

High-Speed Quad-MOSFET Driver

Features

- · 6 ns Rise and Fall Time
- · 2A Peak Output Source and Sink Currents
- 1.8V to 5V Input CMOS Compatible
- 5V to 12V Total Supply Voltage
- · Smart Logic Threshold
- · Low-Jitter Design
- · Four Matched Channels
- · Drives Two P-Channel and Two N-Channel **MOSFETs**
- Outputs can Swing Below Ground
- · Low-Inductance Package
- · High-Performance, Thermally Enhanced Package

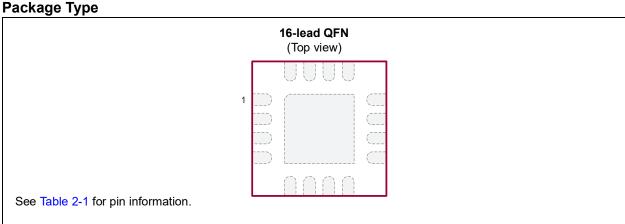
Applications

- · Medical Ultrasound Imaging
- · Piezoelectric Transducer Drivers
- · Non-Destructive Testing
- PIN Diode Driver
- · CCD Clock Driver/Buffer
- · High-Speed Level Translator

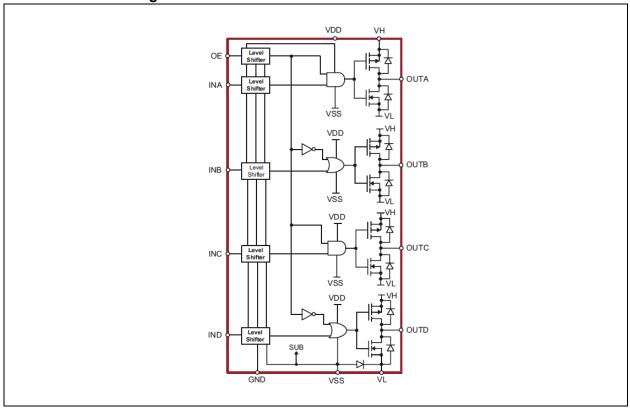
General Description

The MD1811 is a high-speed quad-MOSFET driver designed to drive high-voltage P-channel and N-channel MOSFETs for medical ultrasound applications and other applications requiring a high-output current for a capacitive load. The high-speed input stage of the MD1811 can operate from a 1.8V to 5V logic interface with an optimum operating input signal range of 1.8V to 3.3V. An adaptive threshold circuit is used to set the level translator switch threshold to the average of the input logic 0 and logic 1 levels. The input logic levels may be ground referenced even though the driver is putting out bipolar signals. The level translator uses a proprietary circuit, which provides DC coupling together with high-speed operation.

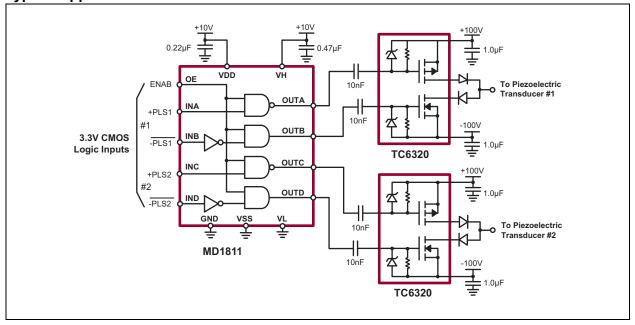
The output stage of the MD1811 has separate power connections, enabling the output signal L and H levels to be chosen independently from the supply voltages used for the majority of the circuit. As an example, the input logic levels may be between 0V and 1.8V, the control logic may be powered by +5V and -5V, and the output L and H levels may be varied anywhere over the range of -5V to +5V. The output stage is capable of peak currents of up to ±2A, depending on the supply voltages used and load capacitance present. The OE pin serves a dual purpose. First, its logic H level is used to compute the threshold voltage level for the channel input level translators. Second, when OE is low, the outputs are disabled, with the A and C outputs high and the B and D outputs low. This assists in properly pre-charging the AC coupling capacitors that may be used in series in the gate drive circuit of an external PMOS and NMOS transistor pair.



