# **ANALOG** GaAs, pHEMT, MMIC, 1 W, Power Amplifier with Power Detector, 37 GHz to 40 GHz

### **Data Sheet**

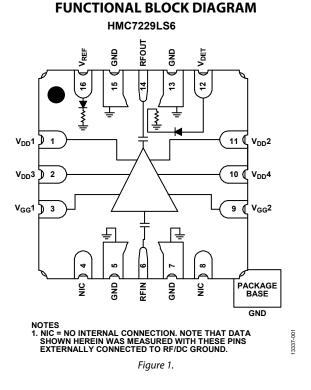
# HMC7229LS6

#### **FEATURES**

32 dBm typical saturated output power (P<sub>SAT</sub>) at 18% power added efficiency (PAE) at 39 GHz
P1dB compression output power: 31.5 dBm typical
High output third-order intercept (IP3): 40 dBm typical
High gain: 24 dB typical
50 Ω matched input/output
Ceramic, 6 mm × 6 mm, high frequency, air cavity package

#### **APPLICATIONS**

Point to point radios Point to multipoint radios Very small aperture terminal (VSAT) and satellite communications (SATCOM)



#### **GENERAL DESCRIPTION**

The HMC7229LS6 is a four stage, gallium arsenide (GaAs), pseudomorphic high electron mobility transfer (pHEMT), monolithic microwave integrated circuit (MMIC), 1 W power amplifier, with an integrated temperature compensated on-chip power detector that operates between 37 GHz to 40 GHz. The HMC7229LS6 provides 24 dB of gain and 32 dBm of saturated output power at 18% PAE at 39 GHz from a 6 V supply. With an excellent IP3 of 40 dBm, the HMC7229LS6 is ideal for linear

applications such as high capacity, point to point or multipoint radios or VSAT/SATCOM applications demanding 32 dBm of efficient saturated output power. The radio frequency (RF) input/outputs are internally matched and dc blocked for ease of integration into higher level assemblies. The HMC7229LS6 is housed in a ceramic, 6 mm × 6 mm, high frequency, air cavity package that exhibits low thermal resistance and is compatible with surface-mount manufacturing techniques.

#### Rev. E

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#### **REVISION HISTORY**

#### 8/2020-Rev. D to Rev. E

Added Minimum Gate Voltage (V <sub>GG1</sub> , V <sub>GG2</sub> ) Parameter,	
Table 2	4
Updated Outline Dimensions	
Changes to Ordering Guide	15
12/2017—Rev. C to Rev. D	

Changes to Figure 3311
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#### 9/2017—Rev. B to Rev. C

Changes to Theory of Operation Section and Figure 33	11
Changes to Ordering Guide	15

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#### 6/2016—Rev. v01.0514 to Rev. B This Hittite Microw 1 n. 1

6/2010—Rev. v01.0514 to Rev. B
This Hittite Microwave Products data sheet has been reformatted
to meet the styles and standards of Analog Devices, Inc.
Updated FormatUniversal
Changes to General Description Section1
Changes to Table 1
Added Theory of Operation Section 11
Added Applications Information Section and Recommended
Bias Sequence Section
Updated Outline Dimensions 15
Changes to Ordering Guide 15

### **SPECIFICATIONS** ELECTRICAL SPECIFICATIONS

 $T_A = 25^{\circ}$ C,  $V_{DD}1$  to  $V_{DD}4$  ( $V_{DD}x$ ) = 6 V,  $I_{DD} = 1200$  mA (adjust  $V_{GG}1/V_{GG}2$ ,  $V_{GG}x$ , between -2 V to 0 V to achieve an  $I_{DD} = 1200$  mA typical), unless otherwise stated.

