

### PRECISION ADJUSTABLE CURRENT-LIMITED POWER SWITCHES

### Description

The AP2552/53 and AP2552A/53A are single channel precision adjustable current-limited switches optimized for applications that require precision current limiting, or to provide up to 2.1A of continuous load current during heavy loads/short circuits. These devices offer a programmable current-limit threshold between 75mA and 2.36A (Typ.) via an external resistor. Current limit accuracy  $\pm 6\%$  can be achieved at high current-limit settings. The rise and fall time are controlled to minimize current surges during turn on/off.

The devices have fast short-circuit response time for improved overall system robustness. They provide a complete protection solution for applications subject to heavy capacitive loads and the prospect of short circuit, offering reverse current blocking and limiting, overcurrent, over-temperature and short-circuit protection, as well as controlled rise time and under-voltage lockout functionality. A 7ms deglitch capability on the open-drain Flag output prevents false over-current reporting and does not require any external components.

AP2552/53 limits the output current to a safe level when the output current exceeds current-limit threshold.

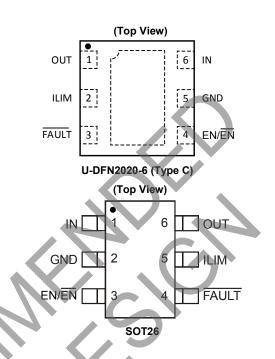
AP2552A/53A provides latch-off function during over-current or reverse-voltage conditions.

All devices are available in SOT26 and U-DFN2020-6 (Type C) packages.

### Features

- Up to 2.1A Maximum Load Current
- Accurate Adjustable Current Limit, 75mA to 2365mA
- ±6% Accurate Adjustable Current Limit, 1.632A with R<sub>LIM</sub> = 15kΩ
- Constant-Current (AP2552/53) During Over-Current
- Output Latch-Off (AP2552A/53A) at Over-Current
- Fast Short-Circuit Response Time: 2µs (Typ.)
- Reverse Current Blocking During Shutdown and Reverse Current Limiting During Enable
- Operating Range: 2.7V to 5.5V
- Built-in Soft-Start with 3ms Typical Rise Time
- Over-Current, Output Over-Voltage and Thermal Protection
- Fault Report (FAULT) with Blanking Time
- ESD Protection: 2kV HBM, 500V CDM
- Active Low (AP2552/52A) or Active High (AP2553/53A) Enable
- Ambient Temperature Range: -40°C to +85°C
- SOT26 and U-DFN2020-6 (Type C) Packages: Available in "Green" Molding Compound (No Br, Sb)
- 15kV ESD Protection per IEC 61000-4-2 (With External Capacitance)
- UL Recognized, File Number E322375, Vol. 1
- 1IEC60950-1 CB Scheme Certified
- Totally Lead-Free & Fully RoHS Compliant (Notes 1 & 2)
- Halogen and Antimony Free. "Green" Device (Note 3)
- For automotive applications requiring specific change control (i.e. parts qualified to AEC-Q100/101/200, PPAP capable, and manufactured in IATF 16949 certified facilities), please <u>contact us</u> or your local Diodes representative. <u>https://www.diodes.com/quality/product-definitions/</u>
- Notes: 1. No purposely added lead. Fully EU Directive 2002/95/EC (RoHS), 2011/65/EU (RoHS 2) & 2015/863/EU (RoHS 3) compliant.
  - 2. See https://www.diodes.com/quality/lead-free/ for more information about Diodes Incorporated's definitions of Halogen- and Antimony-free, "Green" and Lead-free.
    - 3. Halogen- and Antimony-free "Green" products are defined as those which contain <900ppm bromine, <900ppm chlorine (<1500ppm total Br + Cl) and <1000ppm antimony compounds.

## Pin Assignments

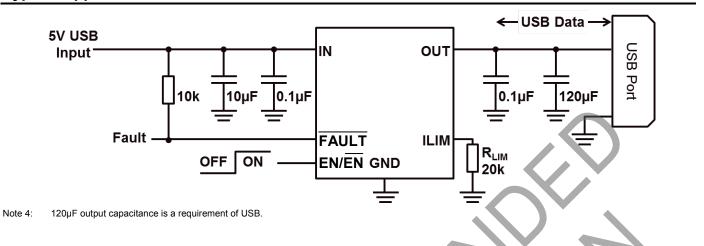


### Applications

- Set-Top Boxes
- LCD TVs & Monitors
- Residential Gateways
- Laptops, Desktops, Servers, e-Readers, Printers, Docking Stations, HUBs