Functional Block Diagram



Typical Application Circuit



1.0 ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings†

Logic Supply Voltage, V _{DD} –V _{SS}	0.5V to +13.5V
Output High Supply Voltage, V _H	
Output Low Supply Voltage, V _L	V_{SS} =0.5V to V_{H} + 0.5V
Low-Side Supply Voltage, V _{SS}	–7V to + 0.5V
Logic Input Levels	V _{SS} -0.5V to GND +7V
Maximum Junction Temperature, T _J	+125°C
Operating Ambient Temperature, T _A	–20°C to +85°C
Storage Temperature, T _S	65°C to +150°C
Power Dissipation	2.2W
ESD Rating (Note 1)	ESD Sensitive

† Notice: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only, and functional operation of the device at those or any other conditions above those indicated in the operational sections of this specification is not intended. Exposure to maximum rating conditions for extended periods may affect device reliability.

Note 1: Device is ESD sensitive. Handling precautions are recommended.

DC ELECTRICAL CHARACTERISTICS

Electrical Specifications: $V_H = V_{DD} = 12V$, $V_L = V_{SS} = GND = 0V$, $V_{OE} = 3.3V$, and $T_A = 25^{\circ}C$							
Parameter	Sym.	Min.	Тур.	Max.	Unit	Conditions	
Logic Supply Voltage	V _{DD} -V _{SS}	4.5		13	V	$2.5V \le V_{DD} \le 13V$	
Low-Side Supply Voltage	V_{SS}	-5.5	_	0	V		
Output High Supply Voltage	V_{H}	V _{SS} +2	_	V_{DD}	V		
Output Low Supply Voltage	V_{L}	V _{SS}	_	V _{DD} –2	V		
V _{DD} Quiescent Current	I_{DDQ}	_	0.8	_	mA	No input transitions, OE = 1	
V _H Quiescent Current	I_{HQ}	_	_	10	μA	No input transitions, OE = 1	
V _{DD} Average Current	I _{DD}	_	8	_	mA	One channel on at 5 MHz,	
V _H Average Current	I _H	_	26	_	mA	no load	
Input Logic Voltage High	V_{IH}	V _{OE} -0.3	_	5	V		
Input logic Voltage Low	V_{IL}	0	_	0.3	V	For logic inputs INA, INB,	
Input Logic Current High	I _{IH}	_	_	1	μA	INC, and IND	
Input Logic Current Low	I _{IL}	_	_	1	μA		
OE Input Logic Voltage High	V_{IH}	1.7	_	5	V		
OE Input Logic Voltage Low	V _{IL}	0	_	0.3	V	For logic input OE	
OE Input Resistance	R _{IN}	10	20	30	kΩ		
Logic Input Capacitance	C _{IN}	_	5	10	pF		
Output Sink Resistance	R _{SINK}	_	_	12.5	Ω	I _{SINK} = 50 mA	
Output Source Resistance	R _{SOURCE}	_	_	12.5	Ω	I _{SOURCE} = 50 mA	
Peak Output Sink Current	I _{SINK}	_	2	_	Α		
Peak Output Source Current	I _{SOURCE}	_	2	_	Α		

AC ELECTRICAL CHARACTERISTICS

Electrical Specifications: $V_H = V_{DD} = 12V$, $V_L = V_{SS} = GND = 0V$, $V_{OE} = 3.3V$ and $T_A = 25^{\circ}C$								
Parameter	Sym.	Min.	Тур.	Max.	Unit	Conditions		
Input or OE Rise and Fall Time	t _{irf}	1	_	10	ns	Logic input edge speed requirement		
Propagation Delay when Output is from Low to High	t _{PLH}		7		ns			
Propagation Delay when Output is from High to Low	t _{PHL}	_	7	_	ns	C _{LOAD} = 1000 pF, input signal rise/fall time of 2 ns (See Tim -		
Output Rise Time	t _r	_	6	_	ns	ing Diagram)		
Output Fall Time	t _f	_	6	_	ns			
Rise and Fall Time Matching	l t _r –t _f l	_	1	_	ns			
Propagation Low-to-High and High-to-Low Matching	l t _{PLH} -t _{PHL} l	_	1	_	ns	For each channel		
Propagation Delay Matching	Δt _{dm}	_	±2	_	ns	Device-to-device delay match		
Output Enable Time	t _{OE}	_	9	_	ns			

TEMPERATURE SPECIFICATIONS

Parameter	Sym.	Min.	Тур.	Max.	Unit	Conditions		
TEMPERATURE RANGE								
Maximum Junction Temperature	TJ	_	_	125	°C			
Operating Ambient Temperature	T _A	-20	_	85	°C			
Storage Temperature	T _S	-65		150	°C			
PACKAGE THERMAL RESISTANCE								
16-lead QFN	θ_{JA}	_	45	_	°C/W	Note 1		

