



### **General Description**

The MAX5392 dual, 256-tap, volatile, low-voltage, linear taper digital potentiometer offers three end-to-end resistance values of  $10k\Omega$ ,  $50k\Omega$ , and  $100k\Omega$ . Operating from a single +1.7V to +5.5V power supply, the device provides a low 35ppm/°C end-to-end temperature coefficient. The device features an I<sup>2</sup>C interface.

The small package size, low supply operating voltage, low supply current, and automotive temperature range of the MAX5392 makes the device uniquely suited for the portable consumer market, battery-backup industrial applications, and the automotive market.

The MAX5392 is specified over the automotive -40°C to +125°C temperature range and is available in a 16-pin TSSOP package.

### **Applications**

Low-Voltage Battery Applications Portable Electronics Mechanical Potentiometer Replacement Offset and Gain Control Adjustable Voltage References/Linear Regulators Automotive Electronics

#### **Features**

- ◆ Dual, 256-Tap Linear Taper Positions
- ♦ Single +1.7V to +5.5V Supply Operation
- ♦ Low 12μA Quiescent Supply Current
- ♦ 10kΩ, 50kΩ, 100kΩ End-to-End Resistance Values
- ♦ I<sup>2</sup>C-Compatible Interface
- ♦ Wiper Set to Midscale on Power-Up
- → -40°C to +125°C Operating Temperature Range

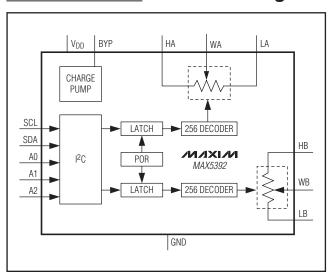
### **Ordering Information**

PART	PIN-PACKAGE	END-TO-END RESISTANCE ( $k\Omega$ )
MAX5392LAUE+	16 TSSOP	10
MAX5392MAUE+	16 TSSOP	50
MAX5392NAUE+	16 TSSOP	100

Note: All devices are specified over the -40°C to +125°C operating temperature range.

+Denotes a lead(Pb)-free/RoHS-compliant package.

### **Functional Diagram**



#### **ABSOLUTE MAXIMUM RATINGS**

VDD to GND	0.3V to +6V
H_, W_, L_ to GND	0.3V to the lower of
	$(V_{DD} + 0.3V)$ and +6V
All Other Pins to GND	-0.3V to +6V
Continuous Current in to H_, W_, and L	_
MAX5392L	±5mA
MAX5392M	±2mA
MAX5392N	±1mA

Continuous Power Dissipation (T <sub>A</sub> = +70°C	C)
16-Pin TSSOP (derate 11.1mW/°C abov	e +70°C)888.9mW
Operating Temperature Range	40°C to +125°C
Junction Temperature	+150°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (soldering, 10s)	+300°C
Soldering Temperature (reflow)	+260°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### **ELECTRICAL CHARACTERISTICS**

 $(V_{DD} = +1.7V \text{ to } +5.5V, V_{H_{-}} = V_{DD}, V_{L_{-}} = 0, T_{A} = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted. Typical values are at } V_{DD} = +1.8V, T_{A} = +25^{\circ}\text{C.}) \text{ (Note 1)}$ 

PARAMETER	SYMBOL		CONDITIONS	MIN	TYP	MAX	UNITS
Resolution	N			256			Тар
DC PERFORMANCE (Voltage Di	vider Mode)	•		'			
Integral Nonlinearity	INL	(Note 2)		-0.5		+0.5	LSB
Differential Nonlinearity	DNL	(Note 2)		-0.5		+0.5	LSB
Dual Code Matching		Register A = R	legister B	-0.5		+0.5	LSB
Ratiometric Resistor Tempco		(ΔVW/VW)/ΔT,	no load		5		ppm/°C
			MAX5392L	-3	-2.2		
Full-Scale Error		Code = FFh	MAX5392M	-1	-0.6		LSB
			MAX5392N	-0.5	-0.3		
			MAX5392L		2.2	3	
Zero-Scale Error		Code = 00h	MAX5392M		0.6	1.0	LSB
			MAX5392N		0.3	0.5	
DC PERFORMANCE (Variable R	esistor Mode	)					•
		MAX5392L (No	ote 3)	-1.5		+1.5	
Integral Nonlinearity	R-INL	MAX5392M (N		-0.75		+0.75	LSB
		MAX5392N (N	ote 3)	-0.5		+0.5	
Differential Nonlinearity	R-DNL	(Note 3)		-0.5		+0.5	LSB
DC PERFORMANCE (Resistor C	haracteristic	s)					
Wiper Resistance	RwL	(Note 4)				200	Ω
Terminal Capacitance	CH_, CL_	Measured to G	AND		10		pF
Wiper Capacitance	Cw_	Measured to G	AND		50		pF
End-to-End Resistor Tempco	TCR	No load			35		ppm/°C
End-to-End Resistor Tolerance	ΔRHL	Wiper not conr	nected	-25		+25	%

