 (BT.565-4) compliant; see Figure 32, Figure 33 for NTSC and Figure 37, Figure 38 for PAL.

## HVSTIM, Horizontal VSYNC Timing, Address 0x31[3]

The HVSTIM bit allows the user to select where the VSYNC signal is asserted within a line of video. Some interface circuitry may require VSYNC to go low while HSYNC is low.
When HVSTIM is 0 (default), the start of the line is relative to HSE.
When HVSTIM is 1 , the start of the line is relative to HSB.

## VSBHO, VSYNC Begin Horizontal Position Odd, Address 0x32[7]

The VSBHO and VSBHE bits select the position within a line at which the VSYNC signal (not the bit in the AV code) becomes active. Some follow-on chips require the VSYNC signal to change state only when HSYNC is high or low.
When VSBHO is 0 (default), the VSYNC signal goes high in the middle of a line of video (odd field).
When VSBHO is 1 , the VSYNC signal changes state at the start of a line (odd field).

## VSBHE, VSYNC Begin Horizontal Position Even, Address 0x32[6]

The VSBHO and VSBHE bits select the position within a line at which the VSYNC signal (not the bit in the AV code) becomes active. Some follow-on chips require the VSYNC signal to change state only when HS is high or low.

When VSBHE is 0 (default), the VSYNC signal goes high in the middle of a line of video (even field).
When VSBHE is 1 , the VSYNC signal changes state at the start of a line (even field).

## VSEHO, VSYNC End Horizontal Position Odd, Address 0x33[7]

The VSEHO and VSEHE bits select the position within a line at which the VSYNC signal (not the bit in the AV code) becomes active. Some follow-on chips require the VSYNC signal to change state only when HSYNC is high or low.

When VSEHO is 0 (default), the VSYNC signal goes low (inactive) in the middle of a line of video (odd field).
When VSEHO is 1 , the VSYNC signal changes state at the start of a line (odd field).

## VSEHE, VSYNC End Horizontal Position Even, Address 0x33[6]

The VSEHO and VSEHE bits select the position within a line at which the VSYNC signal (not the bit in the AV code) becomes active. Some follow-on chips require the VSYNC signal to change state only when HS is high or low.
When VSEHE is 0 (default), the VSYNC signal goes low (inactive) in the middle of a line of video (even field).
When VSEHE is 1 , the VSYNC signal changes state at the start of a line (even field).

## PVS, VSYNC Polarity, Address 0x37[5]

The polarity of the VSYNC signal can be inverted using the PVS bit.
When PVS is 0 (default), VSYNC is active high.
When PVS is 1 , VSYNC is active low.

## PF, FIELD Polarity, Address 0x37[3]

The FIELD pin can be inverted using the PHS bit. When PHS is 0 (default), FIELD pin is active high. When PHS is 1 , FIELD pin is active low.


Figure 32. NTSC Default, ITU-R BT. 656 (Polarity of H, V, and F Embedded in Data)


Figure 33. NTSC Typical VS/FIELD Positions


Figure 34. NTSC VSYNC Begin
NVBEGDELO, NTSC VSYNC Begin Delay on Odd Field, Address 0xE5[7]
When NVBEGDELO is 0 (default), there is no delay.
Setting NVBEGDELO to 1 delays VSYNC going high by one line relative to NVBEG (odd field).
NVBEGDELE, NTSC VSYNC Begin Delay on Even Field, Address 0xE5[6]
When NVBEGDELE is 0 (default), there is no delay.
Setting NVBEGDELE to 1 delays VSYNC going high by one line relative to NVBEG (even field).

## NVBEGSIGN, NTSC VSYNC Begin Sign, Address 0xE5[5]

Setting NVBEGSIGN to 0 delays the start of VSYNC; sets to low when manual programming.
Setting NVBEGSIGN to 1 (default) advances the start of VSYNC; however, it is not suitable for user programming.
NVBEG[4:0], NTSC VSYNC Begin, Address 0xE5[4:0]
The default value of NVBEG is 00101, indicating the NTSC VSYNC begin position. For all NTSC/PAL VSYNC timing controls, both the V bit in the AV code and the VSYNC signal are modified.


Figure 35. NTSC VSYNC End
NVENDDELO, NTSC VSYNC End Delay on Odd Field, Address 0xE6[7]
When NVENDDELO is 0 (default), there is no delay.
Setting NVENDDELO to 1 delays VSYNC going low by one line relative to NVEND (odd field).

## NVENDDELE, NTSC VSYNC End Delay on Even Field, Address 0xE6[6]

When NVENDDELE is 0 (default), there is no delay.
Setting NVENDDELE to 1 delays VSYNC going low by a line relative to NVEND (even field).
NVENDSIGN, NTSC VSYNC End Sign, Address 0xE6[5]
Setting NVENDSIGN to 0 (default) delays the end of VSYNC; sets to low when manual programming.
Setting NVENDSIGN to 1 advances the end of VSYNC; however, it is not suitable for user programming.

## NVEND[4:0], NTSC VSYNC End, Address 0xE6[4:0]

The default value of NVEND is 00100 , indicating the NTSC VSYNC end position.
For all NTSC/PAL VSYNC timing controls, both the V bit in the AV code and the VSYNC signal are modified.


Figure 36. NTSC FIELD Toggle

NFTOGDELO, NTSC FIELD Toggle Delay on Odd Field, Address 0xE7[7]
When NFTOGDELO is 0 (default), there is no delay.
Setting NFTOGDELO to 1 delays the FIELD toggle/transition by one line relative to NFTOG (odd field).
NFTOGDELE, NTSC FIELD Toggle Delay on Even Field, Address 0xE7[6]
When NFTOGDELE is 0 , there is no delay.
Setting NFTOGDELE to 1 (default) delays the FIELD toggle/ transition by one line relative to NFTOG (even field).
NFTOGSIGN, NTSC FIELD Toggle Sign, Address 0xE7[5]
Setting NFTOGSIGN to 0 delays the FIELD toggle/transition; sets to low when manual programming.
Setting NFTOGSIGN to 1 (default) advances the FIELD toggle/transition; however, it is not not suitable for user programming.
NFTOG[4:0], NTSC FIELD Toggle, Address 0xE7[4:0]
The default value of NFTOG is 00011, indicating the NTSC field toggle position.
For all NTSC/PAL field timing controls, both the F bit in the AV code and the field signal are modified.

FIELD 1



Figure 38. PAL Typical VS/FIELD Positions


Figure 39. PAL VSYNC Begin
PVBEGDELO, PAL VSYNC Begin Delay on Odd Field, Address 0xE8[7]
When PVBEGDELO is 0 (default), there is no delay.
Setting PVBEGDELO to 1 delays VSYNC going high by one line relative to PVBEG (odd field).

## PVBEGDELE, PAL VSYNC Begin Delay on Even Field, Address 0xE8[6]

When PVBEGDELE is 0 , there is no delay.
Setting PVBEGDELE to 1 (default) delays VSYNC going high by one line relative to PVBEG (even field).
PVBEGSIGN, PAL VSYNC Begin Sign, Address 0xE8[5]
Setting PVBEGSIGN to 0 delays the beginning of VSYNC; sets to low when manual programming.
Setting PVBEGSIGN to 1 (default) advances the beginning of VSYNC; however, it is not suitable for user programming.
PVBEG[4:0], PAL VSYNC Begin, Address 0xE8[4:0]
The default value of PVBEG is 00101, indicating the PAL VSYNC begin position. For all NTSC/PAL VSYNC timing controls, the V bit in the AV code and the VSYNC signal are modified.


Figure 40. PAL VSYNC End

## PVENDDELO, PAL VSYNC End Delay on Odd Field, Address 0xE9[7]

When PVENDDELO is 0 (default), there is no delay.
Setting PVENDDELO to 1 delays VSYNC going low by one line relative to PVEND (odd field).
PVENDDELE, PAL VSYNC End Delay on Even Field, Address 0xE9[6]

When PVENDDELE is 0 (default), there is no delay.
Setting PVENDDELE to 1 delays VSYNC going low by one line relative to PVEND (even field).
PVENDSIGN, PAL VSYNC End Sign, Address 0xE9[5]
Setting PVENDSIGN to 0 (default) delays the end of VSYNC; sets to low when manual programming.
Setting PVENDSIGN to 1 advances the end of VSYNC; however, it is not suitable for user programming.

## PVEND[4:0], PAL VSYNC End, Address 0xE9[4:0]

The default value of PVEND is 10100 , indicating the PAL VSYNC end position.
For all NTSC/PAL VSYNC timing controls, both the $V$ bit in the AV code and the VSYNC signal are modified.


Figure 41. PAL FIELD Toggle
PFTOGDELO, PAL FIELD Toggle Delay on Odd Field, Address 0xEA[7]

When PFTOGDELO is 0 (default), there is no delay.
Setting PFTOGDELO to 1 delays the FIELD toggle/transition by one line relative to PFTOG (odd field).

## PFTOGDELE, PAL FIELD Toggle Delay on Even Field, Address 0xEA[6]

When PFTOGDELE is 0 , there is no delay.
Setting PFTOGDELE to 1 (default) delays the FIELD toggle/transition by one line relative to PFTOG (even field).

## PFTOGSIGN, PAL FIELD Toggle Sign, Address 0xEA[5]

Setting PFTOGSIGN to 0 delays the field transition and set to low when manual programming.
Setting PFTOGSIGN to 1 (default) advances the field transition; however, it is not suitable for user programming.

## PFTOG, PAL FIELD Toggle, Address 0xEA[4:0]

The default value of PFTOG is 00011 , indicating the PAL field toggle position.

For all NTSC/PAL field timing controls, the F bit in the AV code and the FIELD signal are modified.

## SYNC PROCESSING

The ADV7182 has two additional sync processing blocks that postprocess the raw synchronization information extracted from the digitized input video. If desired, the blocks can be disabled via the following two $\mathrm{I}^{2} \mathrm{C}$ bits: ENHSPLL and ENVSPROC.

ENHSPLL, Enable HSYNC Processor, Address 0x01[6]
The HSYNC processor is designed to filter incoming HSYNCs that have been corrupted by noise, providing improved performance for video signals with stable time bases but poor SNR.
Setting ENHSPLL to 0 disables the HSYNC processor.
Setting ENHSPLL to 1 (default) enables the HSYNC processor.
ENVSPROC, Enable VSYNC Processor, Address 0x01[3]
This block provides extra filtering of the detected VSYNCs to improve vertical lock.
Setting ENVSPROC to 0 disables the VSYNC processor.
Setting ENVSPROC to 1 (default) enables the VSYNC processor.

## VBI DATA DECODE

The VBI data processor (VDP) on the ADV7182 can slice both low bandwidth standards and high bandwidth standards such as teletext.
The VDP is capable of slicing multiple VBI data standards on SD video. It decodes the VBI data on the incoming CVBS and Y/C or YUV data. The decoded results are available as ancillary data in output 656 data stream. For low data rate VBI standards like CC/WSS/CGMS, the decoded data bytes can be read from the $\mathrm{I}^{2} \mathrm{C}$ registers.

The VBI data standards that can be decoded by the VDP are listed in Table 73 and Table 74.

Table 73. PAL

| Feature | Standard |
| :--- | :--- |
| Teletext System A, Teletext System C, or | ITU-R BT.653 |
| Teletext System D |  |
| Teletext System B/Teletext System WST | ITU-R BT.653 |
| Wide Screen Signaling (WSS) | ITU-R BT.1119-1/ |
|  | ETSI EN.300294 |
| Closed Captioning (CCAP) | Not applicable |

Table 74. NTSC

| Feature | Standard |
| :--- | :--- |
| Teletext System B and Teletext System D | ITU-R BT.653 |
| Teletext System C/ Teletext System NABTS | ITU-R BT.653/ |
|  | EIA-516 |
| Copy Generation Management System (CGMS) | EIA-J CPR-1204/ |
|  | IEC 61880 |
| Closed Captioning (CCAP) | EIA-608 |

The VBI data standard that the VDP decodes on a particular line of incoming video has been set by default as described in Table 75. This can be overridden manually and any VBI data can be decoded on any line. The details of manual programming are described in Table 76.

## VDP Default Configuration

The VDP can decode different VBI data standards on a line-toline basis. The various standards supported by default on different lines of VBI are explained in Table 75.

## VDP Manual Configuration

MAN_LINE_PGM, Enable Manual Line Programming of VBI Standards, Address 0x64[7], Interrupt/VDP Map

The user can configure the VDP to decode different standards on a line-to-line basis through manual line programming. For this, set the MAN_LINE_PGM bit. The user must write into all the line programming registers, VBI_DATA_Px_Ny and VBI_DATA_Px (see Register 0x64 to Register 0x77 in Table 97).

When MAN_LINE_PGM is set to 0 (default), the VDP decodes default standards on lines, as shown in Table 75.
When MAN_LINE_PGM is set to 1 , the VBI standards to be decoded are manually programmed.

```
VBI_DATA_Px_Ny[3:0], VBI_DATA_Px[3:0], VBI Standard to be Decoded on Line \(X\) for PAL, Line \(Y\) for NTSC, Address 0x64 to Address 0x77, Interrupt/VDP Map
```

These are related 4-bit clusters in Register 0x64 to Register 0x77 of the Interrupt/VDP Map Details (see Table 94). These 4-bit, line programming registers, VBI_DATA_Px_Ny and VBI_DATA_Px, identify the VBI data standard that is decoded on Line X in PAL mode or on Line Y in NTSC mode. The different types of VBI standards decoded by VBI_DATA_Px_Ny and VBI_DATA_Px are shown in Table 76. Note that the X or Y value depends on whether the ADV7182 is in PAL or NTSC mode.

Table 75. Default Standards on Lines for PAL and NTSC

| PAL-625/50 |  |  |  | NTSC-525/60 |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Line No. | Default VBI Data Decoded | Line No. | Default VBI <br> Data Decoded | Line No. | Default VBI Data Decoded | Line No. | Default VBI Data Decoded |
| 6 | WST | 318 | Reserved | 23 | Reserved | 286 | Reserved |
| 7 | WST | 319 | WST | 24 | Reserved | 287 | Reserved |
| 8 | WST | 320 | WST | 25 | Reserved | 288 | Reserved |
| 9 | WST | 321 | WST | 10 | NABTS | 272 | NABTS |
| 10 | WST | 322 | WST | 11 | NABTS | 273 | NABTS |
| 11 | WST | 323 | WST | 12 | NABTS | 274 | NABTS |
| 12 | WST | 324 | WST | 13 | NABTS | 275 | NABTS |
| 13 | WST | 325 | WST | 14 | Reserved | 276 | NABTS |
| 14 | WST | 326 | WST | 15 | NABTS | 277 | Reserved |
| 15 | WST | 327 | WST | 16 | Reserved | 278 | NABTS |
| 16 | Reserved | 328 | WST | 17 | NABTS | 279 | Reserved |
| 17 | Reserved | 329 | Reserved | 18 | NABTS | 280 | NABTS |
| 18 | Reserved | 332 | Reserved | 19 | NABTS | 281 | NABTS |
| 19 | Reserved | 333 | WST | 20 | CGMS | 282 | NABTS |
| 20 | WST | 334 | WST | 21 | CCAP | 283 | CGMS |
| 21 | WST | 335 | CCAP | $22 \text { + full }$ <br> odd field | NABTS | 284 | CCAP |
| 22 | CCAP | 336 | WST | Reserved | Reserved | $285 \text { + full }$ <br> even field | NABTS |
| 23 | WSS | $\begin{aligned} & 337+\text { full } \\ & \text { even field } \end{aligned}$ | WST | Reserved | Reserved | Reserved | Reserved |
| $24 \text { + full }$ <br> odd field | WST | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved |

Table 76. VBI Data Standards for Manual Configuration

| VBI_DATA_Px_Ny | PAL-625/50 | NTSC_525/60 |
| :--- | :--- | :--- |
| 0000 | Disable VDP | Disable VDP |
| 0001 | Teletext system identified by VDP_TTXT_TYPE | Teletext system identified by VDP_TTXT_TYPE |
| 0010 | Reserved | Reserved |
| 0011 | Reserved | Reserved |
| 0100 | WSS ITU-R BT.1119-1/ETSI.EN.300294 | CGMS EIA-J CPR-1204/IEC 61880 |
| 0101 | Reserved | Reserved |
| 0110 | Reserved | Reserved |
| 0111 | CCAP | CCAP EIA-608 |
| 1000 to $\mathbf{1 1 1 1}$ | Reserved | Reserved |

Table 77.VBI Data Standards to be Decoded on Line Px (PAL) or Line Ny (NTSC)

| Signal Name | Bit Location | Dec Address | Hex Address |
| :---: | :---: | :---: | :---: |
| VBI_DATA_P6_N23 | VDP_LINE_00F[7:4] | 101 | 0x65 |
| VBI_DATA_P7_N24 | VDP_LINE_010[7:4] | 102 | 0x66 |
| VBI_DATA_P8_N25 | VDP_LINE_011[7:4] | 103 | 0x67 |
| VBI_DATA_P9 | VDP_LINE_012[7:4] | 104 | 0x68 |
| VBI_DATA_P10 | VDP_LINE_013[7:4] | 105 | 0x69 |
| VBI_DATA_P11 | VDP_LINE_014[7:4] | 106 | 0x6A |
| VBI_DATA_P12_N10 | VDP_LINE_015[7:4] | 107 | 0x6B |
| VBI_DATA_P13_N11 | VDP_LINE_016[7:4] | 108 | 0x6C |
| VBI_DATA_P14_N12 | VDP_LINE_017[7:4] | 109 | 0x6D |
| VBI_DATA_P15_N13 | VDP_LINE_018[7:4] | 110 | 0x6E |
| VBI_DATA_P16_N14 | VDP_LINE_019[7:4] | 111 | 0x6F |
| VBI_DATA_P17_N15 | VDP_LINE_01A[7:4] | 112 | 0x70 |
| VBI_DATA_P18_N16 | VDP_LINE_01B[7:4] | 113 | 0x71 |
| VBI_DATA_P19_N17 | VDP_LINE_01C[7:4] | 114 | 0x72 |
| VBI_DATA_P20_N18 | VDP_LINE_01D[7:4] | 115 | 0x73 |
| VBI_DATA_P21_N19 | VDP_LINE_01E[7:4] | 116 | 0x74 |
| VBI_DATA_P22_N20 | VDP_LINE_01F[7:4] | 117 | 0x75 |
| VBI_DATA_P23_N21 | VDP_LINE_020[7:4] | 118 | 0x76 |
| VBI_DATA_P24_N22 | VDP_LINE_021[7:4] | 119 | 0x77 |
| VBI_DATA_P318 | VDP_LINE_00E[3:0] | 100 | 0x64 |
| VBI_DATA_P319_N286 | VDP_LINE_00F[3:0] | 101 | 0x65 |
| VBI_DATA_P320_N287 | VDP_LINE_010[3:0] | 102 | 0x66 |
| VBI_DATA_P321_N288 | VDP_LINE_011[3:0] | 103 | 0x67 |
| VBI_DATA_P322 | VDP_LINE_012[3:0] | 104 | 0x68 |
| VBI_DATA_P323 | VDP_LINE_013[3:0] | 105 | 0x69 |
| VBI_DATA_P324_N272 | VDP_LINE_014[3:0] | 106 | 0x6A |
| VBI_DATA_P325_N273 | VDP_LINE_015[3:0] | 107 | 0x6B |
| VBI_DATA_P326_N274 | VDP_LINE_016[3:0] | 108 | 0x6C |
| VBI_DATA_P327_N275 | VDP_LINE_017[3:0] | 109 | 0x6D |
| VBI_DATA_P328_N276 | VDP_LINE_018[3:0] | 110 | 0x6E |
| VBI_DATA_P329_N277 | VDP_LINE_019[3:0] | 111 | 0x6F |
| VBI_DATA_P330_N278 | VDP_LINE_01A[3:0] | 112 | 0x70 |
| VBI_DATA_P331_N279 | VDP_LINE_01B[3:0] | 113 | 0x71 |
| VBI_DATA_P332_N280 | VDP_LINE_01C[3:0] | 114 | 0x72 |
| VBI_DATA_P333_N281 | VDP_LINE_01D[3:0] | 115 | 0x73 |
| VBI_DATA_P334_N282 | VDP_LINE_01E[3:0] | 116 | 0x74 |
| VBI_DATA_P335_N283 | VDP_LINE_01F[3:0] | 117 | 0x75 |
| VBI_DATA_P336_N284 | VDP_LINE_020[3:0] | 118 | 0x76 |
| VBI_DATA_P337_N285 | VDP_LINE_021[3:0] | 119 | 0x77 |

Full field detection (lines other than VBI lines) of any standard can also be enabled by writing to the VBI_DATA_P24_N22[3:0] and VBI_DATA_P337_N285[3:0] bits. So, if VBI_DATA_P24_N22[3:0] is programmed with any teletext standard, teletext is decoded off for the entire odd field. The corresponding register for the even field is VBI_DATA_P337_N285[3:0].
For teletext system identification, VDP assumes that if teletext is present in a video channel, all the teletext lines comply with a single standard system. Therefore, the line programming using the VBI_DATA_Px_Ny and VBI_DATA_Px registers identifies whether the data in line is teletext; the actual standard is identified by the VDP_TTXT_TYPE_MAN bit.

To program the VDP_TTXT_TYPE_MAN bit, the VDP_TTXT_TYPE_MAN_ENABLE bit must be set to 1.

## VDP_TTXT_TYPE_MAN_ENABLE, Enable Manual Selection of Teletext Type, Address 0x60[2], Interrupt/VDP Map

Setting VDP_TTXT_TYPE_MAN_ENABLE to 0 (default), disables manual programming of the teletext type.

Setting VDP_TTXT_TYPE_MAN_ENABLE to 1 , enables manual programming of the teletext type.

VDP_TTXT_TYPE_MAN[1:0], Specify the Teletext Type, Address 0x60[1:0], Interrupt/VDP Map
These bits specify the teletext type to be decoded. These bits are functional only if VDP_TTXT_TYPE_MAN_ENABLE is set to 1 .

Table 78. VDP_TTXT_TYPE_MAN Function

| VDP_TTXT_- |  |  |
| :--- | :--- | :--- |
| TYPE_MAN[1:0] | $\mathbf{6 2 5 / 5 0}$ (PAL) | $\mathbf{5 2 5 / 6 0}$ (NTSC) |
| 00 (default) | Teletext-ITU-BT.653- | Reserved |
| 01 | $625 / 50-A$ |  |
|  | Teletext-ITU-BT.653- | Teletext-ITU-BT.653- |
| 10 | $625 / 50-$ B (WST) | $525 / 60-$ B |
|  | Teletext-ITU-BT.653- | Teletext-ITU-BT.653- |
|  | $625 / 50-C$ | $525 / 60-C$ or EIA516 |
|  |  | (NABTS) |
| 11 | Teletext-ITU-BT.653- | Teletext-ITU-BT.653- |
|  | $625 / 50-$ D | $525 / 60-$ D |

## VDP Ancillary Data Output

Reading the data back via $I^{2} \mathrm{C}$ may not be feasible for VBI data standards with high data rates (for example, teletext). An alternative is to place the sliced data in a packet in the line blanking of the digital output CCIR656 stream. This is available for all standards sliced by the VDP module.
When data is sliced on a given line, the corresponding ancillary data packet is placed immediately after the next EAV code that occurs at the output (that is, data sliced from multiple lines are not buffered up and then emitted in a burst). Note that, due to the vertical delay through the comb filters, the line number on which the packet is placed differs from the line number on which the data was sliced.
The user can enable or disable the insertion of VDP results that have been decoded into the 656 ancillary streams by using the ADF_ENABLE bit.

ADF_ENABLE, Enable Ancillary Data Output Through 656 Stream, Address 0x62[7], User Sub Map 2

Setting ADF_ENABLE to 0 (default) disables the insertion of VBI decoded data into the ancillary 656 stream.
Setting ADF_ENABLE to 1 enables the insertion of VBI decoded data into the ancillary 656 stream.
The user may select the data identification word (DID) and the secondary data identification word (SDID) through programming the ADF_DID[4:0] and ADF_SDID[5:0] bits, respectively.

## ADF_DID[4:0], User-Specified Data ID Word in Ancillary Data, Address 0x62[4:0], Interrupt/VDP Map

These bits select the data ID word to be inserted into the ancillary data stream with the data decoded by the VDP.

The default value of ADF_DID[4:0] is 10101.
ADF_SDID[5:0], User-Specified Secondary Data ID Word in Ancillary Data, Address 0x63[5:0], Interrupt/VDP Map
These bits select the secondary data ID word to be inserted in the ancillary data stream with the data decoded by the VDP.

The default value of ADF_SDID[5:0] is 101010.

## DUPLICATE_ADF, Enable Duplication/Spreading of Ancillary Data over Y and C Buses, Address 0x63[7], Interrupt/VDP Map

This bit determines whether the ancillary data is duplicated over both Y and C buses or if the data packets are spread between the two channels.

When DUPLICATE_ADF to 0 (default) is set, the ancillary data packet is spread across the Y and C data streams.
When DUPLICATE_ADF to 1 is set, the ancillary data packet is duplicated on the Y and C data streams.

ADF_MODE[1:0], Determine the Ancillary Data Output Mode, Address 0x62[6:5], Interrupt/VDP Map
These bits determine whether the ancillary data output mode is in byte mode or nibble mode.

