

#### General Description

The MAX512/MAX513 contain three 8-bit, voltage-output digital-to-analog converters (DAC A, DAC B, and DAC C). Output buffer amplifiers for DACs A and B provide voltage outputs while reducing external component count. The output buffer for DAC A can source or sink 5mA to within 0.5V of V $_{\rm DD}$  or V $_{\rm SS}$ . The buffer for DAC B can source or sink 0.5mA to within 0.5V of V $_{\rm DD}$  or V $_{\rm SS}$ . DAC C is unbuffered, providing a third voltage output with increased accuracy. The MAX512 operates with a single +5V  $\pm 10\%$  supply, and the MAX513 operates with a +2.7V to +3.6V supply. Both devices can also operate with split supplies.

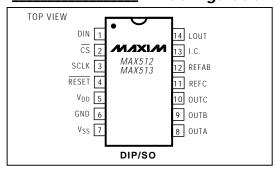
The 3-wire serial interface has a maximum operating frequency of 5MHz and is compatible with SPI™, QSPI™, and Microwire™. The serial input shift register is 16 bits long and consists of 8 bits of DAC input data and 8 bits for DAC selection and shutdown. DAC registers can be loaded independently or in parallel at the positive edge of  $\overline{\mathbb{CS}}$ . A latched logic output is also available for auxiliary control.

Ulfra-low power consumption and small packages (14-pin DIP/SO) make the MAX512/MAX513 ideal for portable and battery-powered applications. Supply current is only 1mA, dropping to less than 1 $\mu$ A in shutdown. Any of the three DACs can be independently shut down. In shutdown mode, the DAC's R-2R ladder network is disconnected from the reference input, minimizing system power consumption.

#### **Applications**

Digital Gain and Offset Adjustment Programmable Attenuators Programmable Current Sources Programmable Voltage Sources RF Digitally Adjustable Bias Circuits VCO Tuning

#### Pin Configuration



#### Features

- Operate from a Single +5V (MAX512) or +3V (MAX513) Supply, or from Bipolar Supplies
- Low Power Consumption
   1mA Operating Current
   <1µA Shutdown Current</li>
- ♦ Unipolar or Bipolar Outputs
- **♦ 5MHz, 3-Wire Serial Interface**
- ♦ SPI, QSPI, and Microwire Compatible
- **♦ Two Buffered, Bipolar-Output DACs (DACs A/B)**
- ♦ Independently Programmable Shutdown Mode
- ♦ Space-Saving 14-Pin SO/DIP Packages
- ♦ Pin and Software Reset

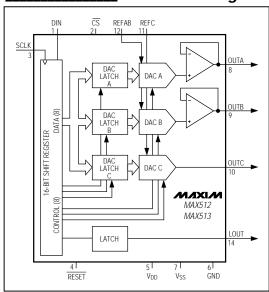
#### Ordering Information

PART	TEMP. RANGE	PIN-PACKAGE
MAX512CPD	0°C to +70°C	14 Plastic DIP
MAX512CSD	0°C to +70°C	14 SO
MAX512C/D	0°C to +70°C	Dice*

Ordering Information continued at end of data sheet.

\* Contact factory for dice specifications.

#### \_Functional Diagram



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#### **ABSOLUTE MAXIMUM RATINGS**

V <sub>DD</sub> to GND	
V <sub>DD</sub> to V <sub>SS</sub>	
Digital Inputs and Outputs to GND	0.3V, (V <sub>DD</sub> + 0.3V)
REFAB	$(V_{SS} - 0.3V), (V_{DD} + 0.3V)$
OUTA, OUTB (Note 1)	V <sub>SS</sub> , V <sub>DD</sub>
OUTC	0.3V, $(V_{DD} + 0.3V)$
REFC	0.3V, (V <sub>DD</sub> + 0.3V)

Continuous Power Dissipation ( $T_A = +70$ °C)
Plastic DIP (derate 10.00mW/°C above +70°C)800mW
SO (derate 8.33mW/°C above +70°C)667mW
CERDIP (derate 9.09mW/°C above +70°C)727mW
Operating Temperature Ranges
MAX51_C0°C to +70°C
MAX51_E40°C to +85°C
MAX51_MJD55°C to +125°C
Storage Temperature Range65°C to +165°C
Lead Temperature (soldering, 10sec)+300°C

Note 1: The outputs may be shorted to V<sub>DD</sub>, V<sub>SS</sub>, or GND if the package power dissipation is not exceeded. Typical short-circuit current to GND is 50mA

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### **ELECTRICAL CHARACTERISTICS**