## Typical Applications Circuit (Note 4)



# Available Options

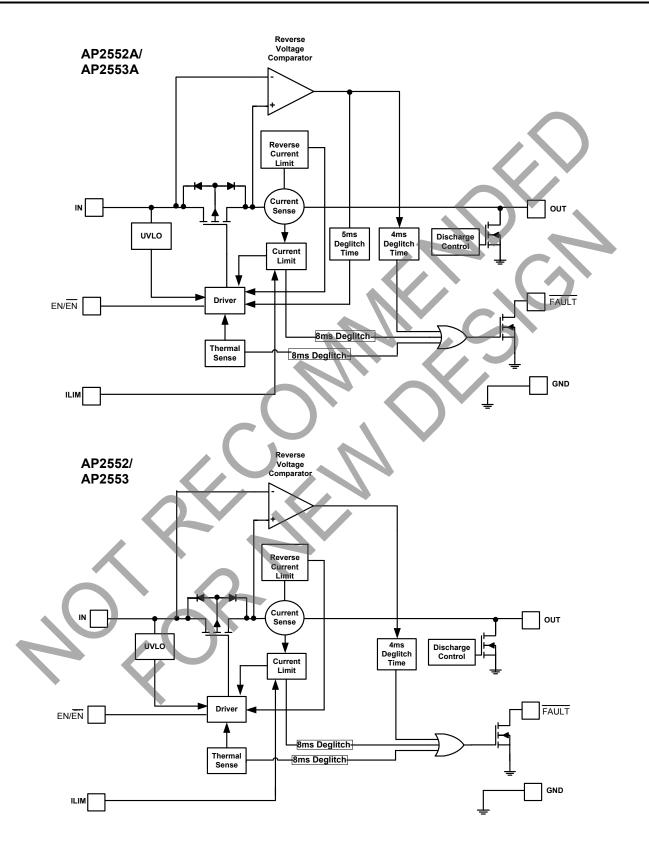
Part Number	Channel	Enabl <u>e P</u> in (EN/EN)	Recommended Maximum Continuous Load Current (A)	Current-Limit Protection	Package
AP2552	1	Active Low	- 21	Constant-Current	U-DFN2020-6 (Type C)
AP2553	1	Active High	2.1	Constant-Current	SOT26
AP2552A	1	Active Low	21	Latch-Off	U-DFN2020-6 (Type C)
AP2553A	1	Active High	2.1	Laiun-On	SOT26

## **Pin Descriptions**

		Pin N	lumber			
Pin	SOT26 U-DFN2020		-6 (Type C)	1/0	Function	
Name	AP2552/ AP2552A	AP2553/ AP2553A	AP2552/ AP2552A	AP2553/ AP2553A		runction
IN	1		6	6	Ι	Input, connect a $0.1\mu F$ or greater ceramic capacitor from IN to GND as close to IC as possible.
GND	2	2	5	5		Ground, connect to external exposed pad.
ĒN	3	—	-4	-	Ι	Enable input, logic low turns on power switch.
EN		3		4	I	Enable input, logic high turns on power switch.
FAULT	4	4	3	3	0	Active-low open-drain output, asserted during over- current, over-temperature, or reverse-voltage conditions.
	5	5	2	2	0	Use external resistor to set current-limit threshold; recommended $10k\Omega \leq R_{LIM} \leq 210k\Omega.$
OUT	6	6	1	1	0	Output
Exposed Pad	_		Pad	Pad	_	No internal connection; recommend to connect to GND externally for improved power dissipation. It should not be used as electrical ground conduction path.



## **Functional Block Diagram**





Symbol		Parameter	Ratings	Unit
	HBM	Human Body Model ESD Protection	2	kV
ESD	CDM	Charged Device Model ESD Protection	500	V
IEC System Level		Surges per IEC61000-4-2. 1999 Applied to Output Terminals of EVM (Note 6)	15	kV
$\begin{array}{c} V_{\text{IN}}, V_{\text{OUT}}, V_{\overline{\text{FAULT}}}, \\ V_{\text{ILIM}}, V_{\overline{\text{EN}}}, V_{\overline{\overline{\text{EN}}}} \end{array}$		Voltage on IN, OUT, FAULT , ILIM, EN, EN	-0.3 to +6.5	V
_		Continuous FAULT Sink Current	25	mA
	_	ILIM Source Current	1	mA
ILOAD		Maximum Continuous Load Current	Internal Limited	A
T <sub>J(MAX)</sub> Maximum Junction Temperature		Maximum Junction Temperature	-40 to +150	°C
	T <sub>STG</sub> Storage Temperature Range (Note 5)		-65 to +150	°C

### **Absolute Maximum Ratings** (@T<sub>A</sub> = +25°C, unless otherwise specified.)

Notes: 5. UL Recognized Rating from -30°C to +70°C (Diodes Incorporated qualified T<sub>STG</sub> from -65°C to +150°C).
 6. External capacitors need to be connected to the output, EVM board was tested with capacitor 2.2µF 50V 0805. This level is a pass test only and not a limit

Caution: Stresses greater than the 'Absolute Maximum Ratings' specified above, can cause permanent damage to the device. These are stress ratings only; functional operation of the device at these or any other conditions exceeding those indicated in this specification is not implied. Device reliability can be affected by exposure to absolute maximum rating conditions for extended periods of time.

Semiconductor devices are ESD sensitive and can be damaged by exposure to ESD events. Suitable ESD precautions should be taken when handling and transporting these devices.

## **Dissipation Rating Table**

Board	Package	Thermal Resistance θ <sub>JA</sub>	Thermal Resistance θ <sub>JC</sub>	T <sub>A</sub> ≤ +25°C Power Rating	Derating Factor Above T <sub>A</sub> = +25°C	T <sub>A</sub> = +70°C Power Rating	T <sub>A</sub> = +85°C Power Rating
High-K (Note 7)	SOT26	160°C/W	55°C/W	625mW	6.25mW/°C	340mW	250mW
High-K (Note 7)	U-DFN2020-6 (Type C)	120°C/W	34°C/W	833mW	8.33mW/°C	450mW	330mW

Note: 7. The JEDEC high-K (2s2p) board used to derive this data was a 3inch × 3inch, multilayer board with 1oz internal power and ground planes with 2oz copper traces on top and bottom of the board.

