

# DAC7641

For most current data sheet and other product information, visit [www.burr-brown.com](http://www.burr-brown.com)

## 16-Bit, Voltage Output DIGITAL-TO-ANALOG CONVERTER

### FEATURES

- **LOW POWER: 2.5mW**
- **UNIPOLAR OR BIPOLAR OPERATION**
- **SETTLING TIME: 10 $\mu$ s to 0.003%**
- **15-BIT LINEARITY AND MONOTONICITY: -40°C to +85°C**
- **PROGRAMMABLE RESET TO MID-SCALE OR ZERO-SCALE**
- **DATA READBACK**
- **DOUBLE-BUFFERED DATA INPUTS**

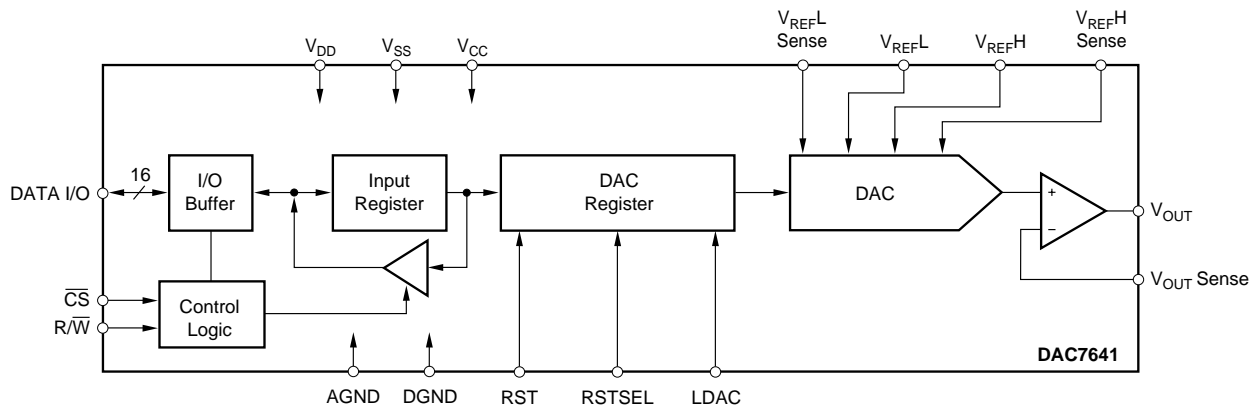
### APPLICATIONS

- **PROCESS CONTROL**
- **ATE PIN ELECTRONICS**
- **CLOSED-LOOP SERVO-CONTROL**
- **MOTOR CONTROL**
- **DATA ACQUISITION SYSTEMS**
- **DAC-PER-PIN PROGRAMMERS**

### DESCRIPTION

The DAC7641 is a 16-bit, voltage output digital-to-analog converter (DAC) with guaranteed 15-bit monotonic performance over the specified temperature range. It accepts 16-bit parallel input data, has double-buffered DAC input logic (allowing asynchronous update), and provides a readback mode of the internal input registers. Programmable asynchronous reset clears all registers to a mid-scale code of 8000<sub>H</sub> or to a zero-scale of 0000<sub>H</sub>. The DAC7641 can operate from a single +5V supply or from +5V and -5V supplies.

Low power and small size per DAC make the DAC7641 ideal for automatic test equipment, DAC-per-pin programmers, data acquisition systems, and closed-loop servo-control. The DAC7641 is available in a TQFP-32 package, and offers guaranteed specifications over the -40°C to +85°C temperature range.



International Airport Industrial Park • Mailing Address: PO Box 11400, Tucson, AZ 85734 • Street Address: 6730 S. Tucson Blvd., Tucson, AZ 85706 • Tel: (520) 746-1111  
Twx: 910-952-1111 • Internet: <http://www.burr-brown.com/> • Cable: BBRCORP • Telex: 066-6491 • FAX: (520) 889-1510 • Immediate Product Info: (800) 548-6132

# SPECIFICATIONS (Dual Supply)

At  $T_A = T_{MIN}$  to  $T_{MAX}$ ,  $V_{DD} = V_{CC} = +5V$ ,  $V_{SS} = -5V$ ,  $V_{REFH} = +2.5V$ , and  $V_{REFL} = -2.5V$ , unless otherwise noted.

PARAMETER	CONDITIONS	DAC7641Y			DAC7641YB			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
<b>ACCURACY</b>								
Linearity Error			±3	±4		±2	±3	LSB
Differential Linearity Error			±2	±3		±1	±2	LSB
Monotonicity, $T_{MIN}$ to $T_{MAX}$		14			15			Bits
Bipolar Zero Error			±1	±3		*	*	mV
Bipolar Zero Error Drift			5	10		*	*	ppm/°C
Full-Scale Error			±1	±3		*	*	mV
Full-Scale Error Drift			5	10		*	*	ppm/°C
Power Supply Rejection Ratio (PSRR)	At Full Scale		10	100		*	*	ppm/V
<b>ANALOG OUTPUT</b>								
Voltage Output	$V_{REF} = -2.5V$ , $R_L = 10k\Omega$ , $V_{SS} = -5V$	$V_{REFL}$ -1.25		$V_{REFH}$ +1.25	*		*	V
Output Current					*		*	mA
Maximum Load Capacitance	No Oscillation		500			*		pF
Short-Circuit Current			-10, +30			*		mA
Short-Circuit Duration	GND or $V_{CC}$ or $V_{SS}$		Indefinite			*		
<b>REFERENCE INPUT</b>								
Ref High Input Voltage Range		$V_{REFL} + 1.25$ -2.5		+2.5 $V_{REFH} - 1.25$	*		*	V
Ref Low Input Voltage Range					*		*	V
Ref High Input Current			500			*		μA
Ref Low Input Current			-500			*		μA
<b>DYNAMIC PERFORMANCE</b>								
Settling Time	To ±0.003%, 5V Output Step		8	10		*	*	μs
Digital Feedthrough			2			*		nV-s
Output Noise Voltage	f = 10kHz		60			*		nV/√Hz
DAC Glitch	7FFF <sub>H</sub> to 8000 <sub>H</sub> or 8000 <sub>H</sub> to 7FFF <sub>H</sub>		40			*		nV-s
<b>DIGITAL INPUT</b>								
$V_{IH}$		$0.7 \cdot V_{DD}$			*		*	V
$V_{IL}$				$0.3 \cdot V_{DD}$			*	V
$I_{IH}$				±10			*	μA
$I_{IL}$				±10			*	μA
<b>DIGITAL OUTPUT</b>								
$V_{OH}$	$I_{OH} = -0.8mA$	3.6	4.5		*	*	*	V
$V_{OL}$	$I_{OL} = 1.2mA$		0.3	0.4	*	*	*	V
<b>POWER SUPPLY</b>								
$V_{DD}$		+4.75	+5.0	+5.25	*	*	*	V
$V_{CC}$		+4.75	+5.0	+5.25	*	*	*	V
$V_{SS}$		-5.25	-5.0	-4.75	*	*	*	V
$I_{CC}$			0.4	0.5		*	*	mA
$I_{DD}$			15			*	*	μA
$I_{SS}$		-0.6	-0.5	-0.4	*	*	*	mA
Power			4	5.5		*	*	mW
<b>TEMPERATURE RANGE</b>								
Specified Performance		-40		+85	*		*	°C

\* Specifications same as DAC7641Y.

The information provided herein is believed to be reliable; however, BURR-BROWN assumes no responsibility for inaccuracies or omissions. BURR-BROWN assumes no responsibility for the use of this information, and all use of such information shall be entirely at the user's own risk. Prices and specifications are subject to change without notice. No patent rights or licenses to any of the circuits described herein are implied or granted to any third party. BURR-BROWN does not authorize or warrant any BURR-BROWN product for use in life support devices and/or systems.

# SPECIFICATIONS (Single Supply)

At  $T_A = T_{MIN}$  to  $T_{MAX}$ ,  $V_{DD} = V_{CC} = +5V$ ,  $V_{SS} = 0V$ ,  $V_{REFH} = +2.5V$ , and  $V_{REFL} = 0V$ , unless otherwise noted.

PARAMETER	CONDITIONS	DAC7641Y			DAC7641YB			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
<b>ACCURACY</b>								
Linearity Error <sup>(1)</sup>			±3	±4		±2	±3	LSB
Differential Linearity Error			±2	±3		±1	±2	LSB
Monotonicity, $T_{MIN}$ to $T_{MAX}$		14			15			Bits
Zero Scale Error			±1	±3		*	*	mV
Zero Scale Error Drift			5	10		*	*	ppm/°C
Full-Scale Error			±1	±3		*	*	mV
Full-Scale Error Drift			5	10		*	*	ppm/°C
Power Supply Rejection Ratio (PSRR)	At Full Scale		10	100		*	*	ppm/V
<b>ANALOG OUTPUT</b>								
Voltage Output	$V_{REFL} = 0V$ , $V_{SS} = 0V$ , $R_L = 10k\Omega$	0		$V_{REFH}$	*		*	V
Output Current		-1.25		+1.25	*		*	mA
Maximum Load Capacitance	No Oscillation		500			*		pF
Short-Circuit Current			±30			*		mA
Short-Circuit Duration	GND or $V_{CC}$		Indefinite			*		
<b>REFERENCE INPUT</b>								
Ref High Input Voltage Range		$V_{REFL} + 1.25$		+2.5	*		*	V
Ref Low Input Voltage Range		0		$V_{REFH} - 1.25$	*		*	V
Ref High Input Current			250			*		μA
Ref Low Input Current			-250			*		μA
<b>DYNAMIC PERFORMANCE</b>								
Settling Time	To ±0.003%, 2.5V Output Step		8	10		*	*	μs
Digital Feedthrough			2			*		nV-s
Output Noise Voltage, $f = 10kHz$			60			*		nV/√Hz
DAC Glitch	$7FFF_H$ to $8000_H$ or $8000_H$ to $7FFF_H$		40			*		nV-s
<b>DIGITAL INPUT</b>								
$V_{IH}$		$0.7 \cdot V_{DD}$			*			V
$V_{IL}$				$0.3 \cdot V_{DD}$			*	V
$I_{IH}$				±10			*	μA
$I_{IL}$				±10			*	μA
<b>DIGITAL OUTPUT</b>								
$V_{OH}$	$I_{OH} = -0.8mA$	3.6	4.5		*	*	*	V
$V_{OL}$	$I_{OL} = 1.2mA$		0.3	0.4	*	*	*	V
<b>POWER SUPPLY</b>								
$V_{DD}$		+4.75	+5.0	+5.25	*	*	*	V
$V_{CC}$		+4.75	+5.0	+5.25	*	*	*	V
$V_{SS}$		0	0	0	*	*	*	V
$I_{CC}$			0.4	0.5		*	*	mA
$I_{DD}$			15			*	*	μA
Power			1.8	2.5		*	*	mW
<b>TEMPERATURE RANGE</b>								
Specified Performance		-40		+85	*		*	°C

\* Specifications same as DAC7641Y.

NOTE: (1) If  $V_{SS} = 0V$  specification applies at Code 0040<sub>H</sub> and above due to possible negative zero-scale error.

