## Highly integrated tuner for AM/FM car radio

## Features

- Fully integrated VCO for world tuning
- High performance PLL for fast RDS system
- AM/FM mixers with high image rejection
- Integrated AM-LNA and AM-PINDIODE
- Automatic self alignment for preselection and image rejection
- Digital IF signal processing, high performance and drift-free
- Integrated IF-filters with high selectivity, high dynamic range and adaptive bandwidth control
- RDS demodulation with group and block synchronization
- High performance stereodecoder with noiseblanker
- $1^{2} \mathrm{C} /$ SPI bus controlled
- Single 5 V supply
- LQFP64 package


## Description

The TDA7705 highly integrated tuner (HIT) is a new generation of high performance tuners for carradio applications.


It contains mixers and IF amplifiers for AM and FM, fully integrated VCO and PLL synthesizer, IF-processing including adaptive bandwidth control, stereo decoder and RDS decoder on a single chip.

The utilization of digital signal processing results in numerous advantages against today's tuners: very low number of external components, very small space occupation and easy application, very high selectivity due to digital filters, high flexibility by software control and automatic alignment.

Table 1. Device summary

| Order code | Package | Packing |
| :---: | :---: | :---: |
| TDA7705 | LQFP64 $(10 \times 10 \times 1.4 \mathrm{~mm})$ | Tray |
| TDA7705TR | LQFP64 $(10 \times 10 \times 1.4 \mathrm{~mm})$ | Tape and reel |

## Contents

1 Block diagram and pins description ..... 6
1.1 Block diagram ..... 6
1.2 Pin description ..... 7
2 Function description ..... 10
2.1 FM - mixers ..... 10
2.2 FM - AGC ..... 10
2.3 AM - LNA ..... 10
2.4 AM - AGC ..... 10
2.5 AM - mixers ..... 10
2.6 IF A/D converters ..... 11
2.7 Audio D/A converters ..... 11
2.8 VCO ..... 11
2.9 PLL ..... 11
2.10 Crystal oscillator ..... 11
2.11 DSP ..... 11
2.12 IO interface pins ..... 12
2.13 Serial interface ..... 12
2.13.1 Serial interface choice / boot mode ..... 12
2.13.2 $I^{2} \mathrm{C}$ bus protocol ..... 13
2.13.3 SPI bus protocol ..... 14
3 Electrical specifications ..... 16
3.1 Absolute maximum ratings ..... 16
3.2 Thermal data ..... 16
3.3 General key parameters ..... 16
3.4 Electrical characteristics ..... 17
3.4.1 FM - section ..... 17
3.4.2 AM - section ..... 18
3.4.3 VCO ..... 19
3.4.4 Phase locked loop ..... 19
3.4.5 Tuning DAC ..... 19
3.4.6 IF ADC ..... 19
3.4.7 Audio DAC ..... 20
3.4.8 IO interface pins ..... 20
3.4.9 $\quad \mathrm{I}^{2} \mathrm{C}$ interface ..... 21
3.4.10 SPI interface ..... 22
3.4.11 Warning ..... 22
3.5 Overall system performance ..... 23
3.5.1 FM overall system performance ..... 23
3.5.2 AM MW overall system performance ..... 25
3.5.3 AM LW overall system performance ..... 27
3.5.4 AM SW overall system performance ..... 28
3.5.5 WX overall system performance ..... 29
4 Front-end processing ..... 30
5 Weak signal processing ..... 33
5.1 FM IF-processing ..... 33
5.1.1 Dynamic channel selection filter (DISS) ..... 33
5.1.2 Soft mute ..... 33
5.1.3 Adjacent channel mute ..... 34
5.1.4 Stereo blend- ..... 34
5.1.5 High cut control ..... 35
5.1.6 Stereo decoder ..... 36
5.2 AM IF-processing ..... 36
5.2.1 Channel selection filter ..... 36
5.2.2 Soft mute ..... 36
5.2.3 High cut control ..... 37
6 Application schematics ..... 38
6.1 Basic application schematic ..... 38
6.2 Application schematic example with SPI-bus and tuned preselection ..... 39
7 Package information ..... 40
8 Revision history ..... 41

## List of tables

Table 1. Device summary ..... 1
Table 2. Pin description ..... 7
Table 3. Boot mode pin configuration ..... 13
Table 4. Absolute maximum ratings ..... 16
Table 5. Thermal data ..... 16
Table 6. General key parameters ..... 16
Table 7. FM - section ..... 17
Table 8. AM - section ..... 18
Table 9. VCO ..... 19
Table 10. Phase locked loop ..... 19
Table 11. Tuning DAC ..... 19
Table 12. IF ADC ..... 19
Table 13. Audio DAC ..... 20
Table 14. IO interface pins ..... 20
Table 15. $\mathrm{I}^{2} \mathrm{C}$ interface ..... 21
Table 16. SPI interface ..... 22
Table 17. FM overall system performance ..... 23
Table 18. AM MW overall system performance ..... 25
Table 19. AM LW overall system performance ..... 27
Table 20. AM SW overall system performance ..... 28
Table 21. WX overall system performance ..... 29
Table 22. Register 0x00 ..... 30
Table 23. Register 0x01 ..... 31
Table 24. Register 0x02 ..... 31
Table 25. Register 0x05 ..... 32
Table 26. Dynamic channel selection filter (DISS) ..... 33
Table 27. Soft mute ..... 33
Table 28. Adjacent channel mute ..... 34
Table 29. Stereo blend ..... 34
Table 30. High cut control ..... 35
Table 31. De-emphasis filter ..... 36
Table 32. Stereo decoder ..... 36
Table 33. Channel selection filter ..... 36
Table 34. Soft mute ..... 36
Table 35. High cut control ..... 37
Table 36. Document revision history ..... 41

## List of figures

Figure 1. Functional block diagram ..... 6
Figure 2. Pin connection (top view) ..... 7
Figure 3. $\mathrm{I}^{2} \mathrm{C}$ "write" sequence ..... 13
Figure 4. $\mathrm{I}^{2} \mathrm{C}$ "read" sequence ..... 14
Figure 5. SPI modes ..... 15
Figure 6. SPI "write" sequence ..... 15
Figure 7. SPI "read" sequence ..... 15
Figure 8. $\quad \mathrm{I}^{2} \mathrm{C}$ bus timing diagram ..... 21
Figure 9. SPI bus timing diagram. ..... 22
Figure 10. FM input set-up. ..... 23
Figure 11. AM MW input set up ..... 25
Figure 12. AM LW input set-up ..... 27
Figure 13. AM SW input set-up ..... 28
Figure 14. WX input set-up ..... 29
Figure 15. FM wide-band application $/ \mathrm{I}^{2} \mathrm{C}$ control ..... 38
Figure 16. Example of FM tuned (narrow-band) application / SPI control ..... 39
Figure 17. LQFP64 (10x10x1.4mm) mechanical data and package dimensions. ..... 40

## 1 <br> Block diagram and pins description

### 1.1 Block diagram

Figure 1. Functional block diagram


### 1.2 Pin description

Figure 2. Pin connection (top view)


Table 2. Pin description

| Pin \# | Pin name |  |
| :---: | :--- | :--- |
| 1 | LF1 | PLL loopfilter output |
| 2 | PLLTEST | PLL test output / GPO |
| 3 | DAC | FM tuning DAC output |
| 4 | TCAGCFM | FM AGC time constant |
| 5 | FMMIX1dec | FM mixer decoupling |
| 6 | FMIX1in | FM mixer input 1 |
| 7 | FMIX2in | FM mixer input 2 |
| 8 | GND-RF | RF Ground |
| 9 | FMPINDRV | FM AGC PIN diode driver |
| 10 | VCC-RF | 5 5 supply for RF section |
| 11 | TCAM | AM AGC time constant |
| 12 | AMPINDRV | AM AGC external PIN diode driver |
| 13 | PINDdec | AM AGC internal PIN diode decoupling |
| 14 | PINDin | AM AGC internal PIN diode input |
| 15 | GND-LNA | AM LNA and internal PIN diode GND |
| 16 | LNAin | AM LNA input |

Table 2. Pin description (continued)

| Pin \# | Pin name | Function |
| :---: | :---: | :---: |
| 17 | LNAdec | AM LNA decoupling |
| 18 | LNAout | AM LNA output first stage |
| 19 | LNAin2 | AM LNA input $2^{\text {nd }}$ stage |
| 20 | LNAout2 | AM LNA output |
| 21 | LNAdec2 | AM LNA decoupling $2^{\text {nd }}$ stage |
| 22 | AMMIXin2 | AM mixer input 2 |
| 23 | AMMIXin1 | AM mixer input 1 |
| 24 | AMMIXdec | AM mixer decoupling |
| 25 | GND-IF | IF and Vref GND |
| 26 | VREF165 | 1.65 V reference voltage decoupling |
| 27 | VREFdec | 3.3V reference voltage decoupling |
| 28 | GND-DIG | Digital GND |
| 29 | VCC-DIG | 5 V supply for digital logic |
| 30 | VCCreg1V2 | VCC of 1.2 V regulator |
| 31 | REG1V2 | 1.2V regulator output |
| 32 | VDD-3V3 | 3.3V VDD output / decoupling |
| 33 | GND-3V3 | 3.3V VDD GND |
| 34 | SPI_CS | SPI chip select |
| 35 | SPI_MISO | SPI Data output |
| 36 | SDA / SPI_MOSI | $\mathrm{I}^{2} \mathrm{C}$ bus data / SPI data input |
| 37 | SCL / SPI_CLK | $1^{2} \mathrm{C}$ bus Clock / SPI clock |
| 38 | VDD-1V2 | 1.2V DSP supply |
| 39 | RDSINT | RDS interrupt |
| 40 | GPIO3 | Reserved |
| 41 | GPIO2 | Reserved |
| 42 | GPIO 1 | Reserved |
| 43 | GPIO 0 | Reserved |
| 44 | MODE | For debug purpose only, connected to GND |
| 45 | RSTN | Reset pin (active low) |
| 46 | TEST | Test input |
| 47 | VDD-1V2 | 1.2V DSP supply |
| 48 | GND-1V2 | Digital GND for 1.2V VDD |
| 49 | VCC-DAC | 5 V supply of audio DAC |
| 50 | OSCout | Xtal osc output |
| 51 | OSCin | Xtal osc input |

Table 2. Pin description (continued)

| Pin \# | Pin name | Function |
| :---: | :--- | :--- |
| 52 | GND-DAC | Audio DAC GND |
| 53 | DACoutL | Audio output left |
| 54 | DACoutR | Audio output right |
| 55 | GND-IFADC | IF ADC GND |
| 56 | LIFrefL | IF ADC reference low |
| 57 | LIFrefH | IF ADC reference high |
| 58 | VCC-IFADC | 5V supply of IF ADC |
| 59 | VCC-PLL | 5V supply of PLL |
| 60 | GND-PLL | PLL GND |
| 61 | VCO-dec | VCO decoupling |
| 62 | LFref | Loopfilter reference |
| 63 | VCC-VCO | 5V supply of VCO |
| 64 | GND-VCO | VCO GND |

## 2 Function description

### 2.1 FM - mixers

The image-rejection mixer has two FM inputs, selectable through software. These inputs feed stages with different gains, noise figures, and IIP3. They are optimized for best performance in case of a passive tuned prestage and for a passive fixed bandpass without tuning for low-cost application respectively.
The second input offers also the possibility of an easy addition of a weather-band preselection filter.