### Recommended Operating Conditions (@TA = +25°C, unless otherwise specified.)

Symbol	Parameter	Min	Max	Unit
Vin	Input Voltage	2.7	5.5	V
lout	Continuous Output Current (-40°C $\leq$ T <sub>A</sub> $\leq$ +85°C)	0	2.1	А
V <sub>EN</sub> , V <sub>EN</sub>	Enable Voltage	0	5.5	V
VIH High-Level Input Voltage on EN or EN		2.0	V <sub>IN</sub>	V
V <sub>IL</sub>	Low-Level Input Voltage on EN or EN	0	0.8	V
R <sub>LIM</sub>	Current-Limit Threshold Resistor Range (1% Initial Tolerance)	10	210	kΩ
lo	Continuous FAULT Sink Current	0	10	mA
_	Input De-Coupling Capacitance, IN to GND	0.1	—	μF
T <sub>A</sub>	Operating Ambient Temperature	-40	+85	°C
TJ	Operating Junction Temperature	-40	+125	°C



Symbol	Parameter	Test Conditions	s (Note 8)	Min	Тур	Max	Unit
Supply							
V <sub>UVLO</sub>	Input UVLO	V <sub>IN</sub> Rising			2.4	2.65	V
$\Delta V_{\text{UVLO}}$	Input UVLO Hysteresis	V <sub>IN</sub> Decreasing			25	_	mV
I <sub>SHDN</sub>	Input Shutdown Current	V <sub>IN</sub> = 5.5V, Disabled, OUT = Ope	n	_	0.1	1	μA
	Input Quicecent Current	V <sub>IN</sub> = 5.5V, Enabled, OUT = Oper	n, R <sub>LIM</sub> = 20kΩ	_	100	140	μA
lq	Input Quiescent Current	V <sub>IN</sub> = 5.5V, Enabled, OUT = Oper	n, R <sub>LIM</sub> = 210kΩ		80	120	μA
I <sub>REV</sub>	Reverse Leakage Current	Disabled, $V_{IN} = 0V$ , $V_{OUT} = 5.5V$ ,	I <sub>REV</sub> at V <sub>IN</sub>	-	0.01	1	μA
ower Swit	ch						
		SOT26 Deekers	T <sub>J</sub> = +25°C, V <sub>IN</sub> = 5.0V	-	70	95	
-	Switch On Desistance	SOT26 Package	-40°C ≤ T <sub>A</sub> ≤ +85°C		_	135	
R <sub>DS</sub> (ON)	Switch On-Resistance		T <sub>J</sub> = +25°C, V <sub>IN</sub> = 5.0V		80	105	mΩ
		U-DFN2020-6 (Type C) Package	-40°C ≤ T <sub>A</sub> ≤ +85°C		_	150	
		V <sub>IN</sub> = 5.5V, C <sub>L</sub> = 1µF, R <sub>LOAD</sub> = 10	0Ω. See Figure 1	_	1.1	1.5	ms
t <sub>R</sub>	Output Turn-On Rise Time	$V_{IN} = 2.7V, C_L = 1\mu F, R_{LOAD} = 100$	0Ω.	—	0.7	1	ms
		$V_{IN} = 5.5V, C_L = 1\mu F, R_{LOAD} = 100$		0.1		0.5	ms
t <sub>F</sub>	Output Turn-Off Fall Time	$V_{IN} = 2.7V, C_L = 1\mu F, R_{LOAD} = 100$		0.1		0.5	ms
urrent Lim	lit						
		R <sub>LIM</sub> = 10kΩ	-40°C ≤ T <sub>A</sub> ≤ +85°C	2200	2365	2542	
		R <sub>LIM</sub> = 15kΩ	-40°C ≤ T <sub>A</sub> ≤ +85°C	1540	1632	1730	
	Current-Limit Threshold (Maximum DC Output Current), V <sub>OUT</sub> = VIN -0.5V		T <sub>J</sub> = +25°C	1180	1251	1326	mA
		$R_{LIM} = 20k\Omega$	-40°C ≤ T <sub>A</sub> ≤ +85°C	1160	1251	1340	
ILIMIT			T <sub>J</sub> = +25°C	500	530	562	
		R <sub>LIM</sub> = 49.9kΩ	-40°C ≤ T <sub>A</sub> ≤ +85°C	485	529	573	
		R <sub>LIM</sub> = 210kΩ		121	142	162	
		I <sub>LIM</sub> Shorted to IN or GND		50	75	100	
		$R_{LIM} = 10k\Omega$		_	2620	_	
		$R_{LIM} = 15k\Omega$		_	1820	_	
	Short-Circuit Current Limit, OUT	$R_{LIM} = 20k\Omega$		_	1380		
I <sub>SHORT</sub>	Connected to GND	$R_{LIM} = 49.9 k\Omega$		_	570		mA
		$R_{LIM} = 210 k\Omega$		_	150	_	
		ILIM Shorted to IN or GND		_	75		
		$V_{OUT} = 0V$ to $I_{OUT} = I_{LIMIT}$ (OUT si	horted to around)		10		
t <sub>SHORT</sub>	Short-Circuit Response Time	See Figure 2	noneu to ground)	—	2	—	μs
nable Pin							
ILEAK-EN	EN Input Leakage Current	V <sub>IN</sub> = 5V, V <sub>EN</sub> = 0V and 6V		-0.5		0.5	μA
ton	Turn-On Time	$C_L = 1\mu F$ , $R_L = 100\Omega$ . See Figure	1	_	_	3	ms
toFF	Turn-Off Time	$C_L = 1\mu F$ , $R_L = 100\Omega$ . See Figure		_	_	1	ms
utput Disc				I	1	I	-
R <sub>DIS</sub>	Discharge Resistance (Note 9)	$V_{IN}$ = 5V, Disabled, $I_{OUT}$ =1mA		_	600		Ω
_	Discharge Resistance During		A Oply				
RDIS_LATCH	Latch-Off	V <sub>IN</sub> = 5V, Latch-Off, AP2552A/53	A Only	_	1000	—	Ω

