

TPS54202HEVM-716 2-A Regulator Evaluation Module

This user's guide contains background information for the TPS54202H as well as support documentation for the TPS54202HEVM-716 evaluation module (PWR716-001). Also included are the performance specifications, the schematic, and the bill of materials for the TPS54202HEVM-716.

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1 Introduction

This user's guide contains background information for the TPS54202H as well as support documentation for the TPS54202HEVM-716 evaluation module (PWR716-001). Also included are the performance specifications, the schematic, and the bill of materials for the TPS54202HEVM-716.

1.1 Background

The TPS54202H dc/dc converter is designed to provide up to a 2-A output from an input voltage source of 4.2 V to 28 V. Rated input voltage and output current range for the evaluation module are given in Table 1. This evaluation module is designed to demonstrate the small printed-circuit-board areas that may be achieved when designing with the TPS54202H regulator. The switching frequency is internally set at a nominal 500 kHz. The high-side and low-side MOSFETs are incorporated inside the TPS54202H package along with the gate-drive circuitry. The low drain-to-source on resistance of the MOSFETs allow the TPS54202H to achieve high efficiencies and helps keep the junction temperature low at high output currents. The compensation components are integrated to the integrated circuit (IC), and an external divider allows for an adjustable output voltage. Additionally, the TPS54202H provides an adjustable undervoltage lockout input. The absolute maximum input voltage is 30 V for the TPS54202HEVM-716.

| EVM | Input Voltage Range | Output Current Range | | |
|------------------|--------------------------|----------------------|--|--|
| TPS54202HEVM-716 | $V_{IN} = 4.2 V$ to 28 V | 0 A to 2 A | | |

1.2 Performance Specification Summary

A summary of the TPS54202HEVM-716 performance specifications is provided in Table 2. Specifications are given for an input voltage of $V_{IN} = 24$ V and an output voltage of 5.0 V, unless otherwise specified. The TPS54202HEVM-716 is designed and tested for $V_{IN} = 4.2$ V to 28 V. The ambient temperature is 25°C for all measurements, unless otherwise noted.

| Specification | Test Conditions | | MIN | TYP | MAX | Unit |
|---|---|--|-----|--------|-----|-----------|
| V _{IN} operating voltage range | | | 4.2 | 24 | 28 | V |
| V _{IN} start voltage | | | | 4.2 | | V |
| V _{IN} stop voltage | | | | 3.7 | | V |
| Output voltage set point | | | | 5 | | V |
| Output current range | V _{IN} = 4.2 V to 28 V | | 0 | | 2 | А |
| Line regulation | $I_0 = 1 \text{ A}, V_{IN} = 8 \text{ V to}$ | $I_0 = 1 \text{ A}, V_{IN} = 8 \text{ V to } 28 \text{ V}$ | | ±0.5% | | |
| Load regulation | $V_{IN} = 12 \text{ V}, I_{O} = 0 \text{ A te}$ | $V_{IN} = 12 V$, $I_{O} = 0 A$ to 2 A | | ±0.5% | | |
| | $I_{\rm O} = 0.5 \text{ A to } 1.5 \text{ A}$ | Voltage change | | -150 | | mV |
| Lood transient reasons | | Recovery time | | 150 | | μs |
| Load transient response | | Voltage change | | 150 | | mV |
| | I ₀ = 1.5 A to 0.5 A | Recovery time | | 150 | | μs |
| Input ripple voltage | I ₀ = 2 A | | | 400 | | mV_{PP} |
| Output ripple voltage | I ₀ = 2 A | I ₀ = 2 A | | <30 | | mV_{PP} |
| Output rise time | time | | | 5 | | ms |
| Center operating frequency | | | | 500 | | kHz |
| Maximum Efficiency TPS54202HEVM-716, V _{IN} = 12 | | 6, V _{IN} = 12 V, I _O = 1 A | | 94.06% | | |

Table 2. TPS54202HEVM-716 Performance Specification Summary



1.3 Modifications

These evaluation modules are designed to provide access to the features of the TPS54202H. Some modifications can be made to this module.