 $(V_{DD}$  = +4.5V to +5.5V for MAX512,  $V_{DD}$  = +2.7V to +3.6V for MAX513,  $V_{SS}$  = GND = 0V, REFAB = REFC =  $V_{DD}$ ,  $T_A$  =  $T_{MIN}$  to  $T_{MAX}$ , unless otherwise noted. Typical values are at  $T_A$  = +25°C.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	
STATIC PERFORMANCE	1		'				
Resolution	N			Bits			
Differential Nonlinearity	DNL	Guaranteed monotonic			±1	LSB	
Integral Nonlinearity	INI	DAC A/B (Note 2)			±1.5	LSB	
integral Nonlinearity	IINL	DAC C		±1			
Total Unadjusted Error	TUE	(Note 2)		±1		LSB	
Zero-Code Temperature		DAC A/B		100		µV/°С	
Coefficient		DAC C		5			
Dower Cumply Delection Detic	PSRR	MAX512, $4.5V \le V_{DD} \le 5.5V$ , REFAB = REFC = $4.096V$ 0.01				%/%	
Power-Supply Rejection Ratio	PSKK	MAX513, $2.7V \le V_{DD} \le 3.6V$ , REFAB = REFC = $2.4V$		0.015			
REFERENCE INPUTS							
Deference Input Voltage Dange		REFAB	V <sub>SS</sub>	V <sub>SS</sub>		V	
Reference Input Voltage Range		REFC	GND		$V_{\mathrm{DD}}$	] <sup>v</sup>	
Reference Input Capacitance				25		pF	
Reference Input Resistance	R <sub>RFF</sub>	REFAB (Note 3)	8			kΩ	
veierence inhai kezizique	1, KEL	REFC (Note 3) 12				7 152	
Reference Input Resistance (shutdown mode)		REFAB, REFC 2				МΩ	

#### **ELECTRICAL CHARACTERISTICS (continued)**

 $(V_{DD}$  = +4.5V to +5.5V for MAX512,  $V_{DD}$  = +2.7V to +3.6V for MAX513,  $V_{SS}$  = GND = 0V, REFAB = REFC =  $V_{DD}$ ,  $V_{AB}$  =  $V_{MAX}$ , unless otherwise noted. Typical values are at  $V_{AB}$  = +25°C.)

PARAMETER	SYMBOL	COI	NDITIONS	MIN	TYP	MAX	UNITS		
DAC OUTPUTS									
Output Voltage Range				0		REF_	V		
		DAC A		0.10					
Capacitive Load		DAC B		0.01			μF		
		DAC C		0			ı		
		DAC A			0.050				
Output Resistance		DAC B			0.500		kΩ		
		DAC C		24					
DIGITAL INPUTS	•								
Input High Voltage	VIH			(0.7)(V <sub>DE</sub>	))		V		
Input Low Voltage	V <sub>IL</sub>				(	0.3)(V <sub>DD</sub> )	V		
Input Current	I <sub>IN</sub>	$V_{IN} = 0V \text{ or } V_{DD}$			0.1	±10	μΑ		
Input Capacitance	CIN	(Notes 4, 5) 10							
DIGITAL OUTPUT									
Output High Voltage	V <sub>OH</sub>	I <sub>SOURCE</sub> ≤ 1.6mA		V <sub>DD</sub> - 0.4			V		
Output Low Voltage	Vol	ISINK ≤ 1.6mA	0.4	V					
DYNAMIC PERFORMANCE	•			•					
Voltage-Output Slew Rate	SR	C <sub>L</sub> = 0.1µF (DAC A)	), C <sub>L</sub> = 0.01µF (DAC B)		0.1		V/µs		
			$C_L = 0.1 \mu F (DAC A)$		70				
Voltage-Output Settling Time		To ±1/2LSB	$C_L = 0.01 \mu F (DAC B)$	$L = 0.01 \mu F (DAC B)$ 70					
			$C_L = 0.1 nF (DAC C)$		35				
Digital Feedthrough and Crosstalk		All 0s to all 1s	•		10		nV-s		
POWER SUPPLIES									
Positive Supply Voltage Range	V <sub>DD</sub>	MAX512		4.5		5.5	V		
Fositive Supply Voltage Range	V DD	MAX513		2.7		3.6	v		
Negative Supply Voltage Range	Vac	MAX512		-5.5		-4.5	V		
(Note 6)	V <sub>SS</sub>	MAX513		-3.6		-2.7	v		
Positive Supply Current	lon	All inputs = 0V	MAX512 (V <sub>DD</sub> = 5.5V)		1.3	2.8	mA		
rosilive supply culteril	I <sub>DD</sub>	All illbuts = UV	MAX513 (V <sub>DD</sub> = 3.6V)		0.9	2.5			
Negative Supply Current	I <sub>SS</sub>	All inputs = 0V, V <sub>SS</sub>	= -5.5V		-1.3		mA		
Shutdown Supply Current		0.1							

Note 2: Digital code from 24 through 232 are due to swing limitations of output amplifiers on DAC A and DAC B. See *Typical Operating Characteristics*.

**Note 3:** Reference input resistance is code dependent. The lowest input resistance occurs at code 55hex. Refer to the reference input section in the *Detailed Description*.

Note 4: Guaranteed by design. Not production tested.

Note 5: Input capacitance is code dependent. The highest capacitance occurs at code 00hex.

Note 6: For single-supply mode, tie  $V_{SS}$  to GND.