### **ELECTRICAL CHARACTERISTICS (continued)**

 $(V_{DD} = +1.7V \text{ to } +5.5V, V_{H_{-}} = V_{DD}, V_{L_{-}} = 0, T_{A} = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted. Typical values are at } V_{DD} = +1.8V, T_{A} = +25^{\circ}\text{C.}) \text{ (Note 1)}$ 

PARAMETER	SYMBOL	С	ONDITIONS	MIN	TYP	MAX	UNITS	
AC PERFORMANCE		•		•				
Crosstalk		(Note 5)			-90		dB	
		Code = 80H,	MAX5392L		600			
-3dB Bandwidth	BW	10pF load,	MAX5392M		100		kHz	
		$V_{DD} = 1.8V$	MAX5392N		50			
Total Harmonic Distortion Plus Noise	THD+N	Measured at W	, V <sub>H</sub> _ = 1V <sub>RMS</sub> at 1kHz		0.02		%	
			MAX5392L		400			
Wiper Settling Time	ts	(Note 6)	MAX5392M		1200		ns	
			MAX5392N		2200			
Charge-Pump Feedthrough at W_	VRW	fclk = 600kHz	, CBYP = 0nF			600	nV <sub>P-P</sub>	
POWER SUPPLIES								
Supply Voltage Range	VDD			1.7		5.5	V	
Chair allas a Command		$V_{DD} = 5.5V$			27			
Standby Current		V <sub>DD</sub> = 1.7V			12		μA	
DIGITAL INPUTS		1		1				
NAI	\ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \	$V_{DD} = 2.6V \text{ to}$	5.5V	70			0/ \/==	
Minimum Input High Voltage	VIH	$V_{DD} = 1.7V \text{ to } 3$	2.6V	75			% x VDD	
	.,	$V_{DD} = 2.6V \text{ to}$	5.5V			30	04 14	
Maximum Input Low Voltage	VIL	$V_{DD} = 1.7V \text{ to } 3$	2.6V			25	% x V <sub>DD</sub>	
Input Leakage Current				-1		+1	μΑ	
Input Capacitance					5		pF	
TIMING CHARACTERISTICS—I <sup>2</sup> C	(Notes 7 a	nd 8)						
Maximum SCL Frequency	fSCL					400	kHz	
Setup Time for START Condition	tsu:sta			0.6			μs	
Hold Time for START Condition	tHD:STA			0.6			μs	
SCL High Time	tHIGH			0.6			μs	
SCL Low Time	tLOW			1.3			μs	

### **ELECTRICAL CHARACTERISTICS (continued)**

 $(V_{DD} = +1.7V \text{ to } +5.5V, V_{H_{-}} = V_{DD}, V_{L_{-}} = 0, T_{A} = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted. Typical values are at } V_{DD} = +1.8V, T_{A} = +25^{\circ}\text{C.}) \text{ (Note 1)}$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Data Setup Time	tsu:dat		100			ns
Data Hold Time	thd:dat		0			μs
SDA, SCL Rise Time	tR				0.3	μs
SDA, SCL Fall	tF				0.3	μs
Setup Time for STOP Condition	tsu:sto		0.6			μs
Bus Free Time Between STOP and START Condition	t <sub>BUF</sub>	Minimum power-up rate = 0.2V/µs	1.3			μs
Pulse Suppressed Spike Width	tsp				50	ns
Capacitive Load for Each Bus	Св	(Note 9)			400	pF