1.3.1 Output Voltage Set Point

The voltage divider, R2 and R3, is used to set the output voltage. To change the output voltage of the EVM, it is necessary to change the value of resistor R3. Changing the value of R3 can change the output voltage above 0.596 V. The value of R3 for a specific output voltage can be calculated using Equation 1. Use 100 k Ω for R2.

$$R3 = \frac{R2 \times 0.596 \text{ V}}{V_{\text{OUT}} - 0.596 \text{ V}}$$

(1)

Table 3 lists the R2 and R3 values for some common output voltages. Note that V_{IN} must be in a range so that the minimum on-time is greater than 150 ns. The values in Table 3 are standard values, not the exact value calculated using Equation 1.

| V _{OUT} (V) | L (µH) | C _{ουτ} (μF) | R2 (kΩ) | R3 (kΩ) | C8 (pF) |
|----------------------|--------|-----------------------|---------|---------|---------|
| 1.8 | 5.6 | 66 | 100 | 49.9 | 47 |
| 2.5 | 8.2 | 44 | 100 | 31.6 | 33 |
| 3.3 | 10 | 44 | 100 | 22.1 | 56 |
| 5 | 15 | 44 | 100 | 13.3 | 75 |
| 12 | 22 | 44 | 100 | 5.23 | 100 |

Table 3. Recommended Component Values

1.3.2 Output Capacitor and Feed-Forward Capacitor

Considering the loop stability and the effect of the internal parasitic parameters, choose a crossover frequency less than 40 kHz, without considering the feed-forward capacitor. A simple estimation for the crossover frequency without feed-forward capacitor C8 is shown in Equation 2, assuming C_{OUT} has small ESR.

$$f_{\rm o} = \frac{3.95}{V_{\rm OUT} \times C_{\rm OUT}}$$

(2)

(3)

3

Depending on V_{OUT} , if the output capacitor, C_{OUT} , is dominated by low-ESR (ceramic types) capacitors, a low phase margin could result. To improve the phase boost, an external feed-forward capacitor, C8, can be added in parallel with R2. C8 is chosen such that phase margin is boosted at the crossover frequency.

C8 is calculated in Equation 3:

$$C8 = \frac{1}{2\pi f_0} \times \frac{1}{R_2}$$

For this design, C8 = 75 pF. C8 is not needed when C_{OUT} has high ESR, and C8 calculated from Equation 3 should be reduced with medium ESR. Use Table 3 as a starting point.



1.3.3 ENABLE

There is an internal 1-M Ω resistor from EN to GND thus EN can be floated to shut down the chip. To enable the chip, a pull-up resistor (range 200 k Ω to 1 M Ω , typical 510 k Ω) is connected between VIN and EN to limit the EN input current for light load efficiency improvement.

2 Test Setup and Results

This section describes how to properly connect, set up, and use the TPS54202HEVM-716 evaluation module. The section also includes test results typical for the evaluation module and covers efficiency, output voltage regulation, load transients, loop response, output ripple, input ripple, and start-up.

2.1 Input/Output Connections

The TPS54202HEVM-716 is provided with input/output connectors and test points as shown in Table 4. A power supply capable of supplying 2 A must be connected to J1 through a pair of 20-AWG wires. The load must be connected to J2 through a pair of 20-AWG wires. The maximum load current capability must be at least 3 A to use the full capability of this EVM. Wire lengths must be minimized to reduce losses in the wires. Test-point TP1 provides a place to monitor the V_{IN} input voltages with TP2 providing a convenient ground reference. TP6 is used to monitor the output voltage with TP7 as the ground reference.

| Reference Designator | Function |
|----------------------|---|
| J1 | VIN (see Table 1 for V _{IN} range) |
| J2 | VOUT, 5 V at 2 A maximum |
| JP1 | 2-pin header for enable. Connect EN to ground to disable, open to enable. |
| TP1 | V _{IN} test point at VIN connector |
| TP2 | GND test point at VIN |
| TP3 | GND test point |
| TP4 | SW test point |
| TP5 | Test point between voltage divider network and output. Used for loop response measurements. |
| TP6 | Output voltage test point at OUT connector |
| TP7 | GND test point at VOUT connector |

Table 4. EVM Connectors and Test Points



2.2 Efficiency

The efficiency of this EVM peaks at a load current of about 0.5 A - 1 A, and then decreases as the load current increases towards full load. Figure 1 shows the efficiency for the TPS54202HEVM-716 at an ambient temperature of 25°C.