Table 79. ADF_MODE

| ADF_MODE[1:0] | Description |
| :--- | :--- |
| 00 (default) | Nibble mode |
| 01 | Byte mode, no code restrictions <br> Byte mode, but 0x00 and 0xFF prevented <br> 10 |
| (0x00 replaced by 0x01, 0xFF replaced by 0xFE) |  |
| 11 | Reserved |

The ancillary data packet sequence is explained in Table 80 and Table 81. The nibble output mode is the default mode of output from the ancillary stream when ancillary stream output is enabled. This format is in compliance with ITU-R BT.1364.
The following abbreviations are used in Table 80 and Table 81:

- EP—Even parity for Bit B8 to Bit B2. The parity bit's EP is set so that an even number of 1s are in Bit B8 to Bit B2, including the parity bit, B 8 .
- CS—Checksum word. The CS word is used to increase confidence of the integrity of the ancillary data packet from the DID, SDID, and dc through user data-words (UDWs). It consists of 10 bits that include the following: a 9-bit calculated value and B 9 as the inverse of B 8 . The checksum value B 8 to B 0 is equal to the nine LSBs of the sum of the nine LSBs of the DID, SDID, and dc, and all UDWs in the packet. Prior to the start of the checksum count cycle, all checksum and carry bits are preset to 0 . Any carry resulting from the checksum count cycle is ignored.
- $\overline{\mathrm{EP}}$-The MSB, B9, is the inverse of EP. This ensures that restricted Code 0x00 and Code 0xFF do not occur.
- LINE_NUMBER[9:0]-The line number of the line that immediately precedes the ancillary data packet. The line number is from the numbering system in ITU-R BT.470. The line number runs from 1 to 625 in a 625 -line system and from 1 to 263 in a 525 -line system. Note that, due to the vertical delay through the comb filters, the line number on which the packet is output differs from the line number on which the VBI data was sliced.
- Data count-The data count specifies the number of UDWs in the ancillary stream for the standard. The total number of user data-words is four times the data count. Padding words can be introduced to make the total number of UDWs divisible by 4 .

Table 80. Ancillary Data in Nibble Output Format

| Byte | B9 | B8 | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 | Description |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | Ancillary data preamble |
| 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |  |
| 2 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |  |
| 3 | $\overline{\mathrm{EP}}$ | EP | 0 | $1{ }^{2} \mathrm{C}$ _DID6_2[4:0] |  |  |  |  | 0 | 0 | DID (data identification word) |
| 4 | $\overline{\mathrm{EP}}$ | EP | 12C_SDID7_2[5:0] |  |  |  |  |  | 0 | 0 | SDID (secondary data identification word) |
| 5 | $\overline{\mathrm{EP}}$ | EP | 0 | DC[4:0] |  |  |  |  | 0 | 0 | Data count |
| 6 | $\overline{\mathrm{EP}}$ | EP | Padding[1:0] |  | VBI_DATA_STD[3:0] |  |  |  | 0 | 0 | ID0 (User Data-Word 1) |
| 7 | $\overline{\mathrm{EP}}$ | EP | 0 | LINE_NUMBER[9:5] |  |  |  |  | 0 | 0 | ID1 (User Data-Word 2) |
| 8 | $\overline{\mathrm{EP}}$ | EP | EVEN_FIELD | LINE_NUMBER[4:0] |  |  |  |  | 0 | 0 | ID2 (User Data-Word 3) |
| 9 | $\overline{\mathrm{EP}}$ | EP | 0 | 0 | 0 | 0 | VDP_TTXT_TYPE[1:0] |  | 0 | 0 | ID3 (User Data-Word 4) |
| 10 | $\overline{\mathrm{EP}}$ | EP | 0 | 0 | VBI_WORD_1[7:4] |  |  |  | 0 | 0 | ID4 (User Data-Word 5) |
| 11 | $\overline{\mathrm{EP}}$ | EP | 0 | 0 | VBI_WORD_1[3:0] |  |  |  | 0 | 0 | ID5 (User Data-Word 6) |
| 12 | $\overline{\mathrm{EP}}$ | EP | 0 | 0 | VBI_WORD_2[7:4] |  |  |  | 0 | 0 | ID6 (User Data-Word 7) |
| 13 | $\overline{\mathrm{EP}}$ | EP | 0 | 0 | VBI_WORD_2[3:0] |  |  |  | 0 | 0 | ID7 (User Data-Word 8) |
| 14 | $\overline{\mathrm{EP}}$ | EP | 0 | 0 | VBI_WORD_3[7:4] |  |  |  | 0 | 0 | ID8 (User Data-Word 9) |
|  |  |  |  |  |  |  |  |  |  |  | Pad 0x200; these padding words may be present, depending on ancillary data type; user data-word |
| $\mathrm{n}-3$ | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |  |
| $\mathrm{n}-2$ | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |  |
| $\mathrm{n}-1$ | $\overline{\mathrm{B} 8}$ | Checksum (CS) |  |  |  |  |  |  | 0 | 0 | CS (checksum word) |

Table 81. Ancillary Data in Byte Output Format ${ }^{1}$

| Byte | B9 | B8 | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 | Description |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | Ancillary data preamble |
| 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |  |
| 2 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |  |
| 3 | $\overline{\mathrm{EP}}$ | EP | 0 | $1^{2}$ C_DID6_2[4:0] |  |  |  |  | 0 | 0 | DID |
| 4 | $\overline{\mathrm{EP}}$ | EP | $1^{2} \mathrm{C}$ _SDID7_2[5:0] |  |  |  |  |  | 0 | 0 | SDID |
| 5 | $\overline{\mathrm{EP}}$ | EP | 0 | DC[4:0] |  |  |  |  | 0 | 0 | Data count |
| 6 | $\overline{\mathrm{EP}}$ | EP | Padding[1:0] |  | VBI_DATA_STD[3:0] |  |  |  | 0 | 0 | ID0 (User Data-Word 1) |
| 7 | $\overline{\mathrm{EP}}$ | EP | 0 | LINE_NUMBER[9:5] |  |  |  |  | 0 | 0 | ID1 (User Data-Word 2) |
| 8 | $\overline{\mathrm{EP}}$ | EP | EVEN_FIELD | LINE_NUMBER[4:0] |  |  |  |  | 0 | 0 | ID2 (User Data-Word 3) |
| 9 | $\overline{\mathrm{EP}}$ | EP | 0 | 0 | 0 | 0 | VDP_TTXT_TYPE[1:0] |  | 0 | 0 | ID3 (User Data-Word 4) |
| 10 | VBI_WORD_1[7:0] |  |  |  |  |  |  |  | 0 | 0 | ID4 (User Data-Word 5) |
| 11 | VBI_WORD_2[7:0] |  |  |  |  |  |  |  | 0 | 0 | ID5 (User Data-Word 6) |
| 12 | VBI_WORD_3[7:0] |  |  |  |  |  |  |  | 0 | 0 | ID6 (User Data-Word 7) |
| 13 | VBI_WORD_4[7:0] |  |  |  |  |  |  |  | 0 | 0 | ID7 (User Data-Word 8) |
| 14 | VBI_WORD_5[7:0] |  |  |  |  |  |  |  | 0 | 0 | ID8 (User Data-Word 9) |
|  |  |  |  |  |  |  |  |  |  |  | Pad 0x200; these padding words may be present, depending on ancillary data type; user data-word |
| $\mathrm{n}-3$ | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |  |
| $\mathrm{n}-2$ | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |  |
| $\mathrm{n}-1$ | $\overline{\mathrm{B} 8}$ | Checksum |  |  |  |  |  |  | 0 | 0 | CS (checksum word) |

${ }^{1}$ This mode does not fully comply with ITU-R BT.1364.

## Structure of VBI Words in the Ancillary Data Stream

Each VBI data standard has been split into a clock-run-in (CRI), a framing code (FC), and a number of data bytes ( n ). The data packet in the ancillary stream includes only the FC and data bytes. Table 82 shows the format of VBI_WORD_x in the ancillary data stream.

Table 82. Structure of VBI Data-Words in the Ancillary Stream

| Ancillary Data Byte No. | Byte Type | Description |
| :--- | :--- | :--- |
| VBI_WORD_1 | FC0 | Framing Code[23:16] |
| VBI_WORD_2 | FC1 | Framing Code[15:8] |
| VBI_WORD_3 | FC2 | Framing Code[7:0] |
| VBI_WORD_4 | DB1 | First data byte |
| $\ldots$ | $\ldots$ | $\ldots$ |
| VBI_WORD_N +3 | DBn | Last $\left(\mathrm{n}^{\text {th }}\right)$ data byte |

## VDP Framing Code

The length of the actual framing code depends on the VBI data standard. For uniformity, the length of the framing code reported in the ancillary data stream is always 24 bits. For standards with a smaller framing code length, the extra LSB bits are set to 0 . The valid length of the framing code can be decoded from the VBI_DATA_STD bits available in ID0 (UDW 1). The framing code is always reported in the inverse-transmission order.

Table 83 shows the framing code and its valid length for VBI data standards supported by VDP.

## Example

For teletext (B-WST), the framing code byte is 11100100 ( 0 xE 4 ), with bits shown in the order of transmission. VBI_WORD_1 = $0 \times 27$, VBI_WORD $\_2=0 \times 00$, and VBI_WORD_3 $=0 \times 00$ translated into UDWs in the ancillary data stream for nibble mode are as follows:

UDW5[5:2] $=0010$
UDW6[5:2] = 0111
UDW7[5:2] $=0000$ (undefined bits set to 0 )
UDW8[5:2] $=0000$ (undefined bits set to 0)
UDW9[5:2] $=0000$ (undefined bits set to 0 )
UDW10[5:2] = 0000 (undefined bits set to 0 )
For byte mode,
UDW5[9:2] = 0010_0111
UDW6[9:2] = 0000_0000 (undefined bits set to 0)
UDW7[9:2] = 0000_0000 (undefined bits set to 0)

## Data Bytes

VBI_WORD_4 to VBI_WORD_N + 3 contain the data-words that were decoded by the VDP in the transmission order. The position of bits in bytes is in the inverse transmission order.

For example, closed captioning has two user data bytes, as shown in Table 88.

The data bytes in the ancillary data stream are as follows:

$$
\begin{aligned}
& \text { VBI_WORD_4 }=\text { Byte } 1[7: 0] \\
& \text { VBI_WORD_5 }=\text { Byte } 2[7: 0]
\end{aligned}
$$

The number of VBI_WORDS for each VBI data standard and the total number of UDWs in the ancillary data stream is shown in Table 84.

Table 83. Framing Code Sequence for Different VBI Standards

| VBI Standard | Length in Bits | Error-Free Framing Code Bits <br> (in Order of Transmission) | Error-Free Framing Code Reported by <br> VDP (in Reverse Order of Transmission) |
| :--- | :--- | :--- | :--- |
| TTXT_SYSTEM_A (PAL) | 8 | 11100111 | 11100111 |
| TTXT_SYSTEM_B (PAL) | 8 | 11100100 | 00100111 |
| TTXT_SYSTEM_B (NTSC) | 8 | 11100100 | 00100111 |
| TTXT_SYSTEM_C (PAL and NTSC) | 8 | 11100111 | 11100111 |
| TTXT_SYSTEM_D (PAL and NTSC) | 8 | 11100101 | 10100111 |
| WSS (PAL) | 24 | 000111100011110000011111 | 111110000011110001111000 |
| CCAP (NTSC and PAL) | 001 | 100 |  |
| CGMS (NTSC) | 0 | 0 |  |

Table 84. Total User Data-Words for Different VBI Standards ${ }^{1}$

| VBI Standard | ADF Mode | Framing Code UDWs | VBI Data-Words | No. of Padding Words | Total UDWs |
| :--- | :--- | :--- | :--- | :--- | :--- |
| TTXT_SYSTEM_A (PAL) | 00 (nibble mode) | 6 | 74 | 84 |  |
|  | 01,10 (byte mode) | 3 | 37 | 0 | 44 |
| TTXT_SYSTEM_B (PAL) | 00 (nibble mode) | 6 | 84 | 96 |  |
| TTXT_SYSTEM_B (NTSC) | 01,10 (byte mode) | 3 | 42 | 52 |  |
|  | 00 (nibble mode) | 6 | 68 | 3 | 80 |
| TTXT_SYSTEM_C (PAL and NTSC) | 01,10 (byte mode) | 3 | 34 | 44 |  |
|  | 00 (nibble mode) | 6 | 66 | 76 |  |
| TTXT_SYSTEM_D (PAL and NTSC) | 01,10 (byte mode) | 3 | 3 | 42 |  |
|  | 00 (nibble mode) | 6 | 3 | 80 |  |
| WSS (PAL) | 01,10 (byte mode) | 3 | 68 | 44 |  |
|  | 00 (nibble mode) | 6 | 34 | 16 |  |
| CCAP (NTSC and PAL) | 01,10 (byte mode) | 3 | 4 | 12 |  |
|  | 00 (nibble mode) | 6 | 2 | 16 |  |
| CGMS (NTSC) | 01,10 (byte mode) | 3 | 4 | 12 |  |
|  | 00 (nibble mode) | 6 | 2 | 16 |  |

[^1]
## $1^{2}$ C Interface

Dedicated $\mathrm{I}^{2} \mathrm{C}$ readback registers are available for CCAP, CGMS, and WSS. Because teletext is a high data rate standard, data extraction is supported only through the ancillary data packet.

## User Interface for $\mathrm{I}^{2} \mathrm{C}$ Readback Registers

The VDP decodes all enabled VBI data standards in real-time. Because the $\mathrm{I}^{2} \mathrm{C}$ access speed is much lower than the decoded rate, when the registers are accessed, they may be updated with data from the next line. To avoid this, VDP has a self-clearing clear bit and an available (AVL) status bit accompanying all $\mathrm{I}^{2} \mathrm{C}$ readback registers.

The user must clear the $\mathrm{I}^{2} \mathrm{C}$ readback register by writing a high to the clear bit. This resets the state of the available bit to low and indicates that the data in the associated readback registers is not valid. After the VDP decodes the next line of the corresponding VBI data, the decoded data is placed into the $\mathrm{I}^{2} \mathrm{C}$ readback register, and the available bit is set to high to indicate that valid data is now available.

Though the VDP decodes this VBI data in subsequent lines, if present, the decoded data is not updated to the readback registers until the clear bit is set high again. However, this data is available through the 656 ancillary data packets.
The clear and available bits are in the VDP_STATUS_CLEAR (0x78, Interrupt/VDP Map, write only) and VDP_STATUS (0x78, Interrupt/VDP Map, read only) registers, respectively.

## VDP—Content-Based Data Update

For certain standards such as WSS and CGMS, the information content in the signal transmitted remains the same over numerous lines, and the user may want to be notified only when there is a change in the information content or loss of the information content. The user must enable content-based updating for the required standard through the WSS_CGMS_CB_CHANGE bit. This bit shows the availability of WSS and CGMS information only when its content has changed.
Content-based updating also applies to lines with lost data. Therefore, for standards like CGMS and WSS, if no data arrives in the next four lines programmed, the corresponding available bit in the VDP_STATUS register is set high, and the content in the $I^{2} \mathrm{C}$ registers for that standard is set to 0 . The user must write high to the corresponding clear bit so that when a valid line is decoded after some time, the decoded results are available in the $I^{2} \mathrm{C}$ registers, with the available status bit set high.
If content-based updating is enabled, the available bit is set high (assuming the clear bit was written) in the following cases:

- The data contents have changed.
- Data was being decoded and four lines with no data have been detected.
- No data was being decoded, and new data is now being decoded.


## WSS_CGMS_CB_CHANGE, Enable Content-Based Updating for WSS/CGMS, Address 0x9C[4], Interrupt/VDP Map

Setting WSS_CGMS_CB_CHANGE to 0 disables content-based updating.

Setting WSS_CGMS_CB_CHANGE to 1 (default) enables content-based updating.

## VDP—Interrupt-Based Reading of VDP I ${ }^{2}$ C Registers

Some VDP status bits are also linked to the interrupt request controller so that the user does not have to poll the available status bit. The user can configure the video decoder to trigger an interrupt request on the INTRQ pin in response to the valid data available in the $\mathrm{I}^{2} \mathrm{C}$ registers. This function is available for the following data types: CGMS or WSS. The user can select either triggering an interrupt request each time sliced data is available or triggering an interrupt request only when the sliced data has changed. Selection is made via the WSS_CGMS_CB_ CHANGE bit.

The sequence for the interrupt-based reading of the VDP I ${ }^{2} \mathrm{C}$ data registers is as follows for the CCAP standard:

1. The user unmasks the CCAP interrupt mask bit (Register 0x50, Bit 0, Interrupt/VDP Map = 1). CCAP data occurs on the incoming video. VDP slices CCAP data and places it into the VDP readback registers.
2. The VDP CCAP available bit, VDP_CCAPD_Q, goes high, and the VDP module signals to the interrupt controller to stimulate an interrupt request (for CCAP in this case).
3. The user reads the interrupt status bits (Interrupt/VDP Map) and sees that new CCAP data is available (Register $0 x 4 \mathrm{E}$, Bit 0 , Interrupt/VDP Map = 1).
4. The user writes 1 to the CCAP interrupt clear bit (Register 0x4F, Bit 0 , Interrupt/VDP Map $=1$ ) in the interrupt $\mathrm{I}^{2} \mathrm{C}$ space (this is a self-clearing bit). This clears the interrupt on the INTRQ pin but does not have an effect in the VDP $I^{2} \mathrm{C}$ area.
5. The user reads the CCAP data from the VDP I ${ }^{2} \mathrm{C}$ area.
6. The user writes to Bit CC_CLEAR in the VDP_STATUS_CLEAR register, (Register 0x78, Bit 0, User Sub Map $2=1$ ) to signify that the CCAP data has been read (therefore the VDP CCAP can be updated at the next occurrence of CCAP).
7. The user goes back to Step 2.

## Interrupt Mask Register Details

The following bits set the interrupt mask on the signal from the VDP VBI data slicer.

## VDP_CCAPD_MSK, Address 0x50[0], Interrupt/VDP Map

Setting $\overline{\text { VDP_CCAPD_MSK }}$ to 0 (default) masks (disables) the interrupt on the VDP_CCAPD_Q signal.
Setting VDP_CCAPD_MSK to 1 unmasks (enables) the interrupt on the VDP_CCAPD_Q signal.

VDP_CGMS_WSS_CHNGD_MSK, Address 0x50[2], Interrupt/VDP Map
Setting $\overline{\text { VDP_CGMS_WSS_CHNGD_MSK }}$ to 0 (default) masks (disables) the interrupt on the VDP_CGMS_WSS_CHNGD_Q signal.
Setting VDP_CGMS_WSS_CHNGD_MSK to 1 unmasks (enables) the interrupt on the VDP_CGMS_WSS_CHNGD_Q signal.

## Interrupt Status Register Details

The following read-only bits contain data detection information from the VDP module since the status bit was last cleared or unmasked.

VDP_CCAPD_Q, Address 0x4E[0], Interrupt/VDP Map
When VDP_CCAPD_Q is 0 (default), CCAP data has not been detected.

When VDP_CCAPD_Q is 1, CCAP data has been detected.
VDP_CGMS_WSS_CHNGD_Q, Address 0x4E[2], Interrupt/VDP Map
When VDP_CGMS_WSS_CHNGD_Q is 0 (default), CGMS or WSS data has not been detected.

VDP_CGMS_WSS_CHNGD_Q is also 0 if the CGMS or WSS data state has not changed since last cleared (see Interrupt Status Clear Register Details).
When VDP_CGMS_WSS_CHNGD_Q is 1, CGM or WSS data has been detected.

## Interrupt Status Clear Register Details

It is not necessary to write 0 to these write-only bits because they automatically reset after they have been set to 1 (self-clearing).

VDP_CCAPD_CLR, Address 0x4F[0], Interrupt/VDP Map
Setting VDP_CCAPD_CLR to 1 clears the VDP_CCAP_Q bit.
VDP_CGMS_WSS_CHNGD_CLR, Address 0x4F[2], Interrupt/VDP Map
Setting VDP_CGMS_WSS_CHNGD_CLR to 1 clears the VDP_CGMS_WSS_CHNGD_Q bit.

## $I^{2} C$ READBACK REGISTERS

## Teletext

Because teletext is a high data rate standard, the decoded bytes are available only as ancillary data. However, a TTXT_AVL bit has been provided in $\mathrm{I}^{2} \mathrm{C}$ so that the user can check whether the VDP has detected teletext. Note that the TTXT_AVL bit is a plain status bit and does not use the protocol identified in the $\mathrm{I}^{2} \mathrm{C}$ Interface section.

TTXT_AVL, Teletext Detected Status, Address 0x78[7], Interrupt/VDP Map, Read Only

When TTXT_AVL is 0 , teletext was not detected.
When TTXT_AVL is 1 , teletext was detected.

## WST Packet Decoding

For WST only, the VDP decodes the magazine and row address of teletext packets and further decodes the packet's $8 \times 4$ hamming coded words. This feature can be disabled using the WST_PKT_DECODE_DISABLE bit (Bit 3, Register 0x60, user sub map). This feature is valid for WST only.

## WST_PKT_DECODE_DISABLE, Disable Hamming Decoding of Bytes in WST, Address 0x60[3], Interrupt/VDP Map

Setting WST_PKT_DECODE_DISABLE to 0 enables hamming decoding of WST packets.
Setting WST_PKT_DECODE_DISABLE to 1 (default) disables hamming decoding of WST packets.

For hamming-coded bytes, the de-hammed nibbles are output along with some error information from the hamming decoder as follows:

- Input hamming coded byte: $\{\mathrm{D} 3, \mathrm{P} 3, \mathrm{D} 2, \mathrm{P} 2, \mathrm{D} 1, \mathrm{P} 1, \mathrm{D} 0, \mathrm{P} 0\}$ (bits in decoded order)
- Output dehammed byte: \{E1, E0, 0, 0, D3', D2', D1', D0'\}
(Di' - corrected bits, Ei error information).
Table 85. Error Bits in the Dehammed Output Byte

| E[1:0] | Error Information | Output Data Bits <br> in Nibble |
| :--- | :--- | :--- |
| 00 | No errors detected | Okay |
| 01 | Error in P4 | Okay |
| 10 | Double error | Bad |
| 11 | Single error found and corrected | Okay |

Table 86 describes the WST packets that are decoded.

Table 86. WST Packet Description

| Packet | Byte | Description |
| :---: | :---: | :---: |
| Header Packet (X/00) | 1st <br> 2nd <br> 3rd <br> 4th <br> 5th to 10th <br> 11th to 42nd | Magazine number—Dehammed Byte 4 <br> Row number-Dehammed Byte 5 <br> Page number-Dehammed Byte 6 <br> Page number-Dehammed Byte 7 <br> Control bytes—Dehammed Byte 8 to Byte 13 <br> Raw data bytes |
| Text Packets (X/01 to X/25) | 1st <br> 2nd <br> 3rd to 42nd | Magazine number—Dehammed Byte 4 Row number-Dehammed Byte 5 Raw data bytes |
| 8/30 (Format 1) Packet Design Code $=0000$ or 0001 UTC | 1st <br> 2nd <br> 3rd <br> 4th to 10th <br> 11th to 23 rd <br> 24th to 42nd | Magazine number—Dehammed Byte 4 <br> Row number-Dehammed Byte 5 <br> Design code—Dehammed Byte 6 <br> De-hammed initial teletext page, Byte 7 to Byte 12 <br> UTC bytes—Dehammed Byte 13 to Byte 25 <br> Raw status bytes |
| 8/30 (Format 2) Packet Design Code $=0010$ or 0011 PDC | 1st <br> 2nd <br> 3rd <br> 4th to 10th <br> 11th to 23rd <br> 24th to 42nd | Magazine number—Dehammed Byte 4 <br> Row number-Dehammed Byte 5 <br> Design code—Dehammed Byte 6 <br> Dehammed initial teletext page, Byte 7 to Byte 12 <br> PDC bytes—Dehammed Byte 13 to Byte 25 <br> Raw status bytes |
| $\mathrm{X} / 26, \mathrm{X} / 27, \mathrm{X} / 28, \mathrm{X} / 29, \mathrm{X} / 30, \mathrm{X} / 31^{1}$ | 1st 2nd 3rd 4th to 42nd | Magazine number—Dehammed Byte 4 <br> Row number-Dehammed Byte 5 <br> Design code—Dehammed Byte 6 <br> Raw data bytes |

${ }^{1}$ For $\mathrm{X} / 26, \mathrm{X} / 28$, and $\mathrm{X} / 29$, further decoding needs $24 \times 18$ hamming decoding. Not supported at present.


Figure 42. WSS Waveform

## CGMS and WSS

The CGMS and WSS data packets convey the same type of information for different video standards. WSS is for PAL and CGMS is for NTSC; therefore, the CGMS and WSS readback registers are shared. WSS is biphase coded; the VDP performs a biphase decoding to produce the 14 raw WSS bits in the CGMS/ WSS readback $I^{2} \mathrm{C}$ registers and to set the CGMS_WSS_AVL bit.

CGMS_WSS_CLEAR, CGMS/WSS Clear, Address 0x78[2], Interrupt/VDP Map, Write Only, Self-Clearing
Setting CGMS_WSS_CLEAR to 0 does not reinitialize the CGMS/WSS readback registers.
Setting CGMS_WSS_CLEAR to 1 reinitializes the CGMS/WSS readback registers.

CGMS_WSS_AVL, CGMS/WSS Available, Address 0x78[2], User Sub Map, Read Only
When CGMS_WSS_AVL is 0, CGMS/WSS was not detected.
When CGMS_WSS_AVL is 1, CGMS/WSS was detected.

## VDP_CGMS_WSS_DATA_0[3:0], Address 0x7D[3:0]; <br> VDP_CGMS_WSS_DATA_1[7:0], Address 0x7E[7:0]; <br> VDP_CGMS_WSS_DATA_2[7:0], Address 0x7F[7:0]; Interrupt/VDP Map, Read Only

These bits hold the decoded CGMS or WSS data.
Refer to Figure 42 and Figure 43 for the $\mathrm{I}^{2} \mathrm{C}$-to-WSS and $\mathrm{I}^{2} \mathrm{C}$-toCGMS bit mapping.


Figure 43. CGMS Waveform
Table 87. CGMS Readback Registers (These registers are readback registers; default value does not apply.)


