

DC Brushless Fan Motor Drivers Multifunction Single-phase Full-wave Fan Motor Driver

BD61243FV

General Description

BD61243FV is a 1chip driver that is composed of H-bridge power DMOS FET. Moreover, the circuit configuration is restructured, and convenience has been improved by reducing the external parts and simplifying the setting compared with the conventional driver.

Features

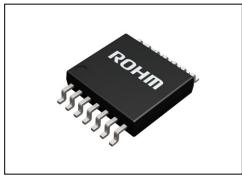
- SSOP Small Package
- Driver Including Power DMOS FET
- Speed Controllable by DC / PWM Input
- I/O Duty Slope Adjust
- PWM Soft Switching
- Current Limit
- Start Duty Assist
- Lock Protection and Automatic Restart
- Quick Start
- Rotation Speed Pulse Signal (FG) Output

Key Specifications

- Operating Voltage Range: 5.5V to 16V
 Operating Temperature Range: -40°C to +105°C
- Operating remperature Range: -40°C to +10;
 - Output Voltage(total): 0.2V(Typ) at 0.2A

Package

W (Typ) x D (Typ) x H (Max) 5.00mm x 6.40mm x 1.35mm



SSOP-B14

Applications

Fan motors for general consumer equipment of desktop PC, Projector, etc.

Typical Application Circuits

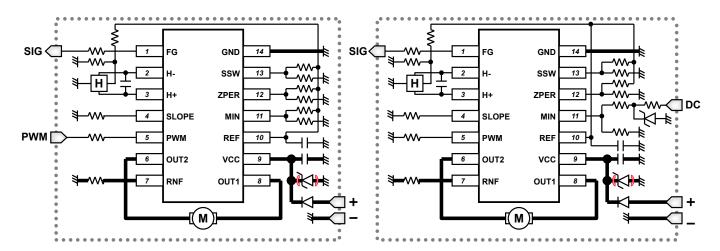
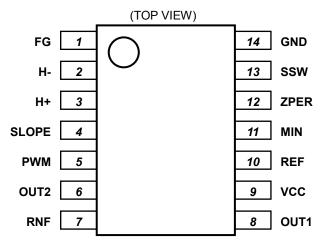


Figure 1. Application of Direct PWM Input

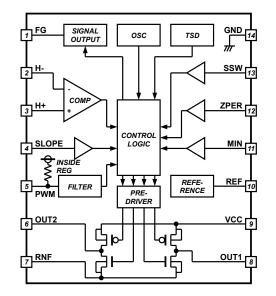
Figure 2. Application of DC Voltage Input

OProduct structure : Silicon monolithic integrated circuit OThis product has no designed protection against radioactive rays

Pin Configuration



Block Diagram



Pin Description

P/No.	T/Name	Function
1	FG	Speed pulse signal output terminal
2	H–	Hall – input terminal
3	H+	Hall + input terminal
4	SLOPE	I/O duty slope setting terminal
5	PWM	PWM input duty terminal
6	OUT2	Motor output terminal 2
7		Output current detecting resistor
1	RNF	connecting terminal (motor ground)
8	OUT1	Motor output terminal 1
9	VCC	Power supply terminal
10	REF	Reference voltage output terminal
11	MIN	Minimum output duty setting terminal
12	ZPER	Re-circulate period setting terminal
13	SSW	Soft switching setting terminal
14	GND	Ground terminal (signal ground)

I/O Truth Table

Hall	Input	Driver Output				
H+ H–		OUT1	OUT2	FG		
Н	H L		Н	Hi-Z		
L	Н	Н	L	L		

H; High, L; Low, Hi-Z; High impedance FG output is open-drain type.

Absolute Maximum Ratings

Parameter	Symbol	Rating	Unit
Supply Voltage	Vcc	18	V
Power Dissipation	Pd	0.87 (Note 1)	W
Operating Temperature Range	Topr	-40 to +105	°C
Storage Temperature Range	Tstg	–55 to +150	°C
Output Voltage	Vomax	18	V
Output Current	IOMAX	1.2 (Note 2)	А
Rotation Speed Pulse Signal (FG) Output Voltage	VFG	18	V
Rotation Speed Pulse Signal (FG) Output Current	IFG	10	mA
Reference Voltage (REF) Output Current	IREF	10	mA
Input Voltage1 (H+, H–,MIN,SSW,ZPER,SLOPE)	V _{IN1}	4	V
Input Voltage2 (PWM)	V _{IN2}	6.5	V
Junction Temperature	Tj	150	°C

(Note 1) Derate by 7.0mW/°C when operating over Ta=25°C. (Mounted on 70.0mm×70.0mm×1.6mm glass epoxy board) (Note 2) Do not exceed Pd and Tj=150°C.

Caution: Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.

Recommended Operating Conditions

Parameter	Symbol	Min	Тур	Max	Unit
Operating Supply Voltage Range	Vcc	5.5	12	16	V
Input Voltage Range1	N	0		V 10.2	Ň
(H+, H–, MIN, SSW, ZPER, SLOPE)	V _{IN1}	0	-	V_{REF} +0.3	v
Input Voltage Range2 (PWM)	VIN2	-0.3	-	5	V
PWM Input Duty Range	D _{PWM}	0	-	100	%
PWM Input Frequency Range	fрwм	15	-	50	kHz

Electrical Characteristics (Unless otherwise specified Ta=25°C, V_{cc}=12V)