Parameter	Test Conditions/Comments	Min	Тур	Max	Unit
FREQUENCY RANGE		37		40	GHz
GAIN		21	24		dB
Gain Variation over Temperature			0.058		dB/°C
RETURN LOSS					
Input			16		dB
Output			14		dB
OUTPUT POWER					
For P1dB Compression		28.5	31.5		dBm
Saturated (P <sub>SAT</sub> )	With 18% PAE at 39 GHz		32		dBm
OUTPUT THIRD-ORDER INTERCEPT (IP3)	Measurement taken at Pout/tone = 20 dBm		40		dBm
TOTAL SUPPLY CURRENT (I <sub>DD</sub> ) <sup>1</sup>	$V_{DD} = 5 \text{ V}, V_{DD} = 5.5 \text{ V}, \text{ and } V_{DD} = 6 \text{ V}$		1200		mA

<sup>1</sup> Adjust  $V_{GG}x$  to achieve  $I_{DD} = 1200$  mA.

# **ABSOLUTE MAXIMUM RATINGS**

#### Table 2.

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Parameter	Rating
Drain Bias Voltage (V <sub>DD</sub> x)	7 V
Minimum Gate Voltage (V <sub>GG1</sub> , V <sub>GG2</sub> )	-3 V
RF Input Power (RFIN)	21 dBm
Channel Temperature	175℃
Continuous Power Dissipation, P <sub>DISS</sub> (T = 85°C, Derates 95 mW/°C Above 85°C)	9.0 W
Thermal Resistance (Channel to Ground Paddle)	10°C/W
Storage Temperature Range	–65°C to +150°C
Operating Temperature Range	-40°C to +85°C
ESD Sensitivity (Human Body Model)	Class 0, passed 150 V

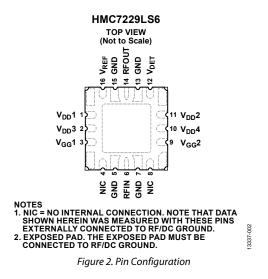
Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

#### **ESD CAUTION**



**ESD (electrostatic discharge) sensitive device.** Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

# **PIN CONFIGURATION AND FUNCTION DESCRIPTIONS**



#### Table 3. Pin Function Descriptions

Pin No.	Mnemonic	Description	
1, 2, 10, 11	$V_{DD}1$ to $V_{DD}4$	Drain Bias Voltages. External bypass capacitors of 100 pF, 10 nF, and 4.7 µF are required for each of these pins. See Figure 3.	
3, 9	$V_{GG}1, V_{GG}2$	Gate Controls for the Power Amplifier. Adjust V <sub>GG</sub> x to achieve the recommended bias current. External by capacitors of 100 pF, 10 nF, and 4.7 μF are required for each of these pins. Apply V <sub>GG</sub> x bias to either Pin 3 or Pir See Figure 4.	
4, 8	NIC	No Internal Connection. Note that data shown herein was measured with these pins externally connected to RF/dc ground.	
5, 7, 13, 15	GND	Ground Pins. Connect these pins and the exposed ground pad to RF/dc ground. See Figure 5.	
6	RFIN	RF Input. This pin is ac-coupled and matched to 50 $\Omega$ . See Figure 6.	
12	Vdet	Detector Voltage. This pin is the dc voltage that represents the RF output power rectified by the diode that is biased through an external resistor. See Figure 8.	
14	RFOUT	RF Output. This pin is ac-coupled and matched to 50 $\Omega$ . See Figure 9.	
16	V <sub>REF</sub>	Detector Reference Voltage. This pin is the dc voltage of the diode biased through an external resistor used for the temperature compensation of $V_{DET}$ . See Figure 7.	
	EPAD	Exposed Pad. The exposed pad must be connected to RF/dc ground.	