Table 88. CCAP Readback Registers (These registers are readback registers; default value does not apply.)

| Signal Name | Register Location | Dec | Hex |
| :--- | :--- | :--- | :--- |
| CCAP_BYTE_1[7:0] | VDP_CCAP_DATA_0[7:0] | 121 | $0 \times 79$ |
| CCAP_BYTE_2[7:0] | VDP_CCAP_DATA_1[7:0] | 122 | $0 \times 7 A$ |

## CCAP

Two bytes of decoded closed caption data are available in the $I^{2} \mathrm{C}$ registers. The field information of the decoded CCAP data can be obtained from the CC_EVEN_FIELD bit (Register 0x78).
CC_CLEAR, Closed Caption Clear, Address 0x78[0], Interrupt/VDP Map, Write Only, Self-Clearing
Setting CC_CLEAR to 0 does not reinitializes the CCAP readback registers.

Setting CC_CLEAR to 1 reinitializes the CCAP readback registers.
CC_AVL, Closed Caption Available, Address 0x78[0], Interrupt/VDP Map, Read Only
When CC_AVL is 0 , closed captioning is not detected.
When CC_AVL is 1 , closed captioning is detected.

## CC_EVEN_FIELD, Address 0x78[1], Interrupt/VDP Map, Read Only

Identifies the field from which the CCAP data was decoded.
When CC_EVEN_FIELD is 0 , closed captioning was detected from an odd field.
When CC_EVEN_FIELD is 1 , closed captioning was detected from an even field.

VDP_CCAP_DATA_0, Address 0x79[7:0], Interrupt/VDP Map, Read Only

Decoded Byte 1 of CCAP data.
VDP_CCAP_DATA_1, Address 0x7A[7:0], Interrupt/VDP Map, Read Only

Decoded Byte 2 of CCAP data.

## Letterbox Detection

Incoming video signals may conform to different aspect ratios (16:9 wide screen or $4: 3$ standard). For certain transmissions in the wide-screen format, a digital sequence (WSS) is transmitted with the video signal. If a WSS sequence is provided, the aspect ratio of the video can be derived from the digitally decoded bits that WSS contains.
In the absence of a WSS sequence, letterbox detection can be used to find wide-screen signals. The detection algorithm examines the active video content of lines at the start and end of a field. If black lines are detected, this may indicate that the currently shown picture is in wide-screen format.

The active video content (luminance magnitude) over a line of video is summed together. At the end of a line, this accumulated value is compared with a threshold, and a decision is made as to whether or not a particular line is black. The threshold value needed may depend on the type of input signal; some control is provided via LB_TH[4:0].

## Detection at the Start of a Field

The ADV7182 expects a section of at least six consecutive black lines of video at the top of a field. After those lines are detected, LB_LCT[7:0] reports the number of black lines that were actually found. By default, the ADV7182 starts looking for those black lines in sync with the beginning of active video, for example, immediately after the last VBI video line. LB_SL[3:0] allows the user to set the start of letterbox detection from the beginning of a frame on a line-by-line basis. The detection window closes in the middle of the field.

## Detection at the End of a Field

The ADV7182 expects at least six continuous lines of black video at the bottom of a field before reporting the number of lines actually found via the LB_LCB[7:0] value. The activity window for letterbox detection (end of field) starts in the middle of an active field. Its end is programmable via LB_EL[3:0].

## Detection at the Midrange

Some transmissions of wide-screen video include subtitles within the lower black box. If the ADV7182 finds at least two black lines followed by some more nonblack video, for example, the subtitle followed by the remainder of the bottom black block, it reports a midcount via LB_LCM[7:0]. If no subtitles are found, LB_LCM[7:0] reports the same number as LB_LCB[7:0].

There is a two-field delay in reporting any line count parameter.
There is no letterbox detected bit. Read the LB_LCT[7:0] and LB_LCB[7:0] register values to determine whether the letterboxtype video is present in the software.

LB_LCT[7:0], Letterbox Line Count Top, Address 0x9B[7:0];
LB_LCM[7:0], Letterbox Line Count Mid, Address 0x9C[7:0];
LB_LCB[7:0], Letterbox Line Count Bottom, Address 0x9D[7:0]
Table 89. LB_LCx Access Information

| Signal Name | Address |
| :--- | :--- |
| LB_LCT[7:0] | $0 \times 9 \mathrm{~B}$ |
| LB_LCM[7:0] | $0 \times 9 \mathrm{C}$ |
| LB_LCB[7:0] | $0 \times 9 \mathrm{D}$ |

LB_TH[4:0], Letterbox Threshold Control, Address 0xDC[4:0]

Table 90. LB_TH Function

| LB_TH[4:0] | Description |
| :--- | :--- |
| 01100 (default) | Default threshold for detection of black lines <br> 01101 to 10000 <br> Increase threshold (need larger active video <br> content before identifying nonblack lines) |
| 00000 to 01011 | Decrease threshold (even small noise levels <br> can cause the detection of nonblack lines) |

LB_SL[3:0], Letterbox Start Line, Address 0xDD[7:4]
The LB_SL[3:0] bits are set at 1100 by default. For an NTSC signal, this window is from Line 31 to Line 294.
By changing the bits to 0100 , the detection window starts on Line 23 and ends on Line 286.

## LB_EL[3:0], Letterbox End Line, Address 0xDD[3:0]

The LB_EL[3:0] bits are set at 1100 by default. This means that the letterbox window ends with the last active video line. For an NTSC signal, this window is from Line 261 to Line 524.

By changing the bits to 1101 , the detection window starts on Line 262 and ends on Line 255.

## ITU-R BT. 656 Tx CONFIGURATION

The ADV7182 receives analog video and outputs digital video according to the ITU-R BT. 656 specification. The ADV7182 outputs the ITU-R BT. 656 video data stream over the P0 to P7 data pins and has a line locked clock (LLC) pin and two synchronization pins (HS and VS/FIELD/SFL).
Video data is output over the P 0 to P 7 pins in $\mathrm{YCrCb} 4: 2: 2$ format. Synchronization signals are automatically embedded in the video data signal in accordance with the ITU-R BT. 656 specification.

The LLC output is used to clock the output data on the P0 to P7 pins at a nominal frequency of 27 MHz .
The two synchronization pins (HS and VS/FIELD/SFL) output a variety of synchronization signals such as horizontal sync, vertical sync, field sync, and color subcarrier frequency lock (SFL) sync. The majority of these synchronization signals are already embedded in the video data. Therefore, the use of the synchronization pins is optional.
The following registers can be used to configure the ITU-R BT. 656 output from the ADV7182. See the Global Control Registers section for further registers that can be used to configure the output from the ADV7182.

## SWPC, Swap Pixel Cr/Cb, Address 0x27[7]

This bit allows Cr and Cb samples to be swapped.
When SWPC is 0 (default), no swapping is allowed. When SWPC is 1 , the Cr and Cb values can be swapped.

## Polarity LLC Pin

PCLK, Address 0x37[0]
The polarity of the clock that leaves the ADV7182 via the LLC pin can be inverted using the PCLK bit. Changing the polarity of
the LLC clock output may be necessary to meet the setup-and-hold time expectations of follow-on chips. When PCLK is 0 , the LLC output polarity is inverted. When PCLK is 1 (default), the LLC output polarity is normal (see the Timing Specifications section.)

## LLC_PAD_SEL[2:0] LLC Output Selection, Address 0x8F[6:4]

The following $\mathrm{I}^{2} \mathrm{C}$ write allows the user to select between LLC (nominally at 27 MHz ) and LLC (nominally at 13.5 MHz ).

When LLC_PAD_SEL is 000, the output is nominally 27 MHz LLC on the LLC pin (default).
When LLC_PAD_SEL is 101, the output is nominally 13.5 MHz LLC on the LLC pin.

## BT.565-4, ITU-R BT.565-3/ITU-R BT.565-4 Enable, Address 0x04[7]

Between Revision 3 and Revision 4 of the ITU-R BT. 656 standards, the toggling position for the V bit within the SAV EAV codes for NTSC has changed. Note that the standard change affects only NTSC and has no bearing on PAL. For further information, visit the International Telecommunication Union website.

The BT.656-4 bit allows the user to select an output mode that is compliant with either the ITU-R BT.656-3 standard or ITU-R BT.656-4 standard.

When the BT.656-4 bit is set to 0 (default), the ITU-R BT.656-3 specification is used. The V bit goes low at EAV of Line 10 and Line 273.

When ITU-R BT.656-4 is 1 , the ITU-R BT.656-4 specification is used. The V bit goes low at EAV of Line 20 and Line 283.


Figure 45. ITU-R BT. 656 Output Stage of the ADV7182

## MPU PORT DESCRIPTION

The ADV7182 supports a 2-wire ( ${ }^{2} \mathrm{C}$-compatible) serial interface. Two inputs, serial data (SDATA) and serial clock (SCLK), carry information between the ADV7182 and the system $\mathrm{I}^{2} \mathrm{C}$ master controller. Each slave device is recognized by a unique address. The ADV7182 $\mathrm{I}^{2} \mathrm{C}$ port allows the user to set up and configure the decoder and to read back the captured VBI data. The ADV7182 has four possible slave addresses for both read and write operations, depending on the logic level of the ALSB pin. The four unique addresses are shown in Table 91. The ADV7182 ALSB pin controls Bit 1 of the slave address. By altering the ALSB, it is possible to control two ADV7182s in an application without the conflict of using the same slave address. The LSB (Bit 0 ) sets either a read or write operation. Logic 1 corresponds to a read operation, and Logic 0 corresponds to a write operation.

Table 91. ${ }^{2}{ }^{2} \mathrm{C}$ Address for ADV7182

|  |  | Slave Address |  |
| :--- | :--- | :--- | :--- |
| ALSB | R/W | Hexadecimal | Binary |
| 0 | 0 (read) | $0 \times 40$ | $0 b^{\prime} 01000000$ |
| 0 | 1 (write) | $0 \times 41$ | $0 b^{\prime} 01000001$ |
| 1 | 0 (read) | $0 \times 42$ | $0 b^{\prime} 01000010$ |
| 1 | 1 (write) | $0 \times 43$ | $0 b^{\prime} 01000011$ |

To control the device on the bus, a specific protocol must be followed. First, the master initiates a data transfer by establishing a start condition, which is defined by a high-to-low transition on SDATA while SCLK remains high. This indicates that an address/ data stream follows. All peripherals respond to the start condition and shift the next eight bits (the 7-bit address plus the R/W bit). The bits are transferred from MSB down to LSB. The peripheral that recognizes the transmitted address responds by pulling the data line low during the ninth clock pulse; this is known as an acknowledge bit. All other devices withdraw from the bus at this point and maintain an idle condition. The idle condition is where the device monitors the SDATA and SCLK lines for the
start condition and the correct transmitted address. The R/ $\overline{\mathrm{W}}$ bit determines the direction of the data. Logic 0 on the LSB of the first byte means that the master writes information to the peripheral. Logic 1 on the LSB of the first byte means that the master reads information from the peripheral.
The ADV7182 acts as a standard slave device on the bus. The data on the SDATA pin is eight bits long, supporting the 7-bit address plus the $\mathrm{R} / \overline{\mathrm{W}}$ bit. The device has subaddresses to enable access to the internal registers. It, therefore, interprets the first byte as the device address and the second byte as the starting subaddress. The subaddresses auto-increment, allowing data to be written to or read from the starting subaddress. A data transfer is always terminated by a stop condition. The user can also access any unique subaddress register on a one-by-one basis without updating all the registers.

Stop and start conditions can be detected at any stage during the data transfer. If these conditions are asserted out of sequence with normal read and write operations, they cause an immediate jump to the idle condition. During a given SCLK high period, only issue one start condition, one stop condition, or a single stop condition followed by a single start condition. If an invalid subaddress is issued, the ADV7182 does not issue an acknowledge and returns to the idle condition.

In autoincrement mode, if the user exceeds the highest subaddress, the following action is taken:

- In read mode, the highest subaddress register contents continue to be output until the master device issues a no acknowledge. This indicates the end of a read. A no acknowledge condition occurs when the SDATA line is not pulled low on the ninth pulse.
- In write mode, the data for the invalid byte is not loaded into any subaddress register. A no acknowledge is issued by the ADV7182, and the part returns to the idle condition.


Figure 46. Bus Data Transfer


Figure 47. Read and Write Sequence

## REGISTER ACCESS

The MPU can write to or read from all of the ADV7182 registers except the subaddress register, which is write only. The subaddress register determines which register the next read or write operation accesses. All communications with the part through the bus start with an access to the subaddress register. A read/write operation is then performed from or to the target address, which increments to the next address until a stop command on the bus is performed. Note that only the registers listed in the $\mathrm{I}^{2} \mathrm{C}$ Register Maps section are supported.

## REGISTER PROGRAMMING

The following sections describe the configuration for each register. The communication register is an 8-bit, write-only register. After the part is accessed over the bus and a read/write operation is selected, the subaddress is set up. The subaddress register determines to or from which register the operation takes place.

## SUB_USR_EN, Address Ox0E[6:5]

The ADV7182 has three $\mathrm{I}^{2} \mathrm{C}$ maps. The main register map (see Table 92) is the map that is available by default. The other two maps are accessed using the SUB_USR_EN bit (Address 0x0E). When programming of these maps is completed, it is necessary to write to the SUB_USR_EN bit to return to the main register map (see Table 92).


Figure 48. Register Access—Main Register Map, Interrupt/VDP Map, and User Sub Map 2

## Register Select (SR7 to SRO)

These bits are set up to point to the required starting address.

## $I^{2} \mathbf{C}$ SEQUENCER

An $\mathrm{I}^{2} \mathrm{C}$ sequencer is used when a parameter exceeds eight bits and is, therefore, distributed over two or more $\mathrm{I}^{2} \mathrm{C}$ registers, for example, HSB[10:0].

When such a parameter is changed using two or more $\mathrm{I}^{2} \mathrm{C}$ write operations, the parameter can hold an invalid value for the time between the first $\mathrm{I}^{2} \mathrm{C}$ being completed and the last $\mathrm{I}^{2} \mathrm{C}$ being completed. In other words, the top bits of the parameter may hold the new value while the remaining bits of the parameter still hold the previous value.
To avoid this problem, the $\mathrm{I}^{2} \mathrm{C}$ sequencer holds the updated bits of the parameter in local memory, and all bits of the parameter are updated together once the last register write operation has completed.
The correct operation of the $\mathrm{I}^{2} \mathrm{C}$ sequencer relies on the following:

- All I ${ }^{2} \mathrm{C}$ registers for the parameter in question must be written to in order of ascending addresses. For example, for HSB[10:0], write to Address $0 \times 34$ first, followed by $0 \times 35$, and so on.
- No other $\mathrm{I}^{2} \mathrm{C}$ can take place between the two (or more) $\mathrm{I}^{2} \mathrm{C}$ writes for the sequence. For example, for HSB[10:0], write to Address $0 \times 34$ first, immediately followed by $0 \times 35$, and so on.


## I²C REGISTER MAPS

To access all the registers listed in Table 92, SUB_USR_EN in Register Address 0x0E must be programmed to 00b. All read only registers are left blank.

Table 92. Main Register Map Details

| Address |  | Register Name | RW | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Reset <br> Value | (Hex) |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Dec | Hex |  |  |  |  |  |  |  |  |  |  |  |  |
| 0 | 00 | Input control | RW |  |  |  | INSEL[4] | INSEL[3] | INSEL[2] | INSEL[1] | INSEL[0] | 00001110 | OE |
| 1 | 01 | Video Selection 1 | RW |  | ENHSPLL | BETACAM |  | ENVSPROC |  |  |  | 11001000 | C8 |
| 2 | 02 | Video Selection 2 | RW | VID_SEL[3] | VID_SEL[2] | VID_SEL[1] | VID_SEL[0] |  |  |  |  | 00000100 | 04 |
| 3 | 03 | Output control | RW | VBI_EN | TOD |  |  |  |  |  |  | 01001100 | 4C |
| 4 | 04 | Extended output control | RW | BT.656-4 |  |  |  | TIM_OE | BL_C_VBI | EN_SFL_PIN | Range | 00110101 | 35 |
| 5 | 05 | Reserved |  |  |  |  |  |  |  |  |  |  |  |
| 6 | 06 | Reserved |  |  |  |  |  |  |  |  |  |  |  |
| 7 | 07 | Autodetect enable | RW | AD_SEC525_EN | AD_SECAM_EN | AD_N443_EN | AD_P60_EN | AD_PALN_EN | AD_PALM_EN | AD_NTSC_EN | AD_PAL_EN | 01111111 | 7F |
| 8 | 08 | Contrast | RW | CON[7] | CON[6] | CON[5] | CON[4] | CON[3] | CON[2] | CON[1] | CON[0] | 10000000 | 80 |
| 9 | 09 | Reserved |  |  |  |  |  |  |  |  |  |  |  |
| 10 | OA | Brightness Adjust | RW | BRI[7] | BRI[6] | BRI[5] | BRI[4] | BRI[3] | BRI[2] | BRI[1] | BRI[0] | 00000000 | 00 |
| 11 | OB | Hue Adjust | RW | HUE[7] | HUE[6] | HUE[5] | HUE[4] | HUE[3] | HUE[2] | HUE[1] | HUE[0] | 00000000 | 00 |
| 12 | OC | Default Value $Y$ | RW | DEF_Y[5] | DEF_Y[4] | DEF_Y[3] | DEF_Y[2] | DEF_Y[1] | DEF_Y[0] | $\begin{array}{\|l} \hline \text { DEF_VAL_ } \\ \text { AUTO_EN } \\ \hline \end{array}$ | DEF_VAL_EN | 00110110 | 36 |
| 13 | OD | Default Value C | RW | DEF_C[7] | DEF_C[6] | DEF_C[5] | DEF_C[4] | DEF_C[3] | DEF_C[2] | DEF_C[1] | DEF_C[0] | 01111100 | $7 C$ |
| 14 | OE | ADI Control 1 | RW |  | SUB_USR_EN[1] | SUB_USR_EN[0] |  |  |  |  |  | 00000000 | 00 |
| 15 | OF | Power management | RW | Reset |  | PWRDWN |  |  |  |  |  | 00100000 | 20 |
| 16 | 10 | Status 1 | R | COL_KILL | AD_RESULT[2] | AD_RESULT[1] | AD_RESULT[0] | FOLLOW_PW | FSC_LOCK | LOST_LOCK | IN_LOCK |  |  |
| 17 | 11 | IDENT | R | IDENT[7] | IDENT[6] | IDENT[5] | IDENT[4] | IDENT[3] | IDENT[2] | IDENT[1] | IDENT[0] | 00011100 | 1C |
| 18 | 12 | Status 2 | R |  |  | FSC NSTD | LL NSTD | MV AGC DET | MV PS DET | MVCS T3 | MVCS DET |  |  |
| 19 | 13 | Status 3 | R | PAL_SW_LOCK | Interlaced | STD FLD LEN | FREE_RUN_ACT | Reserved | SD_OP_50Hz | Reserved | INST_HLOCK |  |  |
| 20 | 14 | Analog clamp control | RW |  |  |  | CCLEN |  | FREE_RUN_- PAT_SEL. 2 | $\begin{aligned} & \text { FREE_RUN_-_ } \\ & \text { PAT_SEL. } 1 \end{aligned}$ | FREE_RUN_- PAT_SEL. 0 | 00010000 | 10 |
| 21 | 15 | Digital Clamp Control 1 | RW |  | DCT[1] | DCT[0] | DCFE |  |  |  |  | 0000xxxx | 00 |
| 22 | 16 | Reserved |  |  |  |  |  |  |  |  |  |  |  |
| 23 | 17 | Shaping Filter Control 1 | RW | CSFM[2] | CSFM[1] | CSFM[0] | YSFM[4] | YSFM[3] | YSFM[2] | YSFM[1] | YSFM[0] | 00000001 | 01 |
| 24 | 18 | Shaping Filter Control 2 | RW | WYSFMOVR |  |  | WYSFM[4] | WYSFM[3] | WYSFM[2] | WYSFM[1] | WYSFM[0] | 10010011 | 93 |
| 25 | 19 | Comb filter control | RW |  |  |  |  | NSFSEL[1] | NSFSEL[0] | PSFSEL[1] | PSFSEL[0] | 11110001 | F1 |
| 29 | 1D | ADI Control 2 | RW | TRI_LLC |  |  |  |  |  |  |  | 11000xxx | C0 |
| 39 | 27 | Pixel delay control | RW | SWPC | AUTO_PDC_EN | CTA[2] | CTA[1] | CTA[0] |  | LTA[1] | LTA[0] | 01011000 | 58 |
| 43 | 2B | Misc gain control | RW |  | CKE |  |  |  |  |  | PW_UPD | 11100001 | E1 |
| 44 | 2 C | AGC mode control | RW |  | LAGC[2] | LAGC[1] | LAGC[0] |  |  | CAGC[1] | CAGC[0] | 10101110 | AE |
| 45 | 2D | Chroma Gain Control 1 | W | CAGT[1] | CAGT[0] |  |  | CMG[11] | CMG[10] | CMG[9] | CMG[8] | 11110100 | F4 |
| 45 | 2D | Chroma Gain 1 | R |  |  |  |  | CG[11] | CG[10] | CG[9] | CG[8] |  |  |
| 46 | 2 E | Chroma Gain Control 2 | W | CMG[7] | CMG[6] | CMG[5] | CMG[4] | CMG[3] | CMG[2] | CMG[1] | CMG[0] | 00000000 | 00 |
| 46 | 2E | Chroma Gain 2 | R | CG[7] | CG[6] | CG[5] | CG[4] | CG[3] | CG[2] | CG[1] | CG[0] |  |  |
| 47 | 2 F | Luma Gain Control 1 | W | LAGT[1] | LAGT[0] |  |  | LMG[11] | LMG[10] | LMG[9] | LMG[8] | 1111xxxx | F0 |
| 47 | 2 F | Luma Gain 1 | R |  |  |  |  | LG[11] | LG[10] | LG[9] | LG[8] |  |  |
| 48 | 30 | Luma Gain Control 2 | W | LMG[7] | LMG[6] | LMG[5] | LMG[4] | LMG[3] | LMG[2] | LMG[1] | LMG[0] | xxxxxxxx | 00 |
| 48 | 30 | Luma Gain 2 | R | LG[7] | LG[6] | LG[5] | LG[4] | LG[3] | LG[2] | LG[1] | LG[0] |  |  |
| 49 | 31 | VS/FIELD Control 1 | RW |  |  |  | NEWAVMODE | HVSTIM |  |  |  | 00000010 | 02 |
| 50 | 32 | VS/FIELD Control 2 | RW | VSBHO | VSBHE |  |  |  |  |  |  | 01000001 | 41 |
| 51 | 33 | VS/FIELD Control 3 | RW | VSEHO | VSEHE |  |  |  |  |  |  | 10000100 | 84 |
| 52 | 34 | HS Position Control 1 | RW |  | HSB[10] | HSB[9] | HSB[8] |  | HSE[10] | HSE[9] | HSE[8] | 00000000 | 00 |
| 53 | 35 | HS Position Control 2 | RW | HSB[7] | HSB[6] | HSB[5] | HSB[4] | HSB[3] | HSB[2] | HSB[1] | HSB[0] | 00000010 | 02 |
| 54 | 36 | HS Position Control 3 | RW | HSE[7] | HSE[6] | HSE[5] | HSE[4] | HSE[3] | HSE[2] | HSE[1] | HSE[0] | 00000000 | 00 |
| 55 | 37 | Polarity | RW | PHS |  | PVS |  | PF |  |  | PCLK | 00010001 | 09 |
| 56 | 38 | NTSC comb control | RW | CTAPSN[1] | CTAPSN[0] | CCMN[2] | CCMN[1] | CCMN[0] | YCMN[2] | YCMN[1] | YCMN[0] | 10000000 | 80 |
| 57 | 39 | PAL comb control | RW | CTAPSP[1] | CTAPSP[0] | CCMP[2] | CCMP[1] | CCMP[0] | YCMP[2] | YCMP[1] | YCMP[0] | 11000000 | C0 |
| 58 | 3A | ADC control | RW |  |  |  |  | PWRDWN_MUX_0 | $\begin{array}{\|l\|} \hline \text { PWRDWN_MUX } \\ \hline-1 \end{array}$ | PWRDWN_MUX  <br>   | MUX PDN override | 00000000 | 00 |
| 61 | 3D | Manual window control | RW |  | CKILLTHR[2] | CKILLTHR[1] | CKILLTHR[0] |  |  |  |  | 00100010 | 22 |
| 65 | 41 | Resample control | RW |  | SFL_INV |  |  |  |  |  |  | 00000001 | 01 |
| 77 | 4D | CTI DNR Control 1 | RW |  |  | DNR_EN |  | CTI_AB[1] | CTI_AB[0] | CTI_AB_EN | CTI_EN | 11101111 | EF |
| 78 | 4 E | CTI DNR Control 2 | RW | CTI_C_TH[7] | CTI_C_TH[6] | CTI_C_TH[5] | CTI_C_TH[4] | CTI_C_TH[3] | CTI_C_TH[2] | CTI_C_TH[1] | CTI_C_TH[0] | 00001000 | 08 |
| 80 | 50 | DNR Noise Threshold 1 | RW | DNR_TH[7] | DNR_TH[6] | DNR_TH[5] | DNR_TH[4] | DNR_TH[3] | DNR_TH[2] | DNR_TH[1] | DNR_TH[0] | 00001000 | 08 |
| 81 | 51 | Lock count | RW | FSCLE | SRLS | COL[2] | COL[1] | COL[0] | CIL[2] | CIL[1] | CIL[0] | 00100100 | 24 |
| 96 | 60 | ADC Switch 3 | RW |  |  |  |  |  | MUX3[2] | MUX3[1] | MUX3[0] | 00010000 | 10 |
| 106 | 6A | Output Sync Select 1 | RW |  |  |  |  |  | HS_OUT_SEL[2] | HS_OUT_SEL[1] | HS_OUT_SEL[0] | 00000000 | 00 |
| 107 | 6B | Output Sync Select 2 | RW |  | FLD_OUT_SEL[2] | FLD_OUT_SEL[1] | FLD_OUT_SEL[0] |  |  |  |  | 00010010 | 12 |
| 143 | 8F | Free-Run Line Length 1 | W |  | LLC_PAD_SEL[2] | $\begin{aligned} & \hline \text { LLC_PAD_ } \\ & \text { SEL[1] } \\ & \hline \end{aligned}$ | $\begin{aligned} & \text { LLC_PAD_ } \\ & \text { SEL[0] } \end{aligned}$ |  |  |  |  | 00000000 | 00 |
| 153 | 99 | CCAP1 | R | CCAP1[7] | CCAP1[6] | CCAP1[5] | CCAP1[4] | CCAP1[3] | CCAP1[2] | CCAP1[1] | CCAP1[0] |  |  |