Deremeter	Symbol	Limit		1.1	Conditions	Characteristic	
Parameter	Symbol	Min	Тур	Max	Unit	Conditions	Data
Circuit Current	Icc	3.0	4.5	6.5	mA		Figure 3
Output Voltage	Vo	-	0.2	0.35	V	I ₀ =±200mA,	Figure 4 to
Culput Voltage	VO					High and low side total	Figure 7
Lock Detection ON Time	ton	0.3	0.5	0.7	S		Figure 8
Lock Detection OFF Time	toff	3.0	5.0	7.0	S		Figure 9
Lock Detection OFF/ON Ratio	RLCK	8.5	10.0	11.5	-	RLCK=tOFF / tON	Figure 10
FG Hysteresis Voltage+	V _{HYS+}	7	12	17	mV	F inance	
FG Hysteresis Voltage-	V _{HYS-}	-5	-10	-15	mV		Figure 11
FG Output Low Voltage	V _{FGL}	-	-	0.30	V	I _{FG} =5mA	Figure 12 to
TO Output Eow Voltage	VFGL						Figure 13
FG Output Leak Current	IFGL	-	-	10	μA	V _{FG} =16V	Figure 14
PWM Input High Level Voltage	VPWMH	2.5	-	5.0	V		-
PWM Input Low Level Voltage	VPWML	0.0	-	1.0	V		-
DWW lanut Current	Iрwмн	-10	0	10	μA	V _{PWM} =5V	Figure 15 to
PWM Input Current	PWML	-50	-25	-12	μA	V _{PWM} =0V	Figure 16
Deference Voltage	N/	2.0	3.3	3.6	V	I _{REF} =-1mA	Figure 17 to
Reference Voltage	VREF	3.0					Figure 18
Current Limit Setting Voltage	VcL	235	265	295	mV		Figure 19

For parameters involving current, positive notation means inflow of current to IC while negative notation means outflow of current from IC.

Typical Performance Curves (Reference Data)

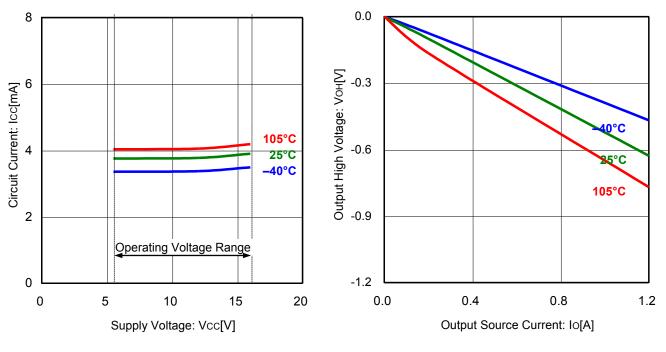


Figure 3. Circuit Current vs Supply Voltage

Figure 4. Output High Voltage vs Output Source Current $(V_{CC}=12V)$

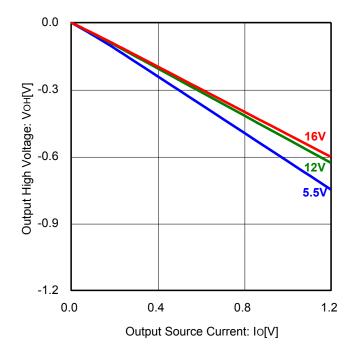


Figure 5. Output High Voltage vs Output Source Current

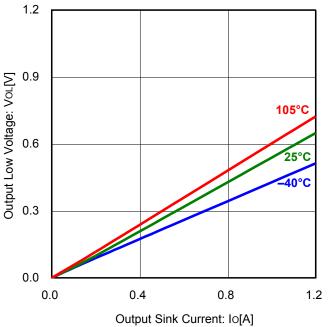


Figure 6. Output Low Voltage vs Output Sink Current (V_{CC} =12V)

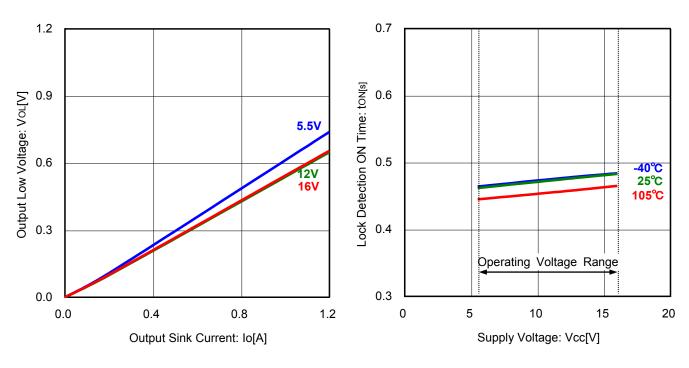


Figure 7. Output Low Voltage vs Output Sink Current (Ta=25°C)

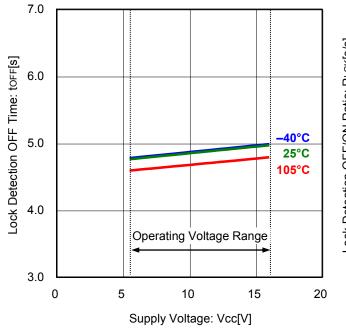


Figure 9. Lock Detection OFF Time vs Supply Voltage

Figure 8. Lock Detection ON Time vs Supply Voltage

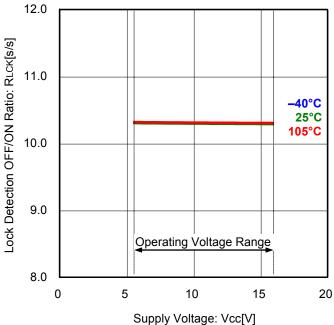


Figure 10. Lock Detection OFF/ON Ratio vs Supply Voltage

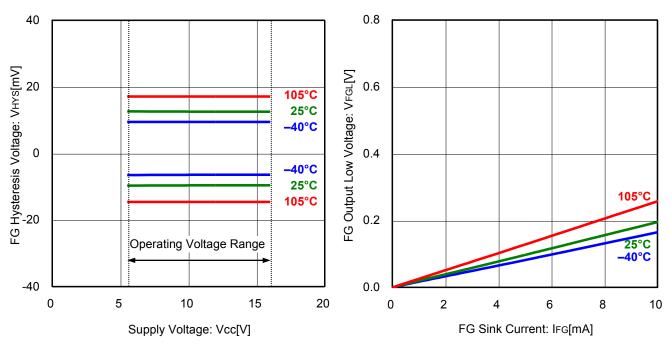


Figure 11. FG Hysteresis Voltage vs Supply Voltage

Figure 12. FG Output Low Voltage vs FG Sink Current (Vcc=12V)