## ABSOLUTE MAXIMUM RATINGS<sup>(1)</sup>

$V_{SS}$ to $V_{SS}$ .....	-0.3V to 11V
$V_{DD}$ to GND .....	-0.3V to 5.5V
$V_{REFL}$ to GND .....	-0.3V to ( $V_{SS} - V_{CC}$ )
$V_{REFH}$ to GND .....	-0.3V to ( $V_{SS} - V_{CC}$ )
$V_{REFH}$ to $V_{REFL}$ .....	-0.3V to +11V
Digital Input Voltage to GND .....	-0.3V to $V_{DD} + 0.3V$
Digital Output Voltage to GND .....	-0.3V to $V_{DD} + 0.3V$
Maximum Junction Temperature .....	+150°C
Operating Temperature Range .....	-40°C to +85°C
Storage Temperature Range .....	-65°C to +150°C
Lead Temperature (soldering, 10s) .....	+300°C

NOTE: (1) Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. Exposure to absolute maximum conditions for extended periods may affect device reliability.



## ELECTROSTATIC DISCHARGE SENSITIVITY

This integrated circuit can be damaged by ESD. Burr-Brown recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

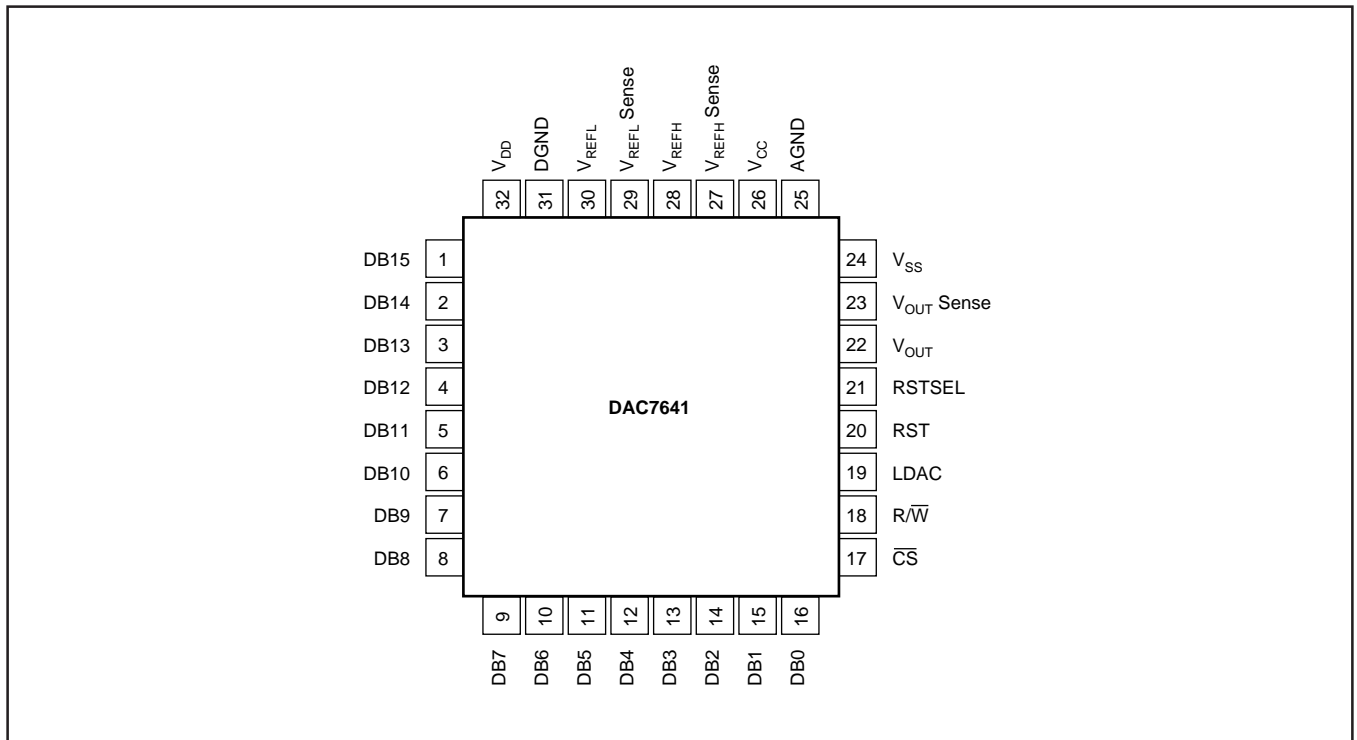
ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

## PACKAGE/ORDERING INFORMATION

PRODUCT	MINIMUM RELATIVE ACCURACY (LSB)	DIFFERENTIAL NONLINEARITY (LSB)	PACKAGE	PACKAGE DRAWING NUMBER	SPECIFICATION TEMPERATURE RANGE	ORDERING NUMBER <sup>(1)</sup>	TRANSPORT MEDIA
DAC7641Y "	±4 "	±3 "	TQFP-32 "	351 "	-40°C to +85°C "	DAC7641Y/250 DAC7641Y/2K	Tape and Reel Tape and Reel
DAC7641YB "	±3 "	±2 "	TQFP-32 "	351 "	-40°C to +85°C "	DAC7641YB/250 DAC7641YB/2K	Tape and Reel Tape and Reel

NOTES: (1) Models with a slash (/) are available only in Tape and Reel in the quantities indicated (e.g., /2K indicates 2000 devices per reel). Ordering 2000 pieces of "DAC7641Y/2K" will get a single 2000-piece Tape and Reel.

## PIN CONFIGURATION



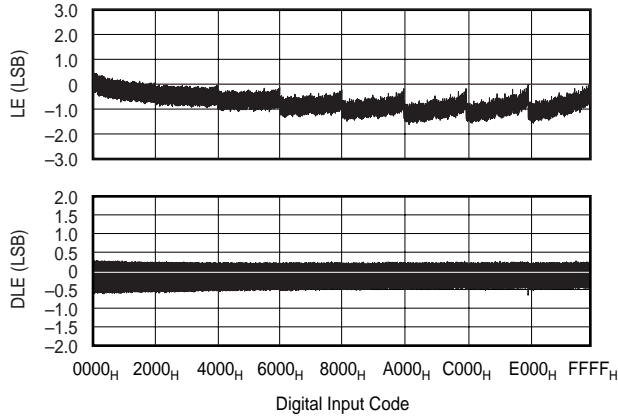
## PIN DESCRIPTIONS

PIN	NAME	DESCRIPTION	PIN	NAME	DESCRIPTION
1	DB15	Data Bit 15, MSB	19	LDAC	DAC Load Strobe, rising-edge triggered.
2	DB14	Data Bit 14	20	RST	Reset, rising-edge triggered. Depending on the state of RSTSEL, the DAC registers are set to either mid-scale or zero.
3	DB13	Data Bit 13	21	RSTSEL	Reset Select. Determines the action of RST. If HIGH, a RST command will set the DAC registers to mid-scale. If LOW, a RST command will set the DAC registers to zero.
4	DB12	Data Bit 12	22	$V_{OUT}$	DAC Voltage Output
5	DB11	Data Bit 11	23	$V_{OUT}$ Sense	DAC Output Amplifier Inverting Input. Used to close the feedback loop at the load.
6	DB10	Data Bit 10	24	$V_{SS}$	Negative Power Supply
7	DB9	Data Bit 9	25	AGND	Analog Ground
8	DB8	Data Bit 8	26	$V_{CC}$	Positive Power Supply
9	DB7	Data Bit 7	27	$V_{REFH}$ Sense	DAC Reference High Sense Input
10	DB6	Data Bit 6	28	$V_{REFH}$	DAC Reference High Input
11	DB5	Data Bit 5	29	$V_{REFL}$ Sense	DAC Reference Low Sense Input
12	DB4	Data Bit 4	30	$V_{REFL}$	DAC Reference Low Input
13	DB3	Data Bit 3	31	DGND	Digital Ground
14	DB2	Data Bit 2	32	$V_{DD}$	Positive Power Supply
15	DB1	Data Bit 1			
16	DB0	Data Bit 0, LSB			
17	$\overline{CS}$	Chip Select, active low.			
18	$R/\overline{W}$	Enabled by $\overline{CS}$ , controls data read and write from the input register.			

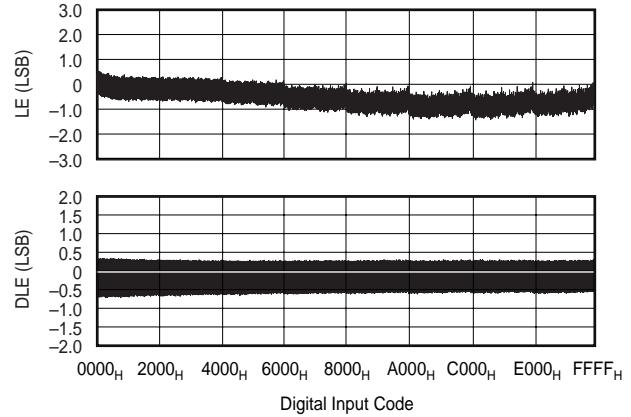
# TYPICAL PERFORMANCE CURVES: $V_{SS} = 0V$

At  $T_A = +25^\circ C$ ,  $V_{DD} = +5V$ ,  $V_{SS} = 0V$ ,  $V_{REFH} = +2.5V$ ,  $V_{REFL} = 0V$ , representative unit, unless otherwise specified.

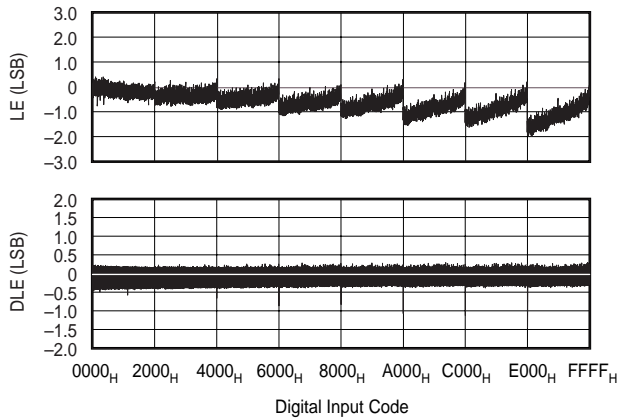
LINEARITY ERROR AND DIFFERENTIAL LINEARITY ERROR vs CODE (+25°C)



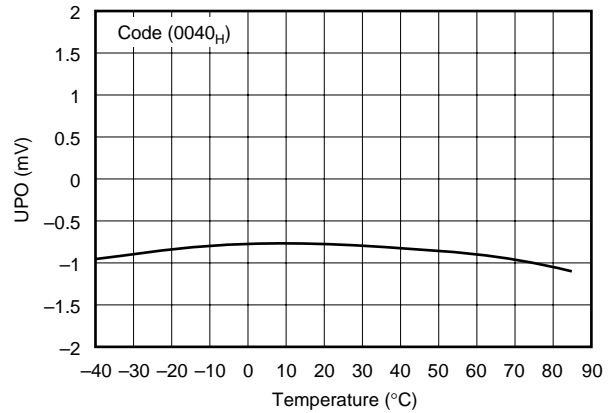
LINEARITY ERROR AND DIFFERENTIAL LINEARITY ERROR vs CODE (+85°C)



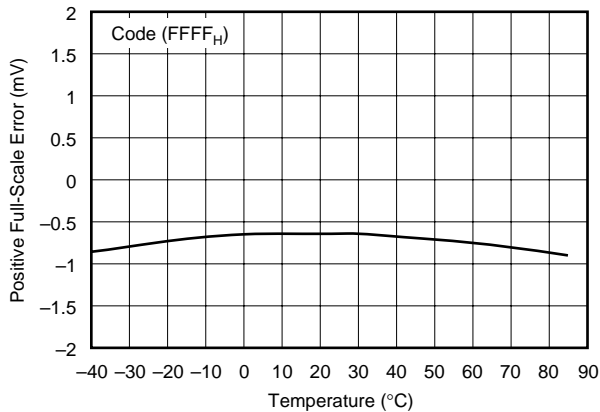
LINEARITY ERROR AND DIFFERENTIAL LINEARITY ERROR vs CODE (-40°C)