| Address |  | Register Name | RW | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | $\begin{array}{\|l\|l\|} \hline \text { Reset } \\ \text { Value } \end{array}$ | (Hex) |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Dec | Hex |  |  |  |  |  |  |  |  |  |  |  |  |
| 154 | 9A | CCAP2 | R | CCAP2[7] | CCAP2[6] | CCAP2[5] | CCAP2[4] | CCAP2[3] | CCAP2[2] | CCAP2[1] | CCAP2[0] |  |  |
| 155 | 9B | Letterbox 1 | R | LB_LCT[7] | LB_LCT[6] | LB_LCT[5] | LB_LCT[4] | LB_LCT[3] | LB_LCT[2] | LB_LCT[1] | LB_LCT[0] |  |  |
| 156 | 9 C | Letterbox 2 | R | LB_LCM[7] | LB_LCM[6] | LB_LCM[5] | LB_LCM[4] | LB_LCM[3] | LB_LCM[2] | LB_LCM[1] | LB_LCM[0] |  |  |
| 157 | 9D | Letterbox 3 | R | LB_LCB[7] | LB_LCB[6] | LB_LCB[5] | LB_LCB[4] | LB_LCB[3] | LB_LCB[2] | LB_LCB[1] | LB_LCB[0] |  |  |
| 178 | B2 | CRC enable | w |  |  |  |  |  | CRC_ENABLE |  |  | 00011100 | 1 C |
| 195 | C3 | ADC Switch 1 | RW | Reserved | MUX1[2] | MUX1[1] | MUX1[0] | Reserved | MUXO[2] | MUXO[1] | MUXO[0] | xxxxxxxx | 00 |
| 196 | C4 | ADC Switch 2 | RW | MAN_MUX_EN |  |  |  | Reserved | MUX2[2] | MUX2[1] | MUX2[0] | 0xxxxxxx | 00 |
| 220 | DC | Letterbox Control 1 | RW |  |  |  | LB_TH[4] | LB_TH[3] | LB_TH[2] | LB_TH[1] | LB_TH[0] | 10101100 | AC |
| 221 | DD | Letterbox Control 2 | RW | LB_SL[3] | LB_SL[2] | LB_SL[1] | LB_SL[0] | LB_EL[3] | LB_EL[2] | LB_EL[1] | LB_EL[0] | 01001100 | 4C |
| 222 | DE | ST Noise Readback 1 | R |  |  |  |  | ST_NOISE_VLD | ST_NOISE[10] | ST_NOISE[9] | ST_NOISE[8] |  |  |
| 223 | DF | ST Noise Readback 2 | R | ST_NOISE[7] | ST_NOISE[6] | ST_NOISE[5] | ST_NOISE[4] | ST_NOISE[3] | ST_NOISE[2] | ST_NOISE[1] | ST_NOISE[0] |  |  |
| 225 | E1 | SD offset Cb channel | RW | SD_OFF_Cb[7] | SD_OFF_Cb[6] | SD_OFF_Cb[5] | SD_OFF_Cb[4] | SD_OFF_Cb[3] | SD_OFF_Cb[2] | SD_OFF_Cb[1] | SD_OFF_Cb[0] | 10000000 | 80 |
| 226 | E2 | SD offset Cr channel | RW | SD_OFF_Cr[7] | SD_OFF_Cr[6] | SD_OFF_Cr[5] | SD_OFF_Cr[4] | SD_OFF_Cr[3] | SD_OFF_Cr[2] | SD_OFF_Cr[1] | SD_OFF_Cr[0] | 10000000 | 80 |
| 227 | E3 | SD saturation Cb channel | RW | SD_SAT_Cb[7] | SD_SAT_Cb[6] | SD_SAT_Cb[5] | SD_SAT_Cb[4] | SD_SAT_Cb[3] | SD_SAT_Cb[2] | SD_SAT_Cb[1] | SD_SAT_Cb[0] | 10000000 | 80 |
| 228 | E4 | SD saturation Cr channel | RW | SD_SAT_Cr[7] | SD_SAT_Cr[6] | SD_SAT_Cr[5] | SD_SAT_Cr[4] | SD_SAT_Cr[3] | SD_SAT_Cr[2] | SD_SAT_Cr[1] | SD_SAT_Cr[0] | 10000000 | 80 |
| 229 | E5 | NTSC V bit begin | RW | NVBEGDELO | NVBEGDELE | NVBEGSIGN | NVBEG[4] | NVBEG[3] | NVBEG[2] | NVBEG[1] | NVBEG[0] | 00100101 | 25 |
| 230 | E6 | NTSC V bit end | RW | NVENDDELO | NVENDDELE | NVENDSIGN | NVEND[4] | NVEND[3] | NVEND[2] | NVEND[1] | NVEND[0] | 00000100 | 04 |
| 231 | E7 | NTSC F bit toggle | RW | NFTOGDELO | NFTOGDELE | NFTOGSIGN | NFTOG[4] | NFTOG[3] | NFTOG[2] | NFTOG[1] | NFTOG[0] | 01100011 | 63 |
| 232 | E8 | PAL V bit begin | RW | PVBEGDELO | PVBEGDELE | PVBEGSIGN | PVBEG[4] | PVBEG[3] | PVBEG[2] | PVBEG[1] | PVBEG[0] | 01100101 | 65 |
| 233 | E9 | PALV bit end | RW | PVENDDELO | PVENDDELE | PVENDSIGN | PVEND[4] | PVEND[3] | PVEND[2] | PVEND[1] | PVEND[0] | 00010100 | 14 |
| 234 | EA | PALF bit toggle | RW | PFTOGDELO | PFTOGDELE | PFTOGSIGN | PFTOG[4] | PFTOG[3] | PFTOG[2] | PFTOG[1] | PFTOG[0] | 01100011 | 63 |
| 235 | EB | Vblank Control 1 | RW | NVBIOLCM[1] | NVBIOLCM[0] | NVBIELCM[1] | NVBIELCM[0] | PVBIOLCM[1] | PVBIOLCM[0] | PVBIELCM[1] | PVBIELCM[0] | 01010101 | 55 |
| 236 | EC | Vblank Control 2 | RW | NVBIOCCM[1] | NVBIOCCM[0] | NVBIECCM[1] | NVBIECCM[0] | PVBIOCCM[1] | PVBIOCCM[0] | PVBIECCM[1] | PVBIECCM[0] | 01010101 | 55 |
| 243 | F3 | AFE_CONTROL 1 | RW |  |  |  | AA_FILT_MAN_ OVR | AA_FlLT_EN[3] | AA_FlLT_EN[2] | AA_FlLT_EN[1] | AA_FILT_EN[0] | 00000000 | 00 |
| 244 | F4 | Drive strength | RW | $\begin{aligned} & \text { GLITCH_FILT_ } \\ & \text { BYP } \end{aligned}$ |  | DR_STR[1] | DR_STR[0] | DR_STR_C[1] | DR_STR_C[0] | DR_STR_S[1] | DR_STR_S[0] | 0x010101 | 15 |
| 248 | F8 | IF comp control | RW |  |  |  |  |  | IFFILTSEL[2] | IFFILTSEL[1] | IFFILTSEL[0] | 00000000 | 00 |
| 249 | F9 | VS mode control | RW |  |  |  |  | $\begin{array}{\|l\|} \hline \text { VS_COAST_ } \\ \text { MODE[1] } \\ \hline \end{array}$ | $\begin{aligned} & \text { VS_COAST_ } \\ & \text { MODE[0] } \\ & \hline \end{aligned}$ | EXTEND_VS_ MIN_FREQ | $\begin{aligned} & \text { EXTEND_VS_ } \\ & \text { MAX_FREQ } \\ & \hline \end{aligned}$ | 00000011 | 03 |
| 251 | FB | Peaking gain | RW | PEAKING_ GAIN[7] | $\begin{array}{\|l} \hline \text { PEAKING_- } \\ \text { GAIN[6] } \\ \hline \end{array}$ | PEAKING_ GAIN[5] | PEAKING_ GAIN[4] | PEAKING GAIN[3] | PEAKING_ GAIN[2] | PEAKING_ GAIN[1] | $\begin{array}{\|l} \hline \text { PEAKING_- } \\ \text { GAIN[0] } \\ \hline \end{array}$ | 01000000 | 40 |
| 252 | FC | DNR Noise Threshold 2 | RW | DNR_TH2[7] | DNR_TH2[6] | DNR_TH2[5] | DNR_TH2[4] | DNR_TH2[3] | DNR_TH2[2] | DNR_TH2[1] | DNR_TH2[0] | 00000100 | 04 |
| 254 | FE | CSI Tx slave address | RW | $\begin{array}{\|l} \left\lvert\, \begin{array}{l} \text { CSI_TX_SLAVE_ } \\ \text { ADDR[6] } \end{array}\right. \\ \hline \end{array}$ | $\begin{aligned} & \hline \begin{array}{l} \text { SSI_TX_SLAVE_ } \\ \text { ADDR[5] } \end{array} \\ & \hline \end{aligned}$ | $\begin{aligned} & \begin{array}{l} \text { CSI_TX_SLAVE_ } \\ \text { ADDR[4] } \end{array} \\ & \hline \end{aligned}$ | $\begin{aligned} & \begin{array}{l} \text { CSI_TX_SLAVE_ } \\ \text { ADDR[3] } \end{array} \\ & \hline \end{aligned}$ | $\begin{aligned} & \hline \text { CSI_TX_SLAVE_ } \\ & \text { ADDR[2] } \\ & \hline \end{aligned}$ | $\begin{aligned} & \begin{array}{l} \text { CSI_TX_SLAVE_ } \\ \text { ADDR[1] } \end{array} \\ & \hline \end{aligned}$ | $\begin{aligned} & \text { CSI_TX_SLAVE_ }^{\text {ADD }} \\ & \text { ADD[0] } \\ & \hline \end{aligned}$ |  | 00000000 | 00 |

To access the registers listed in Table 93, SUB_USR_EN in Register Address 0x0E must be programmed to 10 b . All read only registers are left blank.

Table 93. User Sub Map 2 Register Map Details

| Address |  | Register Name | RW | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Reset <br> Value | (Hex) |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Dec | Hex |  |  |  |  |  |  |  |  |  |  |  |  |
| 128 | 80 | ACE Control 1 | RW | ACE_ENABLE |  |  |  |  |  |  |  | 00000000 | 00 |
| 131 | 83 | ACE Control 4 | RW |  |  |  | $\begin{aligned} & \text { ACE_LUMA_- } \\ & \text { GAIN[4] } \end{aligned}$ UAIN[4] | $\begin{aligned} & \hline \text { ACE_LUMA_GAIN } \\ & \text { [3] } \\ & \hline \end{aligned}$ | $\begin{aligned} & \text { ACE_LUMA } \\ & \text { GAIN[2] } \end{aligned}$ | $\begin{aligned} & \text { ACE_LUMA_ } \\ & \text { GAIN[1] } \\ & \hline \end{aligned}$ | $\begin{aligned} & \text { ACE_LUMA_ } \\ & \text { GAIN[0] } \end{aligned}$ | 00001101 | OD |
| 132 | 84 | ACE Control 5 | RW | ACE_CHROMA MAX[3] | $\begin{aligned} & \hline \text { ACE_CHROMA_ }^{\text {MAX[2] }} \end{aligned}$ | ACE_CHROMA MAX[1] | ACE_CHROMA MAX[0] | ACE_CHROMA GAIN[3] | ACE_CHROMA GAIN[2] | ACE_CHROMA GAIN[1] | ACE_CHROMA GAIN[0] | 10001000 | 88 |
| 133 | 85 | ACE Control 6 | RW | $\begin{aligned} & \hline \begin{array}{l} \text { ACE_RESPONSE_ } \\ \text { SPEED[3] } \end{array} \\ & \hline \end{aligned}$ | $\begin{aligned} & \begin{array}{l} \text { ACE_RESPONSE_ } \\ \text { SPEED[2] } \end{array} \\ & \hline \end{aligned}$ | $\begin{aligned} & \hline \text { ACE_RESPONSE_ } \\ & \text { SPEED[2] } \end{aligned}$ | $\begin{aligned} & \begin{array}{l} \text { ACE_RESPONSE } \\ \text { SPEED[1] } \end{array} \\ & \hline \end{aligned}$ | $\begin{aligned} & \hline \text { ACE_GAMMA } \\ & \text { GAIN[3] } \\ & \hline \end{aligned}$ | $\begin{aligned} & \hline \text { ACE_GAMMA } \\ & \text { GAIN[2] } \\ & \hline \end{aligned}$ | $\begin{aligned} & \hline \text { ACE_GAMMA_ } \\ & \text { GAIN[1] } \\ & \hline \end{aligned}$ | $\begin{aligned} & \hline \begin{array}{l} \text { ACE_GAMMA } \\ \text { GAIN[0] } \end{array} \\ & \hline \end{aligned}$ | 11111000 | F8 |
| 146 | 92 | Dither control | RW |  |  |  |  |  |  |  | $\begin{array}{\|l\|} \hline \begin{array}{l} \text { BR_DITHER_MO } \\ \text { DE } \end{array} \\ \hline \end{array}$ | 00000000 | 00 |
| 217 | D9 | Min Max 0 | RW | MIN_THRESH_Y[7] | MIN_THRESH_Y[6] | MIN_THRESH_Y[5] | $\begin{aligned} & \hline \begin{array}{l} \text { MIN_THRESH_ } \\ \text { Y[4] } \end{array} \\ & \hline \end{aligned}$ | MIN_THRESH_Y[3] | MIN_THRESH_Y[2] | $\begin{aligned} & \hline \text { MIN_THRESH_ } \\ & \mathrm{Y}[1] \end{aligned}$ | $\begin{aligned} & \hline \text { MIN_THRESH_ } \\ & \mathrm{Y}[0] \end{aligned}$ | 00000000 | 00 |
| 218 | DA | Min Max 1 | RW | MAX_THRESH_Y[7] | MAX_THRESH_Y[6] | MAX_THRESH_Y[5] | $\begin{aligned} & \text { MAX_THRESH_ } \\ & \text { Y[4] } \end{aligned}$ | MAX_THRESH_Y[3] | MAX_THRESH_Y[2] | $\begin{aligned} & \text { MAX_THRESH_ } \\ & \text { Y[1] } \end{aligned}$ | $\begin{aligned} & \text { MAX_THRESH_ } \\ & \text { Y[0] } \end{aligned}$ | 11111111 | FF |
| 219 | DB | Min Max 2 | RW | MIN_THRESH_C[7] | MIN_THRESH_C[6] | MIN_THRESH_C[5] | $\begin{aligned} & \hline \text { MIN_THRESH_ } \\ & \mathrm{C}[4] \end{aligned}$ | MIN_THRESH_C[3] | MIN_THRESH_C[2] | $\begin{aligned} & \hline \text { MIN_THRESH_ } \\ & \text { C[1] } \end{aligned}$ | $\begin{aligned} & \hline \text { MIN_THRESH_ } \\ & \text { C[0] } \end{aligned}$ | 00000000 | 00 |
| 220 | DC | Min Max 3 | RW | MAX_THRESH_C[7] | MAX_THRESH_C[6] | MAX_THRESH_C[5] | $\begin{aligned} & \text { MAX_THRESH_ } \\ & \text { C[4] } \end{aligned}$ | MAX_THRESH_C[3] | MAX_THRESH_C[2] | $\begin{aligned} & \text { MAX_THRESH_ } \\ & \text { C[1] } \end{aligned}$ | $\begin{aligned} & \text { MAX_THRESH_ } \\ & \text { C[0] } \end{aligned}$ | 11111111 | FF |
| 221 | DD | Min Max 4 | RW | MIN_SAMPLES_ ALLOWED_Y[3] | MIN_SAMPLES ALLOWED_Y[2] | MIN_SAMPLES_ ALLOWED_Y[1] | MIN_SAMPLES ALLOWED_Y[0] | MAX_SAMPLES ALLOWED_Y[3] | MAX_SAMPLES ALLOWED_Y[2] | MAX_SAMPLES _ALLOWED_Y[1] | MAX_SAMPLES _ALLOWED_Y[0] | 11001100 | CC |
| 222 | DE | Min Max 5 | RW | MIN SAMPLES ALLOWED_C[3] | MIN SAMPLES ALLOWED_C[2] | MIN SAMPLES ALLOWED_C[1] | MIN SAMPLES ALLOWED_C[0] | MAX SAMPLES ALLOWED_C[3] | MAX SAMPLES ALLOWED_C[2] | MAX SAMPLES ALLOWED_C[1] | MAX SAMPLES ALLOWED_C[0] | 11001100 | CC |
| 224 | E0 | FL control | RW |  |  |  |  |  |  |  | FL_ENABLE | 00000000 | 00 |
| 225 | E1 | Y Average 0 | RW | LINE_START[8] | LINE_START[7] | LINE_START[6] | LINE_START[5] | LINE_START[4] | LINE_START[3] | LINE_START[2] | LINE_START[1] | 0001001 | 11 |
| 226 | E2 | Y Average 1 | RW | LINE_END[8] | LINE_END[7] | LINE_END[6] | LINE_END[5] | LINE_END[4] | LINE_END[3] | LINE_END[2] | LINE_END[1] | 10001000 | 88 |
| 227 | E3 | Y Average 2 | RW | SAMPLE_START[9] | SAMPLE_START[8] | SAMPLE_START[7] | SAMPLE START[6] | SAMPLE_START[5] | SAMPLE START[4] | SAMPLE START[3] | SAMPLE START[2] | 00010111 | 1B |
| 228 | E4 | Y Average 3 | RW | SAMPLE_END[9] | SAMPLE_END[8] | SAMPLE_END[7] | SAMPLE_END[6] | SAMPLE_END[5] | SAMPLE_END[4] | SAMPLE_END[3] | SAMPLE_END[2] | 11010111 | D7 |
| 229 | E5 | Y Average 4 | RW | SAMPLE_END[1] | SAMPLE_END[0] | SAMPLE_START[1] | SAMPLE <br> START[0] |  |  | LINE_END[0] | LINE_START[0] | 00100011 | 23 |
| 230 | E6 | Y Average 5 | RW |  |  |  | Y_AVG_TIME_ CONST[2] | Y_AVG_TIME_ CONST[1] | Y_AVG_TIME_ CONST[0] | Y_AVG_FILT_EN | CAPTURE VALUE | 00010000 | 10 |
| 231 | E7 | Y average data MSB | R | Y_AVERAGE[9] | Y_AVERAGE[8] | Y_AVERAGE[7] | Y_AVERAGE[6] | Y_AVERAGE[5] | Y_AVERAGE[4] | Y_AVERAGE[3] | Y_AVERAGE[2] |  |  |
| 232 | E8 | Y average data LSB | R |  |  |  |  |  |  | Y_AVERAGE[1] | Y_AVERAGE[0] |  |  |

To access the registers listed in Table 94, SUB_USR_EN in Register Address 0x0E must be programmed to 01 b . All read only registers are left blank.

Table 94. Interrupt/VDP Map Details ${ }^{1}$

| Address |  | Register Name | RW | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Reset Value | (Hex) |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Dec | Hex |  |  |  |  |  |  |  |  |  |  |  |  |
| 64 | 40 | Interrupt Configuration 1 | RW | $\begin{aligned} & \hline \begin{array}{l} \text { INTRQ_DUR_ } \\ \text { SEL[1] } \end{array} \\ & \hline \end{aligned}$ | $\begin{aligned} & \hline \begin{array}{l} \text { INTRQ_DUR_ } \\ \text { SEL[0] } \end{array} \\ & \hline \end{aligned}$ | $\begin{array}{\|l\|} \hline \begin{array}{l} \text { MV_INTRQ_ } \\ \text { SEL[1] } \end{array} \\ \hline \end{array}$ | $\begin{array}{\|l\|l} \hline \text { MV_INTRQ } \\ \text { SEL[0] } \end{array}$ |  | $\begin{aligned} & \text { MPU_STIM_ } \\ & \text { INTRQ } \end{aligned}$ | $\begin{aligned} & \begin{array}{l} \text { INTRQ_OP_ } \\ \text { SEL[1] } \end{array} \\ & \hline \end{aligned}$ | INTRQ_OP_SEL[0] | 0001×000 | 10 |
| 66 | 42 | Interrupt Status 1 | R |  | MV_PS_CS_Q | SD_FR_CHNG_Q |  |  |  | SD_UNLOCK_Q | SD_LOCK_Q |  |  |
| 67 | 43 | Interrupt Clear 1 | W |  | MV_PS_CS_CLR | $\begin{aligned} & \text { SD_FR_CHNG_ } \\ & \text { CLR } \end{aligned}$ |  |  |  | $\begin{aligned} & \text { SD_UNLOCK_ } \\ & \text { CLR } \end{aligned}$ | SD_LOCK_CLR | $\times 0000000$ | 00 |
| 68 | 44 | Interrupt Mask 1 | RW |  | $\begin{aligned} & \text { MV_PS_CS_ } \\ & \text { MSKB }^{2} \end{aligned}$ | $\begin{aligned} & \hline \text { SD_FR_CHNG_ } \\ & \text { MSKB }^{2} \end{aligned}$ |  |  |  | $\begin{aligned} & \hline \text { SD_UNLOCK_ } \\ & \text { MSKB }^{2} \end{aligned}$ | SD_LOCK_MSKB ${ }^{2}$ | $\times 0000000$ | 00 |
| 69 | 45 | Raw Status 2 | R | $\begin{aligned} & \begin{array}{l} \text { MPU_STIM_ } \\ \text { INTRQ } \end{array} \\ & \hline \end{aligned}$ |  | $\begin{aligned} & \text { CHX_MIN_MAX_ } \\ & \text { INTRQ } \\ & \hline \end{aligned}$ | EVEN_FIELD |  |  |  | CCAPD |  |  |
| 70 | 46 | Interrupt Status 2 | R | MPU_STIM_ INTRQ_Q |  |  | $\begin{aligned} & \hline \text { SD_FIELD_ } \\ & \text { CHNGD_Q } \\ & \hline \end{aligned}$ |  |  |  | CCAPD_Q |  |  |
| 71 | 47 | Interrupt Clear 2 | w | MPU_STIM_ INTRQ_CLR |  | CHX_MIN_MAX INTRQ_CLR | $\begin{aligned} & \hline \text { SD_FIELD_- } \\ & \text { CHNGD_CLR } \end{aligned}$ |  |  |  | CCAPD_CLR | 0xx00000 | 00 |
| 72 | 48 | Interrupt Mask 2 | RW | MPU STIM INTRQ_MSKB ${ }^{2}$ |  | CHX MIN MAX INTRQ_MSKB ${ }^{2}$ | SD_FIELD_- CHNGD_MSKB ${ }^{2}$ |  |  |  | CCAPD_MSKB ${ }^{2}$ | 0xx00000 | 00 |
| 73 | 49 | Raw Status 3 | R |  |  |  | SCM_LOCK |  | SD_H_LOCK | SD_V_LOCK | SD_OP_50Hz |  |  |
| 74 | 4A | Interrupt Status 3 | R |  |  | $\begin{aligned} & \hline \text { PAL_SW_LK_ } \\ & \text { CHNG_Q } \end{aligned}$ | $\begin{aligned} & \hline \begin{array}{l} \text { SCM_LOCK_ } \\ \text { CHNG_Q } \end{array} \\ & \hline \end{aligned}$ | SD_AD_CHNG_Q | $\begin{aligned} & \text { SD_H_LOCK } \\ & \text { CHNG_Q } \end{aligned}$ | $\begin{aligned} & \begin{array}{l} \text { SD_V_LOCK } \\ \text { CHNG_Q } \end{array} \\ & \hline \end{aligned}$ | SD_OP_CHNG_Q |  |  |
| 75 | 4B | Interrupt Clear 3 | w |  |  | PAL_SW_LK CHNG_CLR | $\begin{array}{\|l\|l\|} \hline \text { SCM_LOCK } \\ \text { CHNG_CLR } \\ \hline \end{array}$ | $\begin{aligned} & \text { SD_AD_CHNG_ } \\ & \text { CLR } \end{aligned}$ | $\begin{aligned} & \text { SD_H_LOCK } \\ & \text { CHNG_CLR } \end{aligned}$ | $\begin{aligned} & \text { SD_V_LOCK } \\ & \text { CHNG_CLR } \end{aligned}$ | $\begin{aligned} & \text { SD_OP_CHNG_ } \\ & \text { CLR } \\ & \hline \end{aligned}$ | xx000000 | 00 |
| 76 | 4C | Interrupt Mask 3 | RW |  |  | $\begin{aligned} & \text { PAL_SW_LK- } \\ & \text { CHNG_MSKB } \end{aligned}$ | $\begin{aligned} & \text { SCM_LOCK_ } \\ & \text { CHNG_MSKB } \end{aligned}$ | $\begin{array}{\|l\|} \hline \begin{array}{l} \text { SD_AD_CHNG_ } \\ \text { MSKB²}^{2} \end{array} \\ \hline \end{array}$ | $\begin{aligned} & \text { SD_H_LOCK } \\ & \text { CHNG_MSKB² } \end{aligned}$ | $\begin{aligned} & \text { SD_V_LOCK } \\ & \text { CHNG_MSKB } \\ & \hline \end{aligned}$ | $\begin{aligned} & \hline \begin{array}{l} \text { SD_OP_CHNG_ } \\ \mathrm{MSKB}^{2} \end{array} \\ & \hline \end{aligned}$ | xx000000 | 00 |
| 78 | 4E | Interrupt Status 4 | R |  |  |  |  |  | VDP_CGMS WSS_CHNGD_Q |  | VDP_CCAPD_Q |  |  |
| 79 | 4F | Interrupt Clear 4 | w |  |  |  |  |  | $\begin{aligned} & \text { VDP_CGMS_ } \\ & \text { WSS_CHNG_ } \\ & \text { CLR } \end{aligned}$ |  | VDP_CCAPD_CLR | 00x0x0x0 | 00 |