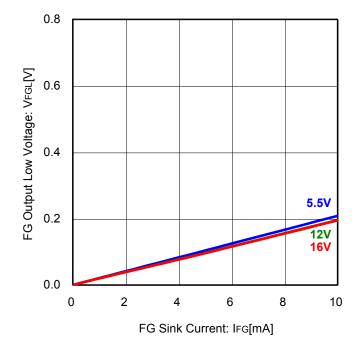


Figure 13. FG Output Voltage vs FG Sink Current (Ta=25°C)

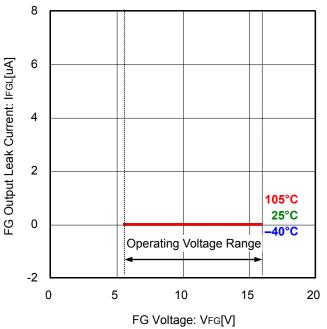
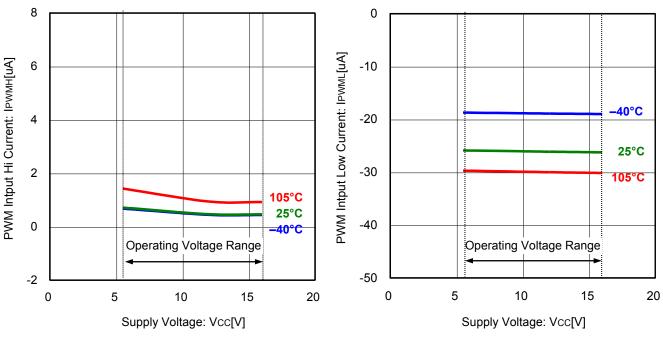


Figure 14. FG Output Leak Current vs FG Voltage



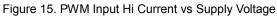


Figure 16. PWM Input Low Current vs Supply Voltage

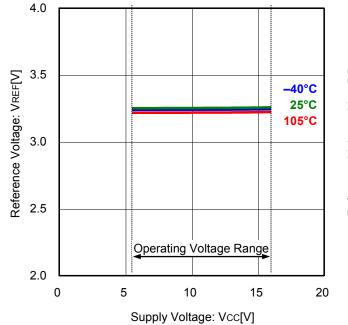


Figure 17. Reference Voltage vs Supply Voltage (IREF=-1mA)

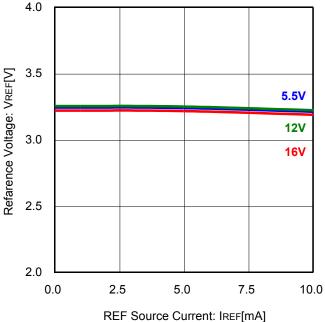


Figure 18. Reference Voltage vs REF Source Current (Ta=25°C)

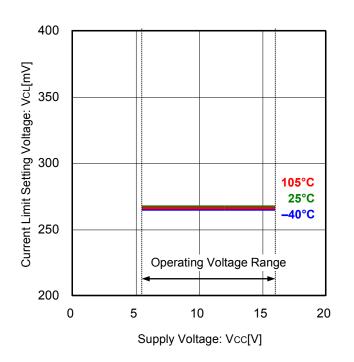


Figure 19. Current Limit Setting Voltage vs Supply Voltage

Application Circuit Examples (Constant Values are for Reference)

- 1. PWM Input Application
 - It is an example of the application of the external PWM input, and controlling the rotational speed. In this application, minimum rotational speed can be set.

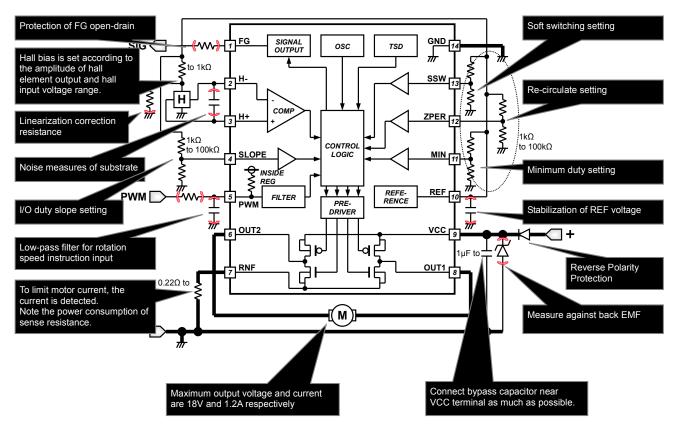


Figure 20. PWM Input Application

Application Design Note

(1) The bypass capacitor connected must be more than the recommended constant value because there is a possibility of the motor start-up failure etc. due to IC malfunction.

Substrate Design Note

- (1) IC power(Vcc), motor power(Vm), motor outputs(OUT1, 2), and motor ground(MGND) lines are made as wide as possible.
- (2) IC ground (GND) line is common with the application ground except motor ground (i.e. hall ground etc.), and arranged near to (-) land.
- (3) The bypass capacitor and/or Zenner diode are placed near to VCC pin.
- (4) H+ and H– lines are arranged side by side and made from the hall element to IC as short as possible, because it is easy for the noise to influence the hall lines.

Application Circuit Examples (Constant Values are for Reference) – continued

- 2. DC Voltage Input Application
 - This is an example application circuit for the rotation speed control by DC voltage. In this application, minimum rotational speed cannot be set.

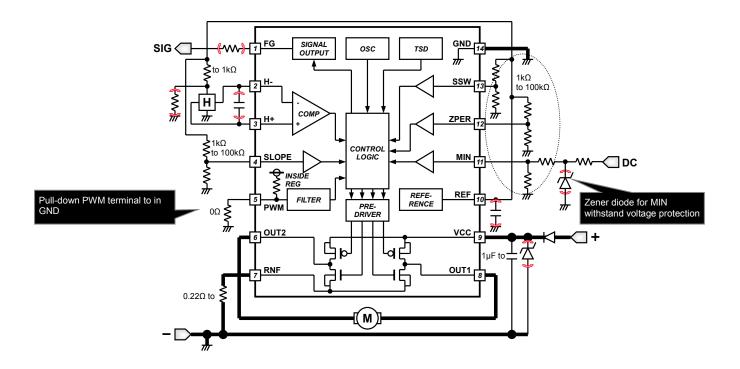


Figure 21. DC Voltage Input Application

Functional Descriptions

1. Variable Speed Operation

The rotational speed of the motor changes by the PWM duty of the motor outputs (OUT1 and OUT2 terminals). However, it provides for the motor's output not by the rotational speed but by the duty in the BD61243FV, because the rotational speed is not uniquely decided by the motor output duty.