The input frequency is downconverted to low IF with high image rejection.
The tuned application is supported by an 8-bit tuning DAC. The alignment of the DAC is performed automatically.

### 2.2 FM - AGC

The programmable RFAGC senses the mixer input whereas the IFAGC senses the IFADC input to avoid overload.

The PIN diode driver is able to drive external PIN diodes with a current value as high as 15 mA .

The time constant of the FM-AGC is defined by an external capacitor.

### 2.3 AM - LNA

The AM-LNA is integrated with low noise and high IIP2 and IIP3. The gain of the LNA is controlled by the AGC. The maximum gain is set with an external resistor, typically 26 dB with $1 \mathrm{k} \Omega$.

### 2.4 AM - AGC

The programmable AM-RF-AGC senses the mixer inputs and controls the internal PIN diode and LNA gain.

First the LNA gain is reduced by about 10 dB , then the PIN diodes are activated to attenuate the signal.
The time constant of the AM-AGC is defined with an external capacitor and programmable internal currents.

### 2.5 AM - mixers

The image-rejection mixer has two AM inputs selectable via software. It easily supports lowcost applications for extended frequency bands like SW, DRM.

The input frequency is converted to low IF with high image rejection.

### 2.6 IF A/D converters

A high performance IQ-IFADC converts the IF-signal to digital IF for subsequent digital signal processing.

### 2.7 Audio D/A converters

A stereo DAC provides the left / right audio signals after IF-processing and stereodecoding by the DSP.

### 2.8 VCO

The VCO is fully integrated without any external tuning component. It covers all FM frequency bands including EU, US , Japan, EastEU, Weatherband and AM-bands including LW, MW, SW.

### 2.9 PLL

The high speed tuning PLL is able to settle within about $300 \mu$ for fast RDS applications.
The frequency step can be as low as 5 kHz in FM and 500 Hz in AM.

### 2.10 Crystal oscillator

The device works with a 37.05 MHz fundamental tone crystal, and can be used also with a $3^{\text {rd }}$ overtone 37.05 MHz crystal.

### 2.11 DSP

The DSP and its hardware accelerators perform all the digital signal processing. The main program is fixed in ROM. Control parameters are copied in RAM and are accessible and modifiable there, thus allowing parametric performance optimization.
It performs:

- digital down-conversion of IF
- bandwidth selection with variable controlled bandwidth
- FM and AM noiseblanking
- FM/AM demodulation with softmute, high-cut, weak signal processing and quality detection
- FM stereo decoding with stereo blend
- RDS demodulation including error correction and block synchronization with generation of an RDS interrupt for the main $\mu \mathrm{P}$
- Autonomous control of RDS-AF tests
- Self alignment of preselection tuning


### 2.12 IO interface pins

The TDA7705 has the following IO pins:

| PLLTEST | pin 2 | general purpose output |
| :--- | :--- | :--- |
| SPI_CS | pin 34 | serial communication with $\mu \mathrm{P}$ |
| SPI_MISO | pin 35 | serial communication with $\mu \mathrm{P}$ |
| SDA/MOSI | pin 36 | serial communication with $\mu \mathrm{P}$ |
| SCL/CLK | pin 37 | serial communication with $\mu \mathrm{P}$ |
| RDSINT | pin 39 | serial communication with $\mu \mathrm{P}$ |
| RSTN | pin 45 | reset pin driven by $\mu \mathrm{P}$ |

The pins labeled GPIO0, 1, 2 and 3 (pins 43 to 40) are reserved.
The pin PLLTEST output voltage can be freely programmed via software and be used to drive switches if needed by the application.

All the inputs are voltage-tolerant up to 3.5 V . The outputs can drive currents up to 0.5 mA from the internal 3.3 V supply line.

### 2.13 Serial interface

The device is controlled with a standard $\mathrm{I}^{2} \mathrm{C}$ bus or SPI interface.
Through the serial bus the processing parameters can be modifed and the signal quality parameters and the RDS information can be read out.

The operation of the device is handled through high level commands sent by the main carradio $\mu \mathrm{P}$ through the serial interface, which allow to simplify the operations carried out in the main $\mu \mathrm{P}$. The high level commands include among others:

- set frequency (which allows to avoid computing the PLL divider factors);
- start seek (the seek operation can be carried out by the TDA7705 in a completely autonomous fashion);
- RDS seek/search (jumps to AF and quality measurements are automatically sequenced).


### 2.13.1 Serial interface choice / boot mode

The device can communicate with the main $\mu \mathrm{P}$ with two different standard serial protocols: SPI and $\mathrm{I}^{2} \mathrm{C}$. The configuration is chosen by setting the proper value ( 0 V or 3.3 V ) at pins 35 and 39 and it is latched (e.g. made effective) when the RSTN line transitions from low to high (when RSTN is low, the IC is in reset mode).

The voltage level forced to pins 35 and 39 must be released to start the system operation a suitable time after the RSTN line has gone high.

The list of configurations is shown in the following table:

Table 3. Boot mode pin configuration

| Configuration: |  | $I^{2} C$ (addr. $0 \times$ C2) |  | $I^{2} C$ (addr. $0 \times$ C8) |  | SPI |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 39 | Rin | at reset | operation | at reset | operation | at reset | operation |
| 37 | SCLINT | 0 <br> in | RDS interrupt <br> out | 0 <br> in | RDS interrupt <br> out | 1 <br> in | RDS interrupt <br> out |
| 36 | SDA | $I^{2} C$ SCL <br> in | $x$ | $I^{2} C$ SCL <br> in | $x$ | SPI CLK <br> in |  |
| 35 | (SPI_MISO) | 0 <br> in | $I^{2} C$ SDA <br> in/out | $x$ | $I^{2} C$ SDA <br> in/out | $x$ | SPI MOSI <br> in |
| 34 | (SPI_CS) | $x$ | - | 1 <br> in | - | 1 <br> in | SPI MISO <br> out |

If $I^{2} C$ serial bus is chosen as means of communication with the controlling device, two chip addresses are possible: $0 \times \mathrm{C} 2 / \mathrm{C} 3$ or $0 \times \mathrm{C} 8 / \mathrm{C} 9$, depending on the initial configuration of pins 35 and 39.

The status of pins 35 and 39 during the reset phase can be set to:
high, through external $<10 \mathrm{k} \Omega$ resistors tied to 3.3 V (pin 32), or
low, by not forcing any voltage on them from outside, as 50 kohm internal pull-down resistors are present on said pins.

To make sure the boot mode is correctly latched up at start-up, it is advisable to keep the RSTN line low until the IC supply pins have reached their steady state, and then for an additional time $\mathrm{T}_{\text {reset }}$ (see Section 3.4.8).

### 2.13.2 $\quad I^{2} \mathrm{C}$ bus protocol

$1^{2} \mathrm{C}$ requires two signals: clock (SCL) and data (SDA - bidirectional). The protocol requires an acknowledge after any 8 -bit transmission.

A "write" communication example is shown in the figure below, for an unspecified number of data bytes (see the relevant technical documentation for frame structure description):

Figure 3. $\quad I^{2} \mathrm{C}$ "write" sequence


The sequence consists of the following phases:

- START: SDA line transitioning from H to L with $\operatorname{SCL}$ fixed H . This signifies a new transmission is starting;
- data latching: on the rising SCL edge. The SDA line can transition only when SCL is low (otherwise its transitions are interpreted as either a START or a STOP transition);
- ACKnowledge: on the $9^{\text {th }}$ SCL pulse the $\mu$ P keeps the SDA line H, and the TDA7705 pulls it down if communication has been successful. Lack of the acknowledge pulse generation from the TDA7705 means that the communication has failed;
- a chip address byte must be sent at the beginning of the transmission. The value can be C2 or C8 (according to the mode chosen at start-up during boot) for "write";
- as many data bytes as needed can follow the address before the communication is terminated. See the next section for details on the frame format;
- STOP: SDA line transitioning from $L$ to $H$ with SCL $H$. This signifies the end of the transmission.

Red lines represent transmissions from the TDA7705 to the $\mu \mathrm{P}$.
A "read" communication example is shown in the figure below, for an unspecified number of data bytes (see later on for frame structure decription):

Figure 4. $\quad I^{2} C$ "read" sequence


The sequence is very similar to the "write" one and has the same constraints for start, stop, data latching. The differences follow:

- a chip address must always be sent by the $\mu \mathrm{P}$ to the TDA7705; the address must be C3 (if C 2 had been selected at boot) or C9 (if C8 had been selected at boot);
- a header is transmitted after the chip address (the same happens for "write") before data are transferred from the TDA7705 to the $\mu$ P. See the relevant technical documentation for details on the frame format;
- when data are transmitted from the TDA7705 to the $\mu \mathrm{P}$, the $\mu \mathrm{P}$ keeps the SDA line H ;
- the ACKnowledge pulse is generated by the $\mu \mathrm{P}$ for those data bytes that are sent by the TDA7705 to the $\mu \mathrm{P}$. Failure of the $\mu \mathrm{P}$ to generate an ACK pulse on the $9^{\text {th }}$ CLK pulse has the same effect on the TDA7705 as a STOP.

The max. clock speed is 500 kbit/s.

### 2.13.3 SPI bus protocol

SPI requires four signals: clock (CLK), master output/slave input (MOSI - for communication from the $\mu \mathrm{P}$ to the TDA7705), master input/slave output (MISO - for communication from the TDA7705 to the $\mu \mathrm{P}$ ), chip select (CS). CLK is generated by the master device and is used for synchronization. MOSI and MISO are the data lines. The CS line is unique for each device in an SPI bus. The $\mu \mathrm{P}$ pulls low the TDA7705 CS line to select it for communication. The protocol does not foresee any transmission acknowledgement.
The SPI protocol has four possible modes of operation as far as data latching is concerned:

Figure 5. SPI modes


In the case of the TDA7705, the data are latched on the clock's rising edge, with CPOL $=1$ and CPHA $=1$ (mode 3 in the figure above). According to the specification of this mode, the polarity of the CLK line when no communication is taking place is high.

A "write" communication example is shown in the figure below, for an unspecified number of bits (see the relevant technical documentation for frame structure description):

Figure 6. SPI "write" sequence


The start condition is signaled by the CS line going low, and the stop condition by the CS line going high. It is not allowed to toggle the CS line while the communication is going on.

A "read" communication example is shown in the figure below, for an unspecified number of bits (see the relevant technical documentation for frame structure description ):

Figure 7. SPI "read" sequence


The red line is controlled by the TDA7705, whereas the black lines are controlled by the $\mu \mathrm{P}$.