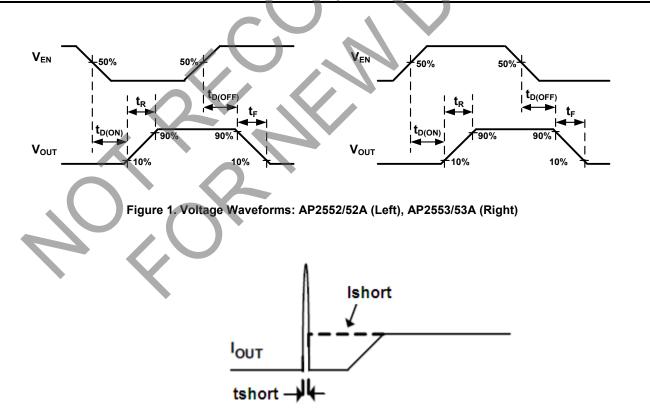
Notes: 8. Pulse-testing techniques maintain junction temperature close to ambient temperature; thermal effects must be taken into account separately.
 9. The discharge function is active when the device is disabled (when enable is de-asserted or during power-up power-down when V<sub>IN</sub> < V<sub>UVLO</sub>). The discharge function offers a resistive discharge path for the external storage capacitor for limited time.

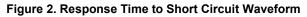


**Electrical Characteristics** (continued) (@T<sub>A</sub> = +25°C, V<sub>IN</sub> = 2.7V to 5.5V, V<sub>EN</sub> = 0V or V<sub>EN</sub> = V<sub>IN</sub>, R<sub>FAULT</sub> = 10k $\Omega$ , unless otherwise specified.)

Symbol	Parameter	Test Conditions (Note 8)	Min	Тур	Max	Unit	
Reverse Vol	tage Protection	·					
V <sub>RVP</sub>	Reverse-Voltage Comparator Trip Point	V <sub>OUT</sub> – V <sub>IN</sub>	95	135	190	mV	
IROCP	Reverse Current Limit	$V_{OUT} - V_{IN} = 200 \text{mV}$	_	0.72	_	А	
t <sub>TRIG</sub>	Time from Reverse-Voltage Condition to MOSFET Turn Off (AP2552A/AP2553A)	V <sub>IN</sub> = 5V		4.75	7	ms	
Fault Flag							
Vol	FAULT Output Low Voltage	I <sub>FAULT</sub> = 1mA	-		180	mV	
I <sub>FOH</sub>	FAULT Off Current	V <sub>FAULT</sub> = 6V			1	μA	
t <sub>BLANK_</sub> OC	FAULT Blanking and Latch Off Time (Over-Current)	Assertion or deassertion due to overcurrent	5	7.5	10	ms	
t <sub>BLANK_RV</sub>	FAULT Blanking Time (Reverse-Voltage)	Assertion or deassertion due to reverse-voltage	2	3.75	6	ms	
Thermal Sh	utdown						
T <sub>SHDN</sub>	Thermal Shutdown Threshold	Enabled, $R_{LOAD} = 1k\Omega$	(	+160	-	°C	
T <sub>SHDN_OCP</sub>	Thermal Shutdown Threshold under Current Limit	Enabled, $R_{LOAD} = 1k\Omega$		+140	_	°C	
T <sub>HYS</sub>	Thermal Shutdown Hysteresis	_	_	+20	_	°C	

## **Typical Performance Characteristics**







# Typical Performance Characteristics (continued)

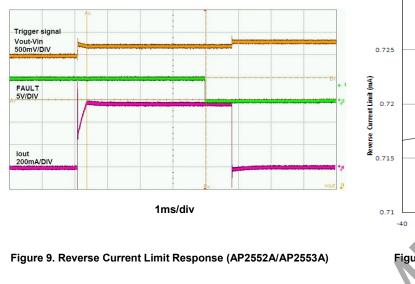


2ms/div Figure 7. Short-Circuit Current Limit Response

20ms/div Figure 8. Extended Short-Circuit into Thermal Cycles



## Typical Performance Characteristics (continued)



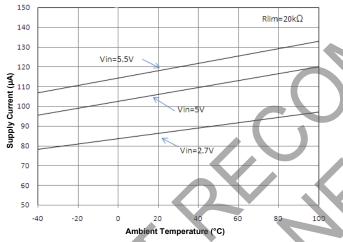


Figure 11. Quiescent Supply Current vs. Ambient Temperature

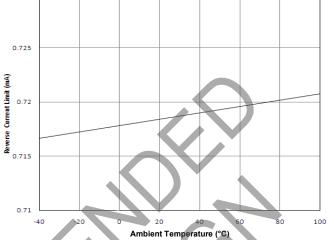
U-DFN2020-6 (Type C)