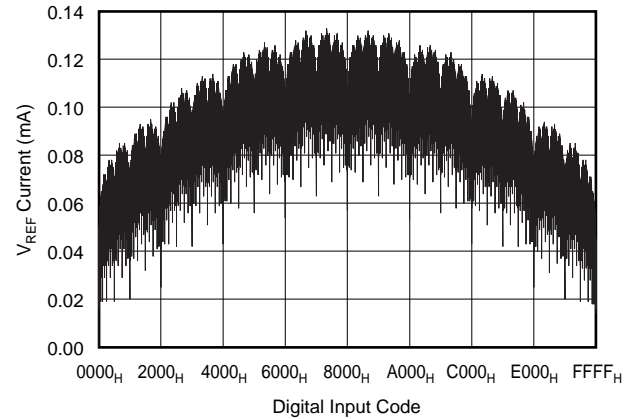
ZERO-SCALE ERROR vs TEMPERATURE



POSITIVE FULL-SCALE ERROR vs TEMPERATURE

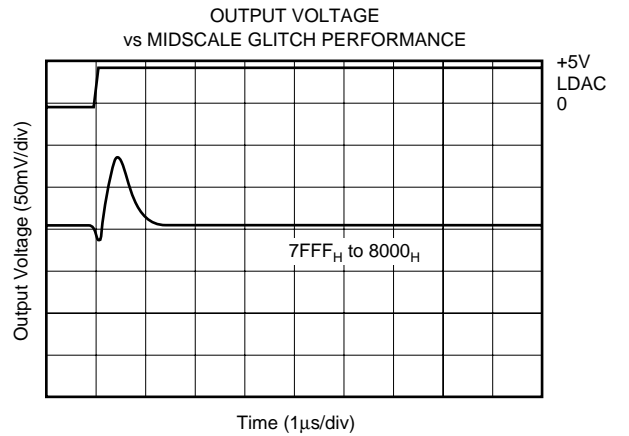
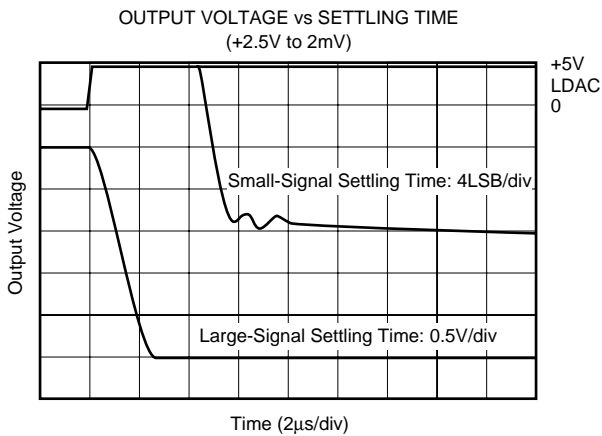
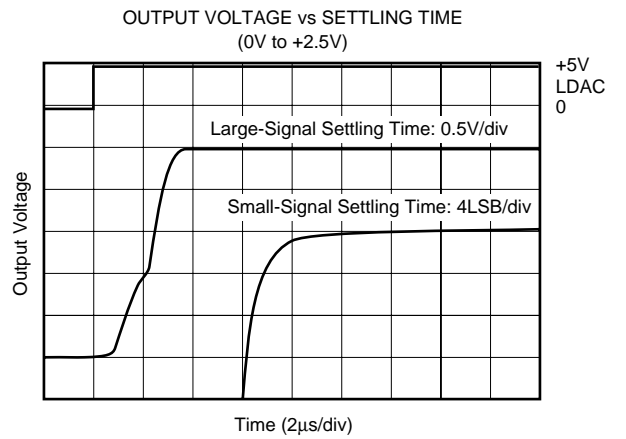
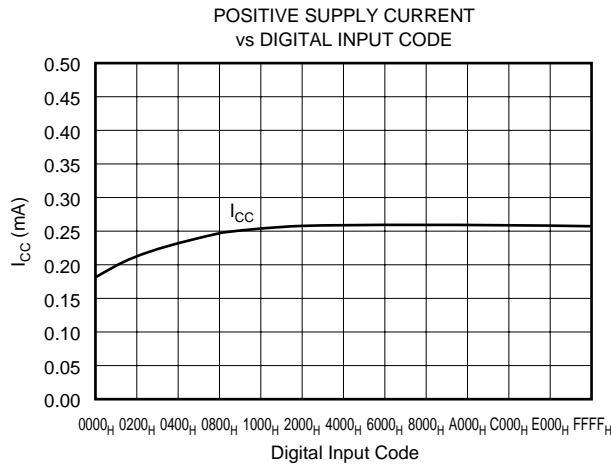
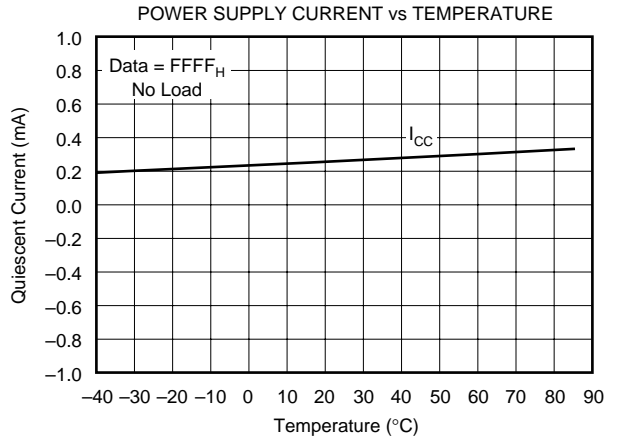
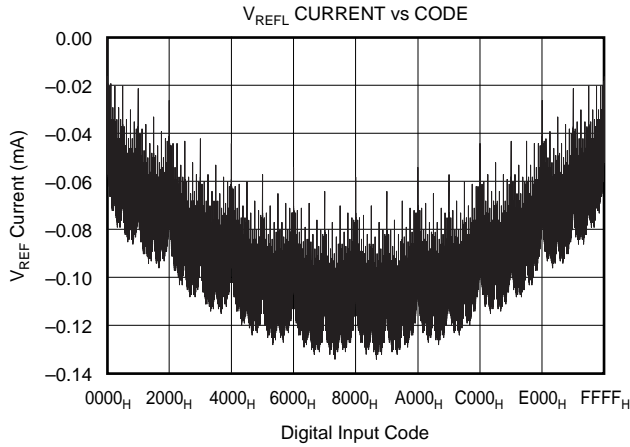


$V_{REFH}$  CURRENT vs CODE



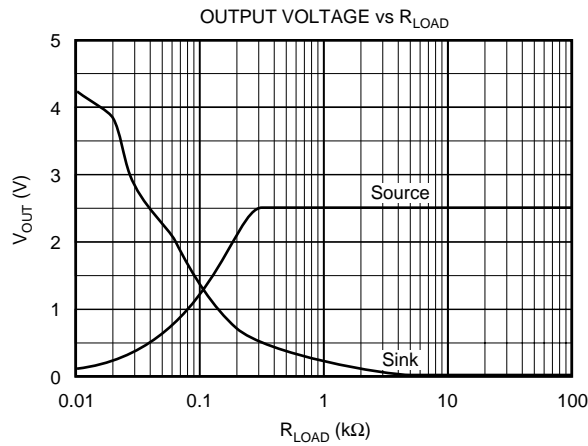
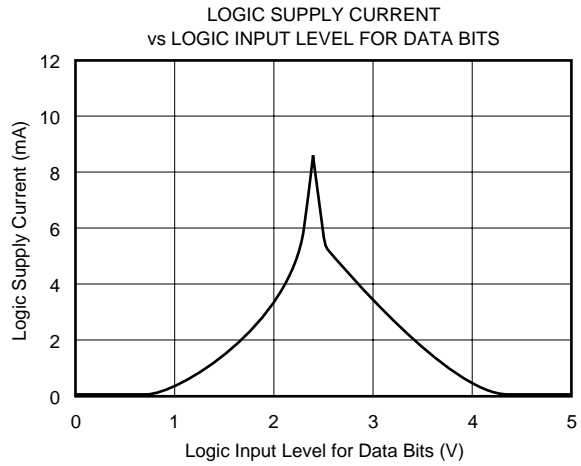
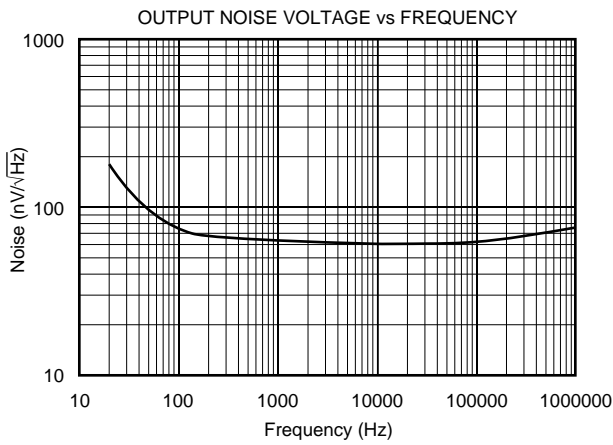
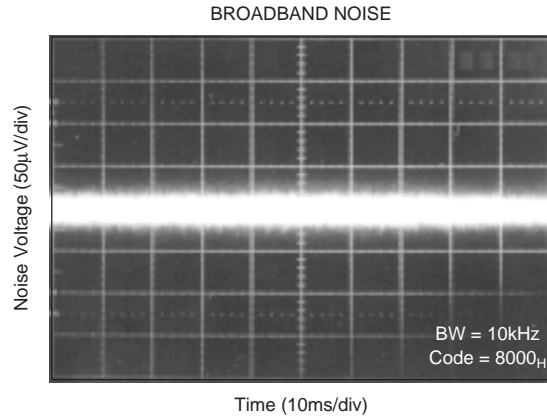
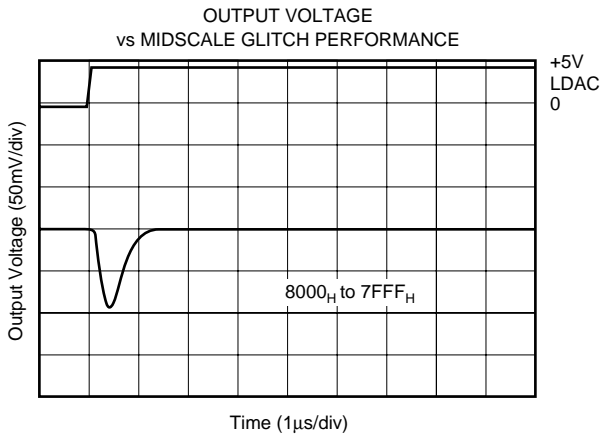
# TYPICAL PERFORMANCE CURVES: $V_{SS} = 0V$ (Cont.)

At  $T_A = +25^\circ C$ ,  $V_{DD} = +5V$ ,  $V_{SS} = 0V$ ,  $V_{REFH} = +2.5V$ ,  $V_{REFL} = 0V$ , representative unit, unless otherwise specified.