Note 1: 1 oz. 4-layer 3" x 4" PCB

Timing Diagram

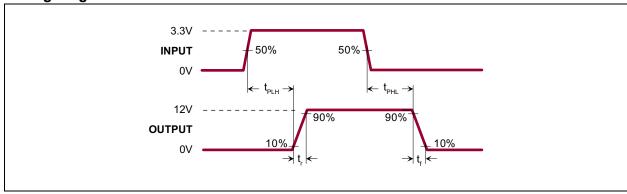


TABLE 1-1: TRUTH FUNCTION TABLE

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	Logic Inputs	Out	outs	
OE	INA	INB	OUTA	OUTB
Н	L	L	V _H	V _H
Н	L	Н	V _H	V _L
Н	Н	L	V _L	V _H
Н	Н	Н	V _L	V_{L}
L	Х	Х	V _H	V_{L}
OE	INC	IND	OUTC	OUTD
Н	L	L	V _H	V_{H}
Н	L	Н	V _H	V_{L}
Н	Н	L	V _L	V _H
Н	Н	Н	V _L	V_{L}
L	X	X	V _H	V_{L}

2.0 PIN DESCRIPTION

Table 2-1 shows details on the pins of MD1811. See **Package Type** for the location of the pins.

TABLE 2-1: PIN FUNCTION TABLE

Pin Number	Pin Name	Description
1	INB	Logic input. Input logic low will cause the output to swing to VH. Input logic high will cause the output to swing to VL. Keep all logic inputs low until IC powers up.
2	VL	Supply voltage for N-channel output stage
3	GND	Logic input ground reference
4	VL	Supply voltage for N-channel output stage
5	INC	Logic input. Input logic low will cause the output to swing to VH. Input logic high will
6	IND	cause the output to swing to VL. Keep all logic inputs low until IC powers up.
7	VSS	Low-side supply voltage. VSS is also connected to the IC substrate. It is required to be connected to the most negative potential of voltage supplies and is powered up first.
8	OUTD	Output driver. Swings from VH to VL. Intended to drive the gate of an external N-channel MOSFET through a series capacitor. When OE is low, the output is disabled. OUTD will swing to VL, turning off the external N-channel MOSFET.
9	OUTC	Output driver. Swings from VH to VL. Intended to drive the gate of an external P-channel MOSFET via a series capacitor. When OE is low, the output is disabled. OUTC will swing to VH, turning off the external P-channel MOSFET.
10 and 11	VH	Supply voltage for P-channel output stage
12	OUTB	Output driver. Swings from VH to VL. Intended to drive the gate of an external N-channel MOSFET via a series capacitor. When OE is low, the output is disabled. OUTB will swing to VL, turning off the external N-channel MOSFET.
13	OUTA	Output driver. Swings from VH to VL. Intended to drive the gate of an external P-channel MOSFET through a series capacitor. When OE is low, the output is disabled. OUTA will swing to VH, turning off the external P-channel MOSFET.
14	VDD	High-side supply voltage
15	INA	Logic input. Input logic low will cause the output to swing to VH. Input logic high will cause the output to swing to VL. Keep all logic inputs low until the IC powers up.
16	OE	Output Enable logic input. When OE is high, $(V_{OE}+V_{GND})/2$ sets the logic threshold level for inputs. When OE is low, OUTA and OUTC are at VH, while OUTB and OUTD are at VL regardless of the inputs INA, INB, INC, or IND. Keep OE low until IC powers up.
Subs	trate	The IC substrate is internally connected to the thermal pad. Thermal Pad and VSS must be connected externally.

3.0 APPLICATION INFORMATION

For proper operation of the MD1811, low-inductance bypass capacitors should be used on the various supply pins. The GND input pin should be connected to the logic ground. The INA, INB, INC, IND, and OE pins should be connected to a logic source with a swing of GND to OE, where OE is 1.8V to 5V. Good trace practices should be followed corresponding to the desired operating speed. The internal circuitry of the MD1811 is capable of operating up to 100 MHz, with the primary speed limitation being the loading effect of the load capacitance. Because of this speed and the high transient currents due to the capacitive loads, the bypass capacitors should be as close to the chip pins as possible. Unless the load specifically requires bipolar drive, the V_{SS} and V_L pins should have low-inductance feed-through connections directly to a ground plane. The power connection V_{DD} should have a ceramic bypass capacitor to the ground plane with short leads and decoupling components to prevent resonance in the power leads.