Package

SOT26

0

Package



0.73

Figure 10. Reverse Current Limit vs. Ambient Temperature

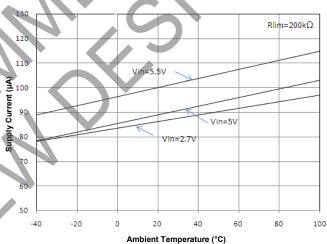


Figure 12. Quiescent Supply Current vs. Ambient Temperature

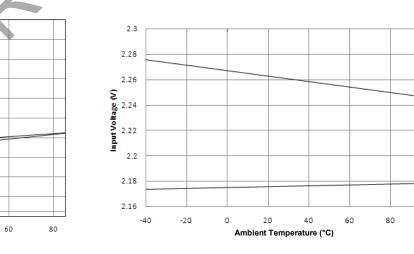


Figure 13. Switch On-Resistance vs. Ambient Temperature

Ambient Temperature (°C)

40

20

Figure 14. Under-Voltage Lock Out vs. Ambient Temperature

AP2552/AP2553/AP2552A/AP2553A Document number: DS35404 Rev. 12 - 3 Downloaded from Arrow.com.

-20

200

180

160

120

100

80

60

40

20

0

-40

g 140

**On-State Resistance** 

8 of 17 www.diodes.com 100



## **Application Information**

The AP2552/53 and AP2552A/53A are integrated high-side power switches optimized for Universal Serial Bus (USB) that require protection functions. The power switches are equipped with a driver that controls the gate voltage and incorporates slew-rate limitation. This, along with the various protection features and special functions, makes these power switches ideal for hot-swap or hot-plug applications.

### **Protection Features**

#### Under-Voltage Lockout (UVLO)

Whenever the input voltage falls below UVLO threshold (~2.5V), the power switch is turned off. This facilitates the design of hot-insertion systems where it is not possible to turn off the power switch before input power is removed.

#### **Over-Current and Short-Circuit Protection**

An internal sensing FET is employed to check for over-current conditions. Unlike current-sense resistors, sense FETs do not increase the series resistance of the current path. When an overcurrent condition is detected, AP2552/53 maintains a constant output current and reduces the output voltage accordingly. Complete shutdown occurs only if the fault stays long enough to activate thermal limiting.

For AP2552A/53A, when an overcurrent condition is detected, the devices will limit the current until the overload condition is removed or the internal deglitch time (7ms typical) is reached, and AP2552A/53A will be turned off. AP2552A/53A will remain latched off until power is cycled or the device enable is toggled.

The different overload conditions and the corresponding response of the AP2552/53 and AP2552A/53A are outlined below:

NO	Conditions	Explanation	Behavior of the AP2552/53
1	Short-circuit condition at start-up	Output is shorted before input voltage is applied or before the part is enabled	The IC senses the short circuit and immediately clamps output current to a certain safe level namely I <sub>SHORT</sub> .
2	Short-circuit or overcurrent condition	Short-Circuit or Overload condition that occurs when the part is enabled.	<ul> <li>At the instance the overload occurs, higher current may flow for a very short period of time before the current limit function can react.</li> <li>After the current limit function has tripped (reached the over-current trip threshold), the device switches into current limiting mode and the current is clamped at I<sub>SHORT</sub> /I<sub>LIMIT</sub>.</li> </ul>
3	Gradual increase from nominal operating current to ILIMIT	Load increases gradually until the current-limit threshold.(I <sub>TRIG</sub> )	The current rises until $I_{\text{LIMIT}}$ or thermal limit. Once the threshold has been reached, the device switches into its current limiting mode and is set at $I_{\text{LIMIT}}.$

NO	Conditions	Explanation	Behavior of the AP2552A/53A
1	Short-circuit condition at start-up	Output is shorted before input voltage is applied or before the part is enabled	The IC senses the short circuit and immediately clamps output current to a certain safe level namely I <sub>SHORT</sub> . When the internal deglitch time (7ms typical) is reached and the devices will be turned off.
2	Short-circuit or overcurrent condition	Short-Circuit or Overload condition that occurs when the part is enabled.	<ul> <li>At the instance the overload occurs, higher current may flow for a very short period of time before the current limit function can react.</li> <li>After the current limit function has tripped (reached the over-current trip threshold), the device switches into current limiting mode and the current is clamped at I<sub>SHORT</sub> /I<sub>LIMIT</sub>. When the internal deglitch time (7ms typical) is reached and the devices will be turned off.</li> </ul>
3	Gradual increase from nominal operating current to ILIMIT	Load increases gradually until the current-limit threshold.(I <sub>TRIG</sub> )	The current rises until $I_{LIMIT}$ or thermal limit. Once the threshold has been reached, the device switches into its current limiting mode and is set at $I_{LIMIT}$ . When the internal deglitch time (7ms typical) is reached and the devices will be turned off.

### **Over-Current FAULT Signal**

The FAULT signal could be asserted in response to OCP before the device reaches its current limit. The output current upon FAULT signal triggered will be lower than the I\_limit value. To implement FAULT signal for precision system protection control, it is recommended to leave enough margin from maximum continuous operating current for each RLIM value condition. For latch-off version, when over current is detected, device current will be clamped and device will latch off once FAULT signal asserted. To implement FAULT signal for precision system protection control, leave enough margin from maximum continuous operating current is highly recommended. Please consult Diodes Incorporated representative about margin recommendation for each RLIM value condition.