# TYPICAL PERFORMANCE CURVES: $V_{SS} = 0V$ (Cont.)

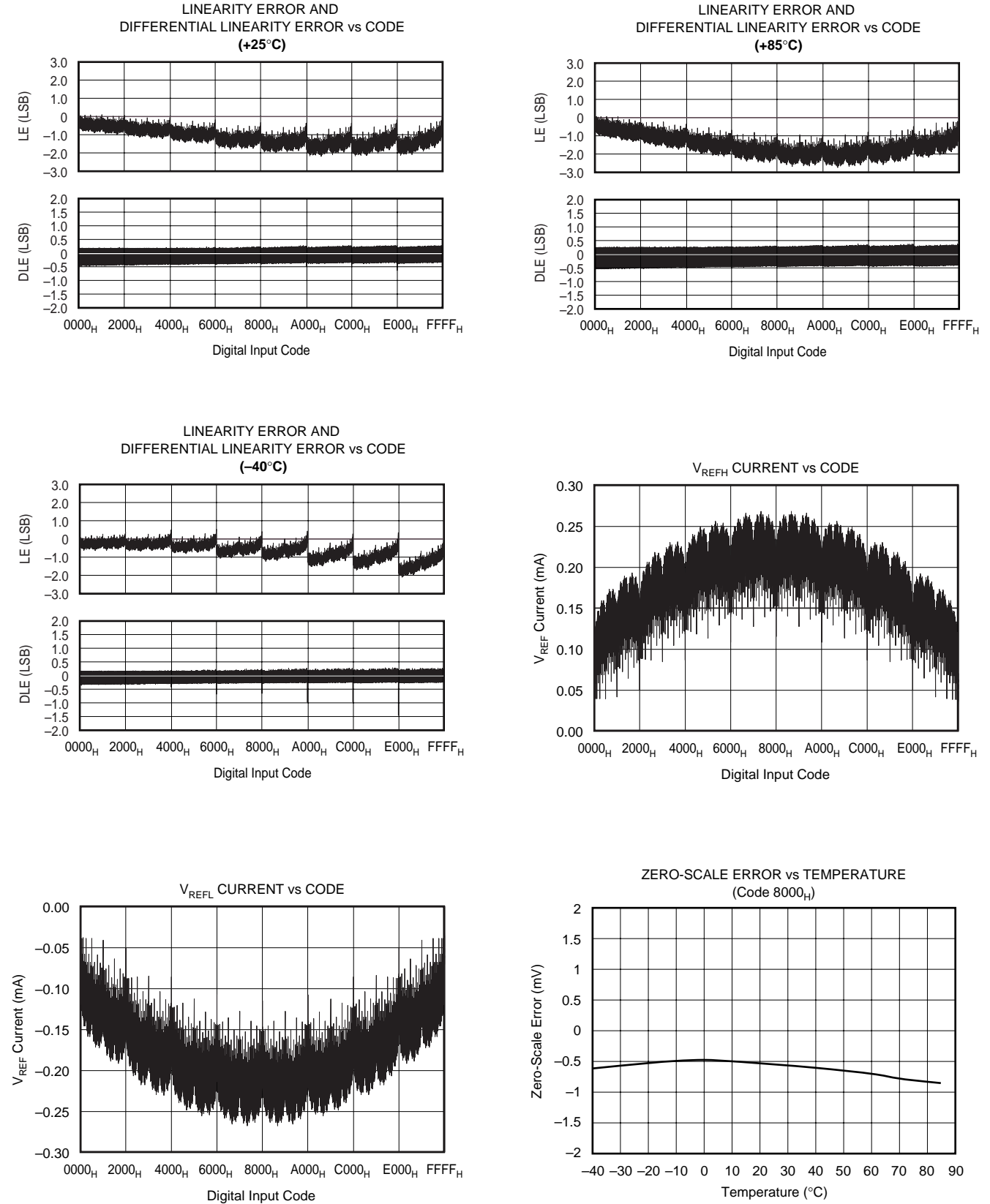
At  $T_A = +25^\circ C$ ,  $V_{DD} = +5V$ ,  $V_{SS} = 0V$ ,  $V_{REFH} = +2.5V$ ,  $V_{REFL} = 0V$ , representative unit, unless otherwise specified.





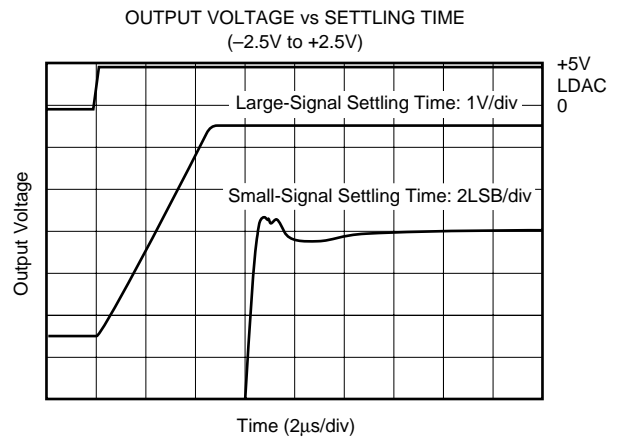
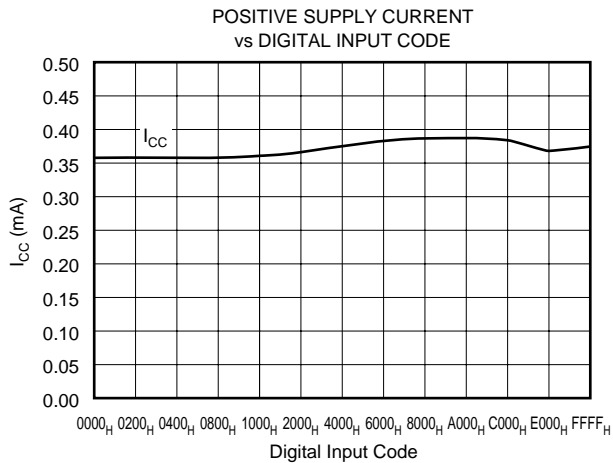
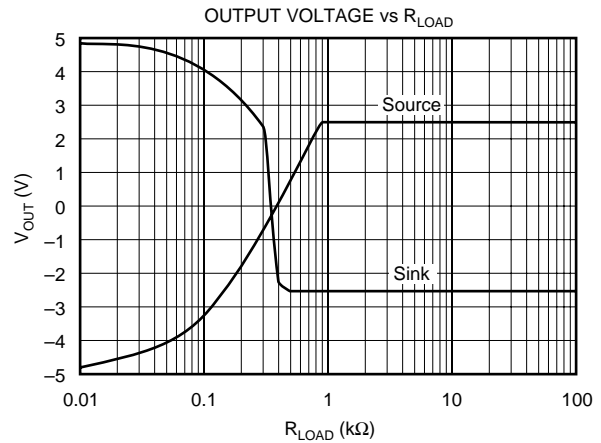
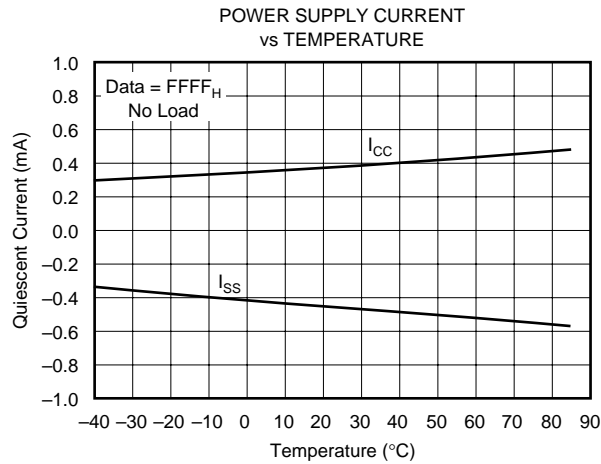
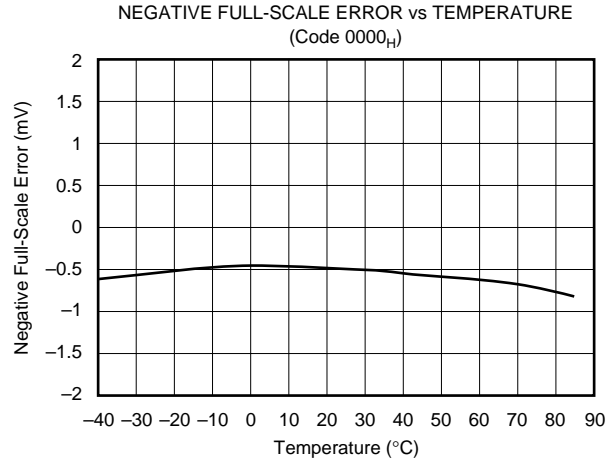
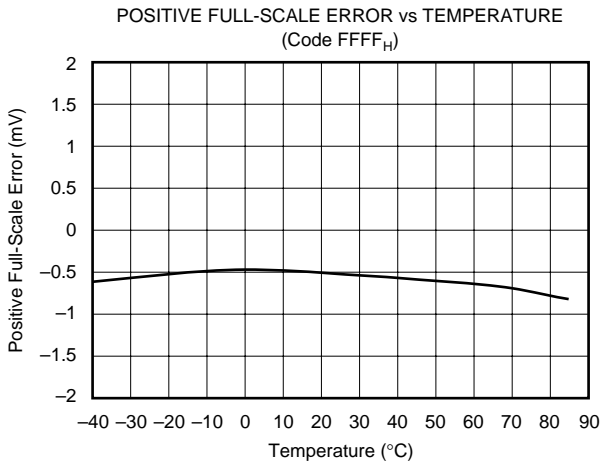
# TYPICAL PERFORMANCE CURVES: $V_{SS} = -5V$

At  $T_A = +25^\circ C$ ,  $V_{DD} = +5V$ ,  $V_{SS} = -5V$ ,  $V_{REFH} = +2.5V$ ,  $V_{REFL} = -2.5V$ , representative unit, unless otherwise specified.



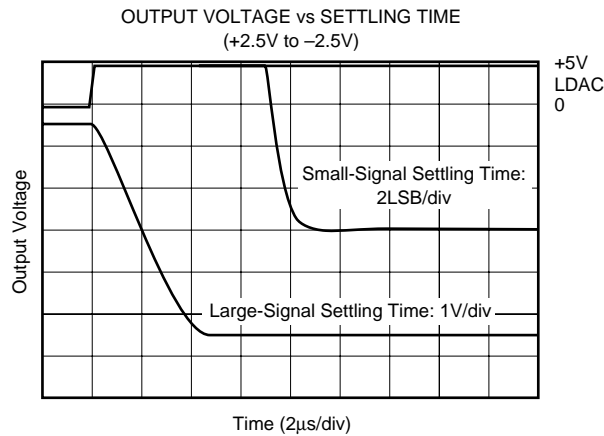
# TYPICAL PERFORMANCE CURVES: $V_{SS} = -5V$ (Cont.)

At  $T_A = +25^\circ C$ ,  $V_{DD} = +5V$ ,  $V_{SS} = -5V$ ,  $V_{REFH} = +2.5V$ ,  $V_{REFL} = -2.5V$ , representative unit, unless otherwise specified.



# TYPICAL PERFORMANCE CURVES: $V_{SS} = -5V$ (Cont.)

At  $T_A = +25^\circ C$ ,  $V_{DD} = +5V$ ,  $V_{SS} = -5V$ ,  $V_{REFH} = +2.5V$ ,  $V_{REFL} = -2.5V$ , representative unit, unless otherwise specified.