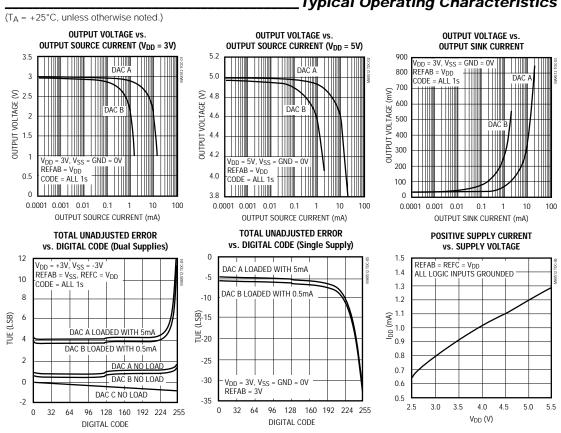
#### **TIMING CHARACTERISTICS** (Note 4)

 $(V_{DD} = +4.5V \text{ to } +5.5V \text{ for MAX512}, V_{DD} = +2.7V \text{ to } +3.6V \text{ for MAX513}, V_{SS} = GND = 0V, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted.})$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
SERIAL INTERFACE TIMING	•		•			
CS Fall to SCLK Rise Setup Time	tcss		150			ns
SCLK Rise to CS Rise Setup Time	tсsн		150			ns
DIN to SCLK Rise Setup Time	t <sub>DS</sub>		50			ns
DIN to SCLK Rise Hold Time	tDH		50			ns
SCLK Pulse Width High	tcH		100			ns
SCLK Pulse Width Low	tCL		100			ns
Output Delay LOUT	t <sub>OD</sub>	C <sub>L</sub> = 100pF			150	ns
CS Pulse Width High	tcspwh		200			ns

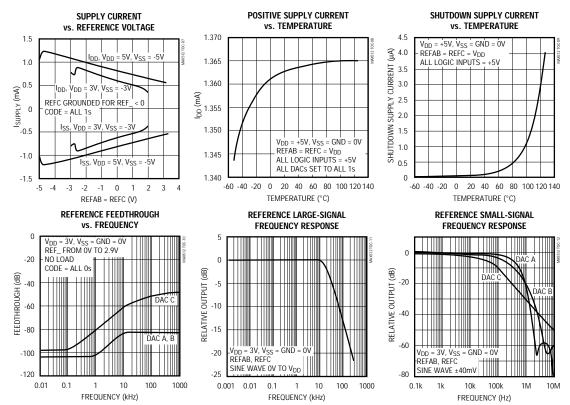
Note 4: Guaranteed by design. Not production tested

#### **Typical Operating Characteristics**



Typical Operating Characteristics (continued)

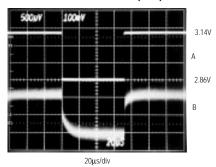
 $(T_A = +25^{\circ}C, \text{ unless otherwise noted.})$ 



\_Typical Operating Characteristics (continued)

 $(T_A = +25^{\circ}C, unless otherwise noted.)$ 

#### LINE-TRANSIENT RESPONSE (OUTA)

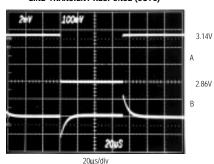


REFAB = 2.56V, NO LOAD, CODE = ALL 1s

 $A:\ V_{DD,}\ 100mV/div$ 

B: OUTA, 500μV/div

#### LINE-TRANSIENT RESPONSE (OUTC)

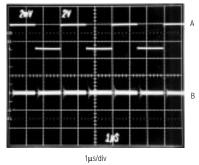


REFC = 2.56V, NO LOAD, CODE = ALL 1s

 $A:\ V_{DD},\ 100mV/div$ 

B: OUTC, 2mV/div

#### CLOCK FEEDTHROUGH (OUTA)

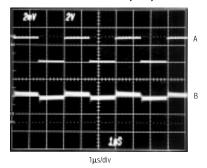


 $V_{SS} = 0V$ ,  $\overline{CS} = HIGH$ 

A: SCLK, 333kHz, 0V TO 2.9V, 2V/div

B: OUTA, 2mV/div

#### CLOCK FEEDTHROUGH (OUTC)



 $V_{SS} = 0V, \overline{CS} = HIGH$ 

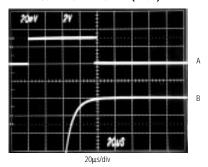
A: SCLK, 333kHz, 0V TO 2.9V, 2V/div

B: OUTC, 2mV/div

\_Typical Operating Characteristics (continued)

 $(T_A = +25^{\circ}C, unless otherwise noted.)$ 

#### POSITIVE SETTLING TIME (DAC A)

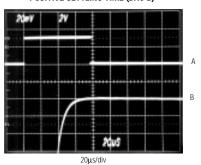


 $V_{DD}$  = 3V,  $V_{SS}$  = 0V, REFAB =  $V_{DD,}$   $R_L$  = 1k  $\Omega,$   $C_L$  = 0.1  $\mu F$  ALL BITS OFF TO ALL BITS ON

A: CS, 2V/div

B: OUTA, 20mV/div

#### POSITIVE SETTLING TIME (DAC B)

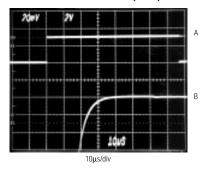


 $V_{DD}$  = 3V,  $V_{SS}$  = 0V, REFAB =  $V_{DD,}$   $R_L$  = 10k  $\Omega,$   $C_L$  =  $0.01 \mu F$  ALL BITS OFF TO ALL BITS ON