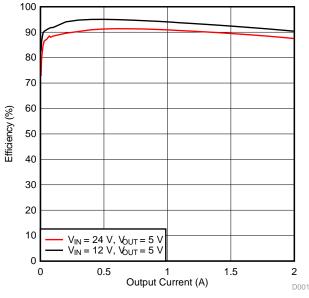


Figure 1. TPS54202HEVM-716 Efficiency

Figure 2 shows the efficiency for the TPS54202HEVM-716 on a semi-log scale to better show light load efficiency. The ambient temperature is 25°C.

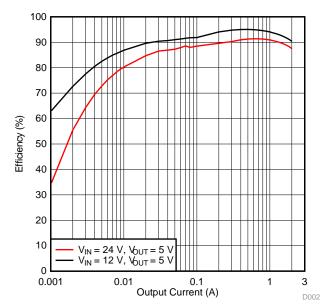


Figure 2. TPS54202HEVM-716 Low Current Efficiency

The efficiency may be lower at higher ambient temperatures, due to temperature variation in the drain-tosource resistance of the internal MOSFET.

2.3 Output Voltage Load Regulation

Figure 3 shows the load regulation for the TPS54202HEVM-716.

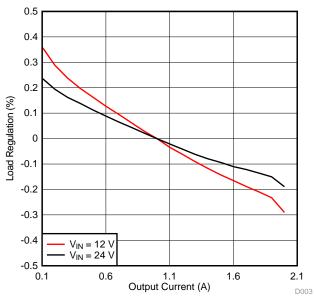


Figure 3. TPS54202HEVM-716 Load Regulation

Measurements are given for an ambient temperature of 25°C.

2.4 Output Voltage Line Regulation

Figure 4 shows the line regulation for the TPS54202HEVM-716.

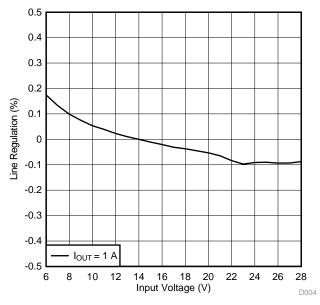


Figure 4. TPS54202HEVM-716 Line Regulation



2.5 Load Transients

Figure 5 shows the TPS54202HEVM-716 response to load transients. The current step is from 25% to 75% of maximum rated load at 24-V input. Total peak-to-peak voltage variation is as shown, including ripple and noise on the output.

Test Setup and Results

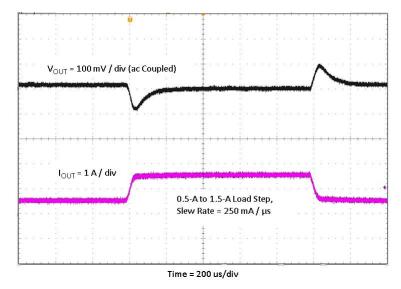


Figure 5. TPS54202HEVM-716 Transient Response



Test Setup and Results

2.6 Output Voltage Ripple

Figure 6, Figure 7, Figure 8, and Figure 9 show the TPS54202HEVM-716 output voltage ripple for full-load, skip-mode, light-load and no-load operation. $V_{IN} = 24$ V. The output The ripple voltage is measured directly across the output capacitors.