- **Note 1:** All devices are 100% production tested at TA = +25°C. Specifications over temperature limits are guaranteed by design and characterization.
- Note 2: DNL and INL are measured with the potentiometer configured as a voltage-divider (Figure 1) with H\_ = V<sub>DD</sub> and L\_ = GND. The wiper terminal is unloaded and measured with a high-input-impedance voltmeter.
- Note 3: R-DNL and R-INL are measured with the potentiometer configured as a variable resistor (Figure 1). DNL and INL are measured with the potentiometer configured as a variable resistor. H\_ is unconnected and L\_ = GND. For V<sub>DD</sub> = +5V, the wiper terminal is driven with a source current of 400μA for the  $10k\Omega$  configuration,  $80\mu$ A for the  $50k\Omega$  configuration, and  $40\mu$ A for the  $100k\Omega$  configuration. For V<sub>DD</sub> = +1.7V, the wiper terminal is driven with a source current of  $150\mu$ A for the  $10k\Omega$  configuration,  $30\mu$ A for the  $50k\Omega$  configuration, and  $15\mu$ A for the  $100k\Omega$  configuration.
- Note 4: The wiper resistance is the worst value measured by injecting the currents given in Note 3 to W\_ with  $L_{-} = GND$ . RW = (VW VH)/IW.
- Note 5: Drive HA with a 1kHz GND to VDD amplitude tone. LA = LB = GND. No load. WB is at midscale with a 10pF load. Measure WB.
- Note 6: The wiper-settling time is the worst-case 0 to 50% rise time, measured between tap 0 and tap 127. H\_ = V<sub>DD</sub>, L\_ = GND, and the wiper terminal is loaded with 10pF capacitance to ground.
- Note 7: Digital timing is guaranteed by design and characterization, not production tested.
- Note 8: The SCL clock period includes rise and fall times ( $t_R = t_F$ ). All digital input signals are specified with  $t_R = t_F = 2$ ns and timed from a voltage level of ( $V_{IL} + V_{IH}$ )/2.
- Note 9: An appropriate bus pullup resistance must be selected depending on board capacitance. For I<sup>2</sup>C-bus specification information from NXP Semiconductor (formerly Philips Semiconductor), refer to the UM10204: I<sup>2</sup>C-Bus Specification and User Manual.

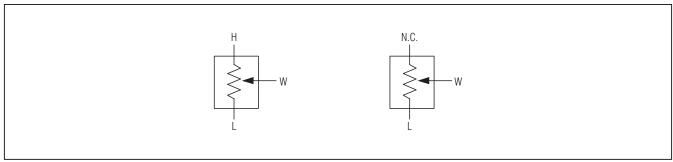
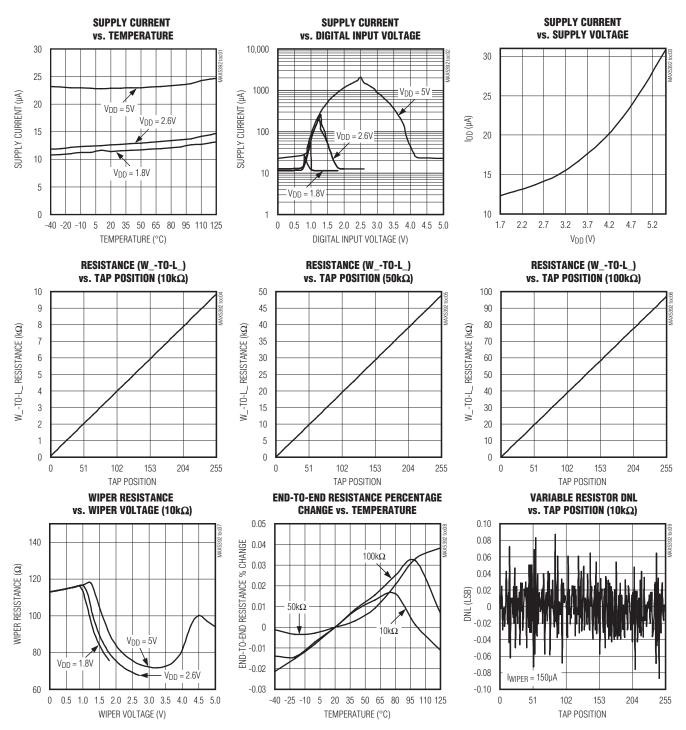