The changeable speed operation is controlled by these two input terminals.

- (1) PWM Operation by Pulse Input in PWM Terminal
- (2) PWM Operation by DC Input in MIN Terminal
- (Note) PWM frequency of output is 50kHz (Typ). Hence, input PWM frequency is not equal to PWM frequency of output.
- (1) PWM Operation by Pulse Input in PWM Terminal

The PWM signal from the controller can be input directly to IC in Figure 22. The output duty is controlled by the input PWM duty (Figure 23). Refer to recommended operating conditions (P.3) and electrical characteristics (P.4) for the input condition.

Internal power-supply voltage (INTERNAL REG; typ 5.0V) is impressed when the PWM terminal is open, it becomes 100% input of the duty and equivalent, and a full torque is driven. There must be a pull- down resistance outside of IC to make it to torque 0 when the PWM terminal opens (However, only at the controller of the complimentary output type.). Insert the protective resistance and capacitor for noise removal if necessary.

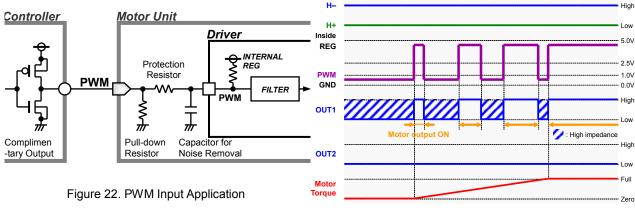


Figure 23. PWM Input Operation Timing Chart

Full torque (V_{PWM} >2.5V) and zero torque(<1.0V) can recognize the DC voltage input of the PWM terminal. However, the variable speed control in the DC voltage between 0V and 5.0V should be not able to be done.

(a) Setting of Minimum Output Duty (MIN)

Minimum rotational speed can be set by MIN terminal in Figure 24. The resolution of the MIN terminal is 128 steps. MIN terminal should be shorted to GND when this function is not used.

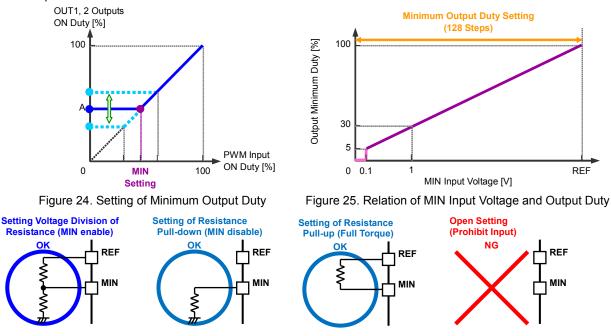


Figure 26. MIN Terminal Setting

(b) Setting of Slope of I/O Duty (SLOPE)

Slope of output duty and the input duty to PWM terminal can be established by SLOPE setting in Figure 27. The resolution of MIN is 128 steps. But if the voltage of the SLOPE terminal is 0.4V to 0.825V (Typ), then the slope of the input and output duty is fixed to 0.5, and if it is less than 0.4V (Typ) the slope is fixed to 1 (Figure 28). SLOPE terminal should be shorted to GND, when this function is not used.

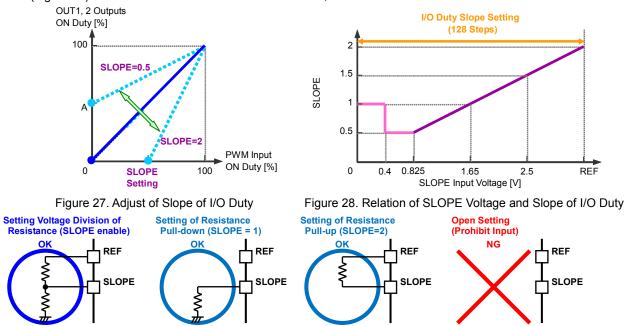


Figure 29. SLOPE Terminal Setting

When you perform the minimum rotation speed setting while making SLOPE setting, please decide MIN voltage based on a lower expression.

MIN Voltage = 3.3 x {(Target Minimum Duty - 100 + 100 x SLOPE) / (100 x SLOPE)} ····Equation 1

(ex.) In the case of SLOPE=1 and Target minimum duty=20%, calculate the SLOPE and MIN voltage in the following.

The SLOPE voltage sets with SLOPE=0V from Figure 28. The MIN voltage from Equation 1; MIN = $3.3 \times \{(20 - 100 + 100 \times 1) / (100 \times 1)\}$ = 0.66[V]

(ex.) In the case of SLOPE=0.75 and Target minimum duty=40%, calculate the SLOPE and MIN voltage in the following.

The SLOPE voltage sets with SLOPE=1.24V from Figure 28. The MIN voltage from Equation 1; MIN = $3.3 \times \{(40 - 100 + 100 \times 0.75) / (100 \times 0.75)\}$ = 0.66[V]