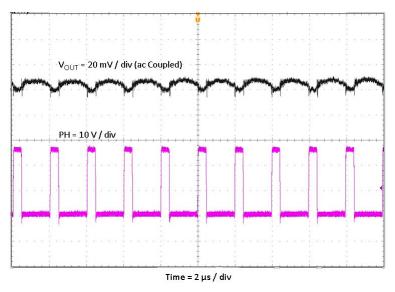


Figure 6. TPS54202HEVM-716 Output Ripple, I_{out} = 2 A

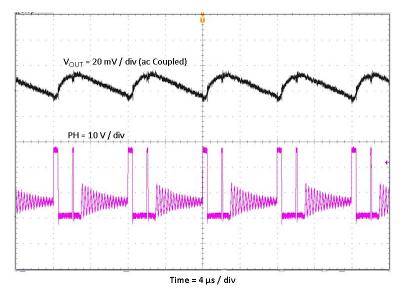


Figure 7. TPS54202HEVM-716 Output Ripple, I_{OUT} = 100 mA



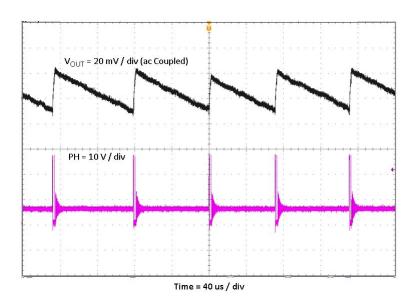


Figure 8. TPS54202HEVM-716 Output Ripple, I_{out} = 10 mA

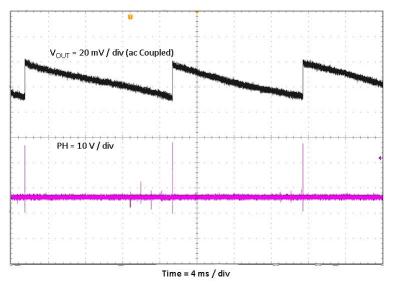


Figure 9. TPS54202HEVM-716 Output Ripple, I_{OUT} = 0 A



Test Setup and Results

2.7 Input Voltage Ripple

Figure 10 shows the TPS54202HEVM-716 input voltage ripple. The output current is the rated full load of 2 A and $V_{IN} = 24$ V. The ripple voltage is measured directly across the input capacitors.

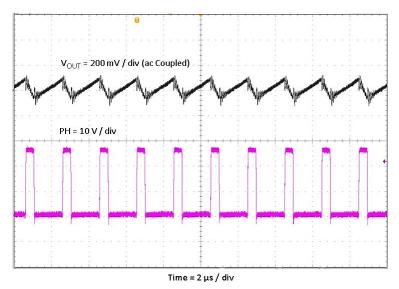


Figure 10. TPS54202HEVM-716 Input Ripple



2.8 Powering Up

Figure 11 and Figure 12 show the start-up waveforms for the TPS54202HEVM-716. In Figure 11, the output voltage ramps up as soon as the input voltage reaches the UVLO threshold as set by the R4 and R5 resistor divider network. In Figure 12, the input voltage is initially applied and the output is inhibited by using a 3.3-V logic signal between EN and GND. When the EN voltage reaches the enable-threshold voltage, the start-up sequence begins and the output voltage ramps up to the externally set value of 5 V. The input voltage for these plots is 24 V and the load is 5 Ω .

Test Setup and Results

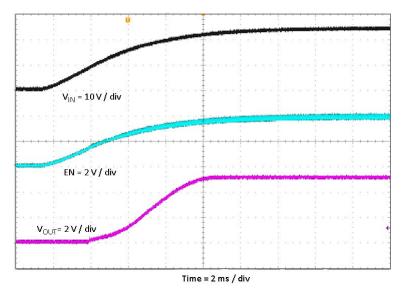


Figure 11. TPS54202HEVM-716 Startup Relative to V_{IN}

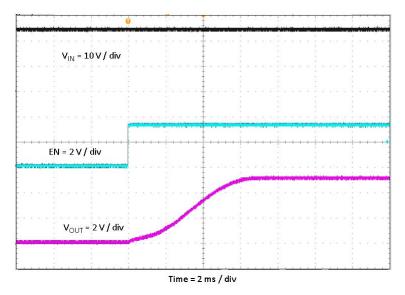


Figure 12. TPS54202HEVM-716 Startup Relative to Enable



Test Setup and Results

2.9 Powering Down

Figure 13 and Figure 14 show the start-up waveforms for the TPS54202HEVM-716. In Figure 13, the output voltage ramps down as soon as the input voltage falls below the UVLO stop threshold as set by the R4 and R5 resistor divider network. In Figure 14, the output is inhibited by using a 3.3-V logic signal between EN and GND. The input voltage for these plots is 24 V and the load is 5 Ω .