A: CS, 2V/div

B: OUTB, 20mV/div

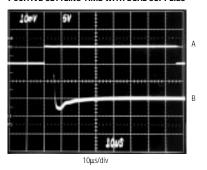
#### POSITIVE SETTLING TIME (DAC C)



 $V_{DD}$  = 3V,  $V_{SS}$  = 0V, REFC =  $V_{DD},\,R_L$  =  $\,$   $\,$   $\,$   $\,$  C  $_L$  = 122pF ALL BITS OFF TO ALL BITS ON

A: CS, 2V/div B: OUTC, 20mV/div

#### POSITIVE SETTLING TIME WITH DUAL SUPPLIES



 $V_{DD}$  = 5V,  $V_{SS}$  = -5V, REFAB = 2.56V,  $R_L$  = 1k  $\Omega,\, C_L$  = 0.1  $\mu F$  ALL BITS OFF TO ALL BITS ON

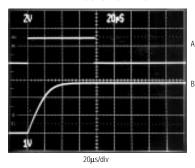
A: CS, 5V/div

B: OUTA, 10mV/div

## \_Typical Operating Characteristics (continued)

 $(T_A = +25^{\circ}C, \text{ unless otherwise noted.})$ 

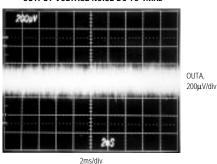
#### TIME EXITING SHUTDOWN MODE



 $V_{DD}$  = 3V,  $V_{SS}$  = 0V, REFAB =  $V_{DD,}$   $R_L$  = 1k  $\Omega,$   $C_L$  = 0.1  $\mu F$  DAC LOADED WITH ALL 1s

A: CS, 2V/div B: OUTA, 1V/div

#### **OUTPUT VOLTAGE NOISE DC TO 1MHz**



 $\mathsf{DIGITAL}\ \mathsf{CODE} = \mathsf{80},\ \mathsf{REFAB} = \mathsf{V}_{\mathsf{DD}},\ \mathsf{NO}\ \mathsf{LOAD}$ 

## \_Pin Description

PIN	NAME	FUNCTION
1	DIN	Serial Data Input of the 16-bit shift register. Data is clocked into the register on the rising edge of SCLK.
2	CS	Chip Select (active low). Enables data to be shifted into the 16-bit shift register. Programming commands are executed at the rising edge of CS.
3	SCLK	Serial Clock Input. Data is clocked in on the rising edge of SCLK.
4	RESET	Asynchronous reset input (active low). Clears all registers to their default state (FFhex for DAC A and DAC B registers); all other registers are reset to 0 (including the input shift register).
5	V <sub>DD</sub>	Positive Power Supply (2.7V to 5.5V). Bypass with 0.22µF to GND.
6	GND	Ground
7	V <sub>SS</sub>	Negative Power Supply 0V or (-1.5V to -5.5V). Tie to GND for single supply operation. If a negative supply is applied, bypass with $0.22\mu F$ to GND.
8	OUTA	DAC A Output Voltage (Buffered). Resets to full scale. Connect 0.1µF capacitor or greater to GND.
9	OUTB	DAC B Output Voltage (Buffered). Resets to full scale. Connect 0.01µF capacitor or greater to GND.
10	OUTC	DAC C Output Voltage (Unbuffered). Resets to zero.
11	REFC	DAC C Reference Voltage
12	REFAB	DAC A/B Reference Voltage
13	I.C.	Internally connected. Do not make connections to this pin.
14	LOUT	Logic Output (latched)

#### Detailed Description

#### **Analog Section**

The MAX512/MAX513 contain three 8-bit, voltage-out-put, digital-to-analog converters (DACs). The DACs are "inverted" R-2R ladder networks using complementary switches that convert 8-bit digital inputs into equivalent analog output voltages in proportion to the applied reference voltages.

The MAX512/MAX513 have two reference inputs: one is shared by DAC A and DAC B and the other is used by DAC C. These inputs allow different full-scale output voltages and different output voltage polarities for the DAC pair A/B and DAC C.

The MAX512/MAX513 include output buffer amplifiers for DACs A and B and input logic for simple microprocessor (µP) and CMOS interfaces.

The MAX512/MAX513 operate in either single-supply or dual-supply mode, as determined by  $V_{SS}$ . If  $V_{SS}$  is within approximately -0.5V of GND, single-supply mode is assumed. If  $V_{SS}$  is below -1.5V, the devices are in dual-supply mode.

#### Reference Inputs and DAC Output Range

The voltage at REF\_ sets the full-scale output of the DACs. The input impedance of the REF\_ inputs is code dependent. The lowest value, approximately  $12k\Omega$  for REFC (8k $\Omega$  for REFAB), occurs when the input code is 01010101 (55hex). The maximum value of infinity occurs when the input code is zero.