### Application Information (continued)

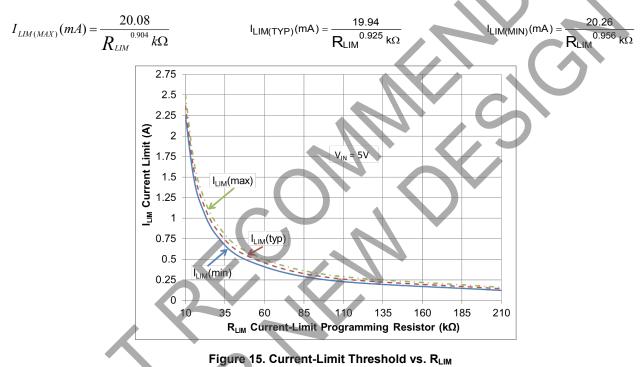
### Current-Limit Threshold Programming

The current-limit threshold can be programmed using an external resistor. The current-limit threshold is proportional to the current sourced out of I<sub>LIM</sub>.

The recommended 1% resistor range for  $R_{LIM}$  is  $10k\Omega \le R_{LIM} \le 210k\Omega$ . Figure 15 includes current-limit tolerance due to variations caused by temperature and process. This graph does not include the external resistor tolerance. The traces routing the RLIM resistor to the AP2552/53 and AP2552A/53A should be as short as possible to reduce parasitic effects on the current-limit accuracy.

To design below a maximum current-limit threshold, find the intersection of  $R_{LIM}$  and the maximum desired load current on the  $I_{OS(MAX)}$  ( $I_{LIM}$ ) curve and choose a value of  $R_{LIM}$  above this value. Programming the current limit below a maximum threshold is important to avoid current limiting upstream power supplies causing the input voltage bus to droop. The resulting minimum current-limit threshold is the intersection of the selected value of  $R_{LIM}$  and the  $I_{OS(MIN)}$  ( $I_{LIM}$ ) curve.

Best Fit Current-Limit Threshold Equations (ILIMIT):



### **Thermal Protection**

Thermal protection prevents the IC from damage when the die temperature exceeds safe margins. This mainly occurs when heavy-overload or short-circuit faults are present for extended periods of time. The AP2552/53 and AP2552A/53A implements a thermal sensing to monitor the operating junction temperature of the power distribution switch. Once the die temperature rises to approximately +160°C (+140°C in case the part is under current limit), the thermal protection feature activates as follows: The internal thermal sense circuitry turns the power switch off and the FAULT output is asserted, thus preventing the power switch from damage. Hysteresis in the thermal sense circuit allows the device to cool down by approximately +20°C before the output is turned back on. This built-in thermal hysteresis feature is an excellent feature, as it avoids undesirable oscillations of the thermal protection circuit. The switch continues to cycle in this manner until the load fault is removed, resulting in a pulsed output.



## Application Information (continued)

#### **Reverse-Current and Reverse-Voltage Protection**

The USB specification does not allow an output device to source current back into the USB port. In a normal MOSFET switch, current will flow in reverse direction (from the output side to the input side) when the output side voltage is higher than the input side. A reverse-current limit (ROCP) feature is implemented in the AP2552/53 and AP2552A/53A to limit such back currents. The ROCP circuit is activated when the output voltage is higher than the input voltage. After the reverse current circuit has tripped (reached the reverse current trip threshold), the current is clamped at this IROCP level.

In addition to ROCP, reverse over-voltage protection (ROVP) is also implemented. The ROVP circuit is activated by the reverse voltage comparator trip point; i.e., the difference between the output voltage and the input voltage.

For AP2552/53, once ROVP is activated, FAULT assertion occurs at a de-glitch time of 4ms. Recovery from ROVP is automatic when the fault is removed. FAULT de-assertion de-glitch time is same as the de-assertion time.

For AP2552A/53A, once ROVP is activated and when the condition exists for more than 5ms (Typ.), output device is disabled and shut down. This is called the "Time from Reverse-Voltage Condition to MOSFET Turn Off". FAULT assertion occurs at a de-glitch time of 4ms after ROVP is reached. Recovery from this fault is achieved by recycling power or toggling EN. FAULT de-assertion de-glitch time is same as the de-assertion time.

#### **Special Functions**

#### **Discharge Function**

When enable is de-asserted, or when the input voltage is under UVLO level, the discharge function is active. The output capacitor is discharged through an internal NMOS that has a discharge resistance of  $100\Omega$ . Hence, the output voltage drops down to zero. The time taken for discharge is dependent on the RC time constant of the resistance and the output capacitor.

#### FAULT Response

The FAULT open-drain output goes active low for any of following faults: Current limit threshold, short- circuit current limit, reverse-voltage condition or thermal shutdown. The time from when a fault condition is encountered to when the FAULT output goes low is 7ms (Typ.). The FAULT output remains low until over-current, short-circuit current limit and over-temperature conditions are removed. Connecting a heavy capacitive load to the output of the device can cause a momentary over-current condition, which does not trigger the FAULT due to the 7ms deglitch timeout. This 7ms timeout is also applicable for over-current recovery and over-temperature recovery. The AP2552/53 and AP2552A/53A are designed to eliminate erroneous over-current reporting without the need for external components, such as an RC delay network.

For the AP2552/53 and AP2552A/53A when the reverse voltage condition is triggered. FAULT output goes low after 4ms (Typ.). This 4ms (Typ.) timeout is also applicable for the recovery from reverse voltage fault.