## 3 Electrical specifications

### 3.1 Absolute maximum ratings

Table 4. Absolute maximum ratings

| Symbol | Parameter | Test condition | Min | Typ | Max | Units |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{CC}}$ | Supply voltage | - | - | - | 5.5 | V |
|  | Storage temperature | - | -55 | - | 150 | ${ }^{\circ} \mathrm{C}$ |
|  | ESD withstand voltage | Human body model | Charged device model | $\geq \pm 2000$ |  |  |
|  |  | Charged device model, corner pins | $\geq \pm 450$ |  | V |  |
|  |  | Machine model | $\geq \pm 150$ |  |  |  |

### 3.2 Thermal data

Table 5. Thermal data

| Symbol | Parameter | Test condition | Value | Units |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{R}_{\text {Th j-amb }}$ | Thermal resistance <br> junction-to-ambient | LQFP64 | $10 \times 10$, double-layer JEDEC PCB | 55 |
| ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |  |  |  |  |

### 3.3 General key parameters

Table 6. General key parameters

| Symbol | Parameter | Test condition | Min | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{CC}}$ | 5 V supply voltage | - | 4.7 | 5 | 5.25 | V |
| $\mathrm{I}_{\mathrm{CC}}$ | Supply current @ 5 V | - | - | 220 | 295 | mA |
| Tamb | Ambient temperature range | - | -40 | - | 85 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{V}_{\text {Vccregi2 }}$ | VCCREG12 supply voltage | see note ${ }^{(1)}$ | 2 | - | - | V |
| $\mathrm{V}_{1 \mathrm{~V} 2}$ | Digital core 1.2 V supply voltage | when supplied externally see note ${ }^{(2)}$ | 1.08 | 1.2 | 1.32 | V |
| $\mathrm{I}_{1 \mathrm{~V} 2}$ | Digital core 1.2 V supply current | $\begin{aligned} & \mathrm{V}_{1 \mathrm{~V} 2}=1.08 \mathrm{~V} \\ & \text { see note }{ }^{(2)} \end{aligned}$ | - | - | 120 | mA |
|  |  | $\begin{aligned} & \mathrm{V}_{1 \mathrm{~V} 2}=1.2 \mathrm{~V} \\ & \text { see note } \end{aligned}$ | - | 80 | 135 | mA |
|  |  | $\begin{aligned} & \mathrm{V}_{1 \mathrm{~V} 2}=1.32 \mathrm{~V} \\ & \text { see note } \end{aligned}$ | - | - | 150 | mA |

[^0]
### 3.4 Electrical characteristics

$\mathrm{V}_{\mathrm{CC}}=4.7 \mathrm{~V}$ to $5.25 \mathrm{~V} ; \mathrm{T}_{\mathrm{amb}}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$; unless otherwise specified.

### 3.4.1 FM - section

Table 7. FM - section

| Symbol | Parameter | Test condition | Min | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| FM IMR mixer |  |  |  |  |  |  |
| $\mathrm{R}_{\text {in }}$ | Input resistance | - | 90 | 130 | 170 | k ת |
| $V_{\text {noise }}$ | Input noise voltage | Mix 1, $\mathrm{R}_{\text {source }}=1.5 \mathrm{k} \Omega$, noiseless | - | 2.5 | 3.1 | $\mathrm{nV} / \sqrt{ } \mathrm{Hz}$ |
|  |  | $\begin{aligned} & \operatorname{Mix} 2, R_{\text {source }}=800 \Omega, \\ & \text { noiseless } \end{aligned}$ | - | 2 | 2.5 |  |
| IIP3 | $33^{\text {rd }}$ order intercept point | Mix 1 <br> up to $\mathrm{V}_{\text {in/tone }}=90 \mathrm{~dB} \mu \mathrm{~V}$ <br> Mix 2 <br> up to $\mathrm{V}_{\text {in/tone }}=85 \mathrm{~dB} \mu \mathrm{~V}$ | $\begin{aligned} & 122 \\ & 118 \end{aligned}$ | $\begin{aligned} & 125 \\ & 121 \end{aligned}$ | - | $\mathrm{dB} \mu \mathrm{V}$ <br> $\mathrm{dB} \mu \mathrm{V}$ |
| FM AGC |  |  |  |  |  |  |
| RFAGC-Thr | RFAGC threshold, referred to mixer input; RF level | Mix 1, min setting | - | 87 | - | $\mathrm{dB} \mu \mathrm{V}$ |
|  |  | Mix 1, max setting | - | 93 | - |  |
|  |  | Mix 2, min setting | - | 85 | - |  |
|  |  | Mix 2, max setting | - | 91 | - |  |
|  | Threshold steps | - | - | 2 | - | dB |
|  | Threshold error | @ $\mathrm{T}_{\mathrm{amb}}=27^{\circ} \mathrm{C}$ | -1.5 |  | 1.5 | dB |
|  | Threshold temperature drift |  | - | 0.016 | - | dB/K |
| IFAGC-Thr | IFAGC threshold, referred to mixer input; at tuned frequency RF level | Mix 1, min setting | - | 81 | - | $\mathrm{dB} \mu \mathrm{V}$ |
|  |  | Mix 1, max setting | - | 85 | - |  |
|  |  | Mix 2, min setting | - | 77 | - |  |
|  |  | Mix 2, max setting | - | 81 | - |  |
|  | Threshold steps | - | - | 2 | - | dB |
|  | Threshold error | @ $\mathrm{T}_{\mathrm{amb}}=27^{\circ} \mathrm{C}$ | -1.5 |  | 1.5 | dB |
|  | Threshold temperature drift | - |  | 0.016 | - | dB/K |
| - | Pin diode source current | @ $\mathrm{T}_{\mathrm{amb}}=27^{\circ} \mathrm{C}$; see note ${ }^{(1)}$ | 12 | - | - | mA |
| - | Pin diode sink current | - T | 3 | - | 20 | $\mu \mathrm{A}$ |
| - | Pin diode source current in constant current mode | $@ \mathrm{~T}_{\text {amb }}=27^{\circ} \mathrm{C}$; see note ${ }^{(1)}$ | 0.4 | - | - | mA |

1. The current is generated by a PTAT (Proportional To Absolute Temperature) source, and has therefore a temperature dependency described by: $\Delta \mathrm{I} / \mathrm{lo}=\Delta \mathrm{T} / \mathrm{To}$, with lo being the current at ambient temperature ( $25^{\circ} \mathrm{C}$ ) and To the ambient temperature $\left(25^{\circ} \mathrm{C}\right)$ expressed in Kelvin, that is 298 K .

### 3.4.2 AM - section

Table 8. AM - section

| Symbol | Parameter | Test condition | Min | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| AM IMR Mixer |  |  |  |  |  |  |
| $\mathrm{R}_{\text {in }}$ | Input resistance | - | 20 | 30 | 45 | $\mathrm{k} \Omega$ |
| $\mathrm{V}_{\text {out_max }}$ | Max. output voltage | without clipping | - | 126 | - | $\mathrm{dB} \mu \mathrm{V}$ |
| $\mathrm{V}_{\mathrm{N}, \mathrm{in}}$ | Input noise voltage | Mix $1, R_{\text {source }}=1 \mathrm{k} \Omega$, noiseless | - | 8.5 | 12 | $\mathrm{nV} / \sqrt{ } \mathrm{Hz}$ |
|  |  | $\text { Mix } 2, R_{\text {source }}=1 \mathrm{k} \Omega \text {, }$ noiseless | - | 8.5 | 12 |  |
| IIP3 | $33^{\text {rd }}$ order intercept point | Mix 1,2 <br> up to $\mathrm{V}_{\text {in/tone }}=90 \mathrm{~dB} \mu \mathrm{~V}$ | 126 | 129 | - | $\mathrm{dB} \mu \mathrm{V}$ |
| IIP2 | $2^{\text {nd }}$ order intercept point | $\begin{array}{\|l\|} \hline \text { Mix1 } 1,2 \\ \text { up to } V_{\text {in/tone }}=90 \mathrm{~dB} \mu \mathrm{~V} \end{array}$ | - | 158 | - | $\mathrm{dB} \mu \mathrm{V}$ |
| LO hsupp | LO harmonic suppression | $\mathrm{N}=2,3,4,5,6$ | - | 100 | - | dB |
|  |  | $\mathrm{N}=7,9$ | - | 85 | - |  |
| AM LNA |  |  |  |  |  |  |
| Gain | Voltage gain | Max Gain, $\mathrm{R}_{\text {ext }}=1 \mathrm{k} \Omega$ | 21 | 25 | 28 | dB |
|  |  | Min Gain (AGC controlled) | - | 12 | - |  |
| $\mathrm{R}_{\text {in }}$ | Input resistance | - | - | 1000 |  | $\mathrm{k} \Omega$ |
| $\mathrm{C}_{\text {in }}$ | Input capacitance | - | - | 20 |  | pF |
| $\mathrm{V}_{\mathrm{N}, \text { in }}$ | Input noise voltage | - | - | 1.0 | 1.4 | $\mathrm{nV} / \sqrt{ } \mathrm{Hz}$ |
| IIP3 | $3^{\text {rd }}$ order intercept point | @ maximum LNA gain | - | 125 | - | $\mathrm{dB} \mu \mathrm{V}$ |
| IIP2 | $2^{\text {nd }}$ order intercept point | @ maximum LNA gain | - | 143 | - | $\mathrm{dB} \mu \mathrm{V}$ |

AM PIN diode

| IIP2 | $2^{\text {nd }}$ order intercept point | Full attenuation, <br> $\mathrm{C}_{\text {source }}=80 \mathrm{pF}, \mathrm{f}=1 \mathrm{MHz}$ | - | 140 | - | $\mathrm{dB} \mu \mathrm{V}$ |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{R}_{\min }$ | Minimum resistance | - | - | 50 | 80 | $\Omega$ |
| $\mathrm{C}_{\text {in }}$ | Input capacitance | High ohmic | - | 12 | - | pF |

AM AGC

| AGC-Thr | Referred to mixer input | Mix 1,2 min setting | - | 87 | - | $\mathrm{dB} \mu \mathrm{V}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | RF level | Mix 1,2 max setting | - | 93 | - |  |
| Thr-steps | Threshold steps | - | - | 1 | - | dB |
|  | Threshold error | @ $\mathrm{T}_{\mathrm{amb}}=27^{\circ} \mathrm{C}$ | -2.5 | - | 2.5 |  |
|  | Threshold temperature drift | - | -3 | - | 3 |  |
| - | Pin diode source current | $@ \mathrm{~T}_{\text {amb }}=27^{\circ} \mathrm{C}$; see note ${ }^{(1)}$ | 2 | - | 10 | mA |
| - | Pin diode sink current | - | 15 | 35 | 50 | $\mu \mathrm{A}$ |
| - | Pin diode source current in constant current mode | @ $\mathrm{T}_{\text {amb }}=27^{\circ} \mathrm{C}$; see note ${ }^{(1)}$ | 1.5 | 2.5 | 3.5 | mA |

[^1]
### 3.4.3 VCO

Table 9. VCO

| Symbol | Parameter | Test condition | Min | Typ | Max | Units |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| F Vco | Frequency range VCO | - | 1100 |  | 1550 | MHz |
| PN | Phase noise of LO | Locked VCO; <br> values referred @ 100 MHz <br> $@ 100 \mathrm{~Hz}$ <br> $@ 1 \mathrm{kHz}$ | - | -100 | - | $\mathrm{dBc} / \mathrm{Hz}$ |
|  |  | @ 10 kHz |  | -115 |  |  |
| dev | Deviation error (rms) | FM reception, deemphasis <br> $50 \mu \mathrm{~s}, \mathrm{f}_{\text {audio }}=20 \mathrm{~Hz} . . .20 \mathrm{kHz}$ | - | 5 | - | Hz |