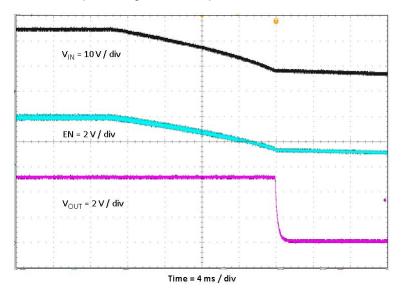


Figure 13. TPS54202HEVM-716 Shutdown Relative to V_{IN}

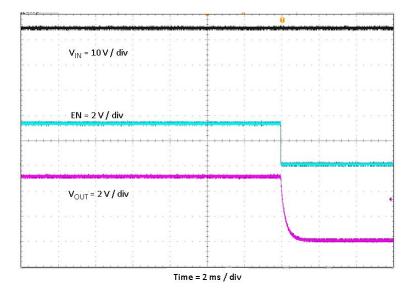


Figure 14. TPS54202HEVM-716 Shutdown Relative to EN



3 Board Layout

This section provides a description of the TPS54202HEVM-716, board layout, and layer illustrations.

3.1 Layout

Figure 15 and Figure 16 show the board layout for the TPS54202HEVM-716. The topside layer of the EVM is laid out in a manner typical of a user application. The top and bottom layers are 2-oz. copper.

The top layer contains the main power traces for V_{IN} , V_{OUT} , and SW. Also on the top layer are connections for the remaining pins of the TPS54202H and a large area filled with ground. To facilitate the placement of the main input bypass capacitor as close to the V_{IN} and GND pins as possible, the trace for SW is routed to the bottom layer immediately at the pin 3 connection. It is routed back to the top layer at the L1 inductor and C4 BOOT capacitor. The bottom layer contains a ground plane plus a copper fill area for SW, an etch run to connect the upper resistor of the voltage set point divider to the regulation point at the J2 output connector, and a trace to connect the upper resistor of the UVLO set point divider network to V_{IN} . The topside ground areas are connected to the bottom and internal ground planes with multiple vias placed around the board to provide a thermal path from the top-side ground area to the bottom-side and internal ground planes.

The input decoupling capacitors (C2, and C1) and bootstrap capacitor (C4) are all located as close to the IC as possible. In addition, the voltage set-point resistor divider components are also kept close to the IC. For the TPS54202H, an additional input bulk capacitor may be required, depending on the EVM connection to the input supply.