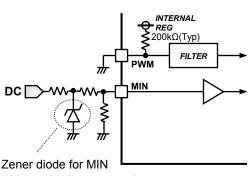
(ex.) In the case of SLOPE=1.75 and Target minimum duty=30%, calculate the SLOPE and MIN voltage in the following.

```
The SLOPE voltage sets with SLOPE=2.9V from Figure 28.
The MIN voltage from Equation 1;
MIN = 3.3 \times \{(30 - 100 + 100 \times 1.75) / (100 \times 1.75)\}
= 1.98[V]
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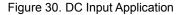
(2) PWM Operation by DC Input in MIN Terminal

The output duty can be varied by inputting DC voltage into MIN terminal. PWM terminal should be shorted to GND when this function is used. Please refer to input voltage range 1(P.3) for the input condition of the MIN terminal. MIN Terminal voltage becomes unsettled when MIN terminal is in an open state. The voltage of the terminal becomes irregular if MIN terminal is open. Input voltages to MIN terminals when you turn on IC power supply (V_{CC}) in Figure 30.

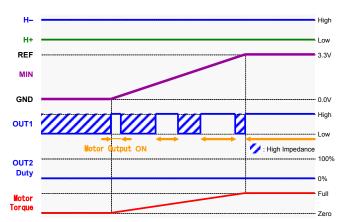
(Note)In the case of DC voltage input, it cannot set the lowest output duty.

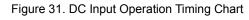


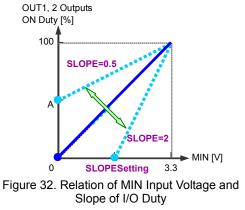
withstand voltage protection



(a) Setting of Slope of I/O Duty (SLOPE) Slope of output duty and the input voltage to MIN terminal can be established by SLOPE setting in Figure 32. The resolution of SLOPE is 128 steps. But if the voltage of the SLOPE terminal is 0.4V to 0.825V (Typ), then the slope of the input and output duty is fixed to 0.5, and if it is less than 0.4V (Typ) the slope is fixed to 1 (Figure 28). SLOPE terminal should be shorted to GND, when this function is not used.







- 2. About Setting of Phase Switching of Output
 - The period of Soft switching and re-circulate can be adjusted by SSW and ZPER setting.
 - (1) Soft Switching Period Setting (SSW)

The soft switching section in the output can be set by SSW terminal. By adjusting SSW voltage, soft switching section can be set from 22.5° to 90° as one period of hall signal 360°. The resolution of SSW is 128 steps in Figure 34. Timing chart is shown in Figure 33.

(Note)A soft switching period is the section where ON duty of the output changes from a target duty into 0% by 16 steps.

Adjust a Soft Switching Period by SSW Setting

Setable Range : Min22.5° to Max90°

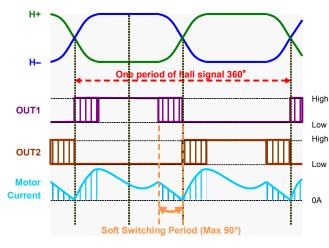


Figure 33. Soft switching Period setting

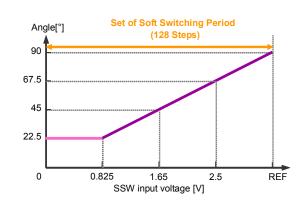


Figure 34. Relation of SSW Input Voltage and Soft Switching Period

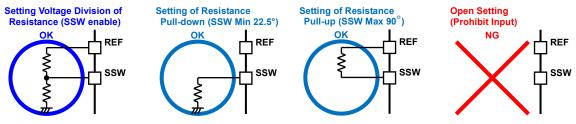


Figure 35. SSW Terminal Setting

(2) Re-circulate Period Setting (ZPER)

The recirculate period in fall of the output can be set by ZPER terminal. By adjusting ZPER voltage, recirculate period can be set from 0° to 90° as one period of hall signal 360° in Figure 37. The resolution of ZPER is 128 steps. Timing chart is shown in Figure 36.

About priority of SSW and ZPER setting, the setting priority of the period to regenerate than a soft switching period is high.

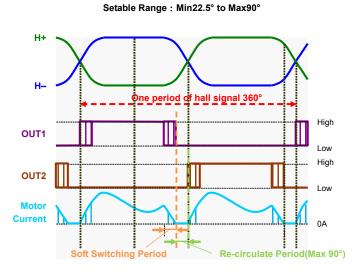
For example, Vssw=1.65V, Vzper = 0.825V

Soft switching period = $(1.65 / 3.3) \times 90^{\circ} - (0.825 / 3.3) \times 90^{\circ} = 45^{\circ} - 22.5^{\circ} = 22.5^{\circ}$ Re-circulate period = $(0.825 / 3.3) \times 90^{\circ} = 22.5^{\circ}$

When you set a period to regenerate for longer than soft switching period, a soft switching section for 5.6° (Typ) enters.

* A recirculate period is a current recirculate period before phase switching of output. In the recirculate period, the logic of the output transistor is decided by the hall input logic.

The phase of output Hi becomes the high impedance, and the phase of output Low is Low.



Adjust a Re-Circulate Period by ZPER Setting

Figure 36. Re-circulate Period Setting

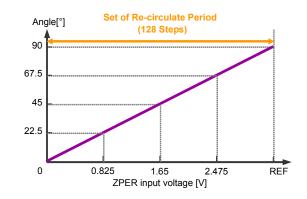


Figure 37. Relation of ZPER Input Voltage and Re-circulate Period

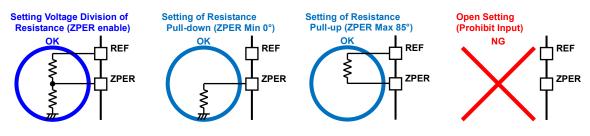


Figure 38. ZPER Terminal Setting

(3) Kickback Restraint Function (Lead angle correction)