Figure 1. Voltage-Divider and Variable Resistor Configurations

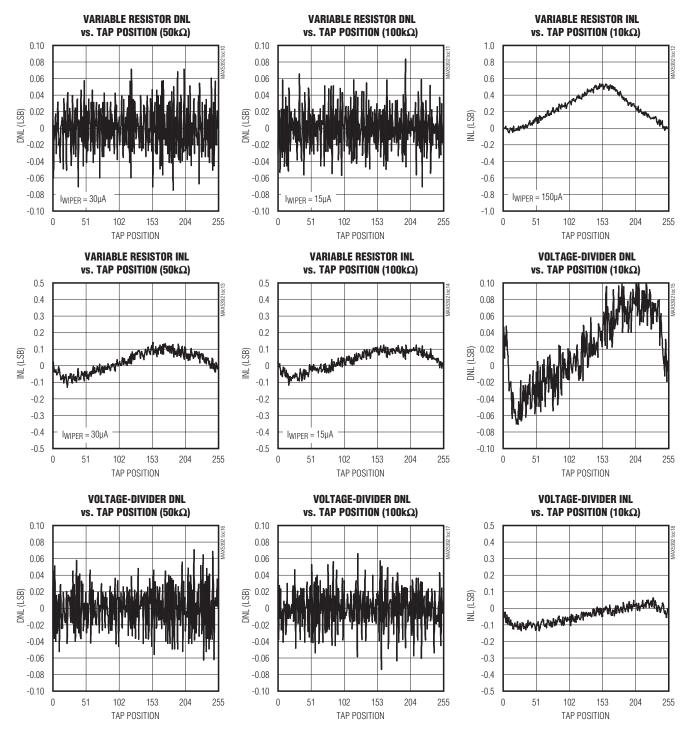
## **Typical Operating Characteristics**

 $(V_{DD} = 1.8V, T_A = +25^{\circ}C, \text{ unless otherwise noted.})$ 



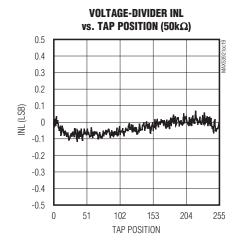
## Typical Operating Characteristics (continued)

 $(V_{DD} = 1.8V, T_A = +25^{\circ}C, unless otherwise noted.)$ 

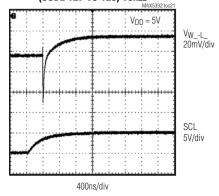


## \_Typical Operating Characteristics (continued)

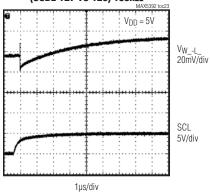
 $(V_{DD} = 1.8V, T_A = +25^{\circ}C, unless otherwise noted.)$ 



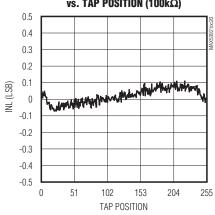
TAP-TO-TAP SWITCHING TRANSIENT (CODE 127 TO 128)  $10k\Omega$ 



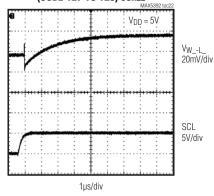
TAP-TO-TAP SWITCHING TRANSIENT (CODE 127 TO 128)  $100k\Omega$ 



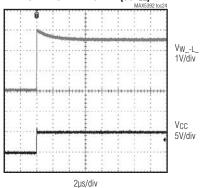
## VOLTAGE-DIVIDER INL vs. TAP POSITION (100k $\Omega$ )