### 3.4.4 Phase locked loop

Table 10. Phase locked loop

| Symbol | Parameter | Test condition | Min | Typ | Max | Units |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{T}_{\text {settle }}$ | Settling time FM | $\Delta \mathrm{f}<10 \mathrm{kHz}$ | - | 300 | - | $\mu \mathrm{s}$ |
| FM step | FM frequency step | - | - | 5 | - | kHz |
| AM step | AM frequency step | - | - | 500 | - | Hz |

### 3.4.5 Tuning DAC

Table 11. Tuning DAC

| Symbol | Parameter | Test condition | Min | Typ | Max | Units |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| Res | Resolution | 8 bit | - | 18 | - | mV |
| $\mathrm{V}_{\text {outmin }}$ | Min output voltage | - | - | 0.6 | 0.7 | V |
| $\mathrm{~V}_{\text {outmax }}$ | Max ouput voltage | - | $\mathrm{VCC}-0.2$ | $\mathrm{VCC}-0.1$ | - | V |
| $\mathrm{R}_{\text {out }}$ | Output impdedance | - | 1.5 | 2.5 | 3.5 | $\mathrm{k} \Omega$ |
| DNL | Diff. Non linearity | - | - | - | 0.5 | LSB |
| $\mathrm{T}_{\text {conv }}$ | Conversion time | - | - | 20 | - | $\mu \mathrm{s}$ |

### 3.4.6 IF ADC

Table 12. IF ADC

| Symbol | Parameter | Test condition | Min | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{DR}_{\text {FM }}$ | Dynamic range in FM | BW $= \pm 200 \mathrm{kHz}$ |  | 90 |  | dB |
| $\mathrm{V}_{\mathrm{N}, \mathrm{in} \text { FM }}$ | Input noise referred to mixer input | mixer 1 <br> mixer 2 |  | $\begin{aligned} & 1.1 \\ & 0.7 \end{aligned}$ | $\begin{aligned} & 1.9 \\ & 1.2 \end{aligned}$ | $\mathrm{nV} / \sqrt{ } \mathrm{Hz}$ |
| $\mathrm{DR}_{\text {AM }}$ | Dynamic range in AM | $\mathrm{BW}= \pm 4 \mathrm{kHz}$ | - | 103 |  | dB |
| $\mathrm{V}_{\mathrm{N}, \mathrm{in} \text { AM }}$ | Input noise referred to mixer input | - | - | 6.9 | 12 | $\mathrm{nV} / \sqrt{ } \mathrm{Hz}$ |

### 3.4.7 Audio DAC

Table 13. Audio DAC

| Symbol | Parameter | Test condition | Min | Typ | Max | Units |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {out }}$ | Max. output voltage | Full scale | - | 1 | - | Vrms |
| BW | Bandwidth | 1 dB attenuation | - | 15 | - | KHz |
| $\mathrm{R}_{\text {out }}$ | Output resistance | - | 600 | 750 | 900 | $\Omega$ |
| $\mathrm{~V}_{\mathrm{N}, \text { out }}$ | Output noise | - | - | 60 | 95 | $\mu \mathrm{Vrms}$ |
| THD | Distortion | -6 dBFS | - | 0.03 | 0.04 | $\%$ |

### 3.4.8 IO interface pins

Table 14. IO interface pins

| Symbol | Parameter | Test condition | Min | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | High level output voltage (all IOs except GPO pin 2) | $\mathrm{l}_{\text {out }}=500 \mu \mathrm{~A}$ | 2.9 | 3.2 | - | V |
| - | GPIOs source current (all IOs in source mode except pin 2) | Total sourced current by all GPIOs | - | - | 1.25 | mA |
| - | Low level output voltage (all IOs except GPO pin 2) | $\mathrm{I}_{\text {out }}=-1 \mathrm{~mA}$ | - | 0.1 | 0.3 | V |
| - | Input voltage range | - | 0 | - | 3.5 | V |
| - | High level input voltage | - | 2.0 | - | - | V |
| - | Low level input voltage | - | - | - | 0.8 | V |
| $\mathrm{T}_{\text {reset }}$ | Reset time | Minimum time during which pin RSTN must be low so as to reset the device | 10 | - | - | $\mu \mathrm{s}$ |
| $\mathrm{T}_{\text {latch }}$ | Boot mode configuration latch time | Minimum time during which the voltage applied at pins 25 and 39 must be kept in order to latch the correct boot mode (serial bus configuration) | 10 | - | - | $\mu \mathrm{s}$ |
| - | GPO PLLTEST (pin 2) max source current | - | - | - | 1 | mA |
| - | GPO PLLTEST (pin 2) max sink current | - | -1 | - |  | mA |
| - | GPO PLLTEST (pin 2) minimum high level output voltage | $\mathrm{I}_{\text {out }}=1 \mathrm{~mA}$ | 2.8 | 3.1 | - | V |
|  | GPO PLLTEST (pin 2) maximum high level output voltage | $\mathrm{I}_{\text {out }}=1 \mathrm{~mA}$ | - | 0.1 | 0.3 | V |

### 3.4.9 $\quad I^{2} \mathrm{C}$ interface

The following parameters apply to the serial bus communication when $I^{2} \mathrm{C}$ protocol has been selected at start-up. For the other electrical characteristics of the pins, Section 3.4.8 applies. The parameters of the following table are defined as in Figure 8.

Table 15. $\quad I^{2} \mathrm{C}$ interface

| Symbol | Parameter | Min | Max | Units |
| :---: | :--- | :---: | :---: | :---: |
| $\mathrm{f}_{\text {SCL }}$ | SCL Clock frequency | - | 500 | kHz |
| $\mathrm{t}_{\text {AA }}$ | SCL low to SDA data valid | 0.3 | - | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\text {buf }}$ | time the bus must be kept free before a new <br> transmisison | 1.3 | - | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\text {HD-STA }}$ | START condition hold time | 0.6 | - | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\text {LOW }}$ | Clock low period | 1.3 | - | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\text {HIGH }}$ | Clock high period | 0.6 | - | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\text {SU-SDA }}$ | START condition setup time | 0.1 | - | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\text {HD-DAT }}$ | Data input hold time | 0 | 0.9 | $\mu \mathrm{~s}$ |
| $\mathrm{t}_{\text {SU-DAT }}$ | Data input setup time | 0.1 | - | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\mathrm{R}}$ | SDA \& SCL rise time | - | 0.3 | $\mu \mathrm{~s}$ |
| $\mathrm{t}_{\mathrm{F}}$ | SDA \& SCL fall time | - | 0.3 | $\mu \mathrm{~s}$ |
| $\mathrm{t}_{\text {SU-STOP }}$ | Stop condition setup time | 0.6 | - | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\text {DH }}$ | Data out time | - | 0.3 | $\mu \mathrm{~s}$ |

Figure 8. $\quad I^{2} \mathrm{C}$ bus timing diagram


### 3.4.10 SPI interface

The following parameters apply to the serial bus communication when SPI protocol has been selected at start-up. For the other electrical characteristics of the pins, Section 3.4.8 applies.

Table 16. SPI interface

| Symbol | Parameter | Min | Max | Unit |
| :---: | :--- | :---: | :---: | :---: |
| $\mathrm{f}_{\text {SCK }}$ | Clock frequency | - | 4.0 | MHz |
| $\mathrm{t}_{\mathrm{SU}}$ | Data setup time | 25 | - | ns |
| $\mathrm{t}_{\mathrm{H}}$ | Data hold time | 25 | - | ns |
| $\mathrm{t}_{\mathrm{WH}}$ | SCK high time | 50 | - | ns |
| $\mathrm{t}_{\mathrm{WL}}$ | SCK low time | 50 | - | ns |
| $\mathrm{t}_{\mathrm{RI}}$ | Input rise time | - | 2 | $\mu \mathrm{~s}$ |
| $\mathrm{t}_{\mathrm{FI}}$ | Input fall time | - | 2 | $\mu \mathrm{~s}$ |
| $\mathrm{t}_{\mathrm{V}}$ | Output valid from clock low | - | 50 | ns |
| $\mathrm{t}_{\mathrm{HO}}$ | Output hold time | 25 | - | ns |
| $\mathrm{t}_{\mathrm{DIS}}$ | Output disable time |  | 25 | ns |
| $\mathrm{t}_{\mathrm{CS}}$ | CS high time | 25 | - | ns |
| $\mathrm{t}_{\mathrm{CSS}}$ | CS setup time | 25 | - | ns |
| $\mathrm{t}_{\mathrm{CSH}}$ | CS hold time | 25 | - | ns |

Figure 9. SPI bus timing diagram


### 3.4.11 Warning

When the TDA7705 is not powered on, the internal ESD protection diodes pull-down keep the $I^{2} \mathrm{C} /$ SPI lines connected to ground. This implies that the $I^{2} \mathrm{C} / \mathrm{SPI}$ bus connected to the TDA7705 may not be used to drive other devices when the TDA7705 is powered off.

### 3.5 Overall system performance

All measurements obtained with application of Figure 16 (FM tuned application / SPI control) unless otherwise specified.

### 3.5.1 FM overall system performance

Antenna level equivalence: $0 \mathrm{~dB} \mu \mathrm{~V}=1 \mu \mathrm{~V}_{\mathrm{rms}}$ (Antenna terminal voltage with $50 \Omega$ source).
Figure 10. FM input set-up


Input level referred to signal generator loaded with $50 \Omega\left(\mathrm{~V}_{\mathrm{rf}}\right.$, node 'A'); no antenna dummy; $A M$ input not connected. $\mathrm{F}_{\mathrm{rf}}=98.1 \mathrm{MHz}, \mathrm{V}_{\mathrm{rf}}=60 \mathrm{~dB} \mu \mathrm{~V}$, mono modulation, $\mathrm{f}_{\mathrm{dev}}=40 \mathrm{kHz}$, $\mathrm{f}_{\text {audio }}=1 \mathrm{kHz}$. De-emphasis $=50 \mu \mathrm{~s}$. Unless otherwise specified

Table 17. FM overall system performance

| Parameter | Test condition | Min | Typ | Max | Units |
| :--- | :--- | :---: | :---: | :---: | :---: |
| Tuning range FM Eu | (can be modified by the user) <br> (automatic FE alignment <br> available) | 87.5 | - | 108 | MHz |
| Tuning step FM Eu | (can be modified by the user) | - | 100 | - | kHz |
| Tuning range FM US | (can be modified by the user) <br> (automatic FE alignment <br> available) | 87.5 | - | 107.9 | MHz |
| Tuning step FM US | (can be modified by the user) | - | 200 | - | kHz |
| Tuning range FM Jp | (can be modified by the user) <br> (automatic FE alignment <br> available) | 76 | - | 90 | MHz |
| Tuning step FM Jp | (can be modified by the user) | - | 100 | - | kHz |
| Tuning range FM EEu | (can be modified by the user) <br> (automatic FE alignment not <br> available) | 65 | - | 74 | MHz |
| Tuning step FM EEu | (can be modified by the user) | - | 100 | - | kHz |
| Sensitivity | S/N =26dB | - | -7 | -4 | $\mathrm{~dB} \mathrm{\mu V}$ |
| S/N | $@ 10 ~ d B \mu \mathrm{~V}$, no highcut, DISS <br> BW $=\# 3$ | - | 55 | - | dB |
|  | @ $60 \mathrm{~dB} \mathrm{\mu V}$, mono | 72 | 75 | - | dB |
|  | $@ 60 \mathrm{~dB} \mu \mathrm{~V}$, <br> Deviation $=75 \mathrm{kHz}$, mono | 78 | 81 | - | dB |
| @ $60 \mathrm{~dB} \mu \mathrm{~V}$, stereo | 70 | 73 | - | dB |  |

