



# Precision Analog-to-Digital Converter (ADC) with 8051 Microcontroller and Flash Memory

# **FEATURES**

#### ANALOG FEATURES

- 24 Bits No Missing Codes
- 22 Bits Effective Resolution at 10Hz
   Low Noise: 75nV
- PGA From 1 to 128
- Precision On-Chip Voltage Reference
- 8 Differential/Single-Ended Channels
- On-Chip Offset/Gain Calibration
- Offset Drift: 0.1ppm/°C
- Gain Drift: 0.5ppm/°C
- On-Chip Temperature Sensor
- Burnout Sensor Detection
- Single-Cycle Conversion
- Selectable Buffer Input

#### **DIGITAL FEATURES**

**Microcontroller Core** 

- 8051-Compatible
- High-Speed Core
  - 4 Clocks per Instruction Cycle
- DC to 33MHz
- Single Instruction 121ns
- Dual Data Pointer

#### Memory

- Up To 32kB Flash Memory
- Flash Memory Partitioning
- Endurance 1M Erase/Write Cycles, 100 Year Data Retention
- In-System Serially Programmable
- External Program/Data Memory (64kB)
- 1,280 Bytes Data SRAM
- Flash Memory Security
- 2kB Boot ROM
- Programmable Wait State Control

**Peripheral Features** 

- 34 I/O Pins
- Additional 32-Bit Accumulator
- Three 16-Bit Timer/Counters
- System Timers
- Programmable Watchdog Timer
- Full-Duplex Dual USARTs
- Master/Slave SPI<sup>™</sup>
- 16-Bit PWM
- Power Management Control
- Idle Mode Current < 1mA
- Stop Mode Current < 1μA
- Programmable Brownout Reset
- Programmable Low Voltage Detect
- 24 Interrupt Sources
- Two Hardware Breakpoints

#### **GENERAL FEATURES**

- Pin-Compatible with MSC1211/12/13/14
- Package: TQFP-64
- Low Power: 4mW
- Industrial Temperature Range: -40°C to +125°C
- Power Supply: 2.7V to 5.25V

# **APPLICATIONS**

- Industrial Process Control
- Instrumentation
- Liquid/Gas Chromatography
- Blood Analysis
- Smart Transmitters
- Portable Instruments
- Weigh Scales
- Pressure Transducers
- Intelligent Sensors
- Portable Applications
- DAS Systems

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#### PACKAGE/ORDERING INFORMATION<sup>(1)</sup>

PRODUCT	FLASH MEMORY	PACKAGE MARKING
MSC1210Y2	4k	MSC1210Y2
MSC1210Y3	8k	MSC1210Y3
MSC1210Y4	16k	MSC1210Y4
MSC1210Y5	32k	MSC1210Y5

(1) For the most current package and ordering information, see the Package Option Addendum at the end of this datasheet, or refer to our web site at www.ti.com.

This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

#### ABSOLUTE MAXIMUM RATINGS(1)

			MSC1210Yx	UNITS
Analog Inputs				
	Momentary		100	mA
Input current	Momentary           urrent         Momentary           continuous         Continuous           supply         continuous           to DGND         o           o AGND         continuous           to DGND         o           o AGND         continuous           input voltage to DGND         output voltage to DGND           output voltage to DGND         um junction temperature           ing temperature range         e temperature range           ge power dissipation         current, all pins           pin short-circuit         Junction to ambient (θ <sub>JA</sub> )           Junction to case (θ <sub>JC</sub> )         Outputs           current         Continuous		10	mA
Input voltage			AGND – 0.3 to AV <sub>DD</sub> + 0.3	V
Power Supply			·	
DV <sub>DD</sub> to DGND			-0.3 to +6	V
AV <sub>DD</sub> to AGND			-0.3 to +6	V
AGND to DGND			-0.3 to +0.3	V
V <sub>REF</sub> to AGND			-0.3 to AV <sub>DD</sub> + 0.3	V
Digital input voltage to D	GND		-0.3 to DV <sub>DD</sub> + 0.3	V
Digital output voltage to	DGND		-0.3 to DV <sub>DD</sub> + 0.3	V
Maximum junction temperature			150	°C
Operating temperature range			-40 to +125	°C
Storage temperature range			-65 to +150	°C
Package power dissipati	ion		(T <sub>J</sub> Max – T <sub>AMBIENT</sub> )/θ <sub>JA</sub>	W
Output current, all pins			200	mA
Output pin short-circuit			10	S
	hundright to problem ( ( )	High K (2s 2p)	62.9	°C/W
Thermal Resistance	Junction to amplient ( $\theta_{JA}$ )	Low K (1s)	78.2	°C/W
	Junction to case ( $\theta_{JC}$ )		13.8	°C/W
Digital Outputs				
<u> </u>			100	mA
I/O source/sink current			100	mA
Power pin maximum			300	mA

(1) Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to absolute maximum conditions for extended periods may affect device reliability.