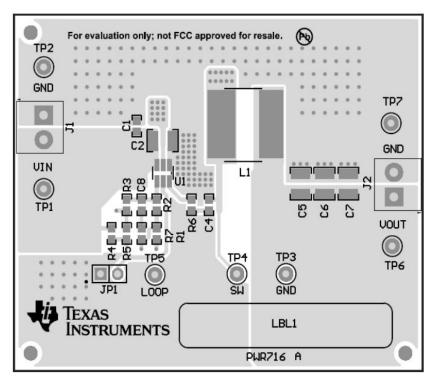


Figure 15. TPS54202HEVM-716 Top-Side Assembly



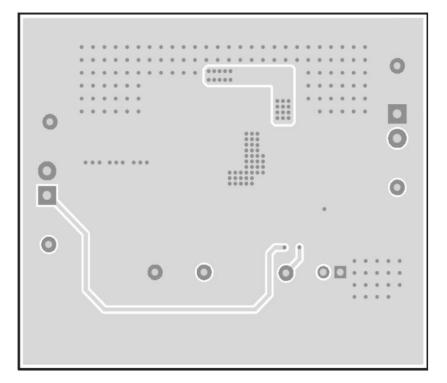


Figure 16. TPS54202HEVM-716 Bottom-Side Layout

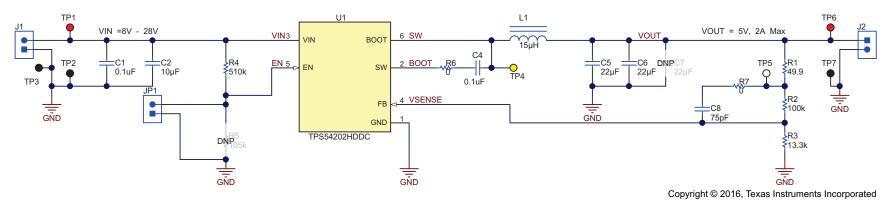


4 Schematic and Bill of Materials

This section presents the TPS54202HEVM-716 schematic and bill of materials.

4.1 Schematic

Figure 17 is the schematic for the TPS54202HEVM-716.







Schematic and Bill of Materials

4.2 Bill of Materials

Table 5 presents the bill of materials for the TPS54202HEVM-716.

Table 5. TPS54202HEVM-716 Bill of Materials

| Designator | Qty | Value | Description | Package Reference | Part Number | Manufacturer |
|---------------|-----|--------|--|-------------------------------|------------------------|---------------------|
| C1, C4 | 2 | 0.1uF | CAP, CERM, 0.1uF, 25V, +/-10%, X5R, 0603 | 0603 | GRM188R61E104KA01 D | Murata |
| C2 | 1 | 10uF | CAP, CERM, 10 μF, 35 V, +/- 10%, X7R, 1210 | 1210 | GRM32ER7YA106KA12 L | Murata |
| C5, C6 | 2 | 22uF | CAP, CERM, 22 µF, 25 V, +/- 10%, X7R, 1210 | 1210 | GRM32ER71E226KE15 L | Murata |
| C8 | 1 | 75pF | CAP, CERM, 75 pF, 50 V, +/- 5%, C0G/NP0, 0603 | 0603 | GRM1885C1H750JA01 D | Murata |
| J1, J2 | 2 | | Terminal Block, 6A, 3.5mm Pitch, 2-Pos, TH | 7.0x8.2x6.5mm | ED555/2DS | On-Shore Technology |
| JP1 | 1 | | Header, 100mil, 2x1, Gold, TH | 2x1 Header | TSW-102-07-G-S | Samtec |
| L1 | 1 | 15uH | Inductor, Shielded Drum Core, Ferrite, 15 µH, 3.5 A, 0.036 ohm, SMD | 10x5x10mm | 7447714150 | Wurth Elektronik |
| R1 | 1 | 49.9 | RES, 49.9, 1%, 0.1 W, 0603 | 0603 | CRCW060349R9FKEA | Vishay-Dale |
| R2 | 1 | 100k | RES, 100 k, 1%, 0.1 W, 0603 | 0603 | CRCW0603100KFKEA | Vishay-Dale |
| R3 | 1 | 13.3k | RES, 13.3 k, 1%, 0.1 W, 0603 | 0603 | CRCW060313K3FKEA | Vishay-Dale |
| R4 | 1 | 510k | RES, 510 k, 5%, 0.1 W, 0603 | 0603 | CRCW0603510KJNEA | Vishay-Dale |
| R6, R7 | 2 | 0 | RES, 0 ohm, 5%, 0.1W, 0603 | 0603 | ERJ-3GEY0R00V | Panasonic |
| TP1, TP6 | 2 | Red | Test Point, Miniature, Red, TH | Red Miniature Testpoint | 5000 | Keystone |
| TP2, TP3, TP7 | 3 | Black | Test Point, Miniature, Black, TH | Black Miniature Testpoint | 5001 | Keystone |
| TP4 | 1 | Yellow | Test Point, Miniature, Yellow, TH | Yellow Miniature Testpoint | 5004 | Keystone |
| TP5 | 1 | White | Test Point, Miniature, White, TH | White Miniature Testpoint | 5002 | Keystone |
| U1 | 1 | | 4.5-V TO 28-V INPUT, 2-A OUTPUT, EMI FRIENDLY SYNCHRONUOUS STEP-DOWN CONVERTOR, DDC0006A | DDC0006A | TPS54202HDDC | Texas Instruments |
| C7 | 0 | 22uF | CAP, CERM, 22 µF, 25 V, +/- 10%, X7R, 1210 | 1210 | GRM32ER71E226KE15 L | Murata |
| R5 | 0 | 105k | RES, 105 k, 1%, 0.1 W, 0603 | 0603 | CRCW0603105KFKEA | Vishay-Dale |