In shutdown mode, the selected DAC output is set to zero while the value stored in the DAC register remains unchanged. This removes the load from the reference input to save power. Bringing the MAX512/MAX513 out of shutdown mode restores the DAC output voltage. Because the input resistance at REF\_ is code dependent, the DAC's reference sources should have an output impedance of no more than  $5\Omega$ . The input capacitance at the REF\_ pins is also code dependent and typically does not exceed 25pF.

The reference voltage on REFAB can range anywhere between the supply rails. In dual-supply mode, a positive reference input voltage on REFAB should be less than ( $V_{DD}$  - 1.5V) to avoid saturating the buffer amplifiers. The reference voltage includes the negative supply rail. See the *Output Buffer Amplifier* section for more information. The REFC input accepts positive voltages up to  $V_{DD}$  and should not be forced below ground.

The absolute difference between any reference voltage and GND should not exceed 6V.

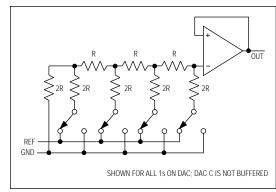


Figure 1. DAC Simplified Circuit Diagram

#### Output Buffer Amplifiers (DAC A / DAC B)

DAC A and DAC B voltage outputs are internally buffered. The buffer amplifiers have a rail-to-rail ( $V_{SS}$  to  $V_{DD}$ ) output voltage range.

In single-supply mode, the DAC outputs A and B are internally divided by two and the buffer is set to a gain of two, eliminating the need for a buffer input voltage range to the positive supply rail.

In dual-supply mode, the DAC outputs are not attenuated and the buffer is set to unity gain.

Although only necessary for negative output voltages, the dual-supply mode may be used even if the desired DAC output voltage is positive. Possible errors associated with the divide-by-two attenuator and gain-of-two buffers in single-supply mode are eliminated in dual-supply mode. In this case, do not use reference voltages higher than  $(V_{\mbox{\scriptsize DD}}$  - 1.5V).

DAC A's output amplifier can source and sink up to 5mA of current (0.5mA for DAC B buffer). See the Total Unadjusted Error vs. Digital Code graph in the *Typical Operating Characteristics* for dual and single supplies. The amplifier is unity-gain stable with a capacitive load of 0.05µF (0.01µF for DAC B buffer) or greater. The slew rate is limited by the load capacitor and is typically 0.1V/µs with a 0.1µF load (0.01µF for DAC B buffer).

#### Unbuffered Output (DAC C)

The output of DAC C is unbuffered and has a typical output impedance of  $24 k\Omega$ . It can be used to drive a high-impedance load, such as an op amp or comparator, and has 35µs typical settling time to 1/2LSB with a single 3V supply. Use DAC C if a quick dynamic response is required.

#### Shutdown Mode

When programmed to shutdown mode, the outputs of DAC A and B go into a high-impedance state. Virtually no current flows into or out of the buffer amplifiers in that state. The output of DAC C goes to 0V when shut down. In shutdown mode, the REF\_ inputs are high impedance (2M $\Omega$  typ) to conserve current drain from the system reference; therefore, the system reference does not have to be powered down. The logic output LOUT remains active in shutdown.

Coming out of shutdown, the DAC outputs return to the values kept in the registers. The recovery time is equivalent to the DAC settling time.

#### Reset

The RESET input is active low. When asserted (RESET = 0), DACs A and B are set to full scale (FFhex) and active, while DAC C is set to zero code (00hex) and active. The 16-bit serial register is cleared to 0000hex. LOUT is reset to zero.

#### Serial Interface

An active-low chip select  $\overline{(CS)}$  enables the shift register to receive data from the serial data input. Data is clocked into the shift register on every rising edge of the serial clock signal (SCLK). The clock frequency can be as high as 5MHz.

Data is sent MSB first and can be transmitted in one 16-bit word. The write cycle can be interrupted at any time when  $\overline{\text{CS}}$  is kept active (low) to allow, for example, two 8-bit-wide transfers. After clocking all 16 bits into

Table 1. Input Shift Register

	-	•						
	B0*	DAC Data Bit 0 (LSB)						
	B1	DAC Data Bit 1						
S	B2	DAC Data Bit 2						
DATA BITS	B3	DAC Data Bit 3						
ATA	B4	DAC Data Bit 4						
ď	B5	DAC Data Bit 5						
	B6	DAC Data Bit 6						
	B7	DAC Data Bit 7 (MSB)						
	LA	Load Reg DAC A, Active High						
()	LB	Load Reg DAC B, Active High						
BITS	LC	Load Reg DAC C, Active High						
0	SA	Shut Down DAC A, Active High						
TR	SB	Shut Down DAC B, Active High						
CONTROL BITS	SC	Shut Down DAC C, Active High						
	Q1	Logic Output						
	Q2**	Uncommitted Bit						

Clocked in last.

the input shift register, the rising edge of  $\overline{CS}$  updates the DAC outputs, the shutdown status, and the status of the logic output. Because of their single buffered structure, DACs cannot be simultaneously updated to different digital values.