| Address |  | Register Name | RW | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Reset Value | (Hex) |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Dec | Hex |  |  |  |  |  |  |  |  |  |  |  |  |
| 80 | 50 | Interrupt Mask 4 | RW |  |  |  |  |  | VDP_CGMS WSS_CHNGD_ MSKB ${ }^{2}$ |  | $\begin{aligned} & \hline \text { VDP_CCAPD_ } \\ & \text { MSKB }^{2} \end{aligned}$ | 00x0x0x0 | 00 |
| 81 | 51 | Interrupt Latch 0 | R |  |  | Y_CHANNEL MIN_VIOLATION | Y_CHANNEL MAX_VIOLATION | CB_CHANNEL MIN_VIOLATION | CB_CHANNEL MAX_VIOLATION | CR_CHANNEL MIN_VIOLATION | CR_CHANNEL MAX_VIOLATION |  |  |
| 96 | 60 | VDP_CONFIG_1 | RW |  |  |  |  | WST_PKT_ DECODE_ DISABLE | VDP_TTXT_ TYPE_MAN_ ENABLE | VDP_TTXT_ TYPE_MAN[1] | VDP_TTXT_ TYPE_MAN[0] | 10001000 | 88 |
| 98 | 62 | $\begin{aligned} & \text { VDP_ADF_CONFIG } \\ & \text { _1 } \end{aligned}$ | RW | ADF_ENABLE | ADF_MODE[1] | ADF_MODE[0] | ADF_DID[4] | ADF_DID[3] | ADF_DID[2] | ADF_DID[1] | ADF_DID[0] | 00010101 | 15 |
| 99 | 63 | $\begin{aligned} & \text { VDP_ADF_CONFIG } \\ & 2 \end{aligned}$ | RW | DUPLICATE_ADF |  | ADF_SDID[5] | ADF_SDID[4] | ADF_SDID[3] | ADF_SDID[2] | ADF_SDID[1] | ADF_SDID[0] | 0x101010 | 2A |
| 100 | 64 | VDP_LINE_00E | RW | MAN_LINE_PGM |  |  |  | $\begin{array}{\|l} \hline \text { VBI_DATA_ } \\ \text { P318[3] } \end{array}$ | $\begin{aligned} & \text { VBI_DATA_ } \\ & \text { P318[2] } \end{aligned}$ | $\begin{aligned} & \text { VBI_DATA_ } \\ & \text { P318[1] } \end{aligned}$ | $\begin{aligned} & \text { VBI_DATA_ } \\ & \text { P318[0] } \end{aligned}$ | 0xxx0000 | 00 |
| 101 | 65 | VDP_LINE_00F | RW | $\begin{aligned} & \text { VBI_DATA_ } \\ & \text { P6_N23[3] } \end{aligned}$ | $\begin{aligned} & \hline \text { VBI_DATA_ } \\ & \text { P6_N23[2] } \end{aligned}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA_ } \\ \text { P6_N23[1] } \end{array}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA } \\ \text { P6_N23[0] } \end{array}$ | VBI_DATA_ <br> P319_N286[3] | $\begin{aligned} & \hline \text { VBI_DATA_- } \\ & \text { P319_N286[2] } \end{aligned}$ | VBI_DATA <br> P319_N286[1] | $\begin{aligned} & \hline \text { VBI_DATA_- } \\ & \text { P319_N286[0] } \end{aligned}$ | 00000000 | 00 |
| 102 | 66 | VDP_LINE_010 | RW | $\begin{aligned} & \text { VBI_DATA_ } \\ & \text { P7_N24[3] } \end{aligned}$ | $\begin{aligned} & \text { VBI_DATA_ } \\ & \text { P7_N24[2] } \end{aligned}$ | $\begin{aligned} & \hline \text { VBI_DATA } \\ & \text { P7_N24[1] } \end{aligned}$ | $\begin{array}{\|l} \text { VBI_DATA } \\ \text { P7_N24[0] } \end{array}$ | $\begin{array}{\|l} \hline \text { VBI_DATA } \\ \text { P320_N287[3] } \end{array}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA_-_ } \\ \text { P320_N287[2] } \end{array}$ | VBI_DATA P320_N287[1] | $\begin{aligned} & \hline \text { VBI_DATA } \\ & \text { P320_N287[0] } \end{aligned}$ | 00000000 | 00 |
| 103 | 67 | VDP_LINE_011 | RW | $\begin{aligned} & \text { VBI_DATA_ } \\ & \text { P8_N25[3] } \end{aligned}$ | $\begin{aligned} & \text { VBI_DATA_ } \\ & \text { P8_N25[2] } \end{aligned}$ | $\begin{array}{\|l} \hline \text { VBI_DATA_ } \\ \text { P8_N25[1] } \end{array}$ | $\begin{array}{\|l} \text { VBI_DATA } \\ \text { P8_N25[0] } \end{array}$ | $\begin{aligned} & \hline \text { VBI_DATA_- } \\ & \text { P321_N288[3] } \end{aligned}$ | VBI_DATA P321_N288[2] | VBI_DATA P321_N288[1] | VBI_DATA_ P321_N288[0] | 00000000 | 00 |
| 104 | 68 | VDP_LINE_012 | RW | $\begin{aligned} & \text { VBI_DATA_ } \\ & \text { P9[3] } \end{aligned}$ | $\begin{aligned} & \text { VBI_DATA_ } \\ & \text { P9[2] } \end{aligned}$ | $\begin{aligned} & \text { VBI_DATA_ } \\ & \text { P9[1] } \end{aligned}$ | $\begin{aligned} & \text { VBI_DATA_ } \\ & \text { P9[0] } \end{aligned}$ | $\begin{aligned} & \text { VBI_DATA_ } \\ & \text { P322[3] } \end{aligned}$ | $\begin{array}{\|l} \begin{array}{l} \text { VBI_DATA_ } \\ \text { P322[2] } \end{array} \\ \hline \end{array}$ | $\begin{aligned} & \text { VBI_DATA_ } \\ & \text { P322[1] } \end{aligned}$ | $\begin{aligned} & \text { VBI_DATA_ } \\ & \text { P322[0] } \end{aligned}$ | 00000000 | 00 |
| 105 | 69 | VDP_LINE_013 | RW | $\begin{aligned} & \hline \text { VBI_DATA_ } \\ & \text { P10[3] } \end{aligned}$ | $\begin{aligned} & \hline \text { VBI_DATA_ } \\ & \text { P10[2] } \end{aligned}$ | $\begin{aligned} & \hline \text { VBI_DATA_ } \\ & \text { P10[1] } \end{aligned}$ | $\begin{aligned} & \text { VBI_DATA_ } \\ & \text { P10[0] } \end{aligned}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA_ } \\ \text { P323[3] } \end{array}$ | $\begin{aligned} & \hline \text { VBI_DATA_ } \\ & \text { P323[2] } \end{aligned}$ | $\begin{aligned} & \hline \text { VBI_DATA_ } \\ & \text { P323[1] } \end{aligned}$ | $\begin{aligned} & \text { VBI_DATA_ } \\ & \text { P323[0] } \end{aligned}$ | 00000000 | 00 |
| 106 | 6A | VDP_LINE_014 | RW | $\begin{aligned} & \hline \text { VBI_DATA_ } \\ & \text { P11[3] } \end{aligned}$ | $\begin{aligned} & \hline \text { VBI_DATA_ } \\ & \text { P11[2] } \end{aligned}$ | $\begin{aligned} & \hline \text { VBI_DATA_ } \\ & \text { P11[1] } \end{aligned}$ | $\begin{aligned} & \hline \text { VBI_DATA_ } \\ & \text { P11[0] } \end{aligned}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA_- } \\ \text { P324_N272[3] } \end{array}$ | $\begin{aligned} & \hline \text { VBI_DATA_- } \\ & \text { P324_N272[2] } \end{aligned}$ | VBI_DATA_ P324_N272[1] | $\begin{aligned} & \hline \text { VBI_DATA_} \\ & \text { P324_N272[0] } \end{aligned}$ | 00000000 | 00 |
| 107 | 6B | VDP_LINE_015 | RW | $\begin{aligned} & \hline \text { VBI_DATA } \\ & \text { P12_N10[3] } \end{aligned}$ | $\begin{array}{\|l} \hline \text { VBI_DATA } \\ \text { P12_N10[2] } \end{array}$ | $\begin{aligned} & \hline \text { VBI_DATA } \\ & \text { P12_N10[1] } \end{aligned}$ | $\begin{aligned} & \hline \text { VBI_DATA } \\ & \text { P12_N10[0] } \end{aligned}$ | VBI_DATA P325_N273[3] | VBI_DATA P325_N273[2] | VBI_DATA P325_N273[1] | $\begin{aligned} & \hline \text { VBI_DATA_ } \\ & \text { P325_N273[0] } \end{aligned}$ | 00000000 | 00 |
| 108 | 6C | VDP_LINE_016 | RW | $\begin{array}{\|l\|} \hline \text { VBI_DATA } \\ \text { P13_N11[3] } \end{array}$ | $\begin{aligned} & \hline \text { VBI_DATA } \\ & \text { P13_N11[2] } \end{aligned}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA } \\ \text { P13_N11[1] } \end{array}$ | $\begin{aligned} & \hline \text { VBI_DATA } \\ & \text { P13_N11[0] } \end{aligned}$ | VBI_DATA P326_N274[3] | $\begin{aligned} & \hline \text { VBI_DATA_- } \\ & \text { P326_N274[2] } \end{aligned}$ | VBI_DATA_ P326_N274[1] | $\begin{array}{\|l\|} \hline \text { VBI_DATA_- } \\ \text { P326_N274[0] } \end{array}$ | 00000000 | 00 |
| 109 | 6D | VDP_LINE_017 | RW | $\begin{aligned} & \text { VBI_DATA } \\ & \text { P14_N12[3] } \end{aligned}$ | $\begin{aligned} & \text { VBI_DATA } \\ & \text { P14_N12[2] } \end{aligned}$ | $\begin{aligned} & \text { VBI_DATA } \\ & \text { P14_N12[1] } \end{aligned}$ | $\begin{aligned} & \text { VBI_DATA } \\ & \text { P14_N12[0] } \end{aligned}$ | VBI_DATA P327_N275[3] | $\begin{aligned} & \text { VBI_DATA_-_ } \\ & \text { P327_N275[2] } \end{aligned}$ | VBI_DATA P327_N275[1] | $\begin{array}{\|l\|} \hline \text { VBI_DATA_- } \\ \text { P327_N275[0] } \end{array}$ | 00000000 | 00 |
| 110 | 6E | VDP_LINE_018 | RW | $\begin{array}{\|l} \text { VBI_DATA } \\ \text { P15_N13[3] } \end{array}$ | $\begin{array}{\|l} \text { VBI_DATA } \\ \text { P15_N13[2] } \end{array}$ | $\begin{array}{\|l} \text { VBI_DATA } \\ \text { P15_N13[1] } \end{array}$ | $\begin{array}{\|l} \text { VBI_DATA } \\ \text { P15_N13[0] } \end{array}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA_- } \\ \text { P328_N276[3] } \end{array}$ | $\begin{aligned} & \hline \text { VBI_DATA_- } \\ & \text { P328_N276[2] } \end{aligned}$ | VBI_DATA P328_N276[1] | $\begin{array}{\|l\|} \hline \text { VBI_DATA_-_ } \\ \text { P328_N276[0] } \end{array}$ | 00000000 | 00 |
| 111 | 6F | VDP_LINE_019 | RW | $\begin{array}{\|l} \text { VBI_DATA } \\ \text { P16_N14[3] } \end{array}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA_ } \\ \text { P16_N14[2] } \\ \hline \end{array}$ | $\begin{array}{\|l} \hline \text { VBI_DATA_ } \\ \text { P16_N14[1] } \end{array}$ | $\begin{array}{\|l} \text { VBI_DATA } \\ \text { P16_N14[0] } \end{array}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA_- } \\ \text { P329_N277[3] } \\ \hline \end{array}$ | $\begin{aligned} & \hline \text { VBI_DATA_- } \\ & \text { P329_N277[2] } \\ & \hline \end{aligned}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA_- } \\ \text { P329_N277[1] } \end{array}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA_- } \\ \text { P329_N277[0] } \\ \hline \end{array}$ | 00000000 | 00 |
| 112 | 70 | VDP_LINE_01A | RW | $\begin{array}{\|l} \text { VBI_DATA } \\ \text { P17_N15[3] } \end{array}$ | $\begin{array}{\|l} \text { VBI_DATA_ } \\ \text { P17_N15[2] } \end{array}$ | $\begin{array}{\|l} \text { VBI_DATA } \\ \text { P17_N15[1] } \end{array}$ | $\begin{array}{\|l} \text { VBI_DATA } \\ \text { P17_N15[0] } \end{array}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA_ } \\ \text { P330_N278[3] } \end{array}$ | $\begin{aligned} & \hline \text { VBI_DATA_- } \\ & \text { P330_N278[2] } \end{aligned}$ | VBI_DATA P330_N278[1] | $\begin{array}{\|l\|} \hline \text { VBI_DATA_- } \\ \text { P330_N278[0] } \\ \hline \end{array}$ | 00000000 | 00 |
| 113 | 71 | VDP_LINE_01B | RW | $\begin{array}{\|l\|} \hline \text { VBI_DATA } \\ \text { P18_N16[3] } \end{array}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA } \\ \text { P18_N16[2] } \end{array}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA } \\ \text { P18_N16[1] } \end{array}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA } \\ \text { P18_N16[0] } \end{array}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA } \\ \text { P331_N279[3] } \end{array}$ | $\begin{aligned} & \hline \text { VBI_DATA_- } \\ & \text { P331_N279[2] } \end{aligned}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA } \\ \text { P331_N279[1] } \end{array}$ | VBI_DATA_ P331_N279[0] | 00000000 | 00 |
| 114 | 72 | VDP_LINE_01C | RW | $\begin{aligned} & \hline \text { VBI_DATA_ } \\ & \text { P19_N17[3] } \end{aligned}$ | $\begin{array}{\|l} \hline \text { VBI_DATA } \\ \text { P19_N17[2] } \end{array}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA_ } \\ \text { P19_N17[1] } \end{array}$ | $\begin{aligned} & \text { VBI_DATA } \\ & \text { P19_N17[0] } \end{aligned}$ | VBI_DATA_ P332_N280[3] | $\begin{aligned} & \hline \text { VBI_DATA_- } \\ & \text { P332_N280[2] } \end{aligned}$ | VBI_DATA_ P332_N280[1] | $\begin{array}{\|l\|} \hline \text { VBI_DATA_- } \\ \text { P332_N280[0] } \end{array}$ | 00000000 | 00 |
| 115 | 73 | VDP_LINE_01D | RW | $\begin{array}{\|l\|} \hline \text { VBI_DATA_ } \\ \text { P20_N18[3] } \\ \hline \end{array}$ | $\begin{array}{\|l} \text { VBI_DATA_ } \\ \text { P20_N18[2] } \\ \hline \end{array}$ | $\begin{array}{\|l} \hline \text { VBI_DATA_ } \\ \text { P20_N18[1] } \\ \hline \end{array}$ | $\begin{array}{\|l} \hline \text { VBI_DATA } \\ \text { P20_N18[0] } \\ \hline \end{array}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA_- } \\ \text { P333_N281[3] } \\ \hline \end{array}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA_- } \\ \text { P333_N281[2] } \\ \hline \end{array}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA_- } \\ \text { P333_N281[1] } \\ \hline \end{array}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA_- } \\ \text { P333_N281[0] } \\ \hline \end{array}$ | 00000000 | 00 |
| 116 | 74 | VDP_LINE_01E | RW | $\begin{aligned} & \hline \text { VBI_DATA_ } \\ & \text { P21_N19[3] } \end{aligned}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA_ } \\ \text { P21_N19[2] } \end{array}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA } \\ \text { P21_N19[1] } \end{array}$ | $\begin{aligned} & \hline \text { VBI_DATA } \\ & \text { P21_N19[0] } \end{aligned}$ | VBI_DATA P334_N282[3] | VBI_DATA_ P334_N282[2] | VBI_DATA P334_N282[1] | $\begin{aligned} & \hline \text { VBI_DATA_- } \\ & \text { P334_N282[0] } \end{aligned}$ | 00000000 | 00 |
| 117 | 75 | VDP_LINE_01F | RW | $\begin{aligned} & \text { VBI_DATA_ } \\ & \text { P22_N20[3] } \end{aligned}$ | $\begin{array}{\|l} \hline \text { VBI_DATA_ } \\ \text { P22_N2O[2] } \end{array}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA_ } \\ \text { P22_N2O[1] } \end{array}$ | $\begin{aligned} & \text { VBI_DATA_ } \\ & \text { P22_N2O[0] } \end{aligned}$ | VBI_DATA_- P335_N283[3] | VBI_DATA P335_N283[2] | VBI_DATA_ <br> P335_N283[1] | VBI_DATA P335_N283[0] | 00000000 | 00 |
| 118 | 76 | VDP_LINE_020 | RW | $\begin{aligned} & \hline \text { VBI_DATA } \\ & \text { P23_N21[3] } \end{aligned}$ | $\begin{aligned} & \hline \text { VBI_DATA } \\ & \text { P23_N21[2] } \end{aligned}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA } \\ \text { P23_N21[1] } \end{array}$ | $\begin{aligned} & \hline \text { VBI_DATA } \\ & \text { P23_N21[0] } \end{aligned}$ | VBI_DATA P336_N284[3] | VBI_DATA P336_N284[2] | VBI_DATA P336_N284[1] | VBI_DATA P336_N284[0] | 00000000 | 00 |
| 119 | 77 | VDP_LINE_021 | RW | $\begin{aligned} & \text { VBI_DATA } \\ & \text { P24_N22[3] } \end{aligned}$ | $\begin{array}{\|l} \text { VBI_DATA } \\ \text { P24_N22[2] } \end{array}$ | $\begin{array}{\|l} \text { VBI_DATA } \\ \text { P24_N22[1] } \end{array}$ | $\begin{array}{\|l} \text { VBI_DATA } \\ \text { P24_N22[0] } \end{array}$ | VBI_DATA P337_N285[3] | $\begin{aligned} & \text { VBI_DATA_- } \\ & \text { P337_N285[2] } \end{aligned}$ | $\begin{array}{\|l\|} \hline \text { VBI_DATA- } \\ \text { P337_N285[1] } \end{array}$ | $\begin{aligned} & \text { VBI_DATA_- } \\ & \text { P337_N285[0] } \end{aligned}$ | 00000000 | 00 |
| 120 | 78 | VDP_STATUS | R | TTXT_AVL |  |  |  |  | CGMS_WSS_AVL | CC_EVEN_FIELD | CC_AVL |  |  |
| 120 | 78 | VDP_STATUS_ CLEAR | W |  |  |  |  |  | $\begin{array}{\|l} \text { CGMS_WSS_ } \\ \text { CLEAR } \end{array}$ |  | CC_CLEAR | 00000000 | 00 |
| 121 | 79 | $\begin{aligned} & \text { VDP_CCAP_ } \\ & \text { DATA_0 } \end{aligned}$ | R | CCAP_BYTE_1[7] | CCAP_BYTE_1[6] | CCAP_BYTE_1[5] | CCAP_BYTE_1[4] | CCAP_BYTE_1[3] | CCAP_BYTE_1[2] | CCAP_BYTE_1[1] | CCAP_BYTE_1[0] |  |  |
| 122 | 7A | $\begin{aligned} & \hline \text { VDP_CCAP_ } \\ & \text { DATA_1 } \end{aligned}$ | R | CCAP_BYTE_2[7] | CCAP_BYTE_2[6] | CCAP_BYTE_2[5] | CCAP_BYTE_2[4] | CCAP_BYTE_2[3] | CCAP_BYTE_2[2] | CCAP_BYTE_2[1] | CCAP_BYTE_2[0] |  |  |
| 125 | 7D | $\begin{aligned} & \hline \text { VDP_CGMS_- } \\ & \text { WSS_DATA_0 } \end{aligned}$ | R |  |  |  |  | CGMS_CRC[5] | CGMS_CRC[4] | CGMS_CRC[3] | CGMS_CRC[2] |  |  |
| 126 | 7E | VDP_CGMS WSS_DATA_1 | R | CGMS_CRC[1] | CGMS_CRC[0] | CGMS_WSS[13] | CGMS_WSS[12] | CGMS_WSS[11] | CGMS_WSS[10] | CGMS_WSS[9] | CGMS_WSS[8] |  |  |
| 127 | 7F | VDP_CGMS WSS_DATA_2 | R | CGMS_WSS[7] | CGMS_WSS[6] | CGMS_WSS[5] | CGMS_WSS[4] | CGMS_WSS[3] | CGMS_WSS[2] | CGMS_WSS[1] | CGMS_WSS[0] |  |  |
| 156 | 9C | VDP_OUTPUT_SEL | RW |  |  |  | WSS_CGMS_CB_ CHANGE |  |  |  |  | 00110000 | 30 |

[^2]
## ADV7182

To access all the registers listed in Table 95, SUB_USR_EN in Register Address 0x0E must be programmed to 00b. The gray shading is the default.

Table 95. Register Map Descriptions (Main Register Map) ${ }^{1}$

| Main Map |  |  | Bits <br> (Shading Indicates Default State) |  |  |  |  |  |  |  | Comments | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Subaddress | Register | Bit Description | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |  |  |
| 0x00 | Input control | INSEL[4:0]; the INSEL bits allow the user to select an input channel and the input format |  |  |  | 0 | 0 | 0 | 0 | 0 | CVBS input on $\mathrm{A}_{\text {IN }} 1$ |  |
|  |  |  |  |  |  | 0 | 0 | 0 | 0 | 1 | CVBS input on $\mathrm{A}_{\text {IN } 2}$ |  |
|  |  |  |  |  |  | 0 | 0 | 0 | 1 | 0 | CVBS input on $\mathrm{A}_{\text {IN }} 3$ |  |
|  |  |  |  |  |  | 0 | 0 | 0 | 1 | 1 | CVBS input on $\mathrm{A}_{\text {IN }} 4$ |  |
|  |  |  |  |  |  | 0 | 1 | 0 | 0 | 0 | Y input on $\mathrm{A}_{\mathbb{1} 1} 1, \mathrm{C}$ input on $\mathrm{A}_{\text {IN }} 2$ |  |
|  |  |  |  |  |  | 0 | 1 | 0 | 0 | 1 | $Y$ input on $A_{\text {IN }} 3, C$ input on $A_{\text {IN }} 4$ |  |
|  |  |  |  |  |  | 0 | 1 | 1 | 0 | 0 | Y input on $\mathrm{A}_{1} 1$, Pb input on $\mathrm{A}_{\mathrm{in} 2}$, Pr input on $\mathrm{A}_{\mathrm{I}} 3$ |  |
|  |  |  |  |  |  | 0 | 1 | 1 | 1 | 0 | Differential positive on $\mathrm{A}_{\mathrm{IN}} 1$, differential negative on $\mathrm{A}_{\mathrm{I}} 2$ |  |
|  |  |  |  |  |  | 0 | 1 | 1 | 1 | 1 | Differential positive on $\mathrm{A}_{\mathrm{N}} 3$, differential negative on $\mathrm{A}_{\mathrm{IN}} 4$ |  |
| 0x01 | Video Selection 1 | Reserved |  |  |  |  |  | 0 | 0 | 0 | Sets to default |  |
|  |  | ENVSPROC |  |  |  |  | 0 |  |  |  | Disables VSYNC processor |  |
|  |  |  |  |  |  |  | 1 |  |  |  | Enables VSYNC processor |  |
|  |  | Reserved |  |  |  | 0 |  |  |  |  | Sets to default |  |
|  |  | BETACAM; enables BETACAM levels |  |  | 0 |  |  |  |  |  | Standard video input |  |
|  |  |  |  |  | 1 |  |  |  |  |  | Betacam input enable |  |
|  |  | ENHSPLL |  | 0 |  |  |  |  |  |  | Disables HSYNC processor |  |
|  |  |  |  | 1 |  |  |  |  |  |  | Enables HSYNC processor |  |
|  |  | Reserved | 1 |  |  |  |  |  |  |  | Sets to default |  |
| 0x02 | Video Selection 2 | Reserved |  |  |  |  | 0 | 1 | 0 | 0 | Set to default |  |
|  |  | VID_SEL[3:0]; the VID_SEL bits allow the user to select the input video standard | 0 | 0 | 0 | 0 |  |  |  |  | Autodetects PAL B/PAL G/PAL H/ PAL I/PAL D, NTSC J (no pedestal), SECAM |  |
|  |  |  | 0 | 0 | 0 | 1 |  |  |  |  | Autodetects PAL B/PAL G/PAL H/ PAL I/PAL D, NTSC M (pedestal), SECAM |  |
|  |  |  | 0 | 0 | 1 | 0 |  |  |  |  | Autodetects PAL N (pedestal), NTSC J (no pedestal), SECAM |  |
|  |  |  | 0 | 0 | 1 | 1 |  |  |  |  | Autodetects PAL N (pedestal), NTSC M (pedestal) SECAM |  |
|  |  |  | 0 | 1 | 0 | 0 |  |  |  |  | NTSC J |  |
|  |  |  | 0 | 1 | 0 | 1 |  |  |  |  | NTSC M |  |
|  |  |  | 0 | 1 | 1 | 0 |  |  |  |  | PAL 60 |  |
|  |  |  | 0 | 1 | 1 | 1 |  |  |  |  | NTSC 4.43 |  |
|  |  |  | 1 | 0 | 0 | 0 |  |  |  |  | PAL B/G/H/I/D |  |
|  |  |  | 1 | 0 | 0 | 1 |  |  |  |  | PAL N = PAL B/PAL G/PAL H/ PAL I/PAL D (with pedestal) |  |
|  |  |  | 1 | 0 | 1 | 0 |  |  |  |  | PAL M (without pedestal) |  |
|  |  |  | 1 | 0 | 1 | 1 |  |  |  |  | PAL M |  |
|  |  |  | 1 | 1 | 0 | 0 |  |  |  |  | PAL Combination N |  |
|  |  |  | 1 | 1 | 0 | 1 |  |  |  |  | PAL Combination N (with pedestal) |  |
|  |  |  | 1 | 1 | 1 | 0 |  |  |  |  | SECAM |  |
|  |  |  | 1 | 1 | 1 | 1 |  |  |  |  | SECAM |  |
| 0x03 | Output control | Reserved |  |  | 0 | 0 | 1 | 1 | 0 | 0 | Reserved |  |
|  |  | TOD; tristate output drivers; this bit allows the user to tristate the output drivers; pixel outputs, HS and VS/FIELD/SFL |  | 0 |  |  |  |  |  |  | Output driverss enabled | See also TIM_OE and TRI_LLC |
|  |  |  |  | 1 |  |  |  |  |  |  | Output drivers tristated |  |
|  |  | VBI_EN; vertical blanking interval data enable; allows VBI data (Line 1 to Line 21) to be passed through with only a minimum amount of filtering performed | 0 |  |  |  |  |  |  |  | All lines filtered and scaled |  |
|  |  |  | 1 |  |  |  |  |  |  |  | Only active video region filtered |  |