## TAP-TO-TAP SWITCHING TRANSIENT (CODE 127 TO 128) $50k\Omega$

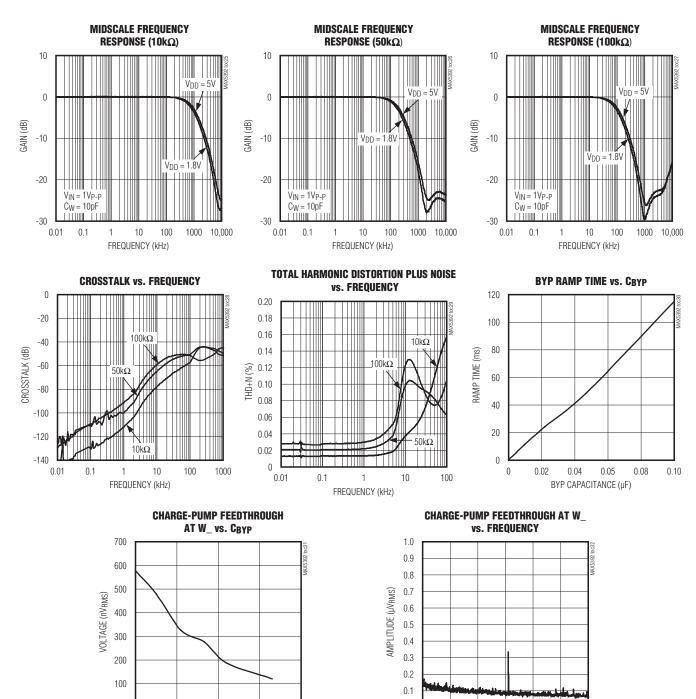


#### POWER-ON TRANSIENT ( $50k\Omega$ )



## Typical Operating Characteristics (continued)

 $(V_{DD} = 1.8V, T_A = +25^{\circ}C, unless otherwise noted.)$ 



300 400

500

600

FREQUENCY (kHz)

800

0

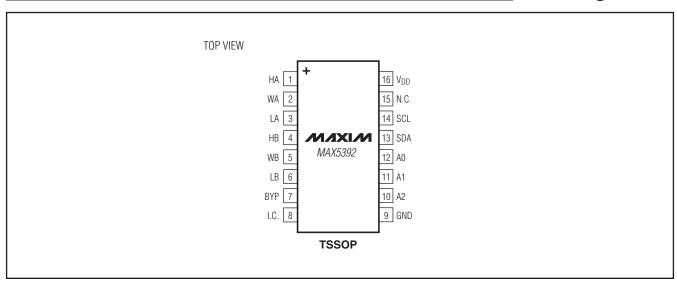
0

200

400

CAPACITANCE (pF)

## Pin Configuration



## Pin Description

PIN	NAME	FUNCTION
1	НА	Resistor A High Terminal. The voltage at HA can be higher or lower than the voltage at LA. Current can flow into or out of HA.
2	WA	Resistor A Wiper Terminal
3	LA	Resistor A Low Terminal. The voltage at LA can be higher or lower than the voltage at HA. Current can flow into or out of LA.
4	НВ	Resistor B High Terminal. The voltage at HB can be higher or lower than the voltage at LB. Current can flow into or out of HB.
5	WB	Resistor B Wiper Terminal
6	LB	Resistor B Low Terminal. The voltage at LB can be higher or lower than the voltage at HB. Current can flow into or out of LB.
7	BYP	Internal Power-Supply Bypass. For additional charge-pump filtering, bypass to GND with a capacitor close to the device.
8	I.C.	Internally Connected. Connect to GND.
9	GND	Ground
10	A2	Address Input 2. Connect to VDD or GND.
11	A1	Address Input 1. Connect to V <sub>DD</sub> or GND.
12	A0	Address Input 0. Connect to VDD or GND.
13	SDA	I <sup>2</sup> C-Compatible Serial-Data Input/Output. A pullup resistor is required.
14	SCL	I <sup>2</sup> C-Compatible Serial-Clock Input. A pullup resistor is required.
15	N.C.	No Connection. Not internally connected.
16	VDD	Power-Supply Input. Bypass VDD to GND with a 0.1µF capacitor close to the device.

### **Detailed Description**

The MAX5392 dual, 256-tap, volatile, low-voltage linear taper digital potentiometer offers three end-to-end resistance values of  $10k\Omega,~50k\Omega,~$  and  $100k\Omega.$  The potentiometer consists of 255 fixed resistors in series between terminals H\_ and L\_. The potentiometer wiper, W\_, is programmable to access any one of the 256 tap points on the resistor string.

The potentiometers are programmable independently of each other. The MAX5392 features an I<sup>2</sup>C interface.

#### **Charge Pump**

TThe MAX5392 contains an internal charge pump that guarantees the maximum wiper resistance, RWL, to be less than  $200\Omega$  for supply voltages down to 1.7V. Pins H\_, W\_, and L\_ are still required to be less than VDD + 0.3V. A bypass input, BYP, is provided to allow additional filtering of the charge-pump output, further reducing clock feedthrough that can occur on H\_, W\_, or L\_. The nominal clock rate of the charge pump is 600kHz. BYP should remain resistively unloaded as any additional load would increase clock feedthrough. See the Charge-Pump Feedthrough at W\_ vs. CBYP graph in the Typical Operating Characteristics for CBYP sizing guidelines with respect to clock feedthrough to the wiper. The value of CBYP does affect the startup time of the charge

pump; however, CBYP does not impact the ability to communicate with the device, nor is there a minimum CBYP requirement. The maximum wiper impedance specification is not guaranteed until the charge pump is fully settled. See the BYP Ramp Time vs. CBYP graph in the *Typical Operating Characteristics* for CBYP impact on charge-pump settling time.