#### **INTERFACE SCHEMATICS**



Figure 3.  $V_{DD}$ 1,  $V_{DD}$ 2,  $V_{DD}$ 3, and  $V_{DD}$ 4 Interface Schematic

13337-003



Figure 4. V<sub>GG</sub>1 and V<sub>GG</sub>2 Interface Schematic

GND 13337-005 ļ

Figure 5. GND Interface Schematic



Figure 6. RFIN Interface Schematic

V<sub>REF</sub> <sup>№</sup> – – – – – – – – Figure 7. V<sub>REF</sub> Interface Schematic

VDET VDET → Figure 8. VDET Interface Schematic



Figure 9. RFOUT Interface Schematic

# **TYPICAL PERFORMANCE CHARACTERISTICS**

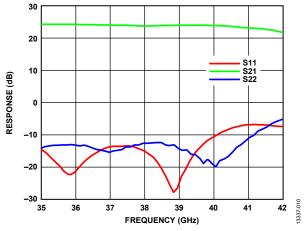


Figure 10. Gain and Return Loss (S11, S21, and S22) vs. Frequency

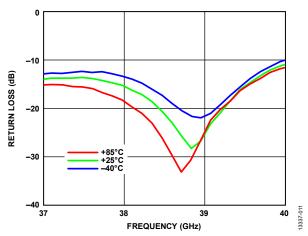


Figure 11. Input Return Loss vs. Frequency for Various Temperatures

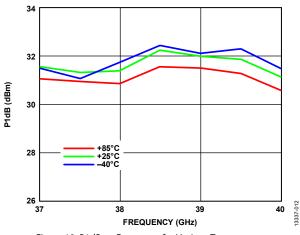
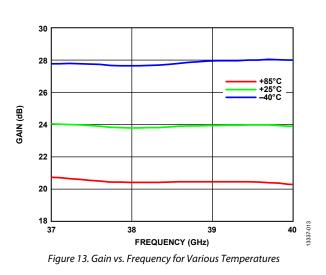


Figure 12. P1dB vs. Frequency for Various Temperatures



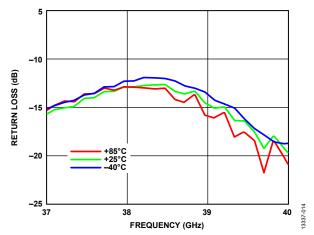


Figure 14. Output Return Loss vs. Frequency for Various Temperatures

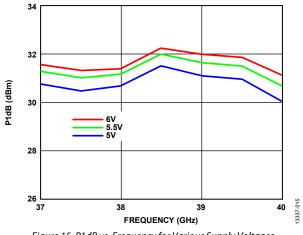
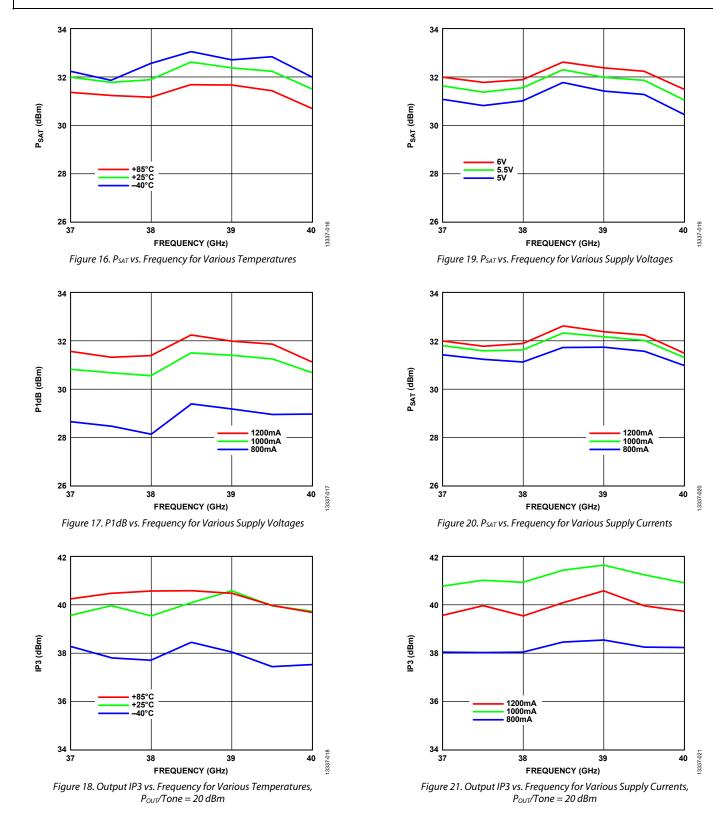
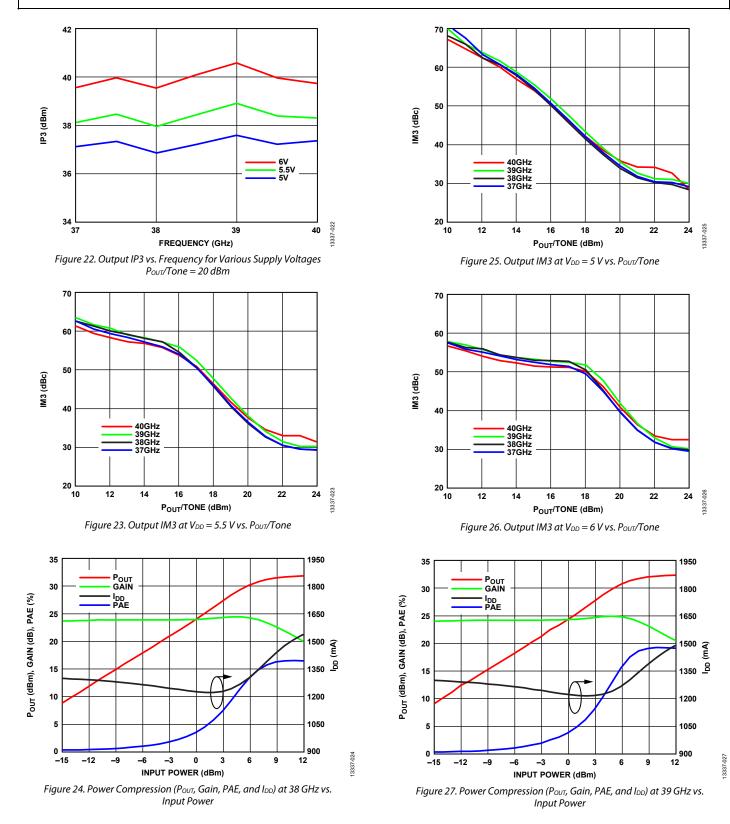