Table 17. FM overall system performance (continued)

| Parameter | Test condition | Min | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Distortion | Deviation $=75 \mathrm{kHz}$ | - | 0.05 | - | \% |
| Max deviation | THD=3\% | - | 140 | - | kHz |
| Adjacent channel selectivity | $\begin{aligned} & \Delta \mathrm{F}=100 \mathrm{kHz}, \mathrm{SINAD}=30 \mathrm{~dB} \\ & \text { desired } 40 \mathrm{~dB} \mu \mathrm{~V}, \mathrm{dev}=40 \mathrm{kHz}, \\ & 400 \mathrm{~Hz} \\ & \text { undesired. } \operatorname{dev}=40 \mathrm{kHz}, 1 \mathrm{KHz} \end{aligned}$ | - | 25 | - | dB |
| Alternate channel selectivity | $\begin{array}{\|l} \Delta \mathrm{F}=200 \mathrm{kHz}, \mathrm{SINAD}=30 \mathrm{~dB} \\ \text { desired } 40 \mathrm{~dB} \mu \mathrm{~V}, \\ \text { dev=}=40 \mathrm{kHz}, 400 \mathrm{~Hz} \\ \text { undesired. } \operatorname{dev}=40 \mathrm{kHz}, 1 \mathrm{kHz} \end{array}$ | - | 63 | - | dB |
| Max. strong signal interferer | $\begin{aligned} & \text { Desired }=10 \mathrm{~dB} \mu \mathrm{~V} \\ & \text { SINAD }=30 \mathrm{~dB} \\ & \text { Undesired } \Delta \mathrm{F}=5 \mathrm{MHz} \\ & \mathrm{dev}=40 \mathrm{kHz}, 1 \mathrm{kHz} \end{aligned}$ | - | 94 | - | $\mathrm{dB} \mu \mathrm{V}$ |
| Max. strong signal interferer no preselection ("wide-band") application | $\begin{aligned} & \text { Desired }=10 \mathrm{~dB} \mu \mathrm{~V} \\ & \text { SINAD }=30 \mathrm{~dB} \\ & \text { Undesired } \Delta \mathrm{F}=5 \mathrm{MHz} \\ & \mathrm{dev}=40 \mathrm{kHz}, 1 \mathrm{kHz} \end{aligned}$ | - | 88 | - | $\mathrm{dB} \mu \mathrm{V}$ |
| 3 signal performance ${ }^{(1)}$ | Desired $=40 \mathrm{~dB} \mu \mathrm{~V}$, <br> $\mathrm{dev}=40 \mathrm{kHz}, 400 \mathrm{~Hz}$, <br> SINAD $=30 \mathrm{~dB}$ <br> Undesired1 $= \pm 400 \mathrm{kHz}$, $\mathrm{dev}=40 \mathrm{kHz}, 1 \mathrm{kHz}$ <br> Undesired2 $= \pm 800 \mathrm{kHz}$, no mod | - | 103 | - | $\mathrm{dB} \mu \mathrm{V}$ |
|  | $\begin{array}{\|l\|} \hline \text { Desired }=40 \mathrm{~dB} \mu \mathrm{~V}, \\ \text { dev }=40 \mathrm{kHz}, 400 \mathrm{~Hz}, \\ \mathrm{SINAD}=30 \mathrm{~dB} \\ \text { Undesired1 }= \pm 1 \mathrm{MHz}, \\ \text { dev=40kHz, } 1 \mathrm{kHz} \\ \text { Undesired2 }= \pm 2 \mathrm{MHz}, \text { no mod } \end{array}$ | - | 106 | - | $\mathrm{dB} \mu \mathrm{V}$ |
| 3 signal performance ${ }^{(1)}$ no preselection ("wide-band") application | $\begin{aligned} & \text { Desired }=40 \mathrm{~dB} \mu \mathrm{~V}, \\ & \text { dev }=40 \mathrm{kHz}, 400 \mathrm{~Hz}, \\ & \mathrm{SINAD}=30 \mathrm{~dB} \\ & \text { Undesired1 }= \pm 400 \mathrm{kHz}, \\ & \text { dev }=40 \mathrm{kHz}, 1 \mathrm{kHz} \\ & \text { Undesired2 }= \pm 800 \mathrm{kHz}, \text { no } \\ & \text { mod } \end{aligned}$ | - | 103 | - | $\mathrm{dB} \mu \mathrm{V}$ |
|  | $\begin{array}{\|l\|} \hline \text { Desired }=40 \mathrm{~dB} \mu \mathrm{~V}, \\ \text { dev }=40 \mathrm{kHz}, 400 \mathrm{~Hz}, \\ \mathrm{SINAD}=30 \mathrm{~dB} \\ \text { Undesired1 }= \pm 1 \mathrm{MHz}, \\ \text { dev }=40 \mathrm{kHz}, 1 \mathrm{kHz} \\ \text { Undesired2 }= \pm 2 \mathrm{MHz}, \text { no mod } \end{array}$ | - | 104 | - | $\mathrm{dB} \mu \mathrm{V}$ |
| AM suppression | m =30 \% | - | 70 | - | dB |

Table 17. FM overall system performance (continued)

| Parameter | Test condition | Min | Typ | Max | Units |
| :--- | :--- | :---: | :---: | :---: | :---: |
| Image rejection | - | - | 80 | - | dB |
| Logarithmic field strength <br> indicator | @ $40 \mathrm{~dB} \mu \mathrm{~V}$ <br> read "FM_Smeter_log" | -0.33 <br> (equiv. <br> to 37 <br> $\mathrm{dB} \mu \mathrm{V})$ | -0.3 | -0.27 <br> (equiv. <br> to 43 <br> $\mathrm{dB} \mu \mathrm{V})$ | $-\mathrm{-}$ |

1. Signal levels referred to combiner output.

### 3.5.2 AM MW overall system performance

Antenna level equivalence: $0 \mathrm{~dB} \mu \mathrm{~V}=1 \mu \mathrm{~V}_{\text {rms }}$.
Figure 11. AM MW input set up


Level referred to SG output before antenna dummy ( $\mathrm{V}_{\mathrm{rf}}$, node ' A '); capacitive dummy $15 p F+68 p F$, $F M$ input not connected. $F_{\text {rf }}=999 \mathrm{kHz}(1000 \mathrm{kHz}$ for US$), \mathrm{V}_{\mathrm{rf}}=74 \mathrm{~dB} \mu \mathrm{~V}$, $\bmod =30 \%, f_{\text {audio }}=400 \mathrm{~Hz}$, unless otherwise specified.

Table 18. AM MW overall system performance

| Parameter | Test condition | Min | Typ | Max | Units |
| :--- | :--- | :---: | :---: | :---: | :---: |
| Tuning range MW Eu/Jp | (can be modified by the user) | 531 | - | 1629 | kHz |
| Tuning step MW Eu/Jp | (can be modified by the user) | - | 9 | - | kHz |
| Tuning range MW US | (can be modified by the user) | 530 | - | 1710 | kHz |
| Tuning step MW US | (can be modified by the user) | - | 10 | - | kHz |
| Sensitivity | $\mathrm{S} / \mathrm{N}=20 \mathrm{~dB}$ | - | 27 | 30 | $\mathrm{~dB} \mu \mathrm{~V}$ |
| Ultimate S/N | @ $80 \mathrm{~dB} \mu \mathrm{~V}$ | 63 | 66 | - | dB |
| AGC F.O.M. | Ref. $=74 \mathrm{~dB} \mu \mathrm{~V}$ <br> -10 dB drop point | 50 | 62 | 65 | dB |
| Distortion | $\mathrm{m}=80 \%$ | - | 0.1 | - | $\%$ |
| Adjacent channel selectivity | $\Delta \mathrm{F}=9 \mathrm{kHz}, \mathrm{SINAD}=26 \mathrm{~dB}$ <br> undesired. $\mathrm{m}=30 \%, 1 \mathrm{kHz}$ | - | 42 | - | dB |
| Alternate channel selectivity | $\Delta \mathrm{F}=18 \mathrm{kHz}, \mathrm{SINAD}=26 \mathrm{~dB}$ <br> undesired. $\mathrm{m}=30 \%, 1 \mathrm{kHz}$ | - | 50 | - | dB |

Table 18. AM MW overall system performance (continued)

| Parameter | Test condition | Min | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Strong signal interferer SNR | $\begin{aligned} & \Delta \mathrm{F}= \pm 40 \mathrm{kHz} \\ & \text { desired }=40 \mathrm{~dB} \mu \mathrm{~V} \\ & \text { undesired }=100 \mathrm{~dB} \mu \mathrm{~V}, \\ & \mathrm{~m}=30 \%, 1 \mathrm{kHz} \end{aligned}$ | - | 15 | - | dB |
|  | $\begin{aligned} & \Delta \mathrm{F}= \pm 400 \mathrm{kHz} \\ & \text { desired }=40 \mathrm{~dB} \mu \mathrm{~V} \\ & \text { undesired }=100 \mathrm{~dB} \mu \mathrm{~V}, \\ & \mathrm{~m}=30 \%, 1 \mathrm{kHz} \end{aligned}$ | 17 | - | - | dB |
| Strong signal interferer suppression | $\begin{aligned} & \Delta \mathrm{F}= \pm 40 \mathrm{kHz} \\ & \text { desired }=40 \mathrm{~dB} \mu \mathrm{~V} \\ & \text { undesired }=110 \mathrm{~dB} \mu \mathrm{~V}, \\ & \mathrm{~m}=30 \%, 1 \mathrm{kHz} \end{aligned}$ | - | 4 | - | dB |
|  | $\begin{aligned} & \Delta \mathrm{F}= \pm 400 \mathrm{kHz} \\ & \text { desired }=40 \mathrm{~dB} \mu \mathrm{~V} \\ & \text { undesired }=110 \mathrm{~dB} \mu \mathrm{~V}, \\ & \mathrm{~m}=30 \%, 1 \mathrm{kHz} \end{aligned}$ | - | 4 | - | dB |
| Strong signal interferer cross-modulation | $\begin{aligned} & \Delta \mathrm{F}= \pm 40 \mathrm{kHz} \\ & \text { desired }=80 \mathrm{~dB} \mu \mathrm{~V} \\ & \text { undesired }=100 \mathrm{~dB} \mu \mathrm{~V}, \\ & \mathrm{~m}=30 \%, 1 \mathrm{kHz} \end{aligned}$ | - | - | 10 | dB |
|  | $\begin{aligned} & \Delta \mathrm{F}= \pm 400 \mathrm{kHz} \\ & \text { desired }=80 \mathrm{~dB} \mu \mathrm{~V} \\ & \text { undesired }=100 \mathrm{~dB} \mu \mathrm{~V}, \\ & \mathrm{~m}=30 \%, 1 \mathrm{kHz} \end{aligned}$ | - | - | 10 | dB |
| Image rejection | - | - | 80 | - | dB |
| Logarithmic field strength indicator | ```@60 dB\muV read "AM_Smeter_log"``` | $\begin{gathered} 0.50 \\ \text { (equiv. } \\ \text { to } 57 \\ \text { dB } \mu \mathrm{V} \text { ) } \end{gathered}$ | 0.47 | $\begin{gathered} 0.43 \\ \text { (equiv. } \\ \text { to } 63 \\ \mathrm{~dB} \mu \mathrm{~V} \text { ) } \end{gathered}$ | - |

### 3.5.3 AM LW overall system performance

Antenna level equivalence: $0 \mathrm{~dB} \mu \mathrm{~V}=1 \mu \mathrm{~V}_{\text {rms }}$
Figure 12. AM LW input set-up


Level referred to SG output before antenna dummy ( $\mathrm{V}_{\mathrm{rf}}$, node ' A '); capacitive dummy $15 p F+68 p F$; FM input not connected. $\mathrm{F}_{\mathrm{rf}}=216 \mathrm{kHz}, \mathrm{V}_{\mathrm{rf}}=74 \mathrm{~dB} \mu \mathrm{~V}, \bmod =30 \%$, $f_{\text {audio }}=400 \mathrm{~Hz}$, unless otherwise specified.