Automatically detects a current phase gap, and an aspect change point is revised to lead angle. When a current phase is delayed for a hall phase, output phase can be changed up to 22.5° automatically. Timing chart is shown in Figure 39 and 40.

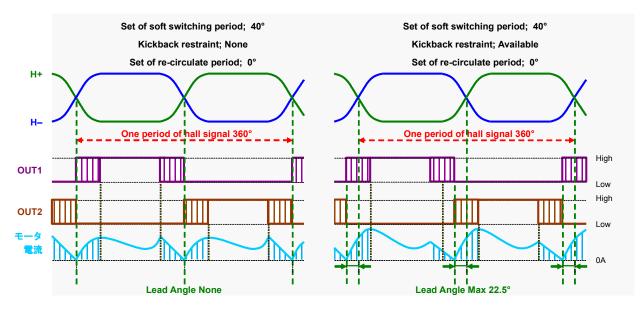


Figure 39. Normal Operation

Figure 40. Kickback Restraint Operation

A kickback restraint function is a miscellaneous function to prevent leaping up of the output voltage to occur at the time of power-on and a torque sudden change. Based on a setting method of SSW and ZPER of figure 41, prevent this function from working in normal operation.

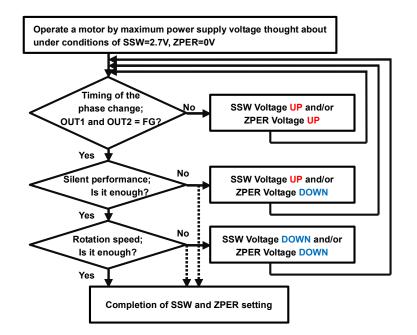


Figure 41. Flow of Setting of SSW and ZPER Terminal

OUT1

High

Functional Descriptions – continued

3. Current Limit

The current limit circuit turns off the output, when the current that flows to the motor coil is detected exceeding a set value. The current value that current limit operates is determined by internal setting voltage and current sense resistor. In Figure 42, I_0 is the current flowed to the motor coil, R_{NF} is the resistance detecting the current, and P_{RMAX} is the power

$$\begin{split} I_{O}[A] &= V_{CL}[V] / R_{NF}[\Omega] \\ &= 265[mV] / 0.33[\Omega] \\ &= 0.803[A] \\ P_{RMAX}[W] &= V_{CL}[V] \times I_{O}[A] \\ &= 265[mV] \times 0.803[A] \\ &= 0.213[W] \end{split}$$

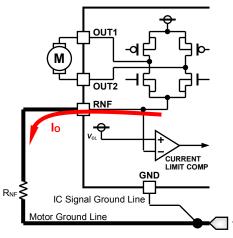


Figure 42. Setting of current limit and grout lines

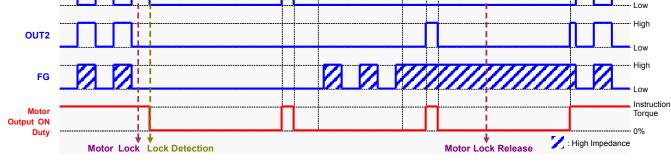


Figure 43. Lock Protection (Incorporated Counter System) Timing Chart

- 5. Quick Start
 - When torque off logic is input by the control signal over a fixed time, the lock protection function is disabled. The motor can restart quickly once the control signal is applied.

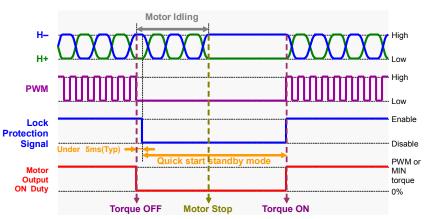


Figure 44. Quick Start Timing Chart (PWM Input Application)

6. Start Duty Assist

Start Duty Assist can secure a constant starting torque even at low duty. The IC is driven by a constant output duty (D_{OHL} ; Typ 50%) within detection of motor rotation. When Output ON duty is less than 50% (Typ), Start Duty Assist function operates under the following conditions:

- (1) Power ON
- (2) Lock Release
- (3) Quick Start
- (4) Thermal Shut Down(TSD) Release

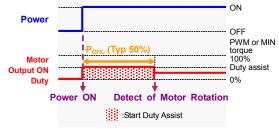


Figure 46. Timing Chart of Power ON

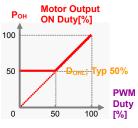


Figure 45. I/O Duty Characteristic in Start Duty Assist

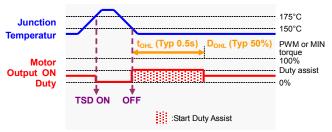
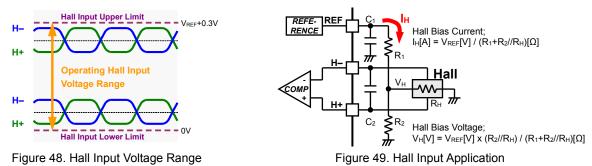


Figure 47. Timing Chart of TSD Release

- 7. Hall Input Setting
 - Hall input voltage range is shown in operating conditions (P.3). Adjust the value of hall element bias resistor R₁,R₂ in Figure 49 so that the input voltage of a hall amplifier is input in "Input Voltage Range 1"(P.3) including signal amplitude. R₂ is resistance to correct the temperature characteristic of the hall element.

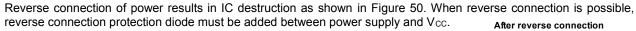


- (1) Reducing the Noise of Hall Signal
 - Vcc noise or the like depending on the wiring pattern of board may affect Hall element. In this case, place a capacitor like C₁ in Figure 49. In addition, when wiring from the hall element output to IC hall input is long, noise may be induced on wiring. In this case, place a capacitor like C₂.
- 8. High-speed detection protection

High-speed detection protection begin lock protection action when it detects that the hall input signal is in an abnormal state (more than Typ 2.5kHz). Noise may be induced on wiring. In this case, place a capacitor like C₂ in Figure 49.

Safety Measure

1. Reverse Connection Protection Diode



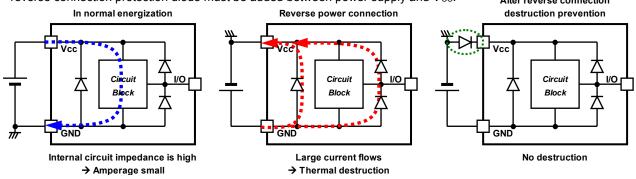


Figure 50. Flow of Current When Power is Connected Reversely

 Protection against V_{CC} Voltage Rise by Back Electromotive Force Back electromotive force (Back EMF) generates regenerative current to power supply. However, when reverse connection protection diode is connected, V_{CC} voltage rises because the diode prevents current flow to power supply.

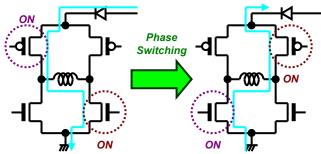


Figure 51. V_{CC} Voltage Rise by Back Electromotive Force