| Main Map |  |  | Bits(Shading Indicates Default State) |  |  |  |  |  |  |  | Comments | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Subaddress | Register | Bit Description | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |  |  |
| 0x04 | Extended output control | Range; allows the user to select the range of output values; can be ITU-R BT. 656 compliant or can fill the whole accessible number range |  |  |  |  |  |  |  | 0 | $16 \leq \mathrm{Y} \leq 235,16 \leq \mathrm{C} / \mathrm{P} \leq 240$ | ITU-R BT. 656 |
|  |  |  |  |  |  |  |  |  |  | 1 | $1 \leq \mathrm{Y} \leq 254,1 \leq \mathrm{C} / \mathrm{P} \leq 254$ | Extended range |
|  |  | EN_SFL_PIN |  |  |  |  |  |  | 0 |  | Disables SFL output | SFL output enables encoder and decoder to be connected directly |
|  |  |  |  |  |  |  |  |  | 1 |  | Outputs SFL information on the SFL pin |  |
|  |  | BL_C_VBI; blank chroma during VBI; if set, it enables data in the VBI region to be passed through the decoder undistorted |  |  |  |  |  | 0 |  |  | Decode and output color during VBI | during VBI |
|  |  |  |  |  |  |  |  | 1 |  |  | Blank Cr and Cb values during VBI |  |
|  |  | TIM_OE; enables timing signals output |  |  |  |  | 0 |  |  |  | HS, VS, FIELD tristated | Controlled by TOD |
|  |  |  |  |  |  |  | 1 |  |  |  | HS, VS, FIELD forced active |  |
|  |  | Reserved |  | 0 | 1 | 1 |  |  |  |  |  |  |
|  |  | BT.656-4; allows the user to select an output mode compatible with ITU-R BT.656-3/-4 | 0 |  |  |  |  |  |  |  | ITU-R BT.656-3 compatible |  |
|  |  |  | 1 |  |  |  |  |  |  |  | ITU-R BT.656-4 compatible |  |
| 0x07 | Autodetect enable | AD_PAL_EN;PAL B/PAL D/PALI/ PAL G/PAL H autodetect enable |  |  |  |  |  |  |  | 0 | Disables |  |
|  |  |  |  |  |  |  |  |  |  | 1 | Enables |  |
|  |  | AD_NTSC_EN; NTSC autodetect enable |  |  |  |  |  |  | 0 |  | Disables |  |
|  |  |  |  |  |  |  |  |  | 1 |  | Enables |  |
|  |  | AD_PALM_EN; PAL M autodetect enable |  |  |  |  |  | 0 |  |  | Disables |  |
|  |  |  |  |  |  |  |  | 1 |  |  | Enables |  |
|  |  | AD_PALN_EN; PAL N autodetect enable |  |  |  |  | 0 |  |  |  | Disables |  |
|  |  |  |  |  |  |  | 1 |  |  |  | Enables |  |
|  |  | AD_P60_EN; PAL 60 autodetect enable |  |  |  | 0 |  |  |  |  | Disables |  |
|  |  |  |  |  |  | 1 |  |  |  |  | Enables |  |
|  |  | AD_N443_EN; NTSC 4.43 autodetect enable |  |  | 0 |  |  |  |  |  | Disables |  |
|  |  |  |  |  | 1 |  |  |  |  |  | Enables |  |
|  |  | AD_SECAM_EN; SECAM autodetect enable |  | 0 |  |  |  |  |  |  | Disables |  |
|  |  |  |  | 1 |  |  |  |  |  |  | Enables |  |
|  |  | AD_SEC525_EN; SECAM 525 autodetect enable | 0 |  |  |  |  |  |  |  | Disables |  |
|  |  |  | 1 |  |  |  |  |  |  |  | Enables |  |
| 0x08 | Contrast | CON[7:0]; contrast adjust; this is the user control for contrast adjustment | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | Luma gain = 1 | $\begin{aligned} & 0 \times 00 \text { gain }=0, \\ & 0 \times 80 \text { gain }=1, \\ & 0 \times F F \text { gain }=2 \end{aligned}$ |
| 0x0A | Brightness Adjust | BRI[7:0]; this register controls the brightness of the video signal | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |  | $\begin{aligned} & 0 \times 00=0 \text { IRE, } \\ & 0 \times 7 \mathrm{~F}=+30 \text { IRE, } \\ & 0 \times 80=-30 \text { IRE } \end{aligned}$ |
| 0x0B | Hue Adjust | HUE[7:0]; this register contains the value for the color hue adjustment | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |  | $\begin{aligned} & \text { Hue range }=-90^{\circ} \text { to } \\ & +90^{\circ} \end{aligned}$ |
| 0x0C | Default Value Y | DEF_VAL_EN; default value enable |  |  |  |  |  |  |  | 0 | Free-run mode dependent on DEF_VAL_AUTO_EN |  |
|  |  |  |  |  |  |  |  |  |  | 1 | Forces free-run mode on |  |
|  |  | DEF_VAL_AUTO_EN; default value automatic enable |  |  |  |  |  |  | 0 |  | Disables free-run mode | When lock is lost, freerun mode can be enabled to output stable timing, clock, and a set color |
|  |  |  |  |  |  |  |  |  | 1 |  | Enables automatic free-run mode |  |
|  |  | DEF_Y[5:0]; default value is $Y$; this register holds the Y default value | 0 | 0 | 1 | 1 | 0 | 1 |  |  | $\mathrm{Y}[7: 0]=\left\{D E F \_Y[5: 0], 0,0\right\}$ | Default $Y$ value output in free-run mode |
| 0x0D | Default Value C | DEF_C[7:0]; default value is $C$; the Cr and Cb default values are defined in this register | 0 | 1 | 1 | 1 | 1 | 1 | 0 | 0 | $\begin{aligned} & \mathrm{Cr}[3: 0]=\{\mathrm{DEF} \text { C[ }[7: 4]\}, \\ & \mathrm{Cb}[3: 0]=\{\mathrm{DEF} \text { _C[3:0] } \end{aligned}$ | Default $\mathrm{Cb} / \mathrm{Cr}$ value output in free-run mode; default values give blue screen output |


| Main Map |  |  | Bits(Shading Indicates Default State) |  |  |  |  |  |  |  | Comments | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Subaddress | Register | Bit Description | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |  |  |
| 0x0E | ADI Control 1 | Reserved |  |  |  | 0 | 0 | 0 | 0 | 0 | Sets as default |  |
|  |  | SUB_USR_EN[1:0]; enables user to access the interrupt/VDP map and User Sub Map 2 |  | 0 | 0 |  |  |  |  |  | Accesses main register space |  |
|  |  |  |  | 0 | 1 |  |  |  |  |  | Accesses interrupt/VDP register space |  |
|  |  |  |  | 1 | 0 |  |  |  |  |  | Accesses User Sub Map 2 |  |
|  |  | Reserved | 0 |  |  |  |  |  |  |  | Sets as default | See Figure 48 |
| 0x0F | Power management | Reserved |  |  |  |  |  |  | 0 | 0 | Sets to default |  |
|  |  | Reserved |  |  |  | 0 | 0 |  |  |  | Sets to default |  |
|  |  | PWRDWN; power-down places the decoder into a full powerdown mode |  |  | 0 |  |  |  |  |  | System functional |  |
|  |  |  |  |  | 1 |  |  |  |  |  | Powered down |  |
|  |  | Reserved |  | 0 |  |  |  |  |  |  | Sets to default |  |
|  |  | Reset; chip reset, loads all ${ }^{1} \mathrm{C}$ bits with default values | 0 |  |  |  |  |  |  |  | Normal operation |  |
|  |  |  | 1 |  |  |  |  |  |  |  | Starts reset sequence | Executing reset takes approximately 2 ms ; this bit is self-clearing |
| 0x10 | Status 1 (read only) | IN_LOCK |  |  |  |  |  |  |  | x | 1 = in lock (now) | Provides info about the internal status of the decoder |
|  |  | LOST_LOCK |  |  |  |  |  |  | x |  | 1 = lost lock (since last read) |  |
|  |  | FSC_LOCK |  |  |  |  |  | x |  |  | 1 = $\mathrm{fsc}_{\text {sc }}$ lock (now) |  |
|  |  | FOLLOW_PW |  |  |  |  | x |  |  |  | 1 = peak white AGC mode active |  |
|  |  | AD_RESULT[2:0]; autodetection result reports the standard of the input video |  | 0 | 0 | 0 |  |  |  |  | NTSC M/NTSC J | Detected standard |
|  |  |  |  | 0 | 0 | 1 |  |  |  |  | NTSC 4.43 |  |
|  |  |  |  | 0 | 1 | 0 |  |  |  |  | PALM |  |
|  |  |  |  | 0 | 1 | 1 |  |  |  |  | PAL 60 |  |
|  |  |  |  | 1 | 0 | 0 |  |  |  |  | PAL B/PAL G/PAL H/PAL I/PAL D |  |
|  |  |  |  | 1 | 0 | 1 |  |  |  |  | SECAM |  |
|  |  |  |  | 1 | 1 | 0 |  |  |  |  | PAL Combination N |  |
|  |  |  |  | 1 | 1 | 1 |  |  |  |  | SECAM 525 |  |
|  |  | COL_KILL | x |  |  |  |  |  |  |  | 1 = color kill is active | Color kill |
| 0x11 | IDENT (read only) | IDENT[7:0]; provides ID on the revision of the part | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 1 |  | Power-up value = $0 \times 41$ |
| 0×12 | Status 2 (read only) | MVCS DET |  |  |  |  |  |  |  | x | MV color striping detected | 1 = detected |
|  |  | MVCS T3 |  |  |  |  |  |  | x |  | MV color striping type | $0=$ Type 2, 1 = Type 3 |
|  |  | MV PS DET |  |  |  |  |  | x |  |  | MV pseudosync detected | 1 = detected |
|  |  | MV AGC DET |  |  |  |  | x |  |  |  | MV AGC pulses detected | 1 = detected |
|  |  | LL NSTD |  |  |  | x |  |  |  |  | Nonstandard line length | 1 = detected |
|  |  | FSC NSTD |  |  | x |  |  |  |  |  | $\mathrm{f}_{\mathrm{sc}}$ frequency nonstandard | 1 = detected |
|  |  | Reserved | x | x |  |  |  |  |  |  |  |  |
| 0x13 | Status 3 (read only) | INST_HLOCK |  |  |  |  |  |  |  | X | 1 = horizontal lock achieved | Unfiltered |
|  |  | Reserved |  |  |  |  |  |  | x |  | Reserved |  |
|  |  | SD_OP_50Hz |  |  |  |  |  | 0 |  |  | SD 60 Hz detected | SD field rate detect |
|  |  |  |  |  |  |  |  | 1 |  |  | SD 50 Hz detected |  |
|  |  | Reserved |  |  |  |  | x |  |  |  |  |  |
|  |  | FREE_RUN_ACT |  |  |  | x |  |  |  |  | 1 = free-run mode active |  |
|  |  | STD FLD LEN |  |  | x |  |  |  |  |  | 1 = field length standard | Correct field length found |
|  |  | Interlaced |  | x |  |  |  |  |  |  | 1 = interlaced video detected | Field sequence found |
|  |  | PAL_SW_LOCK | x |  |  |  |  |  |  |  | 1 = swinging burst detected | Reliable swinging burst sequence |


| Main Map |  |  | Bits(Shading Indicates Default State) |  |  |  |  |  |  |  | Comments | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Subaddress | Register | Bit Description | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |  |  |
| 0x14 | Analog clamp control | FREE_RUN_PAT_SEL[2:0] |  |  |  |  |  | 0 | 0 | 0 | Single color set by DEF_C and DEF_Y; see Color Controls section |  |
|  |  |  |  |  |  |  |  | 0 | 0 | 1 | 100\% color bars |  |
|  |  |  |  |  |  |  |  | 0 | 1 | 0 | Luma ramp |  |
|  |  |  |  |  |  |  |  | 1 | 0 | 1 | Boundary box |  |
|  |  | Reserved |  |  |  |  | 0 |  |  |  | Sets to default |  |
|  |  | CCLEN; current clamp enable allows the user to switch off the current sources in the analog front |  |  |  | 0 |  |  |  |  | Current sources switched off |  |
|  |  |  |  |  |  | 1 |  |  |  |  | Current sources enabled |  |
|  |  | Reserved | 0 | 0 | 0 |  |  |  |  |  | Sets to default |  |
| 0x15 | Digital Clamp Control 1 | Reserved <br> DCFE; digital clamp freeze enable |  |  |  |  | x | x | x | x | Sets to default |  |
|  |  |  |  |  |  | 0 |  |  |  |  | Digital clamp on |  |
|  |  |  |  |  |  | 1 |  |  |  |  | Digital clamp off |  |
|  |  | DCT[1:0]; digital clamp timing determines the time constant of the digital fine clamp circuitry |  | 0 | 0 |  |  |  |  |  | Slow (TC = 1 sec ) |  |
|  |  |  |  | 0 | 1 |  |  |  |  |  | Medium ( $\mathrm{TC}=0.5 \mathrm{sec}$ ) |  |
|  |  |  |  | 1 | 0 |  |  |  |  |  | Fast ( $\mathrm{TC}=0.1 \mathrm{sec}$ ) |  |
|  |  |  |  | 1 | 1 |  |  |  |  |  | TC dependent on video |  |
|  |  | Reserved | 0 |  |  |  |  |  |  |  | Set to default |  |
| 0x17 | Shaping Filter Control 1 | YSFM[4:0]; selects Y shaping filter mode in CVBS-only mode; allows the user to select a wide range of low-pass/notch filters; if either auto mode is selected, the decoder selects the optimum Y filter depending on the CVBS video source quality (good vs. poor) |  |  |  | 0 | 0 | 0 | 0 | 0 | Autowide notch for poor quality sources or wideband filter with comb for good quality input | Decoder selects optimum Y shaping filter depending on CVBS quality |
|  |  |  |  |  |  | 0 | 0 | 0 | 0 | 1 | Autonarrow notch for poor quality sources or wideband filter with comb for good quality input |  |
|  |  |  |  |  |  | 0 | 0 | 0 | 1 | 0 | SVHS 1 | If one of these modes is selected, the decoder does not change filter modes; depending on video quality, a fixed filter response (the one selected) is used for good and bad quality video |
|  |  |  |  |  |  | 0 | 0 | 0 | 1 | 1 | SVHS 2 |  |
|  |  |  |  |  |  | 0 | 0 | 1 | 0 | 0 | SVHS 3 |  |
|  |  |  |  |  |  | 0 | 0 | 1 | 0 | 1 | SVHS 4 |  |
|  |  |  |  |  |  | 0 | 0 | 1 | 1 | 0 | SVHS 5 |  |
|  |  |  |  |  |  | 0 | 0 | 1 | 1 | 1 | SVHS 6 |  |
|  |  |  |  |  |  | 0 | 1 | 0 | 0 | 0 | SVHS 7 |  |
|  |  |  |  |  |  | 0 | 1 | 0 | 0 | 1 | SVHS 8 |  |
|  |  |  |  |  |  | 0 | 1 | 0 | 1 | 0 | SVHS 9 |  |
|  |  |  |  |  |  | 0 | 1 | 0 | 1 | 1 | SVHS 10 |  |
|  |  |  |  |  |  | 0 | 1 | 1 | 0 | 0 | SVHS 11 |  |
|  |  |  |  |  |  | 0 | 1 | 1 | 0 | 1 | SVHS 12 |  |
|  |  |  |  |  |  | 0 | 1 | 1 | 1 | 0 | SVHS 13 |  |
|  |  |  |  |  |  | 0 | 1 | 1 | 1 | 1 | SVHS 14 |  |
|  |  |  |  |  |  | 1 | 0 | 0 | 0 | 0 | SVHS 15 |  |
|  |  |  |  |  |  | 1 | 0 | 0 | 0 | 1 | SVHS 16 |  |
|  |  |  |  |  |  | 1 | 0 | 0 | 1 | 0 | SVHS 17 |  |
|  |  |  |  |  |  | 1 | 0 | 0 | 1 | 1 | SVHS 18 (CCIR 601) |  |
|  |  |  |  |  |  | 1 | 0 | 1 | 0 | 0 | PAL NN1 |  |
|  |  |  |  |  |  | 1 | 0 | 1 | 0 | 1 | PAL NN2 |  |
|  |  |  |  |  |  | 1 | 0 | 1 | 1 | 0 | PAL NN3 |  |
|  |  |  |  |  |  | 1 | 0 | 1 | 1 | 1 | PAL WN1 |  |
|  |  |  |  |  |  | 1 | 1 | 0 | 0 | 0 | PAL WN2 |  |
|  |  |  |  |  |  | 1 | 1 | 0 | 0 | 1 | NTSC NN1 |  |
|  |  |  |  |  |  | 1 | 1 | 0 | 1 | 0 | NTSC NN2 |  |
|  |  |  |  |  |  | 1 | 1 | 0 | 1 | 1 | NTSC NN3 |  |
|  |  |  |  |  |  | 1 | 1 | 1 | 0 | 0 | NTSC WN1 |  |
|  |  |  |  |  |  | 1 | 1 | 1 | 0 | 1 | NTSC WN2 |  |
|  |  |  |  |  |  | 1 | 1 | 1 | 1 | 0 | NTSC WN3 |  |
|  |  |  |  |  |  | 1 | 1 | 1 | 1 | 1 | Reserved |  |





| Main Map |  |  | Bits(Shading Indicates Default State) |  |  |  |  |  |  |  | Comments | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Subaddress | Register | Bit Description | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |  |  |
| 0x37 | Polarity | PCLK; sets polarity of LLC |  |  |  |  |  |  |  | 0 | Inverts polarity |  |
|  |  |  |  |  |  |  |  |  |  | 1 | Normal polarity as per the timing diagrams |  |
|  |  | Reserved |  |  |  |  |  | 0 | 0 |  | Set to 0 |  |
|  |  | PF; sets the FIELD polarity |  |  |  |  | 0 |  |  |  |  |  |
|  |  |  |  |  |  |  | 1 |  |  |  |  |  |
|  |  | Reserved |  |  |  | 0 |  |  |  |  |  |  |
|  |  | PVS; sets the VSYNC polarity |  |  | 0 |  |  |  |  |  | Active high |  |
|  |  |  |  |  | 1 |  |  |  |  |  | Active low |  |
|  |  | Reserved |  | 0 |  |  |  |  |  |  | Sets to 0 |  |
|  |  | PHS; sets HSYNC polarity | 0 |  | 0 |  |  |  |  |  | Active high |  |
|  |  |  | 1 |  | 1 |  |  |  |  |  | Active low |  |
| $0 \times 38$ | NTSC comb control | YCMN[2:0]; luma comb mode, NTSC |  |  |  |  |  | 0 | 0 | 0 | Adaptive three-line, three-tap luma comb |  |
|  |  |  |  |  |  |  |  | 1 | 0 | 0 | Uses low-pass/notch filter |  |
|  |  |  |  |  |  |  |  | 1 | 0 | 1 | Fixed luma comb two-line (two taps) | Top lines of memory |
|  |  |  |  |  |  |  |  | 1 | 1 | 0 | Fixed luma comb three-line (three taps) | All lines of memory |
|  |  |  |  |  |  |  |  | 1 | 1 | 1 | Fixed luma comb two-line (two taps) | Bottom lines of memory |
|  |  | CCMN[2:0]; chroma comb mode, NTSC |  |  | 0 | 0 | 0 |  |  |  | $\begin{aligned} & \text { Three-line adaptive for CTAPSN = 01, } \\ & \text { four-line adaptive for CTAPSN = 10, } \\ & \text { five-line adaptive for CTAPSN = 11 } \end{aligned}$ |  |
|  |  |  |  |  | 1 | 0 | 0 |  |  |  | Disables chroma comb |  |
|  |  |  |  |  | 1 | 0 | 1 |  |  |  | $\begin{aligned} & \text { Fixed two-line for CTAPSN = 01, } \\ & \text { fixed three-line for CTAPSN = 10, } \\ & \text { fixed four-line for CTAPSN = 11 } \end{aligned}$ | Top lines of memory |
|  |  |  |  |  | 1 | 1 | 0 |  |  |  | $\begin{aligned} & \text { Fixed three-line for CTAPSN = 01, } \\ & \text { fixed four-line for CTAPSN = 10, } \\ & \text { fixed five-line for CTAPSN = 11 } \end{aligned}$ | All lines of memory |
|  |  |  |  |  | 1 | 1 | 1 |  |  |  | $\begin{aligned} & \text { Fixed two-line for CTAPSN = 01, } \\ & \text { fixed three-line for CTAPSN = 10, } \\ & \text { fixed four-line for CTAPSN = 11 } \end{aligned}$ | Bottom lines of memory |
|  |  | CTAPSN[1:0]; chroma comb taps, NTSC | 0 | 0 |  |  |  |  |  |  | Not used |  |
|  |  |  | 0 | 1 |  |  |  |  |  |  | Adapts three lines to two lines |  |
|  |  |  | 1 | 0 |  |  |  |  |  |  | Adapts five lines to three lines |  |
|  |  |  | 1 | 1 |  |  |  |  |  |  | Adapts five lines to four lines |  |
| 0x39 | PAL comb control | YCMP[2:0]; luma comb mode, PAL |  |  |  |  |  | 0 | 0 | 0 | Adaptive five-line, three-tap luma comb |  |
|  |  |  |  |  |  |  |  | 1 | 0 | 0 | Use low-pass notch filter |  |
|  |  |  |  |  |  |  |  | 1 | 0 | 1 | Fixed luma comb (three-line) | Top lines of memory |
|  |  |  |  |  |  |  |  | 1 | 1 | 0 | Fixed luma comb (five-line) | All lines of memory |
|  |  |  |  |  |  |  |  | 1 | 1 | 1 | Fixed luma comb (three-line) | Bottom lines of memory |
|  |  | CCMP[2:0]; chroma comb mode, PAL |  |  | 0 | 0 | 0 |  |  |  | $\begin{aligned} & \text { Three-line adaptive for CTAPSN = 01, } \\ & \text { four-line adaptive for CTAPSN = 10, } \\ & \text { five-line adaptive for CTAPSN = 11 } \end{aligned}$ |  |
|  |  |  |  |  | 1 | 0 | 0 |  |  |  | Disable chroma comb |  |
|  |  |  |  |  | 1 | 0 | 1 |  |  |  | ```Fixed two-line for CTAPSN = 01, fixed three-line for CTAPSN = 10, fixed four-line for CTAPSN = 11``` | Top lines of memory |
|  |  |  |  |  | 1 | 1 | 0 |  |  |  | ```Fixed three-line for CTAPSN = 01, fixed four-line for CTAPSN = 10, fixed five-line for CTAPSN = 11``` | All lines of memory |
|  |  |  |  |  | 1 | 1 | 1 |  |  |  | $\begin{aligned} & \text { Fixed two-line for CTAPSN = 01, } \\ & \text { fixed three-line for CTAPSN = 10, } \\ & \text { fixed four-line for CTAPSN = } 11 \end{aligned}$ | Bottom lines of memory |