#### I<sup>2</sup>C Digital Interface

The I<sup>2</sup>C interface contains a shift register that decodes the command and address bytes, routing the data to the appropriate control registers. Data written to a control register immediately updates the wiper position. The wipers A and B power up in midposition, D[7:0] = 80h.

#### Serial Addressing

The MAX5392 operates as a slave device that receives data through an I²C/SMBus<sup>TM</sup>-compatible 2-wire serial interface. The interface uses a serial-data access line (SDA) and a serial-clock line (SCL) to achieve bidirectional communication between master(s) and slave(s). A master, typically a microcontroller, initiates all data transfers to the port and generates the SCL clock that synchronizes the data transfer. See Figure 2. Connect a pullup resistor, typically 4.7k $\Omega$ , between each of the SDA and SCL lines to a voltage between VDD and 5.5V.

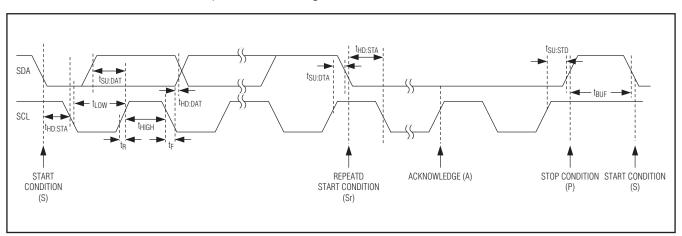


Figure 2. I<sup>2</sup>C Serial-Interface Timing Diagram

Each transmission consists of a START (S) condition sent by a master, followed by a 7-bit slave address plus a NOP/W bit. See Figures 3, 4, and 7.

#### START and STOP Conditions

SCL and SDA remain high when the interface is inactive. A master controller signals the beginning of a transmission with a START condition by transitioning SDA from high to low while SCL is high. The master controller issues a STOP condition by transitioning the SDA from low to high while SCL is high, after finishing communicating with the slave. The bus is then free for another transmission. See Figure 2.

#### Rit Transfer

One data bit is transferred during each clock pulse. The data on the SDA line must remain stable while SCL is high. See Figure 5.

#### Acknowledge

The acknowledge bit is a clocked 9th bit that the recipient uses to handshake receipt of each byte of data. See Figure 6. Each byte transferred requires a total of 9 bits. The master controller generates the 9th clock pulse, and the recipient pulls down SDA during the acknowledge clock pulse, so the SDA line remains stable low during the high period of the clock pulse.