Figure 15. P1dB vs. Frequency for Various Supply Voltages



### **Data Sheet**

# HMC7229LS6



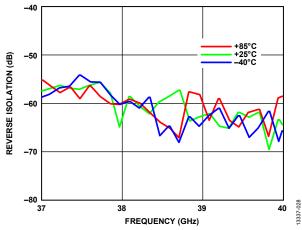


Figure 28. Reverse Isolation vs. Frequency for Various Temperatures

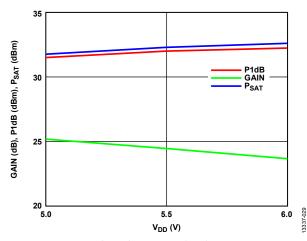


Figure 29. Gain, P1dB, and PSAT vs. Supply Voltage (VDD) at 38.5 GHz

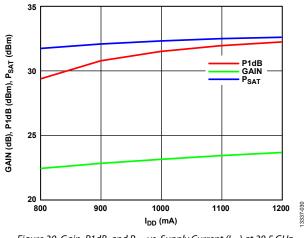
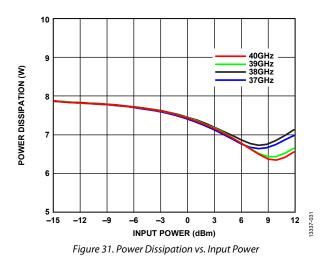


Figure 30. Gain, P1dB, and PSAT vs. Supply Current (IDD) at 38.5 GHz



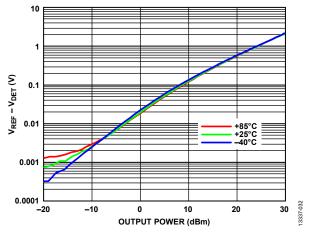


Figure 32. Detector Voltage ( $V_{REF} - V_{DET}$ ) vs. Output Power for Various Temperatures at 38.5 GHz

# THEORY OF OPERATION

The HMC7229LS6 is gallium arsenide (GaAs), pseudomorphic high electron mobility transfer (pHEMT), monolithic microwave integrated circuit (MMIC), 1 W power amplifier consisting of four gain stages that are in series. Figure 33 shows the simplified block diagram.