## THEORY OF OPERATION

The DAC7641 is a voltage output, 16-bit digital-to-analog converter (DAC). The architecture is an R-2R ladder configuration with the three MSBs segmented, followed by an operational amplifier that serves as a buffer (see Figure 1). The minimum voltage output (zero-scale) and maximum voltage output (full-scale) are set by the external voltage

references  $V_{REFL}$  and  $V_{REFH}$ , respectively. The digital input is a 16-bit parallel word and the DAC input register offers a readback capability. The converters can be powered from either a single +5V supply or a dual  $\pm 5V$  supply. The device offers a reset function which immediately sets all DAC output voltages and DAC registers to mid-scale code 8000<sub>H</sub> or to zero-scale code 0000<sub>H</sub>. See Figures 2 and 3 for the basic operation of the DAC7641.

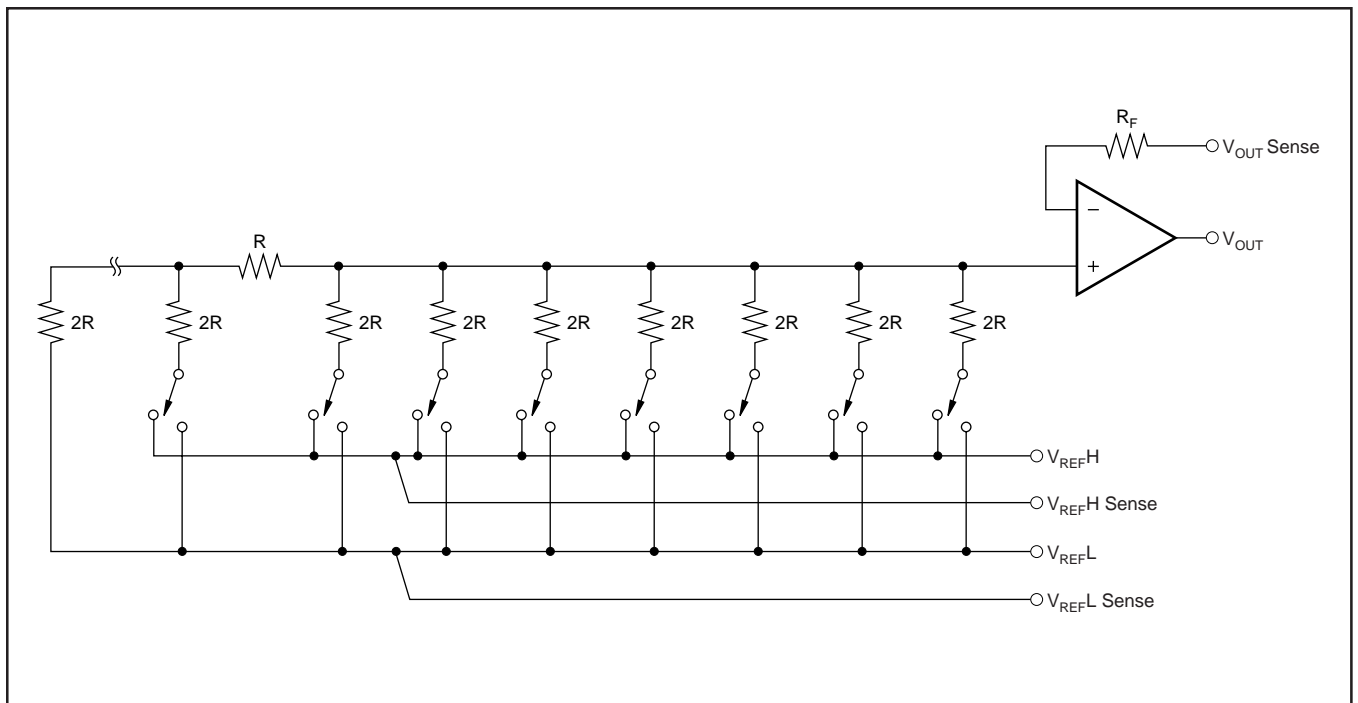


FIGURE 1. DAC7641 Architecture.

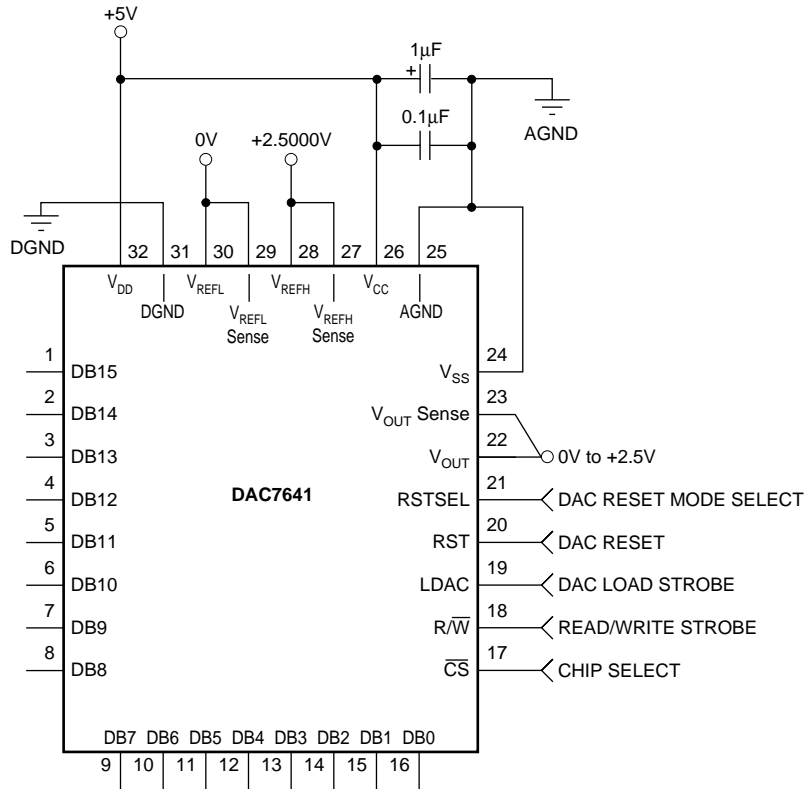


FIGURE 2. Single-Supply Operation.

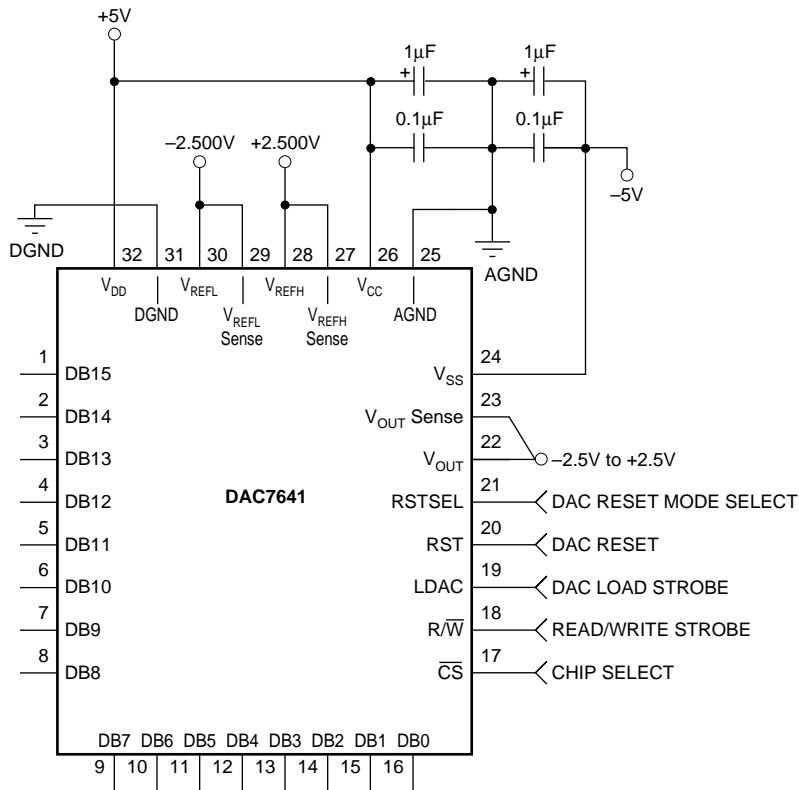


FIGURE 3. Dual-Supply Operation.

## ANALOG OUTPUTS

When  $V_{SS} = -5V$  (dual supply operation), the output amplifier can swing to within 2.25V of the supply rails, guaranteed over the  $-40^{\circ}C$  to  $+85^{\circ}C$  temperature range. With  $V_{SS} = 0V$  (single-supply operation), and with  $R_{LOAD}$  also connected to ground, the output can swing to ground. Care must be taken when measuring the zero-scale error with  $V_{SS} = 0V$ . Since the output voltage cannot swing below ground, the output voltage may not change for the first few digital input codes (0000<sub>H</sub>, 0001<sub>H</sub>, 0002<sub>H</sub>, etc.) if the output amplifier has a negative offset. At the negative limit of  $-2mV$ , the first specified output starts at code 0040<sub>H</sub>.

Due to the high accuracy of these D/A converters, system design problems such as grounding and contact resistance become very important. A 16-bit converter with a 2.5V full-scale range has a 1LSB value of  $38\mu V$ . With a load current of 1mA, series wiring and connector resistance (see Figure 4) of only  $40m\Omega$  ( $R_{W2}$ ) will cause a voltage drop of  $40\mu V$ . To understand what this means in terms of a system layout, the resistivity of a typical 1 ounce copper-clad printed circuit board is  $1/2 m\Omega$  per square. For a 1mA load, a 10 milli-inch wide printed circuit conductor 600 milli-inches long will result in a voltage drop of  $30\mu V$ .

The DAC7641 offers a force and sense output configuration for the high open-loop gain output amplifier. This feature allows the loop around the output amplifier to be closed at the load (see Figure 4), thus ensuring an accurate output voltage.

## REFERENCE INPUTS

The reference inputs,  $V_{REFL}$  and  $V_{REFH}$ , can be any voltage between  $V_{SS} + 2.5V$  and  $V_{CC} - 2.5V$  provided that  $V_{REFH}$  is at least 1.25V greater than  $V_{REFL}$ . The minimum output of each DAC is equal to  $V_{REFL}$  plus a small offset voltage (essentially, the offset of the output op amp). The maximum output is equal to  $V_{REFH}$  plus a similar offset voltage. Note that  $V_{SS}$  (the negative power supply) must either be

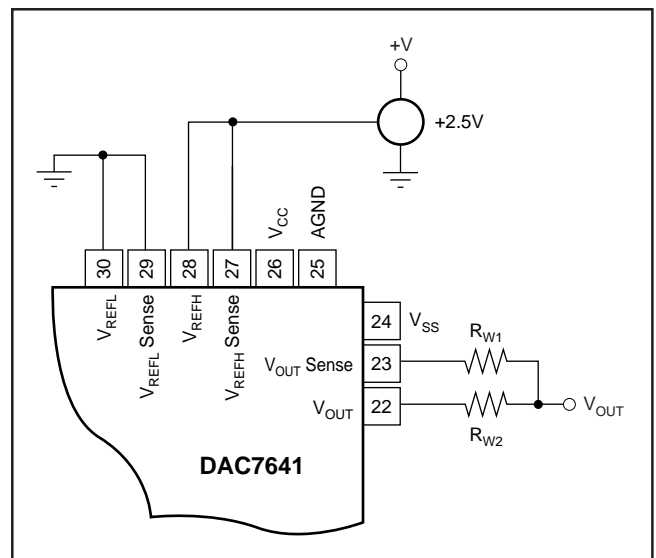


FIGURE 4. Analog Output Closed-Loop Configuration.  $R_W$  represents wiring resistances.

connected to ground or must be in the range of  $-4.75V$  to  $-5.25V$ . The voltage on  $V_{SS}$  sets several bias points within the converter. If  $V_{SS}$  is not in one of these two configurations, the bias values may be in error and proper operation of the device is not guaranteed.