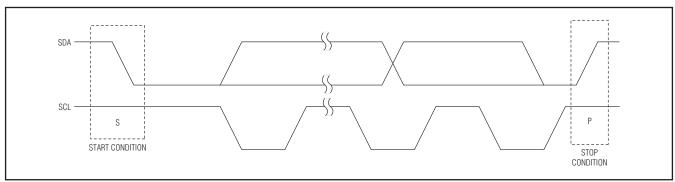


Figure 3. START and STOP Conditions

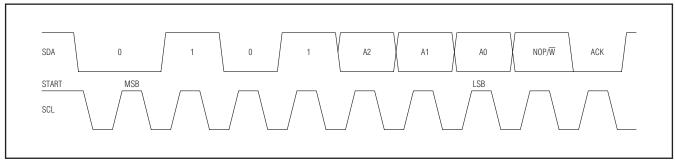


Figure 4. Slave Address

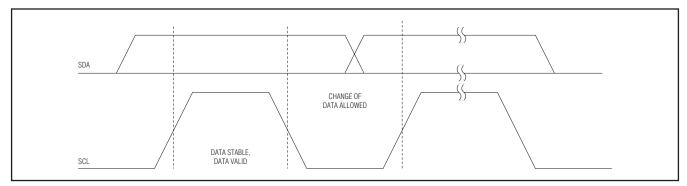


Figure 5. Bit Transfer

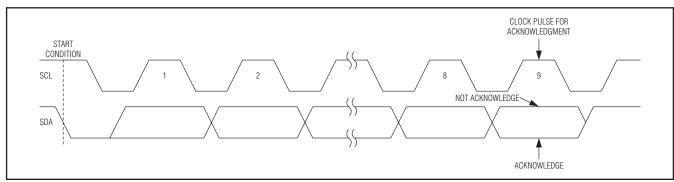


Figure 6. Acknowledge

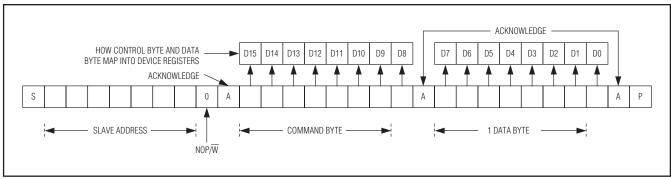


Figure 7. Command and Single Data Byte Received

#### Slave Address

The MAX5392 includes a 7-bit slave address (Figure 4). The 8th bit following the 7th bit of the slave address is the  $NOP/\overline{W}$  bit. Set the  $NOP/\overline{W}$  bit low for a write command and high for a no-operation command. The device does not support readback.

The device provides three address inputs (A0, A1, and A2), allowing up to eight devices to share a common bus (Table 1). The first 4 bits (MSBs) of the factory-set slave addresses are always 0101. A2, A1, and A0 set the next 3 bits of the slave address. Connect each address input to VDD or GND. Each device must have a unique address to share a common bus.

#### Message Format for Writing

Write to the devices by transmitting the device's slave address with  $NOP/\overline{W}$  (8th bit) set to zero, followed by at least 2 bytes of information. The first byte of informa-

tion is the command byte. The second byte is the data byte. The data byte goes into the internal register of the device as selected by the command byte (Figure 7 and Table 2).

**Table 1. Slave Addresses** 

AD	DRESS INP	UTS	SLAVE ADDRESS
A2	A1	A0	SLAVE ADDRESS
GND	GND	GND	0101000
GND	GND	V <sub>DD</sub>	0101001
GND	VDD	GND	0101010
GND	V <sub>DD</sub>	V <sub>DD</sub>	0101011
VDD	GND	GND	0101100
V <sub>DD</sub>	GND	V <sub>DD</sub>	0101101
VDD	VDD	GND	0101110
V <sub>DD</sub>	V <sub>DD</sub>	V <sub>DD</sub>	0101111

Table 2. I<sup>2</sup>C Command Byte Summary

			ADDRESS BYTE							ADDRESS BYTE COMMAND BYTE						DATA BYTE													
		1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	21	22	23	24	25	26	27	
SCL CYCLE NUMBER	START (S)	A6	A5	A4	А3	A2	A1	A0	W	ACK (A)	R7	R6	R5	R4	R3	R2	R1	R0	ACK (A)	D7	D6	D5	D4	D3	D2	D1	D0	ACK (A)	STOP (P)
REG A		0	1	0	1	A2	A1	A0	0		0	0	0	1	0	0	0	1		D7	D6	D5	D4	D3	D2	D1	D0		
REG B		0	1	0	1	A2	A1	A0	0		0	0	0	1	0	0	1	0		D7	D6	D5	D4	D3	D2	D1	D0		
REG A AND B		0	1	0	1	A2	A1	A0	0		0	0	0	1	0	0	1	1		D7	D6	D5	D4	D3	D2	D1	D0		

#### Command Byte

Use the command byte to select the destination of the wiper data. See Table 2.

#### **Command Descriptions**

**REG A:** The data byte writes to register A and the wiper of potentiometer A moves to the appropriate position. D[7:0] indicates the position of the wiper. D[7:0] = 00h moves the wiper to the position closest to LA. D[7:0] = FFh moves the wiper closest to HA. D[7:0] is 80h following power-on.

**REG B:** The data byte writes to register B and the wiper of potentiometer B moves to the appropriate position. D[7:0] indicates the position of the wiper. D[7:0] = 00h moves the wiper to the position closest to LB. D[7:0] = FFh moves the wiper to the position closest to HB. D[7:0] is 80h following power-on.

**REG A and B:** The data byte writes to registers A and B and the wipers of potentiometers A and B move to the appropriate position. D[7:0] indicates the position of the wiper. D[7:0] = 00h moves the wipers to the position closest to L\_. D[7:0] = FFh moves the wipers to the position closest to H\_. D[7:0] is 80h following power-on.

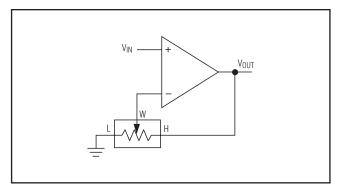


Figure 8. Variable Gain Noninverting Amplifier

### **Applications Information**

#### Variable Gain Amplifier

Figure 8 shows a potentiometer adjusting the gain of a noninverting amplifier. Figure 9 shows a potentiometer adjusting the gain of an inverting amplifier.