The input signal is divided evenly into two, each of these paths are amplified through four independent gain stages, and the amplified signals are then combined at the output. A portion of the RF output signal is directionally coupled to a diode for detection of the RF output power. When the diode is dc biased, it rectifies the RF power and makes it available for measurement as a dc voltage at VDET. To allow for temperature compensation of VDET, a identical and symmetrically located circuit, minus the coupled RF power, is available via VREF. Taking the difference of VREF – VDET provides a temperature compensated signal that is proportional to the RF output. (See Figure 32.) The HMC7229LS6 has single-ended input and output ports whose impedances are nominally matched to 50  $\Omega$  internally over the 37 GHz to 40 GHz frequency range. Consequently, the HMC7229LS6 can directly insert into a 50  $\Omega$  system with no impedance matching circuitry required. In addition, multiple HMC7229LS6 devices can be cascaded back to back without requiring external matching circuitry. Similarly, multiple HMC7229LS6 devices can be used with power dividers at the input and power combiners at the output to obtain higher output power levels.

Because the input and output impedances are sufficiently stable vs. the variations in temperature and supply voltage, no impedance matching compensation is required.

To achieve the best performance and not damage the HMC7229LS6, do not exceed the absolute maximum ratings.

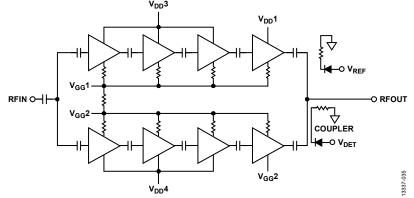


Figure 33. GaAs, pHEMT, 1 W MMIC Power Amplifier Block Diagram

# **APPLICATIONS INFORMATION**

Figure 35 shows the basic connections for operating the HMC7229LS6 and see the Theory of Operation section for additional details. The RF input and RF output are ac-coupled by the internal dc block capacitors. To avoid damaging the HMC7229LS6, follow the recommended bias sequencing during power-up and power-down.

The gate bias of the HMC7229LS6 is supplied by using either the  $V_{GG1}$  pin or the  $V_{GG2}$  pin. While applying drain bias to the HMC7229LS6, all of the  $V_{DD1}$ ,  $V_{DD2}$ ,  $V_{DD3}$ , and  $V_{DD4}$  pins must be used.

#### **RECOMMENDED BIAS SEQUENCE**

#### **During Power-Up**

The recommended bias sequence during power-up is the following:

- 1. Connect the GND pin to ground.
- 2. Set  $V_{GG}x$  to -2 V.
- 3. Set  $V_{DD}x$  to 6 V.
- 4. Increase  $V_{GGX}$  to achieve a typical  $I_{DD} = 1200$  mA.
- 5. Apply the RF signal.

#### **During Power-Down**

The recommended bias sequence during power-down is the following:

- 1. Turn the RF signal off.
- 2. Decrease  $V_{GGX}$  to -2 V to achieve a typical  $I_{DD} = 0 \text{ mA}$ .
- 3. Decrease  $V_{DD}x$  to 0 V.
- 4. Increase  $V_{GG}x$  to 0 V.

The bias conditions previously listed ( $V_{DDx} = 6 \text{ V}$ ,  $I_{DD} = 1200 \text{ mA}$ ), are the recommended operating point to get optimum performance. The data used in this data sheet was taken with the recommended bias condition. When using the HMC7229LS6 with different bias conditions, different performance may result than what is shown in the Typical Performance Characteristics section.

The  $V_{DET}$  and  $V_{REF}$  pins are the output pins for the internal power detector. The  $V_{DET}$  pin is the dc voltage output pin that represents the RF output power rectified by the internal diode, which is biased through an external resistor.