The voltages of V_H and V_L decide the output logic levels. These two pins can draw fast transient currents of up to 2A, so they should be provided with a suitable bypass capacitor located next to the chip pins. A ceramic capacitor of up to 1 μF may be appropriate, with a series ferrite bead to prevent resonance in the power supply lead going to the capacitor.

Pay particular attention to minimizing trace lengths, current loop area, and using sufficient trace width to reduce inductance. Surface-mount components are highly recommended. Since the output impedance of this driver is very low, in some cases, it may be desirable to add a small series resistor in series with the output signal to obtain better waveform transitions at the load terminals. This will reduce the output voltage slew rate at the terminals of a capacitive load.

Ensure that parasitic couplings are minimized from the driver output to the input signal terminals. The parasitic feedback may cause oscillations or spurious waveform shapes on the edges of signal transitions. Since the input operates with signals down to 1.8V, even small coupled voltages may cause problems. The use of a solid ground plane as well as good power and signal layout practices will prevent this problem. Make sure that the circulating ground return current from a capacitive load will not react with common inductance and cause noise voltages in the input logic circuitry.

4.0 PACKAGING INFORMATION

4.1 Package Marking Information

16-lead QFN

Example

XX XXXXXX @YYWW NNN

MD 1811K6 ② 2025 521

Legend: XX...X Product Code or Customer-specific information

Y Year code (last digit of calendar year)
YY Year code (last 2 digits of calendar year)
WW Week code (week of January 1 is week '01')

NNN Alphanumeric traceability code

Pb-free JEDEC® designator for Matte Tin (Sn)

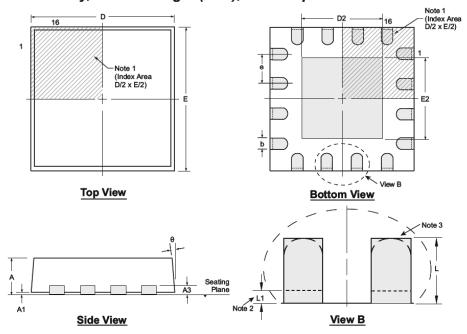
This package is Pb-free. The Pb-free JEDEC designator (e3) can be found on the outer packaging for this package.

Note: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for product code or customer-specific information. Package may or not include

the corporate logo.

16-Lead QFN Package Outline (K6)

4.00x4.00mm body, 1.00mm height (max), 0.65mm pitch



Note: For the most current package drawings, see the Microchip Packaging Specification at www.microchip.com/packaging.

Notes

- A Pin 1 identifier must be located in the index area indicated. The Pin 1 identifier can be: a molded mark/identifier; an embedded metal marker; or a printed indicator.
- 2. Depending on the method of manufacturing, a maximum of 0.15mm pullback (L1) may be present.
- The inner tip of the lead may be either rounded or square.

Symb	ol	Α	A1	А3	b	D	D2	E	E2	е	L	L1	θ
	MIN	0.80	0.00		0.25	3.85*	2.50	3.85*	2.50		0.30 [†]	0.00	0 o
Dimension (mm)	NOM	0.90	0.02	0.20 REF	0.30	4.00	2.65	4.00	2.65	0.65 BSC	0.40 [†]	-	-
()	MAX	1.00	0.05		0.35	4.15*	2.80	4.15*	2.80	230	0.50 [†]	0.15	14º

JEDEC Registration MO-220, Variation VGGC-2, Issue K, June 2006.

Drawings not to scale.

^{*} This dimension is not specified in the JEDEC drawing.

[†] This dimension differs from the JEDEC drawing.

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NOTES:

APPENDIX A: REVISION HISTORY

Revision A (September 2020)

- Converted Supertex Doc# DSFP-MD1811 to Microchip DS20005744A
- Changed the package marking format
- Updated the quantity of the 16-lead QFN K6 package from 3000/Reel to 3300/Reel to align it with the actual BQM
- Made minor text changes throughout the document

PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, contact your local Microchip representative or sales office.

PART NO.	<u> </u>		- <u>x</u> - <u>x</u>	Example:	
Device	Package Options		Environmental Media Type	a) MD1811K6-G:	High-Speed Quad-MOSFET Driver, 16-lead QFN, 3300/Reel
Device:	MD1811	=	High-Speed Quad-MOSFET Driver		
Package:	K6	=	16-lead QFN		
Environmental:	G	=	Lead (Pb)-free/RoHS-compliant Package		
Media Type:	(blank)	=	3300/Reel for a K6 Package		

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