#### **Adjustable Dual Regulator**

Figure 10 shows an adjustable dual linear regulator using a dual potentiometer as two variable resistors.

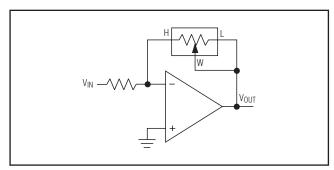


Figure 9. Variable Gain Inverting Amplifier

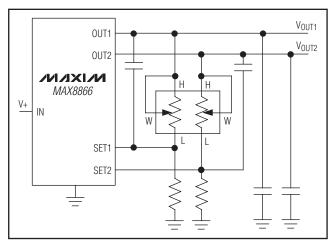


Figure 10. Adjustable Dual Linear Regulator

#### Adjustable Voltage Reference

Figure 11 shows an adjustable voltage reference circuit using a potentiometer as a voltage-divider.

#### Variable Gain Current to Voltage Converter

Figure 12 shows a variable gain current to voltage converter using a potentiometer as a variable resistor.

#### **LCD Bias Control**

Figure 13 shows a positive LCD bias control circuit using a potentiometer as a voltage-divider.

Figure 14 shows a positive LCD bias control circuit using a potentiometer as a variable resistor.

#### **Programmable Filter**

Figure 15 shows a programmable filter using a dual potentiometer.

#### Offset Voltage Adjustment Circuit

Figure 16 shows an offset voltage adjustment circuit using a dual potentiometer.

\_Chip Information

PROCESS: BiCMOS

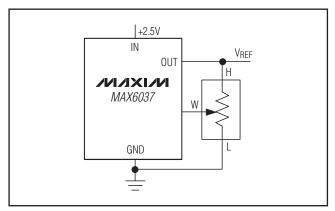


Figure 11. Adjustable Voltage Reference

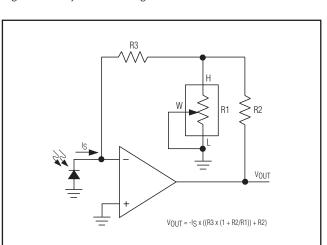


Figure 12. Variable Gain I-to-V Converter

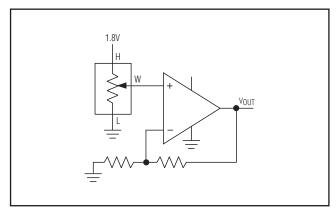


Figure 13. Positive LCD Bias Control Using a Voltage Divider

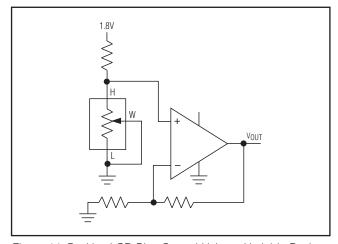
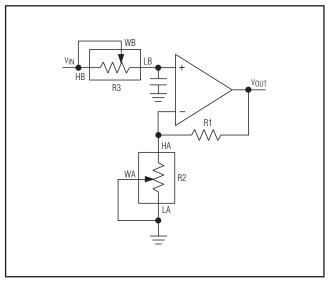
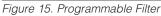


Figure 14. Positive LCD Bias Control Using a Variable Resistor





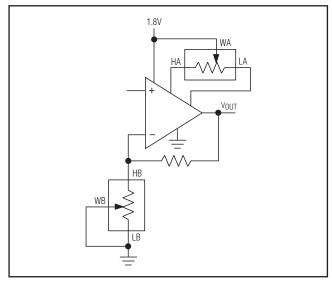


Figure 16. Offset Voltage Adjustment Circuit

## Package Information

For the latest package outline information and land patterns, go to <u>www.maxim-ic.com/packages</u>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE TYPE	PACKAGE CODE	OUTLINE NO.	LAND PATTERN NO.	
16 TSSOP	U16+2	<u>21-0066</u>	90-0117	

## **Revision History**

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	1/10	Initial release	
1	4/10	Added Soldering Temperature in <i>Absolute Maximum Ratings</i> ; corrected code in Conditions of -3dB Bandwidth specification in <i>Electrical Characteristics</i>	2, 3
2	11/10	Changed <i>Electrical Characteristics</i> heading and corrected Figures 9, 12, 14, 15, 16	2, 3, 4, 13, 14, 15

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