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- 3 Regulatory Notices:
 - 3.1 United States
 - 3.1.1 Notice applicable to EVMs not FCC-Approved:

This kit is designed to allow product developers to evaluate electronic components, circuitry, or software associated with the kit to determine whether to incorporate such items in a finished product and software developers to write software applications for use with the end product. This kit is not a finished product and when assembled may not be resold or otherwise marketed unless all required FCC equipment authorizations are first obtained. Operation is subject to the condition that this product not cause harmful interference to licensed radio stations and that this product accept harmful interference. Unless the assembled kit is designed to operate under part 15, part 18 or part 95 of this chapter, the operator of the kit must operate under the authority of an FCC license holder or must secure an experimental authorization under part 5 of this chapter.

3.1.2 For EVMs annotated as FCC – FEDERAL COMMUNICATIONS COMMISSION Part 15 Compliant:

CAUTION

This device complies with part 15 of the FCC Rules. Operation is subject to the following two conditions: (1) This device may not cause harmful interference, and (2) this device must accept any interference received, including interference that may cause undesired operation.

Changes or modifications not expressly approved by the party responsible for compliance could void the user's authority to operate the equipment.

FCC Interference Statement for Class A EVM devices

NOTE: This equipment has been tested and found to comply with the limits for a Class A digital device, pursuant to part 15 of the FCC Rules. These limits are designed to provide reasonable protection against harmful interference when the equipment is operated in a commercial environment. This equipment generates, uses, and can radiate radio frequency energy and, if not installed and used in accordance with the instruction manual, may cause harmful interference to radio communications. Operation of this equipment in a residential area is likely to cause harmful interference in which case the user will be required to correct the interference at his own expense.

FCC Interference Statement for Class B EVM devices

NOTE: This equipment has been tested and found to comply with the limits for a Class B digital device, pursuant to part 15 of the FCC Rules. These limits are designed to provide reasonable protection against harmful interference in a residential installation. This equipment generates, uses and can radiate radio frequency energy and, if not installed and used in accordance with the instructions, may cause harmful interference to radio communications. However, there is no guarantee that interference will not occur in a particular installation. If this equipment does cause harmful interference to radio or television reception, which can be determined by turning the equipment off and on, the user is encouraged to try to correct the interference by one or more of the following measures:

- Reorient or relocate the receiving antenna.
- Increase the separation between the equipment and receiver.
- Connect the equipment into an outlet on a circuit different from that to which the receiver is connected.
- Consult the dealer or an experienced radio/TV technician for help.

3.2 Canada

3.2.1 For EVMs issued with an Industry Canada Certificate of Conformance to RSS-210

Concerning EVMs Including Radio Transmitters:

This device complies with Industry Canada license-exempt RSS standard(s). Operation is subject to the following two conditions: (1) this device may not cause interference, and (2) this device must accept any interference, including interference that may cause undesired operation of the device.

Concernant les EVMs avec appareils radio:

Le présent appareil est conforme aux CNR d'Industrie Canada applicables aux appareils radio exempts de licence. L'exploitation est autorisée aux deux conditions suivantes: (1) l'appareil ne doit pas produire de brouillage, et (2) l'utilisateur de l'appareil doit accepter tout brouillage radioélectrique subi, même si le brouillage est susceptible d'en compromettre le fonctionnement.