Table 19. AM LW overall system performance

| Parameter | Test condition | Min | Typ | Max | Units |
| :--- | :--- | :---: | :---: | :---: | :---: |
| Tuning range LW | (can be modified by the user) | 144 | - | 288 | kHz |
| Tuning step LW | (can be modified by the user) | - | 1 | - | kHz |
| Sensitivity | $\mathrm{S} / \mathrm{N}=20 \mathrm{~dB}$ | - | 30 | 33 | $\mathrm{~dB} \mu \mathrm{~V}$ |
| Ultimate $\mathrm{S} / \mathrm{N}$ | $@ 80 \mathrm{~dB} \mu \mathrm{~V}$ | 63 | 66 | - | dB |
| AGC F.O.M. | Ref. $=74 \mathrm{~dB} \mu \mathrm{~V}$ <br> -10 dB drop point | 50 | 62 | 65 | dB |
| Distortion | $\mathrm{m}=80 \%$ | - | 0.1 | - | $\%$ |
| Image rejection | - | - | 80 | - | dB |

### 3.5.4 AM SW overall system performance

Antenna level equivalence: $0 \mathrm{~dB} \mu \mathrm{~V}=1 \mu \mathrm{~V}_{\mathrm{rms}}$
Figure 13. AM SW input set-up


Level referred to SG output before antenna dummy ( $\mathrm{V}_{\mathrm{rf}}$, node 'A'); capacitive dummy $15 p F+68 p F ;$ FM input not connected. $\mathrm{F}_{\mathrm{rf}}=6000 \mathrm{kHz}, \mathrm{V}_{\mathrm{rf}}=74 \mathrm{~dB} \mu \mathrm{~V}, \bmod =30 \%$, $f_{\text {audio }}=400 \mathrm{~Hz}$, unless otherwise specified.

Table 20. AM SW overall system performance

| Parameter | Test condition | Min | Typ | Max | Units |
| :--- | :--- | :---: | :---: | :---: | :---: |
| Tuning range LW | (can be modified by the user) | 2300 | - | 30000 | kHz |
| Tuning step LW | (can be modified by the user) | - | 1 | - | kHz |
| Sensitivity | $\mathrm{S} / \mathrm{N}=20 \mathrm{~dB}$ | - | 29 | 32 | $\mathrm{~dB} \mu \mathrm{~V}$ |
| Ultimate S/N | $@ 80 \mathrm{~dB} \mu \mathrm{~V}$ | 63 | 66 | - | dB |
| AGC F.O.M. | Ref. $=74 \mathrm{~dB} \mu \mathrm{~V}-10 \mathrm{~dB}$ drop point | 50 | 62 | 65 | dB |
| Distortion | $\mathrm{m}=80 \%$ | - | 0.3 | - | $\%$ |
| Image rejection | - | - | 80 | - | dB |

### 3.5.5 WX overall system performance

Antenna level equivalence: $0 \mathrm{~dB} \mu \mathrm{~V}=1 \mu \mathrm{~V}_{\mathrm{rms}}$ (Antenna terminal voltage with $50 \Omega$ source).
Figure 14. WX input set-up


Input level referred to signal generator loaded with $50 \Omega\left(\mathrm{~V}_{\mathrm{rf}}\right.$, node 'A'); no antenna dummy; AM input not connected. $\mathrm{F}_{\mathrm{rf}}=162.475 \mathrm{MHz}, \mathrm{V}_{\mathrm{rf}}=60 \mathrm{~dB} \mu \mathrm{~V}$, mono modulation, $\mathrm{f}_{\mathrm{dev}}=3 \mathrm{kHz}$, $\mathrm{f}_{\text {audio }}=400 \mathrm{~Hz}$. De-emphasis $=75 \mu \mathrm{~s}$. Application: WX using mixer input 2, in conjunction with FM narrow-band. Unless otherwise specified.

Table 21. WX overall system performance

| Parameter | Test condition | Min | Typ | Max | Units |
| :--- | :--- | :---: | :---: | :---: | :---: |
| Sensitivity | $\mathrm{S} / \mathrm{N}=26 \mathrm{~dB}$ | - | -7 | - | $\mathrm{dB} \mu \mathrm{V}$ |
| Ultimate $\mathrm{S} / \mathrm{N}$ | $@ 60 \mathrm{~dB} \mu \mathrm{~V}$ | - | 81 | - | dB |
| Distortion | Deviation $=4.5 \mathrm{kHz}$ | - | 0.8 | - | $\%$ |
| Max deviation | $\mathrm{THD}=3 \%$ | - | $>5 \mathrm{kHz}$ | - | kHz |
| Adjacent channel Selectivity | $\Delta \mathrm{F}=25 \mathrm{kHz}, \mathrm{SINAD=30dB}$ <br> desired $40 \mathrm{~dB} \mu \mathrm{~V}$, <br> dev =2.0 $\mathrm{kHz}, 400 \mathrm{~Hz}$ <br> undesired. dev=3 kHz, 1 kHz | - | 70 | - | dB |
| Alternate Channel Selectivity | $\Delta \mathrm{F}=50 \mathrm{kHz}, \mathrm{SINAD=30dB}$ <br> desired $40 \mathrm{~dB} \mu \mathrm{~V}$, <br> dev=2.0kHz, 400 Hz <br> undesired. dev=2.0kHz, 1 kHz | - | 70 | - | dB |

## 4 Front-end processing

All the parameters in this section refer to the programmability of the FE part of the device (registers). The part of the registers that are not described here have either fixed values or values written by the tuner drivers, and are described in the proper technical documentation.

Table 22. Register $0 \times 00$

| Register number |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | Register definition |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| MSB LSB |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 23 | 22 | 21 | 20 | 19 | 18 | 17 | 16 | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |  |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | AM mixer input selector |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | 0 | 1 | input \#1 |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | 1 | 0 | input \#2 |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | AM PIN diode |
|  |  |  |  |  |  |  |  |  |  |  | 0 |  |  |  |  |  |  |  |  |  |  |  |  | internal |
|  |  |  |  |  |  |  |  |  |  |  | 1 |  |  |  |  |  |  |  |  |  |  |  |  | external |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | AM AGC mode |
|  |  |  |  |  |  |  |  |  |  | 0 |  |  |  |  |  |  |  |  |  |  |  |  |  | LNA and PIN diode |
|  |  |  |  |  |  |  |  |  |  | 1 |  |  |  |  |  |  |  |  |  |  |  |  |  | PIN diode only |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | AM AGC time constant |
|  |  |  |  |  |  |  |  | 0 | 0 |  |  |  |  |  |  |  |  |  |  |  |  |  |  | slow ( 125 ms with $1 \mu \mathrm{~F}$ ) |
|  |  |  |  |  |  |  |  | 0 | 1 |  |  |  |  |  |  |  |  |  |  |  |  |  |  | medium ( 25 ms with $1 \mu \mathrm{~F}$ ) |
|  |  |  |  |  |  |  |  | 1 | 1 |  |  |  |  |  |  |  |  |  |  |  |  |  |  | fast ( 5 ms with $1 \mu \mathrm{~F}$ ) |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | AM AGC threshold @ mixin |
|  |  |  |  |  | 0 | 0 | 0 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | $90 \mathrm{~dB} \mu \mathrm{~V}$ |
|  |  |  |  |  | 0 | 0 | 1 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | $91 \mathrm{~dB} \mu \mathrm{~V}$ |
|  |  |  |  |  | 0 | 1 | 0 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | $92 \mathrm{~dB} \mu \mathrm{~V}$ |
|  |  |  |  |  | 0 | 1 | 1 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | $93 \mathrm{~dB} \mu \mathrm{~V}$ |
|  |  |  |  |  | 1 | 0 | 0 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | $90 \mathrm{~dB} \mu \mathrm{~V}$ |
|  |  |  |  |  | 1 | 0 | 1 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | $89 \mathrm{~dB} \mu \mathrm{~V}$ |
|  |  |  |  |  | 1 | 1 | 0 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | $88 \mathrm{~dB} \mu \mathrm{~V}$ |
|  |  |  |  |  | 1 | 1 | 1 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | $87 \mathrm{~dB} \mu \mathrm{~V}$ |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | AM AGC attack time constant |
|  |  |  | 0 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | normal |
|  |  |  | 1 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | fast |

Table 23. Register 0x01

| Register number |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | Register definition |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| MSB LSB |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 23 | 22 | 21 | 20 | 19 | 18 | 17 | 16 | 15 | 14 | 13 | 12 | 1 |  | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |  |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | FM mixer input selector |
|  |  |  |  |  |  |  |  |  | 0 | 1 |  |  |  |  |  |  |  |  | 1 | 0 |  |  |  |  | input \#1 |
|  |  |  |  |  |  |  |  |  | 1 | 0 |  |  |  |  |  |  |  |  | 0 | 1 |  |  |  |  | input \#2 |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | FM mixer gain |
|  |  |  |  |  |  |  |  | 0 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | high |
|  |  |  |  |  |  |  |  | 1 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | low |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | FM AGC time constant |
|  |  | 0 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | normal |
|  |  | 1 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | fast |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | FM AGC output mode |
| 0 | 0 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | normal |
| 0 | 1 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | constant 15 mA |
| 1 | 0 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | constant 1 mA |