The current into the  $V_{REFH}$  input and out of  $V_{REFL}$  depends on the DAC output voltages and can vary from a few microamps to approximately 0.5mA. The reference input appears as a varying load to the reference. If the reference can sink or source the required current, a reference buffer is not required. The DAC7641 features a reference drive and sense connection such that the internal errors caused by the changing reference current and the circuit impedances can be minimized. Figures 5 through 13 show different reference configurations and the effect on the linearity and differential linearity.

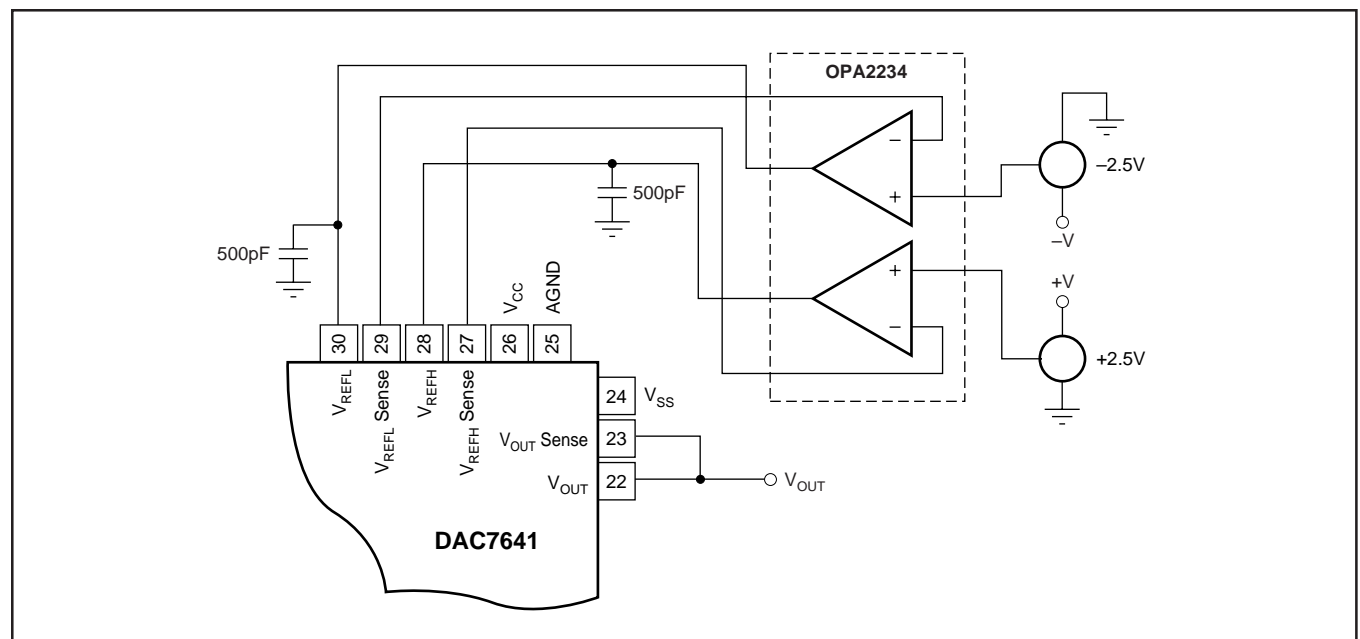


FIGURE 5. Dual Supply Configuration-Buffered References, used for Dual Supply Performance Curves.

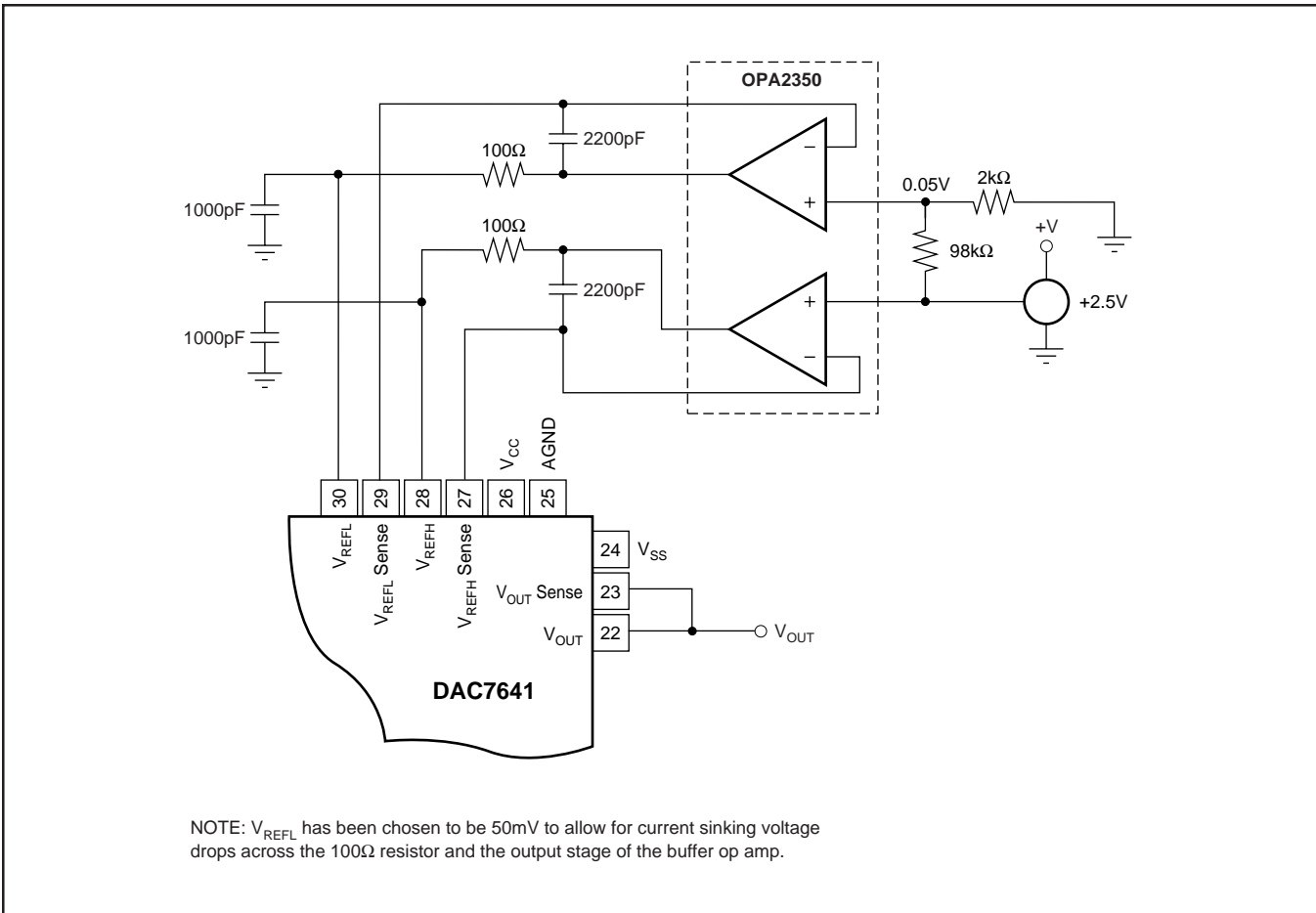


FIGURE 6. Single-Supply Buffered Reference with a Reference Low of 50mV.

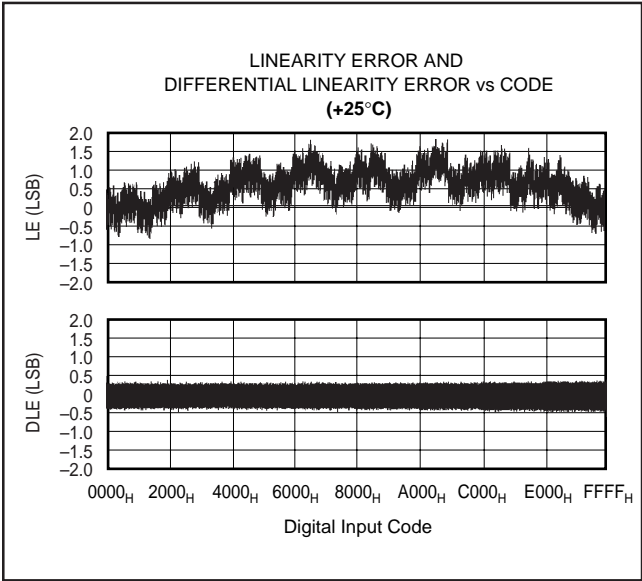


FIGURE 7. Integral Linearity and Differential Linearity Error Curves for Figure 6.

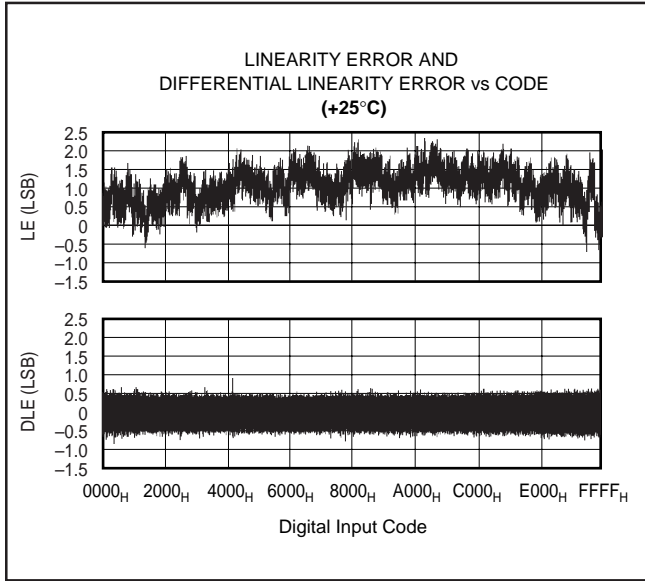


FIGURE 8. Integral Linearity and Differential Linearity Error Curves for Figure 9.