| Main Map |  |  | Bits(Shading Indicates Default State) |  |  |  |  |  |  |  | Comments | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Subaddress | Register | Bit Description | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |  |  |
|  |  | CTAPSP[1:0]; chroma comb taps, PAL | 0 | 0 |  |  |  |  |  |  | Not used |  |
|  |  |  | 0 | 1 |  |  |  |  |  |  | Adapts five lines to three lines (two taps) |  |
|  |  |  | 1 | 0 |  |  |  |  |  |  | Adapts five lines to three lines (three taps) |  |
|  |  |  | 1 | 1 |  |  |  |  |  |  | Adapts five lines to four lines (four taps) |  |
| 0x3A | ADC control | MUX PDN override; mux powerdown override |  |  |  |  |  |  |  | 0 |  | No control over power-down for muxes and associated channel circuit |
|  |  |  |  |  |  |  |  |  |  | 1 |  | Allows power-down of MUXO/MUX1/ MUX2 and associated channel circuit; when INSEL[4:0] is used, unused channels are automatically powered down |
|  |  | PWRDWN_MUX_2; enables power-down of MUX2 and associated channel clamp and buffer |  |  |  |  |  |  | 0 |  | MUX2 and associated channel in normal operation |  |
|  |  |  |  |  |  |  |  |  | 1 |  | Power down MUX2 and associated channel operation | $\begin{aligned} & \text { MUX PDN override = } \\ & 1 \end{aligned}$ |
|  |  | PWRDWN_MUX_1; enables power-down of MUX1 and associated channel clamp and buffer |  |  |  |  |  | 0 |  |  | MUX1 and associated channel in normal operation |  |
|  |  |  |  |  |  |  |  | 1 |  |  | Power down MUX1 and associated channel operation | $\begin{aligned} & \text { MUX PDN override = } \\ & 1 \end{aligned}$ |
|  |  | PWRDWN_MUX_0; enables power-down of MUX0 and associated channel clamp and buffer |  |  |  |  | 0 |  |  |  | MUX0 and associated channel in normal operation |  |
|  |  |  |  |  |  |  | 1 |  |  |  | Power down MUX0 and associated channel operation | $\begin{aligned} & \text { MUX PDN override = } \\ & 1 \end{aligned}$ |
|  |  | Reserved | 0 | 0 | 0 | 0 |  |  |  |  | Sets as default |  |
| 0x3D | Manual window control | Reserved |  |  |  |  | 0 | 0 | 1 | 0 | Sets to default |  |
|  |  | CKILLTHR[2:0]; color kill threshold |  | 0 | 0 | 0 |  |  |  |  | NTSC, PAL color kill at <0.5\%, SECAM no color kill | CKE $=1$ enables the color kill function and must be enabled for CKILLTHR[2:0] to take effect |
|  |  |  |  | 0 | 0 | 1 |  |  |  |  | NTSC, PAL color kill at <1.5\%, SECAM color kill at $<5 \%$ |  |
|  |  |  |  | 0 | 1 | 0 |  |  |  |  | NTSC, PAL color kill at <2.5\%, SECAM color kill at $<7 \%$ |  |
|  |  |  |  | 0 | 1 | 1 |  |  |  |  | NTSC, PAL color kill at $<4 \%$, SECAM color kill at $<8 \%$ |  |
|  |  |  |  | 1 | 0 | 0 |  |  |  |  | NTSC, PAL color kill at <8.5\%, SECAM color kill at <9.5\% |  |
|  |  |  |  | 1 | 0 | 1 |  |  |  |  | NTSC, PAL color kill at <16\%, SECAM color kill at $<15 \%$ |  |
|  |  |  |  | 1 | 1 | 0 |  |  |  |  | NTSC, PAL color kill at <32\%, SECAM color kill at <32\% |  |
|  |  |  |  | 1 | 1 | 1 |  |  |  |  | Reserved |  |
|  |  | Reserved | 0 |  |  |  |  |  |  |  | Sets to default |  |
| 0x41 | Resample control | Reserved |  |  | 0 | 0 | 0 | 0 | 0 | 1 | Sets to default |  |
|  |  | SFL_INV; controls the behavior of the PAL switch bit |  | 0 |  |  |  |  |  |  | SFL-compatible with the ADV717x and ADV73xx video encoders |  |
|  |  |  |  | 1 |  |  |  |  |  |  | SFL-compatible with older video encoders such as the ADV7194. |  |
|  |  | Reserved | 0 |  |  |  |  |  |  |  | Set to default |  |



| Main Map |  |  | Bits(Shading Indicates Default State) |  |  |  |  |  |  |  | Comments | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Subaddress | Register | Bit Description <br> FLD_OUT_SEL[2:0] selects which sync comes out on the VS/FIELD/SFL pin | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |  |  |
| 0x6B | Output Sync Select 2 |  |  |  |  |  |  | 0 | 0 | 0 | HSYNC |  |
|  |  |  |  |  |  |  |  | 0 | 0 | 1 | VSYNC |  |
|  |  |  |  |  |  |  |  | 0 | 1 | 0 | FIELD |  |
|  |  |  |  |  |  |  |  | 0 | 1 | 1 | DE |  |
|  |  |  |  |  |  |  |  | 1 | 0 | 0 | SFL |  |
|  |  | Reserved | 0 | 0 | 0 | 1 | 0 |  |  |  | Set as default |  |
| 0x8F | Free-Run Line Length 1 | Reserved |  |  |  |  | 0 | 0 | 0 | 0 | Set as default |  |
|  |  | LLC_PAD_SEL[2:0]; enables manual selection of the clock for the LLC pin |  | 0 | 0 | 0 |  |  |  |  | LLC (nominal 27 MHz ) selected out on LLC pin |  |
|  |  |  |  | 1 | 0 | 1 |  |  |  |  | LLC (nominal 13.5 MHz ) selected out on LLC pin |  |
|  |  | Reserved | 0 |  |  |  |  |  |  |  | Sets to default |  |
| 0x99 | CCAP1 (read only) | CCAP1[7:0]; closed caption data register | x | x | x | x | x | x | x | x | CCAP1[7] contains parity bit for Byte 0 |  |
| 0x9A | $\begin{aligned} & \text { CCAP2 } \\ & \text { (read only) } \\ & \hline \end{aligned}$ | CCAP2[7:0]; closed caption data register | x | x | X | x | x | x | x | x | CCAP2[7] contains parity bit for Byte 0 |  |
| 0x9B | Letterbox 1 (read only) | LB_LCT[7:0]; letterbox data register | x | x | X | x | x | x | x | x | Reports the number of black lines detected at the top of active video | This feature examines the active video at the start and end of each field; it enables format detection even if the video is not accompanied by a CGMS or WSS sequence |
| 0x9C | Letterbox 2 (read only) | LB_LCM[7:0]; letterbox data register | x | x | x | x | x | x | X | x | Reports the number of black lines detected in the middle half of active video if subtitles are detected |  |
| 0x9D | Letterbox 3 (read only) | LB_LCB[7:0]; letterbox data register | x | x | x | x | x | x | x | x | Reports the number of black lines detected at the bottom of active video |  |
| 0xB2 | CRC enable (write only) | Reserved <br> CRC_ENABLE; enable CRC checksum decoded from FMS packet to validate CGMSD |  |  |  |  |  |  | 0 | 0 | Sets as default |  |
|  |  |  |  |  |  |  |  | 0 |  |  | Turns off CRC check |  |
|  |  |  |  |  |  |  |  | 1 |  |  | CGMSD goes high with valid checksum |  |
|  |  | Reserved | 0 | 0 | 0 | 1 | 1 |  |  |  | Sets as default |  |
| 0xC3 | ADC Switch 1 | MUX0[2:0]; manual muxing control for MUXO; this setting controls which input is routed to the ADC for processing |  |  |  |  |  | 0 | 0 | 0 | No connect | MAN_MUX_EN = 1 |
|  |  |  |  |  |  |  |  | 0 | 0 | 1 | AIn1 |  |
|  |  |  |  |  |  |  |  | 0 | 1 | 0 | AIn2 |  |
|  |  |  |  |  |  |  |  | 0 | 1 | 1 | AIn3 |  |
|  |  |  |  |  |  |  |  | 1 | 0 | 0 | AlN 4 |  |
|  |  | Reserved |  |  |  |  | 0 |  |  |  |  |  |
|  |  | MUX1[2:0]; manual muxing control for MUX1; this setting controls which input is routed to the ADC for processing |  | 0 | 0 | 0 |  |  |  |  | No connect |  |
|  |  |  |  | 0 | 0 | 1 |  |  |  |  | No connect |  |
|  |  |  |  | 0 | 1 | 0 |  |  |  |  | AlN 2 |  |
|  |  |  |  | 0 | 1 | 1 |  |  |  |  | No connect |  |
|  |  |  |  | 1 | 0 | 0 |  |  |  |  | AIN4 |  |
|  |  | Reserved | 0 |  |  |  |  |  |  |  |  | MAN MUX EN = 1 |
| 0xC4 | ADC Switch 2 | MUX2[2:0]; manual muxing control for MUX2; this setting controls which input is routed to the ADC for processing |  |  |  |  |  | 0 | 0 | 0 | No connect |  |
|  |  |  |  |  |  |  |  | 0 | 0 | 1 | No connect |  |
|  |  |  |  |  |  |  |  | 0 | 1 | 0 | AlN 2 |  |
|  |  |  |  |  |  |  |  | 0 | 1 | 1 | $\mathrm{A}_{\mathrm{IN}} 3$ |  |
|  |  |  |  |  |  |  |  | 1 | 0 | 0 | No connect |  |
|  |  | Reserved |  | 0 | 0 | 0 | 0 |  |  |  |  |  |
|  |  | MAN_MUX_EN; enable manual setting of input signal muxing | 0 |  |  |  |  |  |  |  | Disables | This bit must be set to 1 for manual muxing |
|  |  |  | 1 |  |  |  |  |  |  |  | Enables |  |
| 0xDC | Letterbox Control 1 | LB_TH[4:0]; sets the threshold value that determines if a line is black |  |  |  | 0 | 1 | 1 | 0 | 0 | Default threshold for the detection of black lines 01101 to 10000 -increase threshold, 00000 to 01011—decrease threshold |  |
|  |  | Reserved | 1 | 0 | 1 |  |  |  |  |  | Sets as default |  |


| Main Map |  |  | Bits(Shading Indicates Default State) |  |  |  |  |  |  |  | Comments | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Subaddress | Register | Bit Description | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |  |  |
| 0xDD | Letterbox Control 2 | LB_EL[3:0]; programs the end line of the activity window for LB detection (end of field) |  |  |  |  | 1 | 1 | 0 | 0 | LB detection ends with the last line of active video on a field, 1100b: 262/525 |  |
|  |  | LB_SL[3:0]; programs the start line of the activity window for LB detection (start of field) | 1 | 1 | 0 | 0 |  |  |  |  | Letterbox detection aligned with the start of active video, 0100b: 23/286 NTSC |  |
| 0xDE | ST Noise Readback 1 (read only) | ST_NOISE[10:8] |  |  |  |  |  | x | x | x |  | ST noise[10:0] measures the noise on the horizontal sync tip of video source |
|  |  | ST_NOISE_VLD |  |  |  |  | x |  |  |  | When = 1, ST_NOISE[10:0] is valid |  |
| 0xDF | ST Noise Readback 2 (read only) | ST_NOISE[7:0] | x | x | x | x | x | x | x | x |  |  |
| 0xE1 | SD offset Cb channel | SD_OFF_Cb[7:0]; adjusts the hue by selecting the offset for the Cb channel | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | -312 mV offset applied to the Cb channel |  |
|  |  |  | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 mV offset applied to the Cb channel |  |
|  |  |  | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | +312 mV offset applied to the Cb channel |  |
| 0xE2 | SD offset Cr channel | SD_OFF_Cr[7:0]; adjusts the hue by selecting the offset for the Cr channel | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | -312 mV offset applied to the Cr channel |  |
|  |  |  | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 mV offset applied to the Cr channel |  |
|  |  |  | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | +312 mV offset applied to the Cr channel |  |
| 0xE3 | SD saturation Cb channel | SD_SAT_Cb[7:0]; adjusts the saturation by affecting gain on the Cb channel | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | Gain on Cb channel $=-42 \mathrm{~dB}$ |  |
|  |  |  | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | Gain on Cb channel $=0 \mathrm{~dB}$ |  |
|  |  |  | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | Gain on Cb channel $=+6 \mathrm{~dB}$ |  |
| 0xE4 | SD saturation Cr channel | SD_SAT_Cr[7:0]; adjusts the saturation by affecting gain on the Cr channel | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | Gain on Cr channel $=-42 \mathrm{~dB}$ |  |
|  |  |  | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | Gain on Cr channel $=0 \mathrm{~dB}$ |  |
|  |  |  | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | Gain on Cr channel $=+6 \mathrm{~dB}$ |  |
| 0xE5 | NTSC VSYNC begin | NVBEG[4:0]; number of lines after Icount rollover to set V high |  |  |  | 0 | 0 | 1 | 0 | 1 | NTSC default (ITU-R BT.656) |  |
|  |  | NVBEGSIGN |  |  | 0 |  |  |  |  |  | Sets to low when manual programming |  |
|  |  |  |  |  | 1 |  |  |  |  |  | Not suitable for user programming |  |
|  |  | NVBEGDELE; delay V bit going high by one line relative to NVBEG (even field) |  | 0 |  |  |  |  |  |  | No delay |  |
|  |  |  |  | 1 |  |  |  |  |  |  | Additional delay by one line |  |
|  |  | NVBEGDELO; delay V bit going high by one line relative to NVBEG (odd field) | 0 |  |  |  |  |  |  |  | No delay |  |
|  |  |  | 1 |  |  |  |  |  |  |  | Additional delay by one line |  |
| 0xE6 | NTSC VSYNC end | NVEND[4:0]; number of lines after ICount rollover to set V low |  |  |  | 0 | 0 | 1 | 0 | 0 | NTSC default (ITU-R BT.656) |  |
|  |  | NVENDSIGN |  |  | 0 |  |  |  |  |  | Sets to low when manual programming |  |
|  |  |  |  |  | 1 |  |  |  |  |  | Not suitable for user programming |  |
|  |  | NVENDDELE; delay V bit going low by one line relative to NVEND (even field) |  | 0 |  |  |  |  |  |  | No delay |  |
|  |  |  |  | 1 |  |  |  |  |  |  | Additional delay by one line |  |
|  |  | NVENDDELO; delay V bit going low by one line relative to NVEND (odd field) | 0 |  |  |  |  |  |  |  | No delay |  |
|  |  |  | 1 |  |  |  |  |  |  |  | Additional delay by one line |  |
| 0xE7 | NTSC FIELD toggle | NFTOG[4:0]; number of lines after I count rollover to toggle F signal |  |  |  | 0 | 0 | 0 | 1 | 1 | NTSC default |  |
|  |  | NFTOGSIGN |  |  | 0 |  |  |  |  |  | Sets to low when manual programming |  |
|  |  |  |  |  | 1 |  |  |  |  |  | Not suitable for user programming |  |
|  |  | NFTOGDELE; delay F transition by one line relative to NFTOG (even field) |  | 0 |  |  |  |  |  |  | No delay |  |
|  |  |  |  | 1 |  |  |  |  |  |  | Additional delay by one line |  |
|  |  | NFTOGDELO; delay F transition by one line relative to NFTOG (odd field) | 0 |  |  |  |  |  |  |  | No delay |  |
|  |  |  | 1 |  |  |  |  |  |  |  | Additional delay by one line |  |


| Main Map |  |  | Bits(Shading Indicates Default State) |  |  |  |  |  |  |  | Comments | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Subaddress | Register | Bit Description | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |  |  |
| 0xE8 | PAL VSYNC begin | PVBEG[4:0]; number of lines after I Count rollover to set $V$ high |  |  |  | 0 | 0 | 1 | 0 | 1 | PAL default (ITU-R BT.656) |  |
|  |  | PVBEGSIGN |  |  | 0 |  |  |  |  |  | Sets to low when manual programming |  |
|  |  |  |  |  | 1 |  |  |  |  |  | Not suitable for user programming |  |
|  |  | PVBEGDELE; delay V bit going high by one line relative to PVBEG (even field) |  | 0 |  |  |  |  |  |  | No delay |  |
|  |  |  |  | 1 |  |  |  |  |  |  | Additional delay by one line |  |
|  |  | PVBEGDELO; delay V bit going high by one line relative to PVBEG (odd field) | 0 |  |  |  |  |  |  |  | No delay |  |
|  |  |  | 1 |  |  |  |  |  |  |  | Additional delay by one line |  |
| 0xE9 | PAL VSYNC end | PVEND[4:0]; number of lines after Icount rollover to set V low. |  |  |  | 1 | 0 | 1 | 0 | 0 | PAL default (ITU-R BT.656) |  |
|  |  | PVENDSIGN |  |  | 0 |  |  |  |  |  | Sets to low when manual programming |  |
|  |  |  |  |  | 1 |  |  |  |  |  | Not suitable for user programming |  |
|  |  | PVENDDELE; delay V bit going low by one line relative to PVEND (even field) |  | 0 |  |  |  |  |  |  | No delay |  |
|  |  |  |  | 1 |  |  |  |  |  |  | Additional delay by one line |  |
|  |  | PVENDDELO; delay V bit going low by one line relative to PVEND (odd field) | 0 |  |  |  |  |  |  |  | No delay |  |
|  |  |  | 1 |  |  |  |  |  |  |  | Additional delay by one line |  |
| 0xEA | PAL FIELD toggle | PFTOG[4:0]; number of lines after Icount rollover to toggle F signal |  |  |  | 0 | 0 | 0 | 1 | 1 | PAL default (ITU-R BT.656) |  |
|  |  | PFTOGSIGN |  |  | 0 |  |  |  |  |  | Sets to low when manual programming |  |
|  |  |  |  |  | 1 |  |  |  |  |  | Not suitable for user programming |  |
|  |  | PFTOGDELE; delay F transition by one line relative to PFTOG (even field) |  | 0 |  |  |  |  |  |  | No delay |  |
|  |  |  |  | 1 |  |  |  |  |  |  | Additional delay by one line |  |
|  |  | PFTOGDELO; delay F transition by one line relative to PFTOG (odd field) | 0 |  |  |  |  |  |  |  | No delay |  |
|  |  |  | 1 |  |  |  |  |  |  |  | Additional delay by one line |  |
| 0xEB | Vblank Control 1 | PVBIELCM[1:0]; PAL VBI even field line control |  |  |  |  |  |  | 0 | 0 | VBI ends one line earlier (Line 335) | Controls position of first active (comb filtered) line after VBI on even field in PAL |
|  |  |  |  |  |  |  |  |  | 0 | 1 | ITU-R BT.470 compliant (Line 336) |  |
|  |  |  |  |  |  |  |  |  | 1 | 0 | VBI ends one line later (Line 337) |  |
|  |  |  |  |  |  |  |  |  | 1 | 1 | VBI ends two lines later (Line 338) |  |
|  |  | PVBIOLCM[1:0]; PAL VBI odd field line control |  |  |  |  | 0 | 0 |  |  | VBI ends one line earlier (Line 22) | Controls position of first active (comb filtered) line after VBI on odd field in PAL |
|  |  |  |  |  |  |  | 0 | 1 |  |  | ITU-R BT. 470 compliant (Line 23) |  |
|  |  |  |  |  |  |  | 1 | 0 |  |  | VBI ends one line later (Line 24) |  |
|  |  |  |  |  |  |  | 1 | 1 |  |  | VBI ends two lines later (Line 25) |  |
|  |  | NVBIELCM[1:0]; NTSC VBI even field line control |  |  | 0 | 0 |  |  |  |  | VBI ends one line earlier (Line 282) | Controls position of first active (comb filtered) line after VBI on even field in NTSC |
|  |  |  |  |  | 0 | 1 |  |  |  |  | ITU-R BT. 470 compliant (Line 283) |  |
|  |  |  |  |  | 1 | 0 |  |  |  |  | VBI ends one line later (Line 284) |  |
|  |  |  |  |  | 1 | 1 |  |  |  |  | VBI ends two lines later (Line 285) |  |
|  |  | NVBIOLCM[1:0]; NTSC VBI odd field line control | 0 | 0 |  |  |  |  |  |  | VBI ends one line earlier (Line 20) | Controls position of first active (comb filtered) line after VBI on odd field in NTSC |
|  |  |  | 0 | 1 |  |  |  |  |  |  | ITU-R BT. 470 compliant (Line 21) |  |
|  |  |  | 1 | 0 |  |  |  |  |  |  | VBI ends one line later (Line 22) |  |
|  |  |  | 1 | 1 |  |  |  |  |  |  | VBI ends two lines later (Line 23) |  |


| Main Map |  |  | Bits(Shading Indicates Default State) |  |  |  |  |  |  |  | Comments | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Subaddress | Register | Bit Description | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |  |  |
| 0xEC | Vblank Control 2 | PVBIECCM[1:0]; PAL VBI even field color control |  |  |  |  |  |  | 0 | 0 | Color output beginning Line 335 | Controls the position of first line that outputs color after VBI on even field in PAL |
|  |  |  |  |  |  |  |  |  | 0 | 1 | ITU-R BT. 470 compliant color output beginning Line 336 |  |
|  |  |  |  |  |  |  |  |  | 1 | 0 | Color output beginning Line 337 |  |
|  |  |  |  |  |  |  |  |  | 1 | 1 | Color output beginning Line 338 |  |
|  |  | PVBIOCCM[1:0]; PAL VBI odd field color control |  |  |  |  | 0 | 0 |  |  | Color output beginning Line 22 | Controls the position of first line that outputs color after VBI on odd field in PAL |
|  |  |  |  |  |  |  | 0 | 1 |  |  | ITU-R BT.470-compliant color output beginning Line 23 |  |
|  |  |  |  |  |  |  | 1 | 0 |  |  | Color output beginning Line 24 |  |
|  |  |  |  |  |  |  | 1 | 1 |  |  | Color output beginning Line 25 |  |
|  |  | NVBIECCM[1:0]; NTSC VBI even field color control |  |  | 0 | 0 |  |  |  |  | Color output beginning Line 282 | Controls the position of first line that outputs color after VBI on even field in NTSC |
|  |  |  |  |  | 0 | 1 |  |  |  |  | ITU-R BT.470-compliant color output beginning Line 283 |  |
|  |  |  |  |  | 1 | 0 |  |  |  |  | VBI ends one line later (Line 284) |  |
|  |  |  |  |  | 1 | 1 |  |  |  |  | Color output beginning Line 285 |  |
|  |  | NVBIOCCM[1:0]; NTSC VBI odd field color control | 0 | 0 |  |  |  |  |  |  | Color output beginning Line 20 | Controls the position of first line that outputs color after VBI on odd field in NTSC |
|  |  |  | 0 | 1 |  |  |  |  |  |  | ITU-R BT. 470 compliant color output beginning Line 21 |  |
|  |  |  | 1 | 0 |  |  |  |  |  |  | Color output beginning Line 22 |  |
|  |  |  | 1 | 1 |  |  |  |  |  |  | Color output beginning Line 23 |  |
| 0xF3 | AFE CONTROL 1 | AA_FILT_EN[3:0] antialiasing filter enable |  |  |  |  |  |  |  | 0 | Antialiasing Filter 1 disabled | AA_FILT_MAN_OVR must be enabled to change settings defined by INSEL[4:0] |
|  |  |  |  |  |  |  |  |  |  | 1 | Antialiasing Filter 1 enabled |  |
|  |  |  |  |  |  |  |  |  | 0 |  | Antialiasing Filter 2 disabled |  |
|  |  |  |  |  |  |  |  |  | 1 |  | Antialiasing Filter 2 enabled |  |
|  |  |  |  |  |  |  |  | 0 |  |  | Antialiasing Filter 3 disabled |  |
|  |  |  |  |  |  |  |  | 1 |  |  | Antialiasing Filter 3 enabled |  |
|  |  |  |  |  |  |  | 0 |  |  |  | Antialiasing Filter 4 enabled |  |
|  |  |  |  |  |  |  | 1 |  |  |  | Antialiasing Filter 4 enabled |  |
|  |  | AA_FILT_MAN_OVR; antialiasing filter override |  |  |  | 0 |  |  |  |  | Override disabled |  |
|  |  |  |  |  |  | 1 |  |  |  |  | Override enabled |  |
|  |  | Reserved | 0 | 0 | 0 |  |  |  |  |  |  |  |
| 0xF4 | Drive strength | DR_STR_S[1:0]; selects the drive strength for the sync output signals |  |  |  |  |  |  | 0 | 0 | Low drive strength (1×) |  |
|  |  |  |  |  |  |  |  |  | 0 | 1 | Medium low drive strength ( $2 x$ ) |  |
|  |  |  |  |  |  |  |  |  | 1 | 0 | Medium high drive strength ( $3 \times$ ) |  |
|  |  |  |  |  |  |  |  |  | 1 | 1 | High drive strength ( $4 \times$ ) |  |
|  |  | DR_STR_C[1:0]; selects the drive strength for the clock output signal |  |  |  |  | 0 | 0 |  |  | Low drive strength ( $1 \times$ ) |  |
|  |  |  |  |  |  |  | 0 | 1 |  |  | Medium low drive strength ( $2 x$ ) |  |
|  |  |  |  |  |  |  | 1 | 0 |  |  | Medium high drive strength ( $3 \times$ ) |  |
|  |  |  |  |  |  |  | 1 | 1 |  |  | High drive strength ( $4 \times$ ) |  |
|  |  | DR_STR[1:0]; selects the drive strength for the data output signals; can be increased or decreased for EMC or crosstalk reasons |  |  | 0 | 0 |  |  |  |  | Low drive strength ( $1 \times$ ) |  |
|  |  |  |  |  | 0 | 1 |  |  |  |  | Medium low drive strength (2x) |  |
|  |  |  |  |  | 1 | 0 |  |  |  |  | Medium high drive strength ( $3 x$ ) |  |
|  |  |  |  |  | 1 | 1 |  |  |  |  | High drive strength ( $4 \times$ ) |  |
|  |  | Reserved |  | x |  |  |  |  |  |  |  |  |
|  |  | GLITCH_FILT_BYP | 0 |  |  |  |  |  |  |  |  |  |
|  |  |  | 1 |  |  |  |  |  |  |  |  |  |



[^3]To access the registers listed in Table 96, SUB_USR_EN in Register Address 0x0E must be programmed to 10b. The gray shading is the default.