Table 24. Register 0x02

| Register number |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | Register definition |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| MSB LSB |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 23 | 22 | 21 | 20 | 19 | 18 | 17 | 16 | 15 | 14 | 13 | 12 |  | 11 | 10 | 9 |  | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |  |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | FM RF AGC threshold @ mixin |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | 0 | 0 | $87 \mathrm{~dB} \mu \mathrm{~V}$ |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | 0 | 1 | $89 \mathrm{~dB} \mu \mathrm{~V}$ |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | 1 | 0 | $91 \mathrm{~dB} \mu \mathrm{~V}$ |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | 1 | 1 | $93 \mathrm{~dB} \mu \mathrm{~V}$ |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | FM iF AGC threshold @ IFADC in |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | 0 | 0 |  |  | $120 \mathrm{~dB} \mu \mathrm{~V}$ |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | 0 | 1 |  |  | $122 \mathrm{~dB} \mu \mathrm{~V}$ |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | 1 | 0 |  |  | $124 \mathrm{~dB} \mu \mathrm{~V}$ |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | Tuning DAC enable |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | 0 |  |  |  |  |  |  | off |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | 1 |  |  |  |  |  |  | on |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | Tuning DAC programming ${ }^{(1)}$ |
|  |  |  |  |  |  |  |  |  | 0 | 0 | 0 |  | 0 | 0 | 0 |  | 0 | 0 |  |  |  |  |  |  |  | 0 |
|  |  |  |  |  |  |  |  |  | 0 | 0 | 0 |  | 0 | 0 | 0 |  | 0 | 1 |  |  |  |  |  |  |  | 1 |
|  |  |  |  |  |  |  |  |  | $\cdots$ | ... | $\cdots$ |  | $\ldots$ | .. | . |  | $\cdots$ | $\ldots$ |  |  |  |  |  |  |  | $\ldots$ |
|  |  |  |  |  |  |  |  |  | 1 | 1 | 1 |  | 1 | 1 | 1 |  | 1 | 0 |  |  |  |  |  |  |  | 510 |
|  |  |  |  |  |  |  |  |  | 1 | 1 | 1 |  | 1 | 1 | 1 |  | 1 | 1 |  |  |  |  |  |  |  | 511 |

1. Normally handled by tuner drivers.

Table 25. Register 0x05

| Register number |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | Register definition |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| MSB LSB |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 23 | 22 | 21 | 20 | 19 | 18 | 17 | 16 | 15 | 14 | 13 |  | 12 | 11 | 10 | 9 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |  |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | PLLTEST output status |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | 0 | low |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | 1 | high |

## 5 Weak signal processing

All the parameters in this section refer to the programmability of the DSP part of the device. The typical values are those set by default parameters (start-up without parametric change from main $\mu \mathrm{P}$ ); the max and the min values refer to the programmability range. The values are referred to the typical application (Figure 16: Example of FM tuned (narrow-band) application / SPI control). Wherever the possible values are a discrete set, all the possible programmable values are displayed.

### 5.1 FM IF-processing

### 5.1.1 Dynamic channel selection filter (DISS)

Table 26. Dynamic channel selection filter (DISS)
(discrete set)

| Symbol | Parameter | Test condition | Min | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DISS BW | IF filter \#6 | response: - 3dB | - | $\pm 150$ | - | kHz |
|  | IF filter \#5 |  | - | $\pm 110$ | - | kHz |
|  | IF filter \#4 |  | - | $\pm 80$ | - | kHz |
|  | IF filter \#3 |  | - | $\pm 60$ | - | kHz |
|  | IF filter \#2 |  | - | $\pm 45$ | - | kHz |
|  | IF filter \#1 |  | - | $\pm 35$ | - | kHz |
|  | IF filter \#0 |  | - | $\pm 25$ | - | kHz |

### 5.1.2 Soft mute

Table 27. Soft mute

| (continuous set) |  |  |  |  |  |  |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| Symbol | Parameter | Test condition | Min | Typ | Max | Units |
| SMsp | Start point vs. field strength | audio atten =1 dB <br> read "FM_softmute" <br> no adjacent channel present | 0 | 6 | 20 | $\mathrm{~dB} \mu \mathrm{~V}$ |
| SMep | End point vs. field strength | audio atten = SMd + 1 dB <br> read "FM_softmute" <br> no adjacent channel present | -6 | -6 | 10 | $\mathrm{~dB} \mu \mathrm{~V}$ |
| SMd | Depth | - | -30 | -15 | 0 | dB |
| SMtauatt | Field strength LPF cut-off <br> frequency for soft mute <br> activation | - | 0.1 | 100 | 4000 | Hz |
| SMtaurel | Field strength LPF cut-off <br> frequency for soft mute <br> release | - | 0.1 | 1 | 4000 | Hz |

### 5.1.3 Adjacent channel mute

Table 28. Adjacent channel mute
(continuous set)

| Symbol | Parameter | Test condition | Min | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ACMd | Depth |  | SMd | 0 | 0 | $d B$ |

### 5.1.4 Stereo blend-

Table 29. Stereo blend
(continuous set)

| Symbol | Parameter | Test condition | Min | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| MaxSep | Maximum stereo separation | field strength $=80 \mathrm{~dB} \mu \mathrm{~V}$, pilot deviation $=6.75 \mathrm{kHz}$ | 0 | 40 | 50 | dB |
| SBFSsp | Start point vs. field strength | separation $=$ MaxSep -1 dB no multipath present | 20 | 50 | 60 | $\mathrm{dB} \mu \mathrm{V}$ |
| SBFSep | End point vs. field strength | separation $=1 \mathrm{~dB}$ no multipath present | 20 | 30 | 60 | $\mathrm{dB} \mu \mathrm{V}$ |
| SBFStM2S | Field strength-related transition time from mono to stereo | $\mathrm{V}_{\mathrm{rf}}$ step-like variation from <br> $20 \mathrm{~dB} \mu \mathrm{~V}$ to $80 \mathrm{~dB} \mu \mathrm{~V}$ | 0.001 | 3 | 20 | s |
| SBFStS2M | Field strength-related transition time from stereo to mono | $\mathrm{V}_{\mathrm{rf}}$ step-like variation from $80 \mathrm{~dB} \mu \mathrm{~V}$ to $20 \mathrm{~dB} \mu \mathrm{~V}$ | 0.001 | 0.5 | 20 | s |
| SBMPsp | Start point vs. multipath | $\begin{aligned} & \text { separation }=\text { MaxSep }-1 \mathrm{~dB} \\ & \text { equivalent } 19 \mathrm{kHz} \mathrm{AM} \\ & \text { modulation depth; } \\ & \text { field strength }=80 \mathrm{~dB} \mu \mathrm{~V} \end{aligned}$ | 5 | 10 | 80 | \% |
| SBMPep | End point vs. multipath | separation $=1 \mathrm{~dB}$ <br> equivalent 19 kHz AM <br> modulation depth; <br> field strength $=80 \mathrm{~dB} \mu \mathrm{~V}$ | 5 | 30 | 80 | \% |
| SBMPtM2S | Multipath -related transition time from mono to stereo | $\mathrm{V}_{\mathrm{rf}}$ step-like variation from $20 \mathrm{~dB} \mu \mathrm{~V}$ to $80 \mathrm{~dB} \mu \mathrm{~V}$ | 0.001 | 1 | 20 | s |
| SBMPtS2M | Multipath -related transition time from stereo to mono | $\mathrm{V}_{\mathrm{ff}}$ step-like variation from $80 \mathrm{~dB} \mu \mathrm{~V}$ to $20 \mathrm{~dB} \mu \mathrm{~V}$ | 0.001 | 0.001 | 20 | s |
| Pil ThrM2S | Pilot detector stereo threshold | Threshold on pilot tone deviation for mono-stereo transition | 0.8 | 2.74 | 7 | kHz |
| Pil ThrHyst | Pilot detector threshold hysteresis | Difference in pil. det. deviation threshold for stereo to mono transition compared to PilThrM2S | - | 0.01 | - | kHz |

### 5.1.5 High cut control

## Table 30. High cut control

(continuous set)

| Symbol | Parameter | Test condition | Min | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| HCFSsp | Start point vs. field strength | minimum RF level for widest HC filter (filter \# 7) no multipath present | 0 | 50 | 50 | $\mathrm{dB} \mu \mathrm{V}$ |
| HCFSep | End point vs. field strength | maximum RF level for narrowest HC filter (filter \# 0) no multipath present | 0 | 30 | 40 | $\mathrm{dB} \mu \mathrm{V}$ |
| HCFStW2N | Field strength-related transition time from wide to narrow band | $\mathrm{V}_{\mathrm{rf}}$ step-like variation from $60 \mathrm{~dB} \mu \mathrm{~V}$ to $10 \mathrm{~dB} \mu \mathrm{~V}$ | (1) |  |  | - |
| HCFStN2W | Field strength-related transition time from narrow to wide band | $\mathrm{V}_{\mathrm{rf}}$ step-like variation from $0 \mathrm{~dB} \mu \mathrm{~V}$ to $60 \mathrm{~dB} \mu \mathrm{~V}$ | (1) | 14 | 100 | s |
| HCMPsp | Start point vs. multipath | minimum RF level for widest HC filter (filter \# 7) equivalent 19 kHz AM modulation depth; field strength $=80 \mathrm{~dB} \mu \mathrm{~V}$ | 5 | 10 | $150{ }^{(2)}$ | \% |
| HCMPep | End point vs. multipath | maximum RF level for narrowest HC filter (filter \# 0) <br> equivalent 19 kHz AM <br> modulation depth; <br> field strength $=80 \mathrm{~dB} \mu \mathrm{~V}$ | 5 | 30 | $150{ }^{(2)}$ | \% |
| HCMPtN2W | Multipath -related transition time from narrow to wide band | $\mathrm{V}_{\mathrm{rf}}$ step-like variation from <br> $20 \mathrm{~dB} \mu \mathrm{~V}$ to $80 \mathrm{~dB} \mu \mathrm{~V}$ | 0.001 | 0.001 | 20 | s |
| HCMPtW2N | Multipath -related transition time from wide to narrow | $\mathrm{V}_{\mathrm{ff}}$ step-like variation from $80 \mathrm{~dB} \mu \mathrm{~V}$ to $20 \mathrm{~dB} \mu \mathrm{~V}$ | 0.001 | 0.001 | 20 | S |
| HCmaxBW | Maximum cut-off frequency of high cut filter bank | Filter \#7, -3 dB response frequency, input signal with pre-emphasis | HCmin BW | 14 | 18 | kHz |
| HCminBW | Minimum cut-off frequency of high cut filter bank | Filter \#0, -3 dB response frequency, input signal with pre-emphasis | 0.1 | 3 | HCma xBW | kHz |
| HCnumFilt | Number of discrete HC filters | - | - | $8{ }^{(3)}$ | - | - |

1. Depends only on field strength filter time constant.
2. Means that $100 \%$ equivalent 19 kHz AM modulation depth will not achieve full band narrowing
3. Intermediate filters (\#6-\#1) cut-off frequencies exponentially spaced between HCmaxBW and HCminBW.