#### **MSC1210YX FAMILY FEATURES**

FEATURES <sup>(1)</sup>	MSC1210Y2 <sup>(2)</sup>	MSC1210Y3 <sup>(2)</sup>	MSC1210Y4(2)	MSC1210Y5 <sup>(2)</sup>
Flash Program Memory (Bytes)	Up to 4k	Up to 8k	Up to 16k	Up to 32k
Flash Data Memory (Bytes)	Up to 4k	Up to 8k	Up to 16k	Up to 32k
Internal Scratchpad RAM (Bytes)	256	256	256	256
Internal MOVX RAM (Bytes)	1024	1024	1024	1024
Externally Accessible Memory (Bytes)	64k Program, 64k Data	64k Program, 64k Data	64k Program, 64k Data	64k Program, 64k Data

(1) All peripheral features are the same on all devices; the flash memory size is the only difference.

(2) The last digit of the part number (N) represents the onboard flash size =  $(2^N)$ kBytes.

**ELECTRICAL CHARACTERISTICS:**  $AV_{DD} = 5V$ All specifications from T<sub>MIN</sub> to T<sub>MAX</sub>,  $DV_{DD} = +2.7V$  to 5.25V, f<sub>MOD</sub> = 15.625kHz, PGA = 1, Buffer ON, f<sub>DATA</sub> = 10Hz, Bipolar, and  $V_{REF} \equiv (REF IN+) - (REF IN-) = +2.5V$ , unless otherwise noted.

			MSC1210Yx			
PA	RAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
Analog Input (Al	N0-AIN7, AINCOM)		•			•
Analog Input Range		Buffer OFF	AGND – 0.1		AV <sub>DD</sub> + 0.1	V
		Buffer ON	AGND + 50mV		AV <sub>DD</sub> – 1.5	V
ull-Scale Input Voltage Range		(ln+) – (ln–)			±V <sub>REF</sub> /PGA	V
Differential Input I	mpedance	Buffer OFF		7/PGA <sup>(5)</sup>		MΩ
Input Current		Buffer ON		0.5		nA
	Fast Settling Filter	-3dB		0.469 • f <sub>DATA</sub>		
Bandwidth	Sinc <sup>2</sup> Filter	-3dB		0.318 • f <sub>DATA</sub>		
	Sinc <sup>3</sup> Filter	-3dB		0.262 • f <sub>DATA</sub>		
Programmable Ga	ain Amplifier	User-Selectable Gain Range	1		128	
Input Capacitance	e	Buffer On		9		pF
Input Leakage Cu	irrent	Multiplexer channel OFF, T = +25°C		0.5		pА
Burnout Current S	Sources	Buffer On		2		μΑ
Offset DAC			<b>L</b>			1
Offset DAC Range				±V <sub>REF</sub> /(2•PGA)		V
Offset DAC Monotonicity			8			Bits
Offset DAC Gain Error				±1.5		% of Range
Offset DAC Gain Error Drift				1		ppm/°C
System Perform	ance	-				1
Resolution			24			Bits
ENOB				See Typical Ch	naracteristics	1
Output Noise				See Typical Ch	naracteristics	
No Missing Codes	S	Sinc <sup>3</sup> Filter, Decimation > 360	24			Bits
Integral Nonlinear	ity	End Point Fit, Differential Input			±0.0015	% of FSR
Offset Error		After Calibration		7.5		ppm of FS
Offset Drift <sup>(1)</sup>		Before Calibration		0.1		ppm of FS/°C
Gain Error <sup>(2)</sup>		After Calibration		0.002		%
Gain Error Drift(1)		Before Calibration		0.5		ppm/°C
Dffset Drift <sup>(1)</sup> Gain Error <sup>(2)</sup> Gain Error Drift <sup>(1)</sup> System Gain Calibration Range			80		120	% of FS
System Offset Ca	libration Range		-50		50	% of FS
		At DC	100	115		dB
ADC Common-Mode Rejection		$f_{CM} = 60Hz$ , $f_{DATA} = 10Hz$		130		dB
		$f_{CM} = 50Hz$ , $f_{DATA} = 50Hz$		120		dB
		$f_{CM} = 60Hz$ , $f_{DATA} = 60Hz$		120		dB
		$f_{SIG} = 50Hz$ , $f_{DATA} = 50Hz$		100		dB
Normal-Mode Rej	ection	$f_{SIG} = 60Hz$ , $f_{DATA} = 60Hz$		100		dB
Power-Supply Re	iection	At DC, dB = $-20\log(\Delta V_{OUT}/\Delta V_{DD})^{(3)}$	80	88		dB

 $^{(1)}$  Calibration can minimize these errors.

 $^{(2)}$  The self-gain calibration cannot have a REF IN+ of more than AV<sub>DD</sub> –1.5V with Buffer ON. To calibrate gain, turn Buffer OFF.