Table 96. Register Map Descriptions User Sub Map 2


| User Sub Map 2 |  |  | Bits(Shading Indicates Default State) |  |  |  |  |  |  |  | Comments | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Subaddress | Register | Bit Description | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |  |  |
| 0xE5 | Y Average 4 | LINE_START[0] |  |  |  |  |  |  |  | 1 |  |  |
|  |  | LINE_END[0] |  |  |  |  |  |  | 1 |  |  |  |
|  |  | Reserved |  |  |  |  | 0 | 0 |  |  |  |  |
|  |  | SAMPLE_START[1:0] |  |  | 1 | 0 |  |  |  |  |  |  |
|  |  | SAMPLE_END[1:0] | 0 | 0 |  |  |  |  |  |  |  |  |
| 0xE6 | Y Average 5 | CAPTURE_VALUE |  |  |  |  |  |  |  | 0 | Trigger used to store the readback value. |  |
|  |  | Y_AVG_FILT_EN |  |  |  |  |  |  | 0 |  | Enable low pass filtering of the y_averaged signal. |  |
|  |  | Y_AVG_TIME_CONST[2:0] |  |  |  | 1 | 0 | 0 |  |  | Selects the filter cutoff to be used for filtering the $y$ averaged data. <br> $3^{\prime} \mathrm{b} 1 \mathrm{xx}=$ least filtered. <br> $3^{\prime} \mathrm{b} 000=$ next least. <br> 3'b011 = heavily filtered. |  |
|  |  | Reserved | 0 | 0 | 0 |  |  |  |  |  |  |  |
| 0xE7 | Y Average Data MSB | Y_AVERAGE[9:2] | x | x | x | x | x | x | x | x | Contains the averaged video data. | Note these are read only registers |
| 0xE8 | Y Average Data LSB | Y_AVERAGE[1:0] |  |  |  |  |  |  | x | x |  |  |

To access the registers listed in Table 97, SUB_USR_EN in Register Address 0x0E must be programmed to 01 b . The gray shading is the default.
Table 97. Interrupt/VDP Map Descriptions ${ }^{1}$


| Interrupt/VDP Map |  |  | Bit (Shading Indicates Default State) |  |  |  |  |  |  |  | Comments | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Address | Register | Bit Description | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |  |  |
| 0x45 | Raw Status 2 (read only) | CCAPD |  |  |  |  |  |  |  | 0 | No CCAPD data detected-VBI System 2 | These bits are status bits only; they cannot be cleared or masked; Register 0x46 is used for this purpose |
|  |  |  |  |  |  |  |  |  |  | 1 | CCAPD data detected-VBI System 2 |  |
|  |  | Reserved |  |  |  |  | x | x | x |  |  |  |
|  |  | EVEN_FIELD |  |  |  | 0 |  |  |  |  | Current SD field is odd numbered |  |
|  |  |  |  |  |  | 1 |  |  |  |  | Current SD field is even numbered |  |
|  |  | CHX_MIN_MAX_INTRQ |  |  | 0 |  |  |  |  |  | If the input to the ADC is within the correct range this is 0 |  |
|  |  |  |  |  | 1 |  |  |  |  |  | If the input to the ADC is outside the range this is set to 1. The range is set by User Sub Map 2 |  |
|  |  | Reserved |  | x |  |  |  |  |  |  | Not used |  |
|  |  | MPU_STIM_INTRQ | 0 |  |  |  |  |  |  |  | MPU_STIM_INTRQ = 0 |  |
|  |  |  | 1 |  |  |  |  |  |  |  | MPU_STIM_INTRQ = 1 |  |
| 0x46 | Interrupt Status 2 (read only) | CCAPD_Q |  |  |  |  |  |  |  | 0 | Closed captioning not detected in the input video signal-VBI System 2 | These bits can be cleared or masked by Register 0x47 and Register 0x48, respectively; note that the interrupt in Register $0 \times 46$ for the CCAP, CGMS, and WSS data uses the Mode 1 data slicer |
|  |  |  |  |  |  |  |  |  |  | 1 | Closed captioning data detected in the video input signal-VBI System 2 |  |
|  |  | Reserved |  |  |  |  |  |  | x |  | Not used |  |
|  |  | Reserved |  |  |  |  | x | x |  |  | Not used |  |
|  |  | SD_FIELD_CHNGD_Q |  |  |  | 0 |  |  |  |  | SD signal has not changed field from odd to even or vice versa |  |
|  |  |  |  |  |  | 1 |  |  |  |  | SD signal has changed Field from odd to even or vice versa |  |
|  |  | Reserved |  | x | x |  |  |  |  |  | Not used |  |
|  |  | MPU_STIM_INTRQ_Q | 0 |  |  |  |  |  |  |  | Manual interrupt not set |  |
|  |  |  | 1 |  |  |  |  |  |  |  | Manual interrupt set |  |
| 0x47 | Interrupt Clear 2 (write only) | CCAPD_CLR |  |  |  |  |  |  |  | 0 | Do not clear-VBI System 2 | Note that interrupt in Register $0 \times 46$ for the CCAP, CGMS, and WSS data uses the Mode 1 data slicer |
|  |  |  |  |  |  |  |  |  |  | 1 | Clears CCAPD_Q bit—VBI System 2 |  |
|  |  | Reserved |  |  |  |  |  |  | x |  | Not used |  |
|  |  | Reserved |  |  |  |  | x | x |  |  |  |  |
|  |  | SD_FIELD_CHNGD_CLR |  |  |  | 0 |  |  |  |  | Do not clear |  |
|  |  |  |  |  |  | 1 |  |  |  |  | ```Clears SD_FIELD_CHNGD_Q``` |  |
|  |  | CHX_MIN_MAX_INTRQ_CLR |  |  | 0 |  |  |  |  |  | Do not clear |  |
|  |  |  |  |  | 1 |  |  |  |  |  | Clears CHX_MIN_MAX_INTRQ bit |  |
|  |  | Reserved |  | x |  |  |  |  |  |  | Not used |  |
|  |  | MPU_STIM_INTRQ_CLR | 0 |  |  |  |  |  |  |  | Do not clear |  |
|  |  |  | 1 |  |  |  |  |  |  |  | Clears MPU_STIM_INTRQ_Q bit |  |




| Interrupt/VDP Map |  |  | Bit (Shading Indicates Default State) |  |  |  |  |  |  |  | Comments | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Address | Interrupt Clear 4 (write only) | Bit Description | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |  |  |
| 0x4F |  | VDP_CCAPD_CLR |  |  |  |  |  |  |  | 0 | Do not clear | In Register 0x4E, CCAP/CGMS/ WSS data uses VDP data slicer |
|  |  |  |  |  |  |  |  |  |  | 1 | Clears VDP_CCAPD_Q |  |
|  |  | Reserved |  |  |  |  |  |  | 0 |  |  |  |
|  |  | VDP_CGMS_WSS_CHNGD_CLR |  |  |  |  |  | 0 |  |  | Do not clear |  |
|  |  |  |  |  |  |  |  | 1 |  |  | Clears <br> VDP_CGMS_WSS_CHNGD_Q |  |
|  |  | Reserved |  |  |  |  | 0 |  |  |  |  |  |
|  |  | Reserved |  |  |  | 0 |  |  |  |  |  |  |
|  |  | Reserved |  |  | 0 |  |  |  |  |  |  |  |
|  |  | Reserved |  | 0 |  |  |  |  |  |  | Do not clear |  |
|  |  | Reserved | 0 |  |  |  |  |  |  |  |  |  |
|  |  | VDP_CCAPD_CLR |  |  |  |  |  |  |  | 0 | Do not clear |  |
| 0x50 | Interrupt Mask 4 | VDP_CCAPD_MSK |  |  |  |  |  |  |  | 0 | Masks VDP_CCAPD_Q | Note that an interrupt in Register $0 \times 4 E$ for the CCAP, CGMS, and WSS data uses the VDP data slicer |
|  |  |  |  |  |  |  |  |  |  | 1 | Unmasks VDP_CCAPD_Q |  |
|  |  | /7DP_CGMS_WSS_CHNGD_MSK |  |  |  |  |  |  | 0 |  |  |  |
|  |  |  |  |  |  |  |  | 0 |  |  | Masks VDP_CGMS_WSS_CHNGD_Q |  |
|  |  |  |  |  |  |  |  | 1 |  |  | $\begin{array}{\|l\|} \hline \text { Unmasks } \\ \text { VDP_CGMS_WSS_CHNGD_Q } \end{array}$ |  |
|  |  | Reserved |  |  |  |  | 0 |  |  |  |  |  |
|  |  | Reserved |  |  |  | 0 |  |  |  |  |  |  |
|  |  | Reserved |  |  | 0 |  |  |  |  |  |  |  |
|  |  | Reserved |  | 0 |  |  |  |  |  |  |  |  |
|  |  | Reserved | 0 |  |  |  |  |  |  |  |  |  |
| 0x51 | Interrupt Latch 0 (read only) | CR_CHANNEL_MAX_VIOLATION |  |  |  |  |  |  |  | 0 | Cr value is below programmed maximum value | This register is cleared by CHX_MIN_MAX_INTRQ_CLR |
|  |  |  |  |  |  |  |  |  |  | 1 | Cr value is above programmed maximum value |  |
|  |  | CR_CHANNEL_MIN_VIOLATION |  |  |  |  |  |  | 0 |  | Cr value is above programmed minimum value |  |
|  |  |  |  |  |  |  |  |  | 1 |  | Cr value is below programmed minimum value |  |
|  |  | CB_CHANNEL_MAX_VIOLATION |  |  |  |  |  | 0 |  |  | Cb value is below programmed maximum value |  |
|  |  |  |  |  |  |  |  | 1 |  |  | Cb value is above programmed maximum value |  |
|  |  | CB_CHANNEL_MIN_VIOLATION |  |  |  |  | 0 |  |  |  | Cb value is above programmed minimum value |  |
|  |  |  |  |  |  |  | 1 |  |  |  | Cb value is below programmed minimum value |  |
|  |  | Y_CHANNEL_MAX_VIOLATION |  |  |  | 0 |  |  |  |  | Y value is below programmed maximum value |  |
|  |  |  |  |  |  | 1 |  |  |  |  | Y value is above programmed maximum value |  |
|  |  | Y_CHANNEL_MIN_VIOLATION |  |  | 0 |  |  |  |  |  | Y value is above programmed minimum value |  |
|  |  |  |  |  | 1 |  |  |  |  |  | Y value is below programmed minimum value |  |
|  |  | Reserved | 0 | 0 |  |  |  |  |  |  |  |  |



| Interrupt/VDP Map |  |  | Bit (Shading Indicates Default State) |  |  |  |  |  |  |  | Comments | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Address | Register | Bit Description | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |  |  |
| 0x66 | VDP_LINE_010 | VBI_DATA_P320_N287[3:0] |  |  |  |  | 0 | 0 | 0 | 0 | Sets VBI standard to be decoded from Line 320 (PAL), Line 287 (NTSC) | MAN_LINE_PGM must be set to 1 for these bits to be effective |
|  |  | VBI_DATA_P7_N24[3:0] | 0 | 0 | 0 | 0 |  |  |  |  | Sets VBI standard to be decoded from Line 7 (PAL), Line 24 (NTSC) |  |
| 0x67 | VDP_LINE_011 | VBI_DATA_P321_N288[3:0] |  |  |  |  | 0 | 0 | 0 | 0 | Sets VBI standard to be decoded from Line 321 (PAL), Line 288 (NTSC) | MAN_LINE_PGM must be set to 1 for these bits to be effective |
|  |  | VBI_DATA_P8_N25[3:0] | 0 | 0 | 0 | 0 |  |  |  |  | Sets VBI standard to be decoded from Line 8 (PAL), Line 25 (NTSC) |  |
| 0x68 | VDP_LINE_012 | VBI_DATA_P322[3:0] |  |  |  |  | 0 | 0 | 0 | 0 | Sets VBI standard to be decoded from Line 322 (PAL), NTSC—not applicable | MAN_LINE_PGM must be set to 1 for these bits to be effective |
|  |  | VBI_DATA_P9[3:0] | 0 | 0 | 0 | 0 |  |  |  |  | Sets VBI standard to be decoded from Line 9 (PAL), NTSC-not applicable |  |
| 0x69 | VDP_LINE_013 | VBI_DATA_P323[3:0] |  |  |  |  | 0 | 0 | 0 | 0 | Sets VBI standard to be decoded from Line 323 (PAL), NTSC—not applicable | MAN_LINE_PGM must be set to 1 for these bits to be effective |
|  |  | VBI_DATA_P10[3:0] | 0 | 0 | 0 | 0 |  |  |  |  | Sets VBI standard to be decoded from Line 10 (PAL), NTSC—not applicable |  |
| 0x6A | VDP_LINE_014 | VBI_DATA_P324_N272[3:0] |  |  |  |  | 0 | 0 | 0 | 0 | Sets VBI standard to be decoded from Line 324 (PAL), Line 272 (NTSC) | MAN_LINE_PGM must be set to 1 for these bits to be effective |
|  |  | VBI_DATA_P11[3:0] | 0 | 0 | 0 | 0 |  |  |  |  | Sets VBI standard to be decoded from Line 11 (PAL); NTSC—not applicable |  |
| 0x6B | VDP_LINE_015 | VBI_DATA_P325_N273[3:0] |  |  |  |  | 0 | 0 | 0 | 0 | Sets VBI standard to be decoded from Line 325 (PAL), Line 273 (NTSC) | MAN_LINE_PGM must be set to 1 for these bits to be effective |
|  |  | VBI_DATA_P12_N10[3:0] | 0 | 0 | 0 | 0 |  |  |  |  | Sets VBI standard to be decoded from Line 12 (PAL), Line 10 (NTSC) |  |
| 0x6C | VDP_LINE_016 | VBI_DATA_P326_N274[3:0] |  |  |  |  | 0 | 0 | 0 | 0 | Sets VBI standard to be decoded from Line 326 (PAL), Line 274 (NTSC) | MAN_LINE_PGM must be set to 1 for these bits to be effective |
|  |  | VBI_DATA_P13_N11[3:0] | 0 | 0 | 0 | 0 |  |  |  |  | Sets VBI standard to be decoded from Line 13 (PAL), Line 11 (NTSC) |  |
| 0x6D | VDP_LINE_017 | VBI_DATA_P327_N275[3:0] |  |  |  |  | 0 | 0 | 0 | 0 | Sets VBI standard to be decoded from Line 327 (PAL), Line 275 (NTSC) | MAN_LINE_PGM must be set to 1 for these bits to be effective |
|  |  | VBI_DATA_P14_N12[3:0] | 0 | 0 | 0 | 0 |  |  |  |  | Sets VBI standard to be decoded from Line 14 (PAL), Line 12 (NTSC) |  |
| 0x6E | VDP_LINE_018 | VBI_DATA_P328_N276[3:0] |  |  |  |  | 0 | 0 | 0 | 0 | Sets VBI standard to be decoded from Line 328 (PAL), Line 276 (NTSC) | MAN_LINE_PGM must be set to 1 for these bits to be effective |
|  |  | VBI_DATA_P15_N13[3:0] | 0 | 0 | 0 | 0 |  |  |  |  | Sets VBI standard to be decoded from Line 15 (PAL), Line 13 (NTSC) |  |
| 0x6F | VDP_LINE_019 | VBI_DATA_P329_N277[3:0] |  |  |  |  | 0 | 0 | 0 | 0 | Sets VBI standard to be decoded from Line 329 (PAL), Line 277 (NTSC) | MAN_LINE_PGM must be set to 1 for these bits to be effective |
|  |  | VBI_DATA_P16_N14[3:0] | 0 | 0 | 0 | 0 |  |  |  |  | Sets VBI standard to be decoded from Line 16 (PAL), Line 14 (NTSC) |  |



| Interrupt/VDP Map |  |  | Bit (Shading Indicates Default State) |  |  |  |  |  |  |  |  | Comments | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Address | Register | Bit Description | 7 | 6 | 5 | 4 | 3 |  | 2 | 1 | 0 |  |  |
| 0x78 | VDP_STATUS (read only) | CC_AVL |  |  |  |  |  |  |  |  | 0 | Closed captioning not detected | CC_CLEAR resets the CC_AVL bit |
|  |  |  |  |  |  |  |  |  |  |  | 1 | Closed captioning is detected |  |
|  |  | CC_EVEN_FIELD |  |  |  |  |  |  |  | 0 |  | Closed captioning decoded from odd field |  |
|  |  |  |  |  |  |  |  |  |  | 1 |  | Closed captioning decoded from even field |  |
|  |  | CGMS_WSS_AVL |  |  |  |  |  |  | 0 |  |  | CGMS/WSS is not detected | CGMS_WSS_CLEAR resets the CGMS_WSS_AVL bit |
|  |  |  |  |  |  |  |  |  | 1 |  |  | CGMS/WSS detected |  |
|  |  | Reserved |  | 0 | 0 | 0 | 0 |  |  |  |  |  |  |
|  |  | TTXT_AVL | 0 |  |  |  |  |  |  |  |  | Teletext not detected |  |
|  |  |  | 1 |  |  |  |  |  |  |  |  | Teletext detected |  |
|  | VDP_STATUS_CLEAR (write only) | CC_CLEAR |  |  |  |  |  |  |  |  | 0 | Does not reinitialize the CCAP readback registers | This is a self-clearing bit |
|  |  |  |  |  |  |  |  |  |  |  | 1 | Reinitializes the CCAP readback registers |  |
|  |  | Reserved |  |  |  |  |  |  |  | 0 |  |  |  |
|  |  | CGMS_WSS_CLEAR |  |  |  |  |  |  | 0 |  |  | Does not reinitialize the CGMS/WSS readback registers | This is a self-clearing bit |
|  |  |  |  |  |  |  |  |  | 1 |  |  | Reinitializes the CGMS/WSS readback registers |  |
|  |  | Reserved | 0 | 0 | 0 | 0 | 0 |  |  |  |  |  |  |
| 0x79 | $\begin{aligned} & \hline \text { VDP_CCAP_DATA_0 } \\ & \text { (read only) } \end{aligned}$ | CCAP_BYTE_1[7:0] | x | x | x | x | x |  | x | x | x | Decoded Byte 1 of CCAP |  |
| 0x7A | $\begin{aligned} & \text { VDP_CCAP_DATA_1 } \\ & \text { (read only) } \end{aligned}$ | CCAP_BYTE_2[7:0] | x | x | x | x | x |  | x | x | x | Decoded Byte 2 of CCAP |  |
| 0x7D | VDP_CGMS_WSS_DATA_0 (read only) | CGMS_CRC[5:2] |  |  |  |  | x |  | x | x | x | Decoded CRC sequence for CGMS |  |
|  |  | Reserved | 0 | 0 | 0 | 0 |  |  |  |  |  |  |  |
| 0x7E | VDP_CGMS_WSS_DATA_1 (read only) | CGMS_WSS[13:8] |  |  | x | x | x |  | x | x | x | Decoded CGMS/WSS data |  |
|  |  | CGMS_CRC[1:0] | x | x |  |  |  |  |  |  |  | Decoded CRC sequence for CGMS |  |
| 0x7F | VDP_CGMS_WSS_DATA_2 (read only) | CGMS_WSS[7:0] | x | x | x | x | x |  | x | x | x | Decoded CGMS/WSS data |  |
| 0x9C | VDP_OUTPUT_SEL | Reserved |  |  |  |  | 0 |  | 0 | 0 | 0 |  |  |
|  |  | WSS_CGMS_CB_CHANGE |  |  |  | 0 |  |  |  |  |  | Disable content-based updating of CGMS and WSS data | The available bit shows the availability of data only when its content has changed |
|  |  |  |  |  |  | 1 |  |  |  |  |  | Enable content-based updating of CGMS and WSS data |  |
|  |  | Reserved | 0 | 0 | 0 |  |  |  |  |  |  |  |  |

[^4]
## PCB LAYOUT RECOMMENDATIONS

The ADV7182 is a high precision, high speed, mixed-signal device. To achieve the maximum performance from the part, it is important to have a well laid out PCB. The following is a guide for designing a board using the ADV7182.

## ANALOG INTERFACE INPUTS

Take care when routing the inputs on the PCB. Keep track lengths to a minimum and use $75 \Omega$ trace impedances when possible because trace impedances other than $75 \Omega$ increase the chance of reflections.

Place the resistor divider and ac coupling capacitor circuit described in the Input Networks section as close as possible to the $A_{\text {IN }}$ pins of the ADV7182. Place the resistor divider and ac coupling capacitor circuit on the same side of the PCB as the ADV7182.

## POWER SUPPLY DECOUPLING

It is recommended that each power supply pin be decoupled with $0.1 \mu \mathrm{~F}$ and 10 nF capacitors. The fundamental idea is to have a decoupling capacitor within about 0.5 cm of each power pin. In addition, avoid placing the capacitor on the opposite side of the PCB from the ADV7182 because doing so interposes inductive vias in the path. Locate the decoupling capacitors between the power plane and the power pin. Flow current from the power plane to the capacitor and then to the power pin. Do not apply the power connection between the capacitor and the power pin. Placing a via underneath the 100 nF capacitor pads, down to the power plane, is the best approach (see Figure 49).


Figure 49. Recommended Power Supply Decoupling
It is particularly important to maintain low noise and good stability of PVDD. Pay careful attention to regulation, filtering, and decoupling. It is highly desirable to provide separate regulated supplies for each of the analog circuitry groups (AVDD, DVDD, DVDDIO, and PVDD).
Some graphic controllers use substantially different levels of power when active (during active picture time) and when idle (during horizontal and vertical sync periods). This can result in a measurable change in the voltage supplied to the analog supply regulator, which can in turn produce changes in the regulated analog supply voltage. This can be mitigated by regulating the analog supply, or at least $\mathrm{P}_{\mathrm{VDD}}$, from a different, cleaner power source, for example, from a 12 V supply.
Using a single ground plane for the entire board is also recommended.

Experience has repeatedly shown that the noise performance is the same or better with a single ground plane. Using multiple ground planes can be detrimental because each separate ground plane is smaller, and long ground loops can result.

## VREFN AND VREFP

Place the circuit associated with theVREFN and VREFP pins as close as possible and on the same side of the PCB as the ADV7182.
DIGITAL OUTPUTS (BOTH DATA AND CLOCKS)
Try to minimize the trace length that the digital outputs have to drive. Longer traces have higher capacitance, requiring more current and, in turn, causing more internal digital noise. Shorter traces reduce the possibility of reflections.
Adding a $30 \Omega$ to $50 \Omega$ series resistor can suppress reflections, reduce EMI, and reduce the current spikes inside the ADV7182. If series resistors are used, place them as close as possible to the ADV7182 pins. However, try not to add vias or extra length to the output trace to place the resistors closer.

If possible, limit the capacitance that each of the digital outputs drives to less than 15 pF . This can be accomplished easily by keeping traces short and by connecting the outputs to only one device. Loading the outputs with excessive capacitance increases the current transients inside the ADV7182, creating more digital noise on its power supplies.
The ADV7182 has an exposed metal paddle on the bottom of the package. This paddle must be soldered to PCB ground for proper heat dissipation and for noise and mechanical strength benefits.

## DIGITAL INPUTS

The digital inputs on the ADV7182 are designed to work with 1.8 V to 3.3 V signals and are not tolerant of 5 V signals. Extra components are needed if 5 V logic signals are required to be applied to the decoder.

## Data Sheet

## TYPICAL CIRCUIT CONNECTION

Figure 50 provides an example of how to connect the ADV7182. For a detailed schematic of the ADV7182 evaluation boards, contact a local Analog Devices field applications engineer or Analog Devices distributor.


Figure 50. ADV7182 Typical Connection Diagram

## OUTLINE DIMENSIONS



THE EXPOSED PAD, REFER TO
THE PIN CONFIGURATION AND
FUNCTION DESCRIPTIONS


SECTION OF THIS DATA SHEET.
*COMPLIANT TO JEDEC STANDARDS MO-220-WHHD-5
WITH THE EXCEPTION OF THE EXPOSED PAD DIMENSION.


Figure 51. 32-Lead Lead Frame Chip Scale Package [LFCSP_WQ]
$5 \mathrm{~mm} \times 5 \mathrm{~mm}$ Body, Very Very Thin Quad
(CP-32-12)
Dimensions shown in millimeters

ORDERING GUIDE

| Model ${ }^{1,2}$ | Temperature Range | Package Description | Package Option |
| :--- | :--- | :--- | :--- |
| ADV7182WBCPZ | $-40^{\circ} \mathrm{C}$ to $+105^{\circ} \mathrm{C}$ | 32 -Lead Lead Frame Chip Scale Package [LFCSP_WQ] | CP-32-12 |
| ADV7182WBCPZ-RL | $-40^{\circ} \mathrm{C}$ to $+105^{\circ} \mathrm{C}$ | 32 -Lead Lead Frame Chip Scale Package [LFCSP_WQ] |  |
| ADV7182BCPZ | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 32 -Lead Lead Frame Chip Scale Package [LFCSP_WQ] | CP-12 |
| ADV7182BCPZ-RL | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 32-Lead Lead Frame Chip Scale Package [LFCSP_WQ] | Evaluation Board for the 32-Lead LFCSP |
| EVAL-ADV7182EBZ |  | CP-32-12 |  |

${ }^{1} \mathrm{Z}=$ RoHS Compliant Part.
${ }^{2} \mathrm{~W}=$ Qualified for Automotive Applications.

## AUTOMOTIVE PRODUCTS

The ADV7182W models are available with controlled manufacturing to support the quality and reliability requirements of automotive applications. Note that these automotive models may have specifications that differ from the commercial models; therefore, designers should review the Specifications section of this data sheet carefully. Only the automotive grade products shown are available for use in automotive applications. Contact your local Analog Devices account representative for specific product ordering information and to obtain the specific Automotive Reliability reports for these models.


[^0]:    ${ }^{1}$ These specifications apply for all CVBS input types (NTSC, PAL, SECAM) as well as for single-ended and differential CVBS inputs.
    ${ }^{2}$ The common-mode rejection (CMR) of this circuit design is critically dependent on the external resistor matching on its inputs. This measurement was performed with $0.1 \%$ tolerant resistors, a common-mode voltage of 1 V , and a common-mode frequency of 10 kHz .
    ${ }^{3}$ This is the time that it takes the ADV7182 to detect which video format is present at its input, for example, PAL I or NTSC M.
    ${ }^{4}$ This is the time that it takes the ADV7182 to switch from one (single-ended or differential) analog input to another, for example, switching from Ais 1 to Ais2.

[^1]:    ${ }^{1}$ The first four UDWs are always the ID.

[^2]:    ${ }^{1} \mathrm{x}$ in a reset value indicates do not care.
    ${ }^{2} B$ at the end of the bit name equals an overbar for the whole bit name.

[^3]:    ${ }^{1} x$ indicates a bit that keeps the last written value.

[^4]:    ${ }^{1} x$ indicates a bit that keeps the last written value.