When the ILIM pin is shorted to IN or GND, current-limit threshold and short-circuit current limit will be clamped at typically 75mA. When the ILIM pin is shorted to IN or GND, the AP2552/53 and AP2552A/53A FAULT pin will not assert during current limiting conditions; The FAULT pin will assert during short circuit conditions.

For latch-off version, once Fault signal is triggered, device will stay off until EN pin is toggled or power is restarted.

#### Power Supply Considerations

A  $0.01\mu$ F to  $0.1\mu$ F X7R or X5R ceramic bypass capacitor between IN and GND, close to the device, is recommended. This limits the input voltage drop during line transients. Placing a high-value electrolytic capacitor on the input ( $10\mu$ F minimum) and output pin ( $120\mu$ F) is recommended when the output load is heavy. This precaution also reduces power-supply transients that may cause ringing on the input. Additionally, bypassing the device output with a  $0.1\mu$ F to  $4.7\mu$ F ceramic capacitor improves the immunity of the device to short-circuit transients. This capacitor also prevents output from going negative during turn-off due to parasitic inductance.

#### **Power Dissipation and Junction Temperature**

The low on-resistance of the internal MOSFET allows the small surface-mount packages to pass large current. Using the maximum operating ambient temperature (T<sub>A</sub>) and R<sub>DS(ON)</sub>, the power dissipation can be calculated by:

 $P_D = R_{DS(ON)} \times I^2$ 

Finally, calculate the junction temperature:

 $T_J = P_D \times \theta_{JA} + T_A$ 

Where:

 $T_A$  = Ambient temperature °C  $\theta_{JA}$  = Thermal resistance  $P_D$  = Total power dissipation



### Application Information (continued)

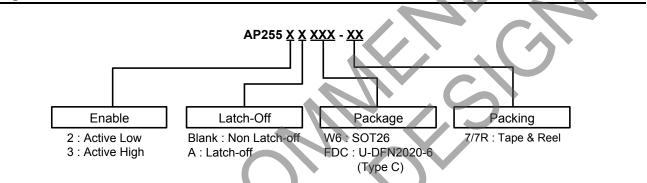
### **Generic Hot-Plug Applications**

In many applications it may be necessary to remove modules or PC boards while the main unit is still operating. These are considered hot-plug applications. Such implementations require the control of current surges seen by the main power supply and the card being inserted. The most effective way to control these surges is to limit and slowly ramp the current and voltage being applied to the card, similar to the way in which a power supply normally turns on. Due to the controlled rise and fall time of the AP2552/53 and AP2552A/53A, these devices can be used to provide a softer start-up to devices being hot-plugged into a powered system. The UVLO feature of the AP2552/53 and AP2552A/53A also ensures that the switch is off after the card has been removed, and that the switch is off during the next insertion.

#### Generic Hot-Plug Applications

By placing the AP2552/53 and AP2552A/53A between the  $V_{CC}$  input and the rest of the circuitry, the input power reaches these devices first after insertion. The typical rise time of the switch is approximately 1ms, providing a slow voltage ramp at the output of the device. This implementation controls system surge current and provides a hot-plugging mechanism for any device.