When the absolute maximum rated voltage may be exceeded due to voltage rise by back electromotive force, place (A) Capacitor or (B) Zener diode between V_{CC} and GND. If necessary, add both (C).

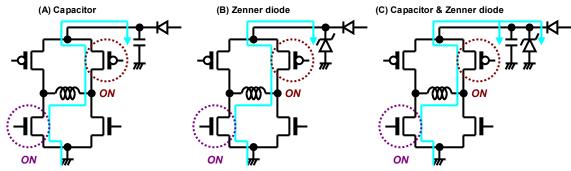
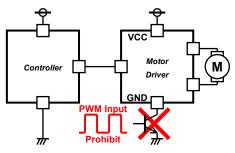


Figure 52. Measure against V_{CC} and Motor Driving Outputs Voltage

- 3. Problem of GND line PWM Switching Do not perform PWM switching of GND line because GND terminal potential cannot be kept to a minimum.
- 4. Protection of Rotation Speed Pulse (FG) Open-Drain Output FG output is an open drain and requires pull-up resistor. Adding resistor can protect the IC. Exceeding the absolute maximum rating, when FG terminal is directly connected to power supply, could damage the IC.



Motor Unit Driver Protection Resistor SIG Connector

Figure 53. GND Line PWM Switching Prohibited

Figure 54. Protection of FG Terminal

Power Consumption

1. Current Pathway

- The current pathways that relates to driver IC are the following, and shown in Figure 55.
- (1) Circuit Current (Icc)
- (2) Motor Current (\hat{I}_M)
- (3) Reference Bias Current to the Resistors (I_{REF})
- (4) FG Output Sink Current (IFG)

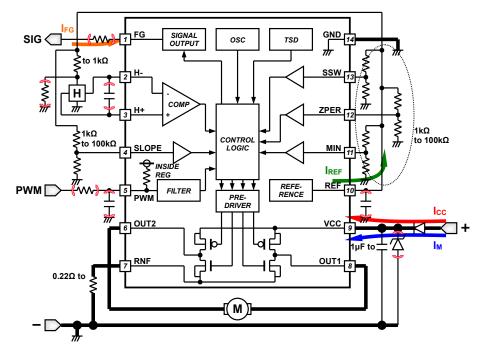


Figure 55. Current Pathway of IC

2. Calculation of Power Consumption

- (1) Circuit Current (I_{CC}) $P_{W1}[W] = V_{CC}[V] \times I_{CC}[A]$ (Icc current doesn't include I_M, I_{REF}) (ex.) Vcc = 11.3[V], Icc = 4.5[mA] $P_{W1}[W] = 11.3[V] \times 4.5[mA] = 50.85 [mW]$
- (2) Motor Driving Current (I_M) V_{OH} is the output saturation voltage of OUT1 or OUT2 high side, V_{OL} is the other low side voltage, $P_{W2}[W] = (V_{OH}[V] + V_{OL}[V]) \times I_M[A]$ (ex.) $V_{OH} = 0.10[V]$, $V_{OL} = 0.10[V]$, $I_M = 200[mA]$ $P_{W2}[W] = (0.10[V] + 0.10[V]) \times 200[mA] = 40.0[mW]$
- (3) Reference Bias Current to the LPF and Resistors (I_{REF}) $P_{W3}[W] = (V_{CC}[V] - V_{REF}[V]) \times I_{REF}[A]$ (ex.) I_{REF} = 6.0[mA] $P_{W3}[W] = (11.3[V] - 3.3[V]) \times 6.0[mA] = 48.0[mW]$
- $\begin{array}{l} \mbox{Total power consumption of driver IC becomes the following by the above (1) to (4).} \\ P_{Wttl}[W] = P_{W1}[W] + P_{W2}[W] + P_{W3}[W] + P_{W4}[W] \\ (ex.) \ P_{Wttl}[W] = 50.85[mW] + 40.0[mW] + 48.0[mW] + 0.5[mW] = 139.35[mW] \\ \end{array}$

Refer to next page, when you calculate the chip surface temperature (Tj) and the package surface temperature (Tc) by using the power consumption value.

Power Dissipation

1. Power Dissipation

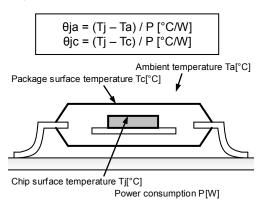
Power dissipation (total loss) indicates the power that can be consumed by IC at Ta=25°C (normal temperature). IC is heated when it consumes power, and the temperature of IC chip becomes higher than ambient temperature. The temperature that can be accepted by IC chip into the package, that is junction temperature of the absolute maximum rating, depends on circuit configuration, manufacturing process, etc. Power dissipation is determined by this maximum joint temperature, the thermal resistance in the state of the substrate mounting, and the ambient temperature. Therefore, when a power dissipation that provides by the absolute maximum rating is exceeded, the operating temperature range is not a guarantee. The maximum junction temperature is in general equal to the maximum value in the storage temperature range.

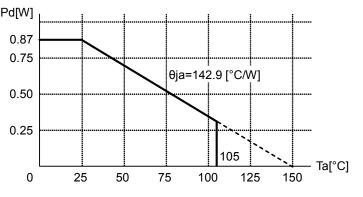
 $\theta_{ja} = (T_j - T_a) / P [^{\circ}C/W]$

2. Thermal Resistance

Heat generated by consumed power of IC is radiated from the mold resin or lead frame of package. The parameter which indicates this heat dissipation capability (hardness of heat release) is called thermal resistance. In the state of the substrate mounting, thermal resistances from the chip junction to the ambience and to the package surface are shown respectively with $\theta_{ja}[^{\circ}C/W]$ and $\theta_{jc}[^{\circ}C/W]$. Thermal resistance is classified into the package part and the substrate part, and thermal resistance in the package part depends on the composition materials such as the mold resins and the lead frames. On the other hand, thermal resistance in the substrate part depends on the substrate heat dissipation capability of the material, the size, and the copper foil area etc. Therefore, thermal resistance can be decreased by the heat radiation measures like installing a heat sink etc. in the mounting substrate. The thermal resistance model and calculations are shown in Figure 56.