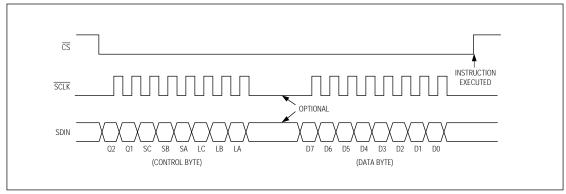


Figure 2. MAX512/MAX513 3-Wire Serial-Interface Timing Diagram

<sup>\*\*</sup>Clocked in first.

**Table 2. Serial-Interface Programming Commands** 

	CONTROL										DA	TA			FUNCTION			
								MSB LSB										
Q2	Q1	sc	SB	SA	LC	LB	LA	В7	В6	B5	B4	В3	B2	B1	B0			
*	*	*	*	*	0	0	0	Х	Х	Х	Х	Х	Х	Х	Х	No Operation to DAC Registers		
*	*	*	*	*	1	0	0			8-	Bit DA	AC Da	ita			Load Register to DAC C		
*	*	*	*	*	0	1	0			8-	Bit DA	AC Da	ıta			Load Register to DAC B		
*	*	*	*	*	0	0	1			8-	Bit DA	AC Da	ıta		Load Register to DAC A			
*	*	*	*	*	1	1	1			8-	Bit DA	AC Da	ıta			Load All DAC Registers		
*	*	0	0	0	*	*	*	Х	Х	Х	Х	Х	Х	Х	Х	All DACs Active		
*	*	1	0	0	*	*	*	Х	Х	Х	Х	Х	Х	Х	Х	Shut Down DAC C		
*	*	0	1	0	*	*	*	Х	Х	Х	Х	Х	Х	Х	Х	Shut Down DAC B		
*	*	0	0	1	*	*	*	Х	Х	Х	Х	Х	Х	Х	Х	Shut Down DAC A		
*	*	1	1	1	*	*	*	Х	Х	Х	Х	Х	Х	Х	Х	Shut Down All DACs		
Х	0	*	*	*	*	*	*	Х	Х	Χ	Х	Χ	Χ	Χ	Χ	Reset LOUT		
Х	1	*	*	*	*	*	*	Х	Х	Х	Х	Х	Х	Х	Х	Set LOUT		

X Don't care.

#### Serial-Input Data Format and Control Codes

Table 2 lists the serial-input data format. The 16-bit input word consists of an 8-bit control byte and an 8-bit data byte. The 8-bit control byte is not decoded internally. Every control bit performs one function. Data is clocked in starting with Q2 (uncommitted bit), followed by the remaining control bits and the data byte. The LSB of the data byte (B0) is the last bit clocked into the shift register (Figure 2).

Example of a 16-bit input word:

	ade First												_	oac	
Q2	Q1	sc	SB	SA	LC	LB	LA	В7	В6	В5	В4	ВЗ	B2	В1	В0
Х	0	1	0	0	0	1	1	1	0	0	0	0	0	0	0

The example above performs the following functions:

- 80hex (128 decimal) loaded into DAC registers A and B.
- · Content of the DAC C register remains unchanged.
- · DAC A and DAC B are active.
- · DAC C is shut down.
- · LOUT is reset to 0.

#### **Digital Inputs**

The digital inputs are compatible with CMOS logic. Supply current increases slightly when toggling the logic inputs through the transition zone between  $(0.3)(V_{DD})$  and  $(0.7)(V_{DD})$ .

#### **Digital Output**

The latched digital output (LOUT) has a 1.6mA source capability while maintaining a (V<sub>DD</sub> - 0.4V) output level. With a 1.6mA sink current, the output voltage is guaranteed to be no more than 0.4V. The output can be used for digital auxiliary control. Please note that the digital output remains fully active during shutdown mode.

#### Microprocessor Interfacing

The MAX512/MAX513 serial interface is compatible with Microwire, SPI, and QSPI. For SPI and QSPI, clear the CPOL and CPHA bits (CPOL = 0 and CPHA = 0). CPOL = 0 sets the inactive state of clock to zero and CPHA = 0 changes data at the falling edge of SCLK. This setting allows both SPI and QSPI to run at full clock speeds (0.5MHz and 4MHz, respectively). If a serial port is not available on your  $\mu P$ , three bits of a parallel port can be used to emulate a serial port by bit manipulation. Minimize digital feedthrough at the voltage outputs by operating the serial clock only when necessary.

<sup>\*</sup> Not shown for clarity. The functions of loading and shutting down the DACs and programming the logic can be combined in a single command.

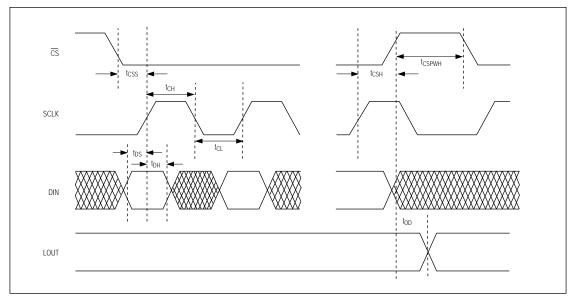


Figure 3. MAX512/MAX513 Detailed Serial-Interface Timing Diagram

#### **Applications Information**

#### Power-Supply and Reference Operating Ranges

The MAX512 is fully specified to operate with  $V_{DD}=5V\pm10\%$  and  $V_{SS}=GND=0V$ . The MAX513 is specified for single-supply operation with  $V_{DD}$  ranging from 2.7V to 3.6V, covering all commonly used supply voltages in 3V systems. The MAX512/MAX513 can also be used with a negative supply ranging from -1.5V to -5.5V. Using a negative supply typically improves zero-code error and settling time (as shown in the *Typical Operating Characteristics* graphs).