The  $V_{REF}$  pin is the dc voltage output pin that represents the reference diode voltage, which is biased through an external resistor. This voltage then compensates for the temperature variation effects on both diodes. A typical circuit is shown in the Typical Application Circuit section that reads out the output voltage and represents the RF output power shown in Figure 35.

#### **EVALUATION PRINTED CIRCUIT BOARD (PCB)**

Use RF circuit design techniques to create the circuit board for this application. Ensure that signal lines have 50  $\Omega$  impedance, and connect the package ground leads and exposed paddle directly to the ground plane similar to that shown in Figure 35. Use a sufficient number of via holes to connect the top and bottom ground planes. The evaluation circuit board shown in Figure 34 is available from Analog Devices, Inc., upon request.

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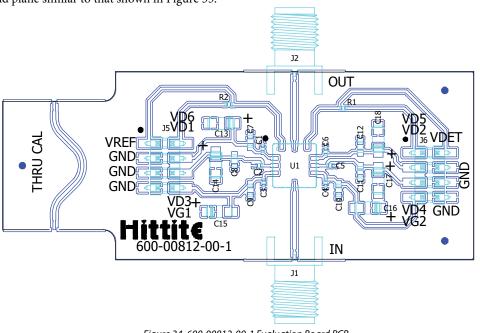


Figure 34. 600-00812-00-1 Evaluation Board PCB

Table 4. List of Materials f	for Evaluation PCB
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ltem	Description
J1, J2	K connector, SRI
J5, J6	DC pin
C1 to C6	100 pF capacitors, 0402 package
C7 to C12	10 nF capacitors, 0603 package
C13 to C18	4.7 μF capacitors, Case A package
R1, R2	40.2 k $\Omega$ resistors, 0402 package
U1	HMC7229LS6 amplifier
РСВ	600-00812-00-1 evaluation board PCB, circuit board material is Rogers 4350 or Arlon 25FR

# **TYPICAL APPLICATION CIRCUIT**

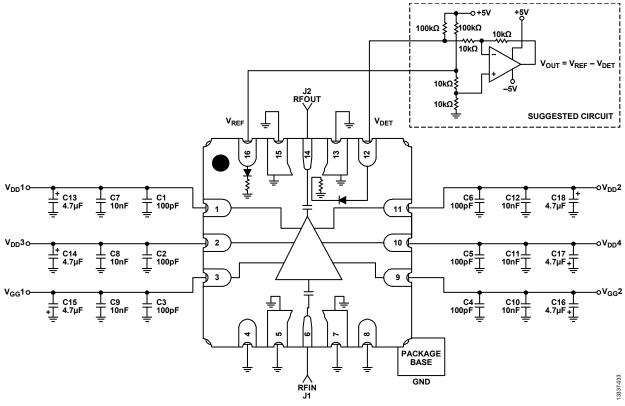


Figure 35. Typical Application Circuit

# **OUTLINE DIMENSIONS**

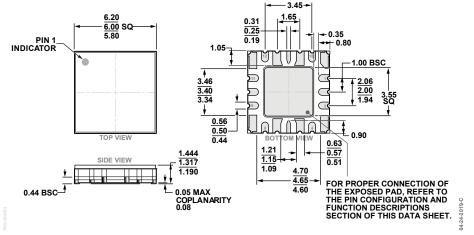


Figure 36. 16-Terminal Ceramic Leadless Chip with Heat Sink [LCC\_HS] (EH-16-1) Dimensions shown in millimeters

#### **ORDERING GUIDE**

Model <sup>1</sup>	Package Body Material	Lead Finish	Temperature Range	Package Description	Package Option
HMC7229LS6	Alumina White	Gold over Nickel	-40°C to +85°C	16-Terminal LCC_HS	EH-16-1
HMC7229LS6TR	Alumina White	Gold over Nickel	-40°C to +85°C	16-Terminal LCC_HS	EH-16-1
EVAL01-HMC7229LS6				<b>Evaluation Board</b>	

<sup>1</sup> All Models are RoHS-Compliant.

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