θic = 36 [°C/W] (Reference Value)



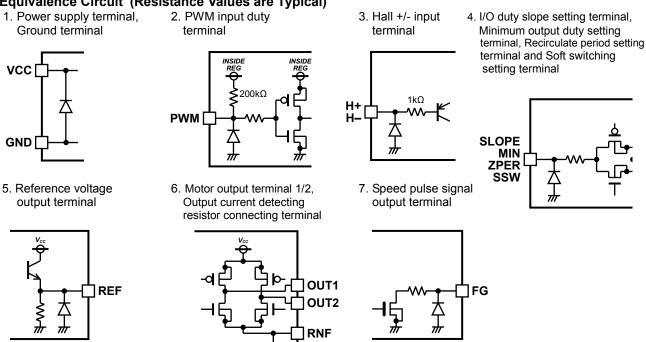


(Note)Reduce Reduce by 7.0mW/°C when operating over Ta=25°C (Mounted on 70.0mm x 70.0mm x 1.6mm glass epoxy board)

Figure 57. Power Dissipation vs Ambient Temperature

Figure 56. Thermal Resistance Model of Surface Mount

I/O Equivalence Circuit (Resistance Values are Typical)



 $\frac{1}{2}$

Operational Notes

1. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply pins.

2. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

3. Ground Voltage

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition. However, pins that drive inductive loads (e.g. motor driver outputs, DC-DC converter outputs) may inevitably go below ground due to back EMF or electromotive force. In such cases, the user should make sure that such voltages going below ground will not cause the IC and the system to malfunction by examining carefully all relevant factors and conditions such as motor characteristics, supply voltage, operating frequency and PCB wiring to name a few.

4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

5. Thermal Consideration

Should by any chance the power dissipation rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. In case of exceeding this absolute maximum rating, increase the board size and copper area to prevent exceeding the Pd rating.

6. Recommended Operating Conditions

These conditions represent a range within which the expected characteristics of the IC can be approximately obtained. The electrical characteristics are guaranteed under the conditions of each parameter.

7. Inrush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

8. Operation Under Strong Electromagnetic Field

Operating the IC in the presence of a strong electromagnetic field may cause the IC to malfunction.

9. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

Operational Notes – continued

10. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

11. Unused Input Pins

Input pins of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

12. Regarding the Input Pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode. When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.

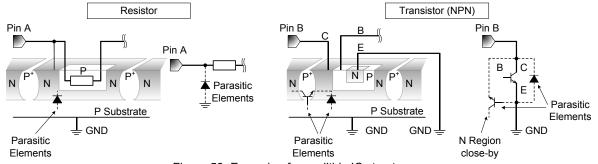


Figure 58. Example of monolithic IC structure

13. Ceramic Capacitor

When using a ceramic capacitor, determine the dielectric constant considering the change of capacitance with temperature and the decrease in nominal capacitance due to DC bias and others.

14. Area of Safe Operation (ASO)

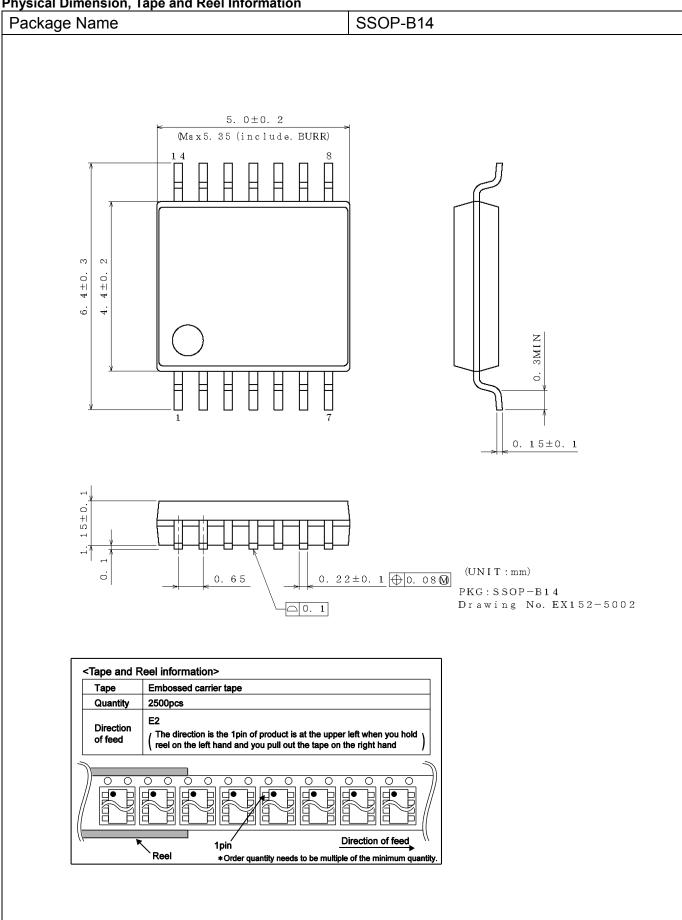
Operate the IC such that the output voltage, output current, and power dissipation are all within the Area of Safe Operation (ASO).

15. Thermal Shutdown (TSD) Circuit

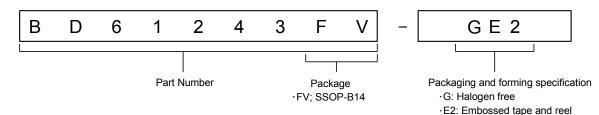
This IC has a built-in thermal shutdown circuit that prevents heat damage to the IC. Normal operation should always be within the IC's power dissipation rating. If however the rating is exceeded for a continued period, the junction temperature will rise which will activate the TSD circuit that will turn OFF all output pins. When the junction temperature falls below the TSD threshold, the circuits are automatically restored to normal operation.

Note that the TSD circuit operates in a situation that exceeds the absolute maximum ratings and therefore, under no circumstances, should the TSD circuit be used in a set design or for any purpose other than protecting the IC from heat damage.

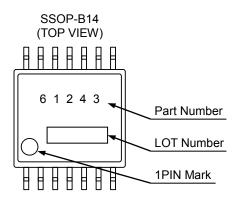
Physical Dimension, Tape and Reel Information



Ordering Information



Marking Diagram



Revision History

Date	Revision	Changes				
Jun.19.2015	001	New Release				

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