The two separate reference inputs for the DAC pair A/B and the unbuffered output C allow different full-scale output voltages and, if a negative supply is used, also allow different polarity. In dual-supply mode, REFAB can vary from  $V_{SS}$  to  $(V_{DD}$  - 1.5V). In single-supply mode, the specified range for REFAB is 0V to  $V_{DD}.$  REFC can range from GND to  $V_{DD}.$  Do not force REFC below ground.

Power-supply sequencing is not critical. If a negative supply is used, make sure  $V_{SS}$  is never more than 0.3V above ground. Do not apply signals to the digital inputs until the device is powered-up. If this is not possible, add current-limiting resistors to the digital inputs.

#### Power-Supply Bypassing and Ground Management

In single-supply operation (V<sub>SS</sub> = GND), GND and V<sub>SS</sub> should be connected to the highest quality ground available. Bypass V<sub>DD</sub> with a 0.1µF to 0.22µF capacitor to GND. For dual-supply operation, bypass V<sub>SS</sub> with a 0.1µF to 0.22µF capacitor to GND. Reference inputs can be used without bypassing. For optimum line/load-transient response and noise performance, bypass the reference inputs with 0.1µF to 4.7µF to GND. Careful PC board layout minimizes crosstalk among DAC outputs, reference inputs, and digital inputs. Separate analog lines with ground traces between them. Make sure that high-frequency digital lines are not routed in parallel to analog lines.

#### **Unipolar Output**

With unipolar output, the output voltage and the reference voltage are the same polarity. The MAX512/MAX513 can be used with a single supply if the reference voltages are positive. With a negative supply, the REFAB voltage can vary from  $\rm V_{SS}$  to approximately ( $\rm V_{DD}$  - 1.5V), allowing two-quadrant multiplication.

**Table 3. Unipolar Code Table** 

	ı	DAC	СО	NTE	NTS	3		ANALOG
В7	В6	В5	В4	ВЗ	B2	В1	В0	OUTPUT
1	1	1	1	1	1	1	1	$+REF_{-} \times \left(\frac{255}{256}\right)$
1	0	0	0	0	0	0	1	$+REF_{-} \times \left(\frac{129}{256}\right)$
1	0	0	0	0	0	0	0	$+REF_{-} \times \left(\frac{128}{256}\right) = +\frac{REF_{-}}{2}$
0	1	1	1	1	1	1	1	$+REF_{-} \times \left(\frac{127}{256}\right)$
0	0	0	0	0	0	0	1	$+REF_{-} \times \left(\frac{1}{256}\right)$
0	0	0	0	0	0	0	0	OV

Note:

1LSB = REF\_ 
$$\times$$
 2<sup>-8</sup> = REF\_  $\times$   $\left(\frac{1}{256}\right)$   
ANALOG OUTPUT = REF\_  $\times$   $\left(\frac{D}{256}\right)$ 

#### Bipolar Output

Using Figure 4's circuit, the MAX512/MAX13 can be configured for bipolar outputs. Table 4 lists the bipolar codes and corresponding output voltages. There are two ways to achieve rail-to-rail outputs: 1) Operate the MAX512/MAX513 with a single supply and positive reference voltages or 2) Use dual supplies with a positive or negative voltage at REFAB and a positive voltage at REFC. In either case, the op amps need dual supplies. When using the dual-supply mode, possible errors associated with the divide-by-two attenuator and gain-of-two buffer are eliminated (see the *Output Buffer Amplifier* section). For maximum output swing of all outputs in idual-supply mode, connect REFAB to V<sub>SS</sub> and REFC to V<sub>DD</sub>. In single-supply mode, connect REFAB, REFC, and V<sub>DD</sub> together.

With dual supplies, DACs A and B can perform four-quadrant multiplication. Please note that in dual-supply mode, the REFAB input ranges from  $V_{SS}$  to ( $V_{DD}$  - 1.5V). Because REFC accepts only positive inputs, DAC C performs two-quadrant multiplication.

Figure 4 shows Maxim's ICL7612A with rail-to-rail input common-mode range and rail-to-rail output voltage swing—ideal for a high output voltage swing from low supply voltages.