Table 31. De-emphasis filter
(continuous set)

| Symbol | Parameter | Test condition | Min | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DEtc | De-emphasis time constant 1 | - | - | 50 | - | us |
|  | De-emphasis time constant 2 | - | - | 75 | - |  |

### 5.1.6 Stereo decoder

Table 32. Stereo decoder

| Symbol | Parameter | Test condition | Min | Typ | Max | Units |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| PilSup | Pilot signal suppression | Pilot $9 \%, 19 \mathrm{kHz}$, ref=40 kHz | - | 60 | - | dB |
| SubcSup | Subcarrier suppression | $\mathrm{f}=38 \mathrm{kHz}$ | - | 70 | - | dB |
|  |  | - | 70 | - | dB |  |
|  |  | - | 80 | - | dB |  |

### 5.2 AM IF-processing

### 5.2.1 Channel selection filter

Table 33. Channel selection filter

| Symbol | Parameter | Test condition | Min | Typ | Max | Units |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| CSF BW | Channel selection filter BW | response: -3 dB | - | $\pm 3.7$ | - | kHz |

### 5.2.2 Soft mute

Table 34. Soft mute
(continuous set)

| Symbol | Parameter | Test condition | Min | Typ | Max | Units |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| SMsp | Start point vs. field strength | audio atten =1 dB <br> read "FM_softmute" <br> no adjacent channel present | 0 | 25 | 40 | $\mathrm{~dB} \mu \mathrm{~V}$ |
| SMep | End point vs. field strength | audio atten = SMd + 1 dB <br> read "FM_softmute" <br> no adjacent channel present | 0 | 0 | 30 | $\mathrm{~dB} \mu \mathrm{~V}$ |
| SMd | Depth | - | -40 | -24 | 0 | dB |
| SMtauatt | Transition time for field <br> strength-dependent soft mute <br> activation | - | 0.001 | 0.1 | 10 | s |
| SMtaurel | Transition time for field <br> strength-dependent soft mute <br> release | - | 0.001 | 3 | 10 | s |

### 5.2.3 High cut control

## Table 35. High cut control

(continuous set)

| Symbol | Parameter | Test condition | Min | Typ | Max | Units |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| HCFSsp | Start point vs. field strength | minimum RF level for widest <br> HC filter (filter \# 7) <br> no multipath present | 0 | 40 | 50 | $\mathrm{~dB} \mu \mathrm{~V}$ |
| HCFSep | End point vs. field strength | maximum RF level for <br> narrowest HC filter (filter \# 0) <br> no multipath present | 0 | 30 | 50 | $\mathrm{~dB} \mu \mathrm{~V}$ |
| HCFStW2N | Field strength-related <br> transition time from wide to <br> narrow band | $\mathrm{V}_{\text {rf }}$ step-like variation from <br> 60 dB $\mu \mathrm{V}$ to 10 dB $\mu \mathrm{V}$ | 0.001 | 0.2 | 20 | s |
| HCFStN2W | Field strength-related <br> transition time from narrow to <br> wide band | $V_{\text {rf }}$ step-like variation from <br> 0 dB $\mu \mathrm{V}$ to 60 dB $\mu \mathrm{V}$ | 0.001 | 10 | 20 | s |
| HCmaxBW | Maximum cut-off frequency of <br> high cut filter bank | Filter \#7, -3 dB response <br> frequency, input signal with <br> pre-emphasis | HCmin |  |  |  |
| BW | 14 | 18 | kHz |  |  |  |
| HCminBW | Minimum cut-off frequency of <br> high cut filter bank | Filter \#0, -3 dB response <br> frequency, input signal with <br> pre-emphasis | 1 | 3 | HCma <br> xBW | kHz |
| HCnumFilt | Number of discrete HC filters |  | - | 8 | - | - |

## 6 Application schematics

### 6.1 Basic application schematic

Figure 15. FM wide-band application $/ \mathrm{I}^{2} \mathrm{C}$ control


1. Note: components marked with $a^{*}$ are being considered for replacement with resistors, pending optimization test results.

### 6.2 Application schematic example with SPI-bus and tuned preselection

Figure 16. Example of FM tuned (narrow-band) application / SPI control


1. Note: components marked with a * are being considered for replacement with resistors, pending optimization test results.

## 7 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK ${ }^{\circledR}$ packages, depending on their level of environmental compliance. ECOPACK ${ }^{\circledR}$ specifications, grade definitions and product status are available at: www.st.com.
ECOPACK ${ }^{\circledR}$ is an ST trademark.

Figure 17. LQFP64 (10x10x1.4mm) mechanical data and package dimensions

| DIM. | mm |  |  | inch |  |  |
| :---: | :---: | :---: | :---: | :--- | :--- | :--- |
|  | MIN. | TYP. | MAX. | MIN. | TYP. | MAX. |
| A |  |  | 1.60 |  |  | 0.063 |
| A1 | 0.05 |  | 0.15 | 0.002 |  | 0.006 |
| A2 | 1.35 | 1.40 | 1.45 | 0.053 | 0.055 | 0.057 |
| B | 0.17 | 0.22 | 0.27 | 0.0066 | 0.0086 | 0.0106 |
| C | 0.09 |  | 0.20 | 0.0035 |  | 0.0079 |
| D | 11.80 | 12.00 | 12.20 | 0.464 | 0.472 | 0.480 |
| D1 | 9.80 | 10.00 | 10.20 | 0.386 | 0.394 | 0.401 |
| D3 |  | 7.50 |  |  | 0.295 |  |
| e |  | 0.50 |  |  | 0.0197 |  |
| E | 11.80 | 12.00 | 12.20 | 0.464 | 0.472 | 0.480 |
| E1 | 9.80 | 10.00 | 10.20 | 0.386 | 0.394 | 0.401 |
| E3 |  | 7.50 |  |  | 0.295 |  |
| L | 0.45 | 0.60 | 0.75 | 0.0177 | 0.0236 | 0.0295 |
| L1 |  | 1.00 |  |  | 0.0393 |  |
| K |  | $0 \circ$ (min.), $3.5^{\circ}(\min ),. 7{ }^{\circ}(m a x)$. |  |  |  |  |
| ccc |  |  | 0.080 |  |  | 0.0031 |



LQFP64 (10 x $10 \times 1.4 \mathrm{~mm}$ )


## 8 Revision history

Table 36. Document revision history

| Date | Revision | Changes |
| :---: | :---: | :--- |
| 31-Jul-2007 | 1 | Initial release. |
| 01-Aug-2008 | 2 | Full update datasheet. |
| 08-May-2009 | 3 | Document status promoted from preliminary data to datasheet. <br> Updated Table 1: Device summary on page 1. <br> Updated Section 3: Electrical specifications on page 16. <br> Updated Section 4: Front-end processing on page 30. <br> Updated Section 5: Weak signal processing on page 33. <br> Updated Section 6: Application schematics on page 38. |
| 09-Jun-2009 | 4 | Updated Table 5: Thermal data on page 16. <br> Updated the value of "Adjacent channel selectivity" parameter in the <br> Table 17: FM overall system performance. |
| 01-Jul-2009 | 5 | Updated Figure 17: LQFP64 (10x10x1.4mm) mechanical data and <br> package dimensions on page 40. |
| 13-Jan-2010 | 6 | Modified Table 1: Device summary on page 1 <br> Modified Table 5: Thermal data on page 16. <br> Modified Section 3.5.5: WX overall system performance on page 29. <br> Modified Section 7: Package information on page 40. |
| 29-Jan-2010 | 7 | Minor text changes in Section 2.13. <br> Modified min. value of "thD-DAT" parameter in Table 15: I 2 C interface <br> on page 21. |
| 22-Mar-2010 | 8 | Added Section 3.4.11: Warning on page 22. |
| 17-Sep-2013 | 9 | Updated Disclaimer |

## Please Read Carefully:

Information in this document is provided solely in connection with ST products. STMicroelectronics NV and its subsidiaries ("ST") reserve the right to make changes, corrections, modifications or improvements, to this document, and the products and services described herein at any time, without notice.

All ST products are sold pursuant to ST's terms and conditions of sale.
Purchasers are solely responsible for the choice, selection and use of the ST products and services described herein, and ST assumes no liability whatsoever relating to the choice, selection or use of the ST products and services described herein.
No license, express or implied, by estoppel or otherwise, to any intellectual property rights is granted under this document. If any part of this document refers to any third party products or services it shall not be deemed a license grant by ST for the use of such third party products or services, or any intellectual property contained therein or considered as a warranty covering the use in any manner whatsoever of such third party products or services or any intellectual property contained therein.

UNLESS OTHERWISE SET FORTH IN ST'S TERMS AND CONDITIONS OF SALE ST DISCLAIMS ANY EXPRESS OR IMPLIED WARRANTY WITH RESPECT TO THE USE AND/OR SALE OF ST PRODUCTS INCLUDING WITHOUT LIMITATION IMPLIED WARRANTIES OF MERCHANTABILITY, FITNESS FOR A PARTICULAR PURPOSE (AND THEIR EQUIVALENTS UNDER THE LAWS OF ANY JURISDICTION), OR INFRINGEMENT OF ANY PATENT, COPYRIGHT OR OTHER INTELLECTUAL PROPERTY RIGHT.
ST PRODUCTS ARE NOT DESIGNED OR AUTHORIZED FOR USE IN: (A) SAFETY CRITICAL APPLICATIONS SUCH AS LIFE SUPPORTING, ACTIVE IMPLANTED DEVICES OR SYSTEMS WITH PRODUCT FUNCTIONAL SAFETY REQUIREMENTS; (B) AERONAUTIC APPLICATIONS; (C) AUTOMOTIVE APPLICATIONS OR ENVIRONMENTS, AND/OR (D) AEROSPACE APPLICATIONS OR ENVIRONMENTS. WHERE ST PRODUCTS ARE NOT DESIGNED FOR SUCH USE, THE PURCHASER SHALL USE PRODUCTS AT PURCHASER'S SOLE RISK, EVEN IF ST HAS BEEN INFORMED IN WRITING OF SUCH USAGE, UNLESS A PRODUCT IS EXPRESSLY DESIGNATED BY ST AS BEING INTENDED FOR "AUTOMOTIVE, AUTOMOTIVE SAFETY OR MEDICAL" INDUSTRY DOMAINS ACCORDING TO ST PRODUCT DESIGN SPECIFICATIONS. PRODUCTS FORMALLY ESCC, QML OR JAN QUALIFIED ARE DEEMED SUITABLE FOR USE IN AEROSPACE BY THE CORRESPONDING GOVERNMENTAL AGENCY.

Resale of ST products with provisions different from the statements and/or technical features set forth in this document shall immediately void any warranty granted by ST for the ST product or service described herein and shall not create or extend in any manner whatsoever, any liability of ST.

ST and the ST logo are trademarks or registered trademarks of ST in various countries.
Information in this document supersedes and replaces all information previously supplied.
The ST logo is a registered trademark of STMicroelectronics. All other names are the property of their respective owners.

$$
\text { © } 2013 \text { STMicroelectronics - All rights reserved }
$$

STMicroelectronics group of companies
Australia - Belgium - Brazil - Canada - China - Czech Republic - Finland - France - Germany - Hong Kong - India - Israel - Italy - Japan Malaysia - Malta - Morocco - Philippines - Singapore - Spain - Sweden - Switzerland - United Kingdom - United States of America
www.st.com


[^0]:    1. In the typical application supplied from 5 V with a series resistor.
    2. When the 1.2 V supply is applied externally, and not using the internal 1.2 V regulator.
[^1]:    1. The current is generated by a PTAT (Proportional To Absolute Temperature) source, and has therefore a temperature dependency described by: $\Delta I / l o=\Delta T / T o$, with lo being the current at ambient temperature $\left(25^{\circ} \mathrm{C}\right)$ and To the ambient temperature ( $25^{\circ} \mathrm{C}$ ) expressed in Kelvin, that is 298 K .