### Ordering Information

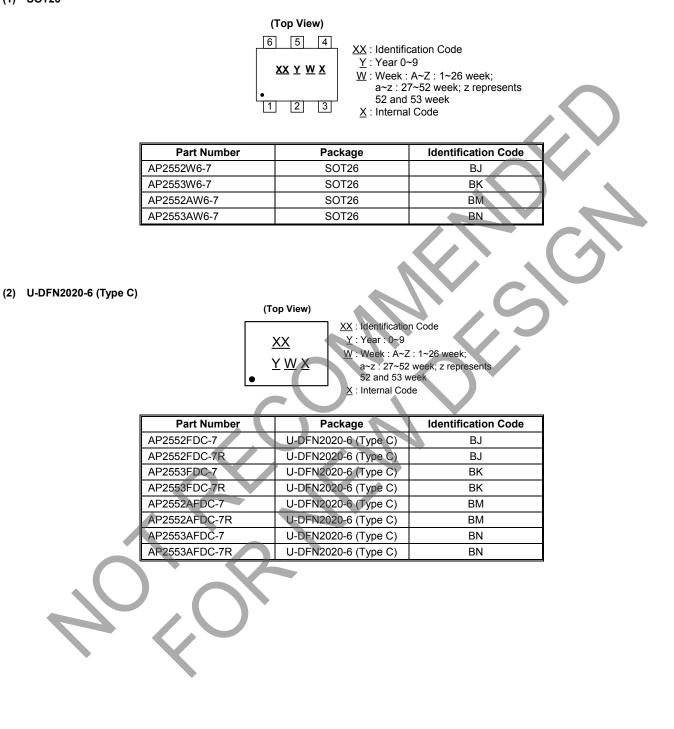


Part Number	Enable	Output Fault	Package Bostosing		7" Tape and Reel		
Part Number	Active	Condition	Code	Packaging	Quantity	Part Number Suffix	
AP2552W6-7			W6	SOT26	3000/Tape & Reel	-7	
AP2552FDC-7	Low		FDC	U-DFN2020-6 (Type C)	3000/Tape & Reel	-7	
AP2552FDC-7R		Output Current	FDC	U-DFN2020-6 (Type C)	3000/Tape & Reel	-7R	
AP2553W6-7		Limits	W6	SOT26	3000/Tape & Reel	-7	
AP2553FDC-7	High		FDC	U-DFN2020-6 (Type C)	3000/Tape & Reel	-7	
AP2553FDC-7R			FDC	U-DFN2020-6 (Type C)	3000/Tape & Reel	-7R	
AP2552AW6-7			W6	SOT26	3000/Tape & Reel	-7	
AP2552AFDC-7	Low		FDC	U-DFN2020-6 (Type C)	3000/Tape & Reel	-7	
AP2552AFDC-7R		Output Latches	FDC	U-DFN2020-6 (Type C)	3000/Tape & Reel	-7R	
AP2553AW6-7		Off	W6	SOT26	3000/Tape & Reel	-7	
AP2553AFDC-7	High		FDC	U-DFN2020-6 (Type C)	3000/Tape & Reel	-7	
AP2553AFDC-7R			FDC	U-DFN2020-6 (Type C)	3000/Tape & Reel	-7R	



## **Marking Information**

### (1) SOT26

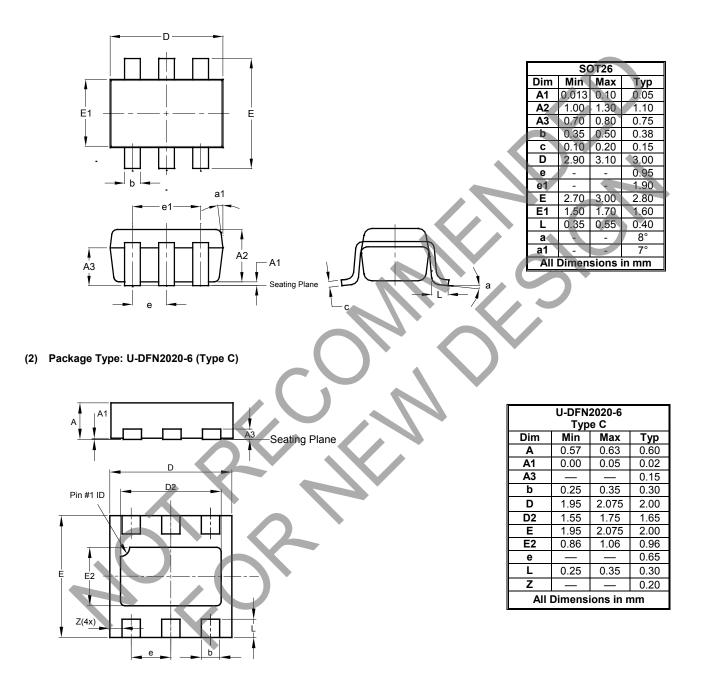




### **Package Outline Dimensions**

Please see http://www.diodes.com/package-outlines.html for the latest version.

### (1) Package Type: SOT26

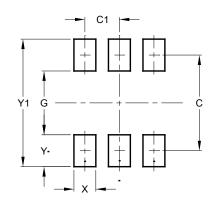




## **Suggested Pad Layout**

Please see http://www.diodes.com/package-outlines.html for the latest version.

### (1) Package Type: SOT26

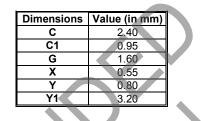


X2 X1

(2) Package Type: U-DFN2020-6 (Type C)

Y2 -

Х

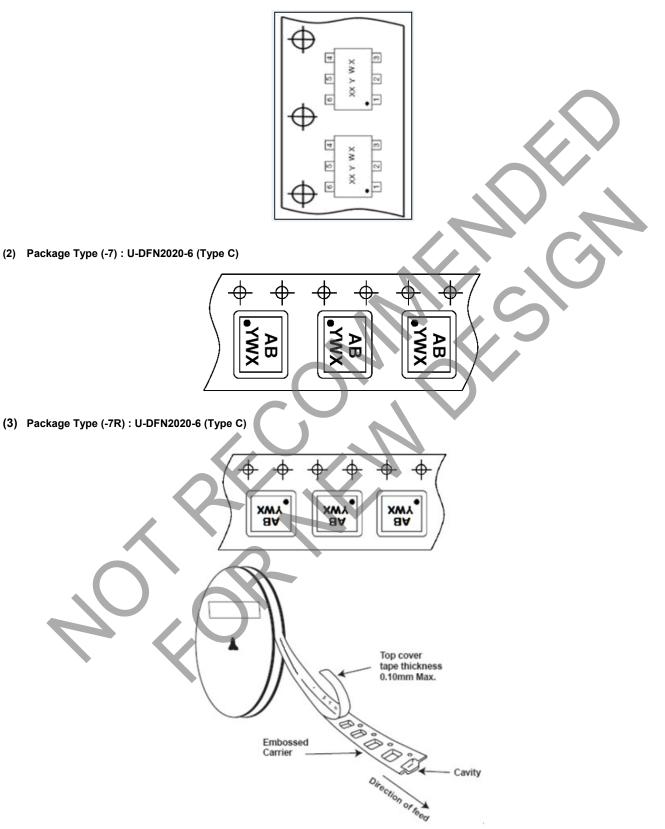


Dimensions	Value		
	(in mm)		
С	0.650		
X	0.350		
X1	1.650		
X2	1.700		
Y	0.525		
Y1	1.010		
Y2	2.400		



## Taping Orientation (Note 10)

(1) Package Type: SOT26



Note: 10. The taping orientation of the other package type can be found on our website at https://www.diodes.com/assets/Packaging-Support-Docs/Ap02007.pdf.



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