**Table 4. Bipolar Code Table** 

	ı	DAC	СО	NTE	NTS	3		ANALOG
В7	В6	В5	В4	ВЗ	B2	В1	В0	OUTPUT
1	1	1	1	1	1	1	1	$+REF_{-} \times \left(\frac{127}{128}\right)$
1	0	0	0	0	0	0	1	$+REF_{-} \times \left(\frac{1}{128}\right)$
1	0	0	0	0	0	0	0	OV
0	1	1	1	1	1	1	1	$-REF_{-} \times \left(\frac{1}{128}\right)$
0	0	0	0	0	0	0	1	$-REF_{-} \times \left(\frac{127}{128}\right)$
0	0	0	0	0	0	0	0	$-REF_{-} \times \left(\frac{128}{128}\right) = -REF_{-}$

Note:

1LSB = REF\_ 
$$\times$$
 2<sup>-(8 · 1)</sup> = REF\_  $\times$   $\left(\frac{1}{128}\right)$   
ANALOG OUTPUT = REF\_  $\times$   $\left(\frac{D}{128} - 1\right)$ 

#### **RF Applications**

Both the MAX512 and MAX513 can bias GaAs FETs, where the gate of the FETs must be negatively biased to ensure that there is no drain current. In a typical application, power to the RF amplifiers should not be turned on until the bias voltages provided by DAC A and DAC B are fully established; likewise, the supply should be turned off before the bias voltage is switched off. Figure 5 shows how DAC B supplies the negative bias  $V_{\rm GG1}$  for the driver stage and DAC A provides the negative bias  $V_{\rm GG2}$  for the output stage [1].

The DAC A and DAC B outputs are also ideal for controlling VCOs in mobile radios or cellular phones. Other applications include varactor and PIN diode circuits.

The unbuffered DAC C provides a span within GND and  $V_{DD}$  and is individually set at REF C. DAC C typically adjusts offset and gain in the system.

<sup>1 [</sup>John Wachsmann. \*A High-Efficiency GaAs MMIC Power Amplifier for 1.9GHz PCS Applications," Proceedings of the First Annual Wireless Symposium, pp. 375, Penton Publishing, Jan. 1993.]

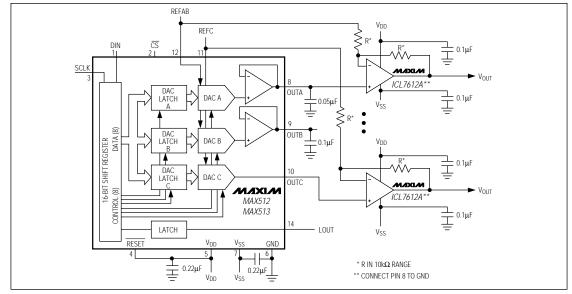


Figure 4. Bipolar Output Circuit

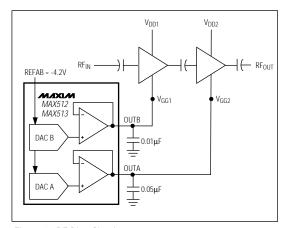


Figure 5. RF Bias Circuit

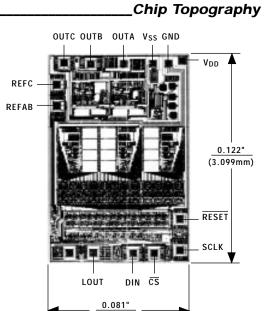
# MAX512/MAX513

# Low-Cost, Triple, 8-Bit Voltage-Output DACs with Serial Interface

## \_Ordering Information (continued)

PART	TEMP. RANGE	PIN-PACKAGE
MAX512EPD	-40°C to +85°C	14 Plastic DIP
MAX512ESD	-40°C to +85°C	14 SO
MAX512MJD	-55°C to +125°C	14 CERDIP
MAX513CPD	0°C to +70°C	14 Plastic DIP
MAX513CSD	0°C to +70°C	14 SO
MAX513C/D	0°C to +70°C	Dice*
MAX513EPD	-40°C to +85°C	14 Plastic DIP
MAX513ESD	-40°C to +85°C	14 SO
MAX513MJD	-55°C to +125°C	14 CERDIP

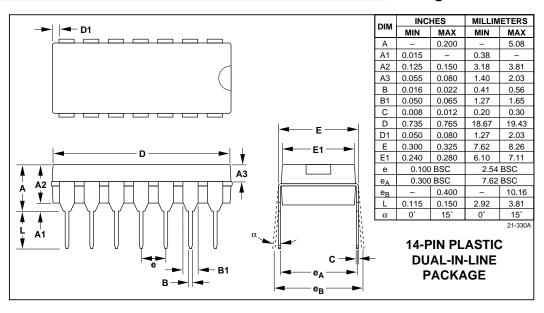
<sup>\*</sup> Contact factory for dice specifications.

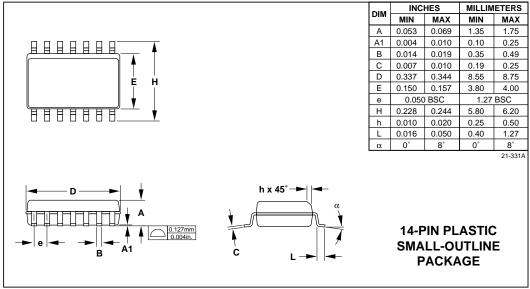


TRANSISTOR COUNT: 1910
SUBSTRATE CONNECTED TO V<sub>DD</sub>

(2.057mm)

#### \_Package Information





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