Concerning EVMs Including Detachable Antennas:

Under Industry Canada regulations, this radio transmitter may only operate using an antenna of a type and maximum (or lesser) gain approved for the transmitter by Industry Canada. To reduce potential radio interference to other users, the antenna type and its gain should be so chosen that the equivalent isotropically radiated power (e.i.r.p.) is not more than that necessary for successful communication. This radio transmitter has been approved by Industry Canada to operate with the antenna types listed in the user guide with the maximum permissible gain and required antenna impedance for each antenna type indicated. Antenna types not included in this list, having a gain greater than the maximum gain indicated for that type, are strictly prohibited for use with this device.

Concernant les EVMs avec antennes détachables

Conformément à la réglementation d'Industrie Canada, le présent émetteur radio peut fonctionner avec une antenne d'un type et d'un gain maximal (ou inférieur) approuvé pour l'émetteur par Industrie Canada. Dans le but de réduire les risques de brouillage radioélectrique à l'intention des autres utilisateurs, il faut choisir le type d'antenne et son gain de sorte que la puissance isotrope rayonnée équivalente (p.i.r.e.) ne dépasse pas l'intensité nécessaire à l'établissement d'une communication satisfaisante. Le présent émetteur radio a été approuvé par Industrie Canada pour fonctionner avec les types d'antenne énumérés dans le manuel d'usage et ayant un gain admissible maximal et l'impédance requise pour chaque type d'antenne. Les types d'antenne non inclus dans cette liste, ou dont le gain est supérieur au gain maximal indiqué, sont strictement interdits pour l'exploitation de l'émetteur

3.3 Japan

- 3.3.1 Notice for EVMs delivered in Japan: Please see http://www.tij.co.jp/lsds/ti_ja/general/eStore/notice_01.page 日本国内に 輸入される評価用キット、ボードについては、次のところをご覧ください。 http://www.tij.co.jp/lsds/ti_ja/general/eStore/notice_01.page
- 3.3.2 Notice for Users of EVMs Considered "Radio Frequency Products" in Japan: EVMs entering Japan may not be certified by TI as conforming to Technical Regulations of Radio Law of Japan.

If User uses EVMs in Japan, not certified to Technical Regulations of Radio Law of Japan, User is required by Radio Law of Japan to follow the instructions below with respect to EVMs:

- 1. Use EVMs in a shielded room or any other test facility as defined in the notification #173 issued by Ministry of Internal Affairs and Communications on March 28, 2006, based on Sub-section 1.1 of Article 6 of the Ministry's Rule for Enforcement of Radio Law of Japan,
- 2. Use EVMs only after User obtains the license of Test Radio Station as provided in Radio Law of Japan with respect to EVMs, or
- 3. Use of EVMs only after User obtains the Technical Regulations Conformity Certification as provided in Radio Law of Japan with respect to EVMs. Also, do not transfer EVMs, unless User gives the same notice above to the transferee. Please note that if User does not follow the instructions above, User will be subject to penalties of Radio Law of Japan.

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- 4 EVM Use Restrictions and Warnings:
 - 4.1 EVMS ARE NOT FOR USE IN FUNCTIONAL SAFETY AND/OR SAFETY CRITICAL EVALUATIONS, INCLUDING BUT NOT LIMITED TO EVALUATIONS OF LIFE SUPPORT APPLICATIONS.
 - 4.2 User must read and apply the user guide and other available documentation provided by TI regarding the EVM prior to handling or using the EVM, including without limitation any warning or restriction notices. The notices contain important safety information related to, for example, temperatures and voltages.
 - 4.3 Safety-Related Warnings and Restrictions:
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 - 4.3.2 EVMs are intended solely for use by technically qualified, professional electronics experts who are familiar with the dangers and application risks associated with handling electrical mechanical components, systems, and subsystems. User assumes all responsibility and liability for proper and safe handling and use of the EVM by User or its employees, affiliates, contractors or designees. User assumes all responsibility and liability to ensure that any interfaces (electronic and/or mechanical) between the EVM and any human body are designed with suitable isolation and means to safely limit accessible leakage currents to minimize the risk of electrical shock hazard. User assumes all responsibility and liability for any improper or unsafe handling or use of the EVM by User or its employees, affiliates, contractors or designees.
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