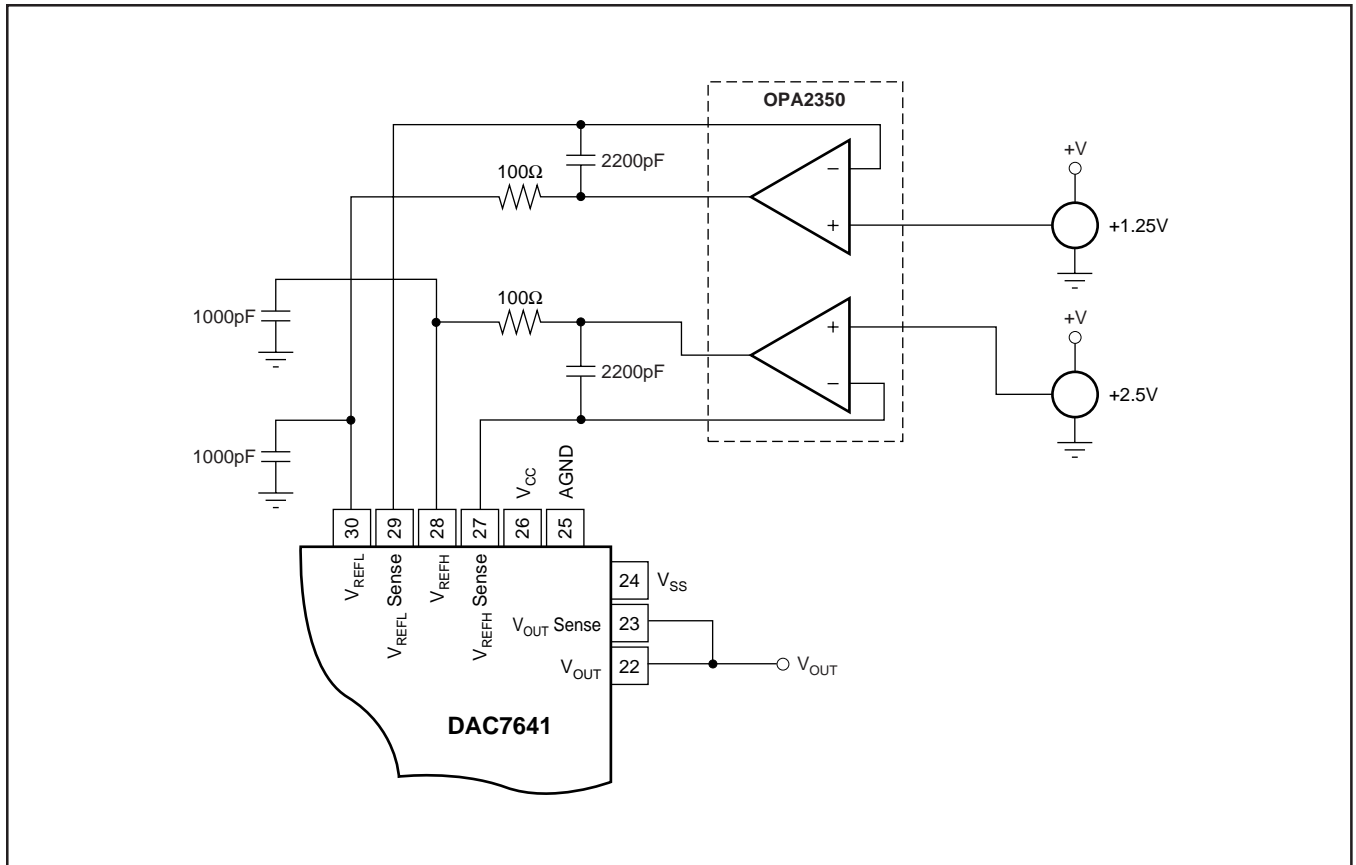


FIGURE 9. Single-Supply Buffered Reference with  $V_{REFL} = +1.25V$  and  $V_{REFH} = +2.5V$ .

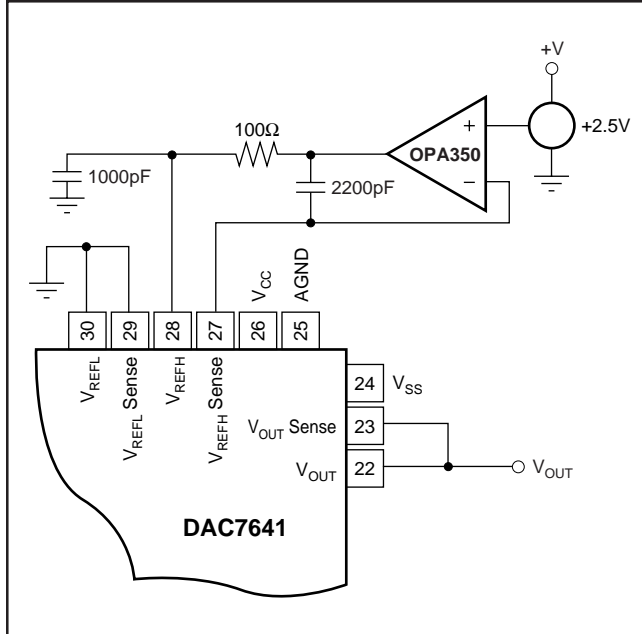


FIGURE 10. Single-Supply Buffered  $V_{REFH}$ .

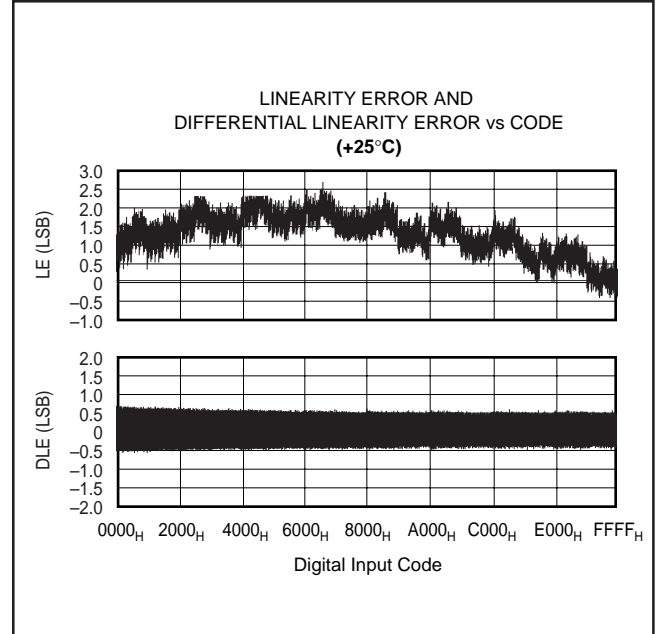


FIGURE 11. Linearity and Differential Linearity Error Curves for Figure 10.

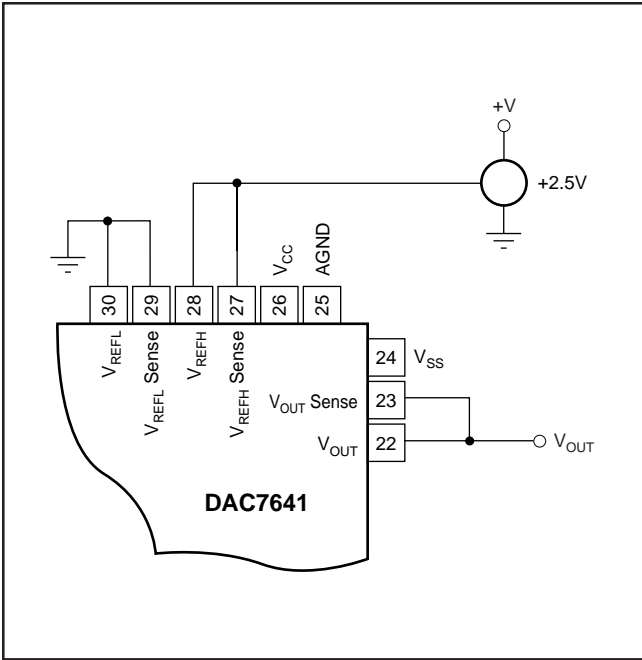


FIGURE 12. Low Cost Single-Supply Configuration.

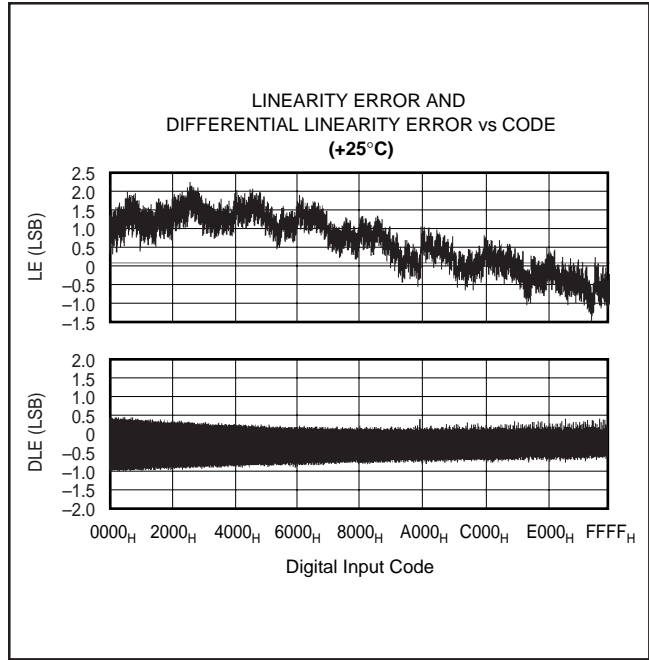


FIGURE 13. Linearity and Differential Linearity Error Curves for Figure 12.

## DIGITAL INTERFACE

Table I shows the basic control logic for the DAC7641. Note that the internal register is edge triggered and not level triggered. When the LDAC signal is transitioned to HIGH, the digital word currently in the register is latched.

The double-buffered architecture is designed so that the DAC input register can be written to at any time.

R/W	CS	RST	RSTSEL	LDAC	REGISTER	REGISTER	INPUT MODE
L	L	H	X	X	Write	Hold	Write Input
H	L	H	X	X	Read	Hold	Read Input
X	H	H	X	↑	Hold	Write	Update
X	H	H	X	H	Hold	Hold	Hold
X	X	↑	L	X		Reset to Zero	Reset to Zero
X	X	↑	H	X		Reset to Midscale	Reset to Midscale

TABLE I. DAC7641 Logic Truth Table.

## DIGITAL TIMING

Figure 14 and Table II provide detailed timing for the digital interface of the DAC7641.

## DIGITAL INPUT CODING

The DAC7641 input data is in Straight Binary format. The output voltage is given by Equation 1.

$$V_{OUT} = V_{REF}L + \frac{(V_{REFH} - V_{REFL}) \cdot N}{65,536} \quad (1)$$

where N is the digital input code. This equation does not include the effects of offset (zero-scale) or gain (full-scale) errors.



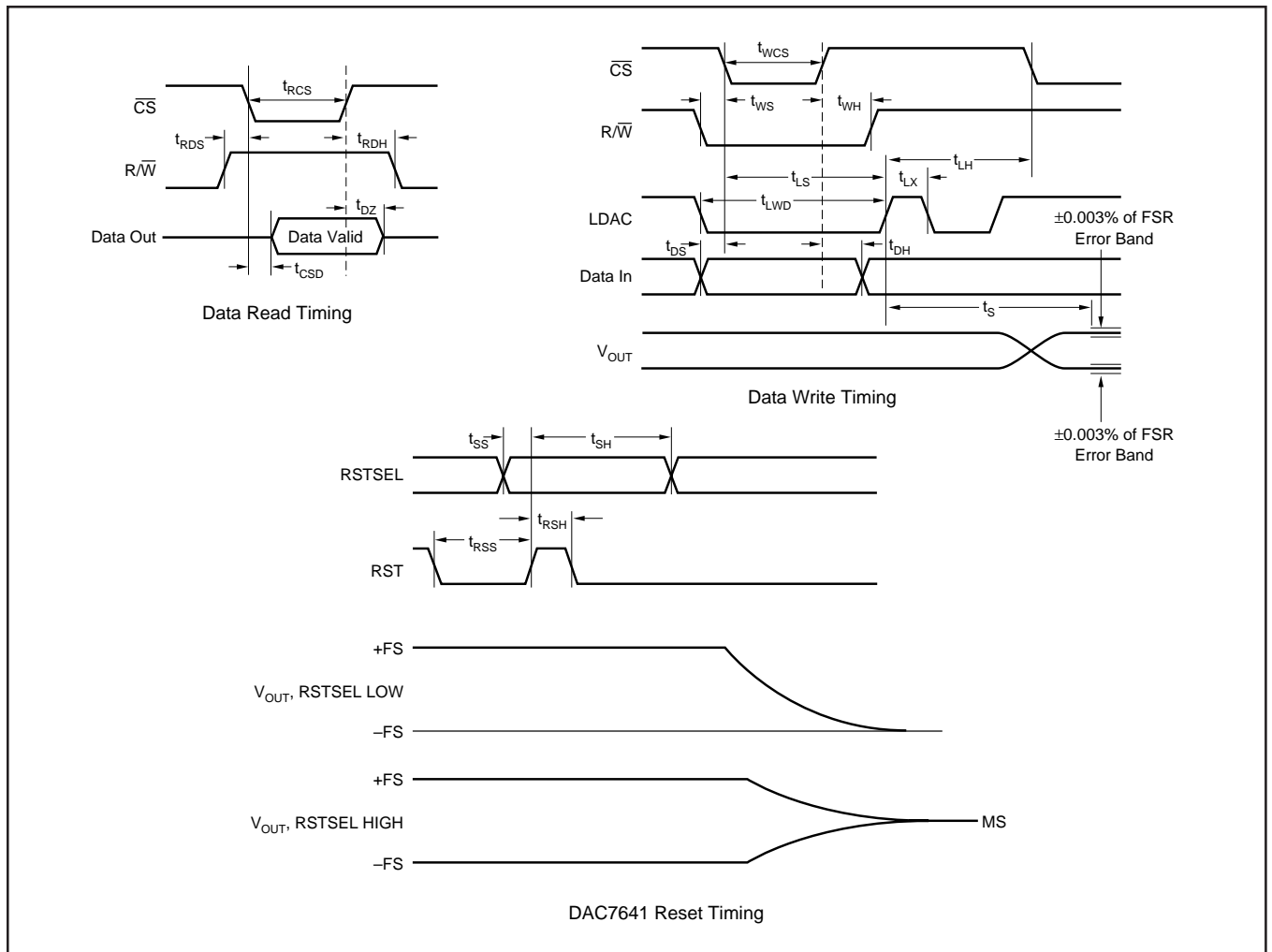


FIGURE 14. Digital Input and Output Timing.

SYMBOL	DESCRIPTION	MIN	TYP	MAX	UNITS
$t_{RCS}$	$\overline{CS}$ LOW for Read	150			ns
$t_{RDS}$	R/W HIGH to $\overline{CS}$ LOW	10			ns
$t_{RDH}$	R/W HIGH after $\overline{CS}$ HIGH	10			ns
$t_{DZ}$	$\overline{CS}$ HIGH to Data Bus in High Impedance	10		100	ns
$t_{CSD}$	$\overline{CS}$ LOW to Data Bus Valid		100	150	ns
$t_{WCS}$	$\overline{CS}$ LOW for Write	40			ns
$t_{WS}$	R/W LOW to $\overline{CS}$ LOW	0			ns
$t_{WH}$	R/W LOW after $\overline{CS}$ HIGH	10			ns
$t_{LS}$	$\overline{CS}$ LOW to LDAC HIGH	30			ns
$t_{LH}$	$\overline{CS}$ LOW after LDAC HIGH	100			ns
$t_{LX}$	LDAC HIGH	100			ns
$t_{DS}$	Data Valid to $\overline{CS}$ LOW	0			ns
$t_{DH}$	Data Valid after $\overline{CS}$ HIGH	10			ns
$t_{LWD}$	LDAC LOW	100			ns
$t_{SS}$	RSTSEL Valid Before RESET HIGH	0			ns
$t_{SH}$	RSTSEL Valid After RESET HIGH	200			ns
$t_{RSS}$	RESET LOW Before RESET HIGH	10			ns
$t_{RSH}$	RESET LOW After RESET HIGH	10			ns
$t_s$	Settling Time			10	$\mu$ s

TABLE II. Timing Specifications ( $T_A = -40^\circ\text{C}$  to  $+85^\circ\text{C}$ ).

## DIGITALLY-PROGRAMMABLE CURRENT SOURCE

The DAC7641 offers a unique set of features that allows a wide range of flexibility in designing applications circuits such as programmable current sources. The DAC7641 offers both a differential reference input as well as an open-loop configuration around the output amplifier. The open-loop configuration around the output amplifier allows transistor to be placed within the loop to implement a digitally-programmable, uni-directional current source. The availability of a differential reference also allows programmability for both the full-scale and zero-scale currents. The output current is calculated as:

$$I_{OUT} = \left( \left( \frac{V_{REFH} - V_{REFL}}{R_{SENSE}} \right) \cdot \left( \frac{N \text{ Value}}{65,536} \right) \right) + (V_{REFL} / R_{SENSE}) \quad (2)$$

Figure 15 shows a DAC7641 in a 4mA to 20mA current output configuration. The output current can be determined by Equation 3:

(3)

$$I_{OUT} = \left( \left( \frac{2.5V - 0.5V}{125\Omega} \right) \cdot \left( \frac{N \text{ Value}}{65,536} \right) \right) + \left( \frac{0.5V}{125\Omega} \right)$$

At full-scale, the output current is 16mA plus the 4mA for the zero current. At zero scale the output current is the offset current of 4mA (0.5V/125Ω).

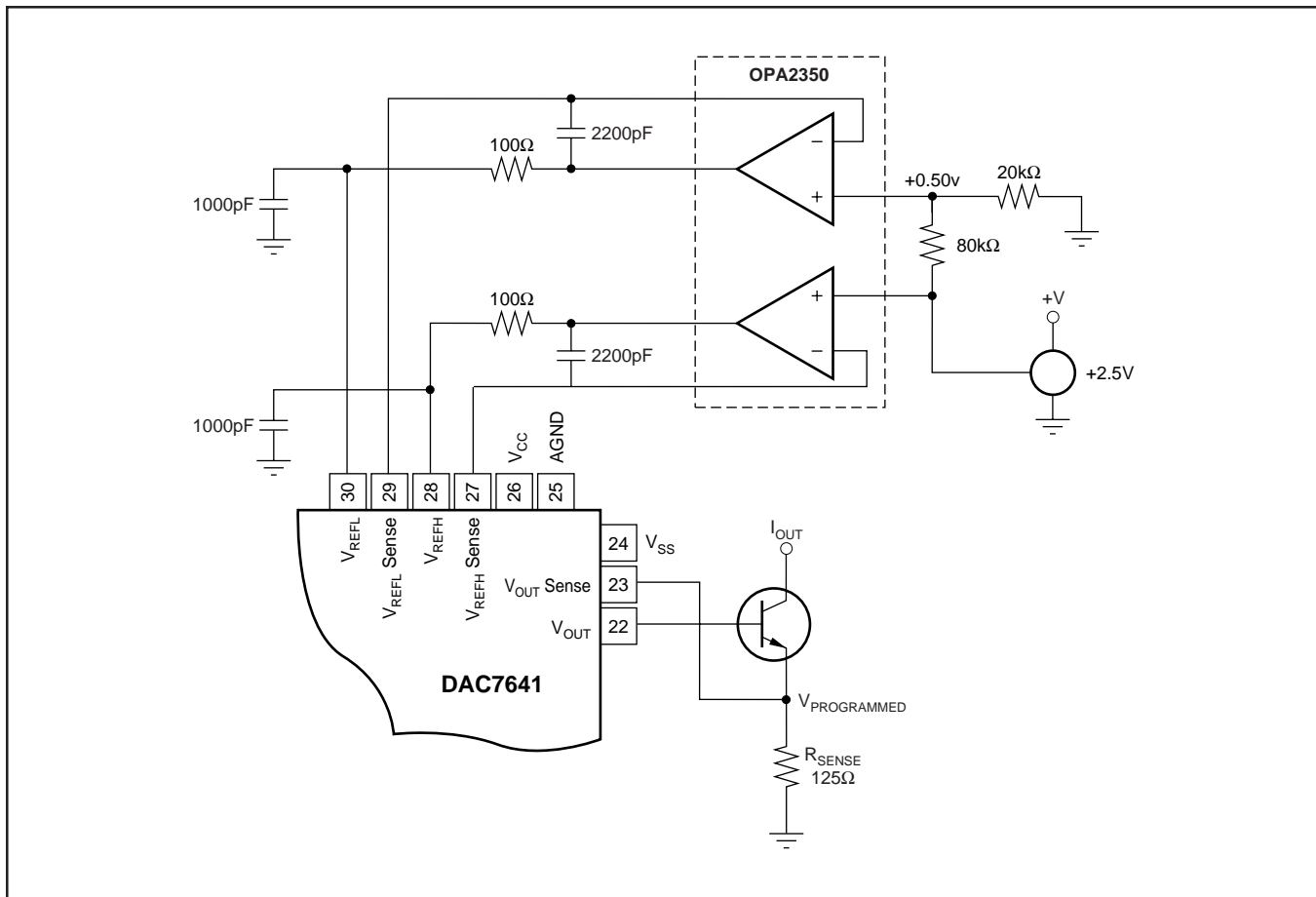




FIGURE 15. 4-to-20mA Digitally Controlled Current Source.

**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
DAC7641Y/250	ACTIVE	TQFP	PBS	32	250	RoHS & Green	NIPDAU	Level-3-260C-168 HR	-40 to 85	7641Y	
DAC7641YB/250	ACTIVE	TQFP	PBS	32	250	RoHS & Green	NIPDAU	Level-3-260C-168 HR	-40 to 85	7641Y B	

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "-" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

**Important Information and Disclaimer:**The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.



## IMPORTANT NOTICE AND DISCLAIMER

TI PROVIDES TECHNICAL AND RELIABILITY DATA (INCLUDING DATASHEETS), DESIGN RESOURCES (INCLUDING REFERENCE DESIGNS), APPLICATION OR OTHER DESIGN ADVICE, WEB TOOLS, SAFETY INFORMATION, AND OTHER RESOURCES "AS IS" AND WITH ALL FAULTS, AND DISCLAIMS ALL WARRANTIES, EXPRESS AND IMPLIED, INCLUDING WITHOUT LIMITATION ANY IMPLIED WARRANTIES OF MERCHANTABILITY, FITNESS FOR A PARTICULAR PURPOSE OR NON-INFRINGEMENT OF THIRD PARTY INTELLECTUAL PROPERTY RIGHTS.

These resources are intended for skilled developers designing with TI products. You are solely responsible for (1) selecting the appropriate TI products for your application, (2) designing, validating and testing your application, and (3) ensuring your application meets applicable standards, and any other safety, security, or other requirements. These resources are subject to change without notice. TI grants you permission to use these resources only for development of an application that uses the TI products described in the resource. Other reproduction and display of these resources is prohibited. No license is granted to any other TI intellectual property right or to any third party intellectual property right. TI disclaims responsibility for, and you will fully indemnify TI and its representatives against, any claims, damages, costs, losses, and liabilities arising out of your use of these resources.

TI's products are provided subject to TI's Terms of Sale (<https://www.ti.com/legal/termsofsale.html>) or other applicable terms available either on [ti.com](https://www.ti.com) or provided in conjunction with such TI products. TI's provision of these resources does not expand or otherwise alter TI's applicable warranties or warranty disclaimers for TI products.

Mailing Address: Texas Instruments, Post Office Box 655303, Dallas, Texas 75265  
Copyright © 2021, Texas Instruments Incorporated