**ELECTRICAL CHARACTERISTICS:**  $AV_{DD} = 5V$  (continued) All specifications from  $T_{MIN}$  to  $T_{MAX}$ ,  $DV_{DD} = +2.7V$  to 5.25V,  $f_{MOD} = 15.625$ kHz, PGA = 1, Buffer ON,  $f_{DATA} = 10$ Hz, Bipolar, and  $V_{REF} = (REF IN+) - (REF IN-) = +2.5V$ , unless otherwise noted.

				MSC1210Yx		
PARAMETER		CONDITIONS	MIN	ТҮР	MAX	UNITS
Voltage Referen	ce Input	•		-		•
Reference Input	Range	REF IN+, REF IN-	AGND		AV <sub>DD</sub> <sup>(2)</sup>	V
VREF		$V_{REF} \equiv (REF IN+) - (REF IN-)$	0.1	2.5	AV <sub>DD</sub>	V
		At DC		130		dB
V <sub>REF</sub> Common-Mode Rejection	$f_{CM} = 60Hz, f_{DATA} = 60Hz$		120		dB	
Input Current <sup>(4)</sup>		V <sub>REF</sub> = 2.5V		3		μA
On-Chip Voltage	e Reference	•		-		•
Output Voltage		VREFH = 1 at +25°C, ACLK = 1MHz	2.495	2.5	2.505	V
		VREFH = 0 at +25°C, ACLK = 1MHz		1.25		V
Power-Supply Rejection Ratio Short-Circuit Current Source				65		dB
Short-Circuit Current Source				8		mA
Short-Circuit Current Sink				50		μA
Short-Circuit Duration		Sink or Source		Indefinite		
Drift				5		ppm/°C
Drift Output Impedance		Sourcing 100µA		3		Ω
Output Impedance Startup Time from Power On		$C_{REF} = 0.1 \mu F$		8		ms
Temperature Sensor Voltage		T = +25°C		115		mV
Temperature Ser	sor Coefficient			375		μV/°C
Analog Power-Su	upply Requirements	-	•		1	
Analog Power-Su	ipply Voltage	AV <sub>DD</sub>	4.75	5.0	5.25	V
	Analog Current (I <sub>ADC</sub> + I <sub>VREF</sub> )	PDADC = 1, ALVDIS = 1, DAB = 1		< 1		nA
		PGA = 1, Buffer OFF		200		μΑ
Analog Power-Supply Current	ADC Current	PGA = 128, Buffer OFF		500		μΑ
	(I <sub>ADC</sub> )	PGA = 1, Buffer ON		240		μΑ
		PGA = 128, Buffer ON		850		μΑ
	V <sub>REF</sub> Supply Current (I <sub>VREF</sub> )	ADC ON, V <sub>REF</sub> = 2.5V		250		μΑ

<sup>(1)</sup> Calibration can minimize these errors.

 $^{(2)}$  The self-gain calibration cannot have a REF IN+ of more than AV<sub>DD</sub> –1.5V with Buffer ON. To calibrate gain, turn Buffer OFF.

 $^{(3)}$   $\Delta V_{OUT}$  is change in digital result.

(4) 9pF switched capacitor at  $f_{SAMP}$  clock frequency (see Figure 14). (5) The input impedance for PGA = 128 is the same as that for PGA = 64 (that is, 7MΩ/64).

**ELECTRICAL CHARACTERISTICS:**  $AV_{DD} = 3V$ All specifications from  $T_{MIN}$  to  $T_{MAX}$ ,  $DV_{DD} = +2.7V$  to 5.25V,  $f_{MOD} = 15.625$ kHz, PGA = 1, Buffer ON,  $f_{DATA} = 10$ Hz, Bipolar, and  $V_{REF} = (REF IN+) - (REF IN-) = +1.25V$ , unless otherwise noted.

ANALOG INPUT (AINO-AIN7, AINCOM)         Analog Input Range       Buffer OFF       AGND - 0.1         Buffer ON       AGND - 50mV       Suffer ON         Full-Scale Input Voltage Range       (In+) - (In-)       70         Differential Input Impedance       Buffer OFF       70         Input Current       Fast Settling Filter       -3dB       0.469         Bandwidth       Sinc <sup>2</sup> Filter       -3dB       0.262         Programmable Gain Amplifier       User-Selectable Gain Range       1       0         Input Capacitance       Multiplexer channel OFF, T = +25°C       0       0         Input Leakage Current       Multiplexer channel OFF, T = +25°C       0       0         Offset DAC Gain Error       Sensor Input Open Circuit       1       0       0         Offset DAC Gain Error Drift       Sensor Input Open Circuit       1       0       0         System PERFORMANCE       Resolution       4       1       0       0       0       0         Resolution       Inerga Nonlinearity       End Point Fit, Differential Input       1       0       0       0       0         Offset DAC Gain Error Drift       End Point Fit, Differential Input       0       0       0       0       0 <td< th=""><th colspan="2">MSC1210Yx</th><th></th></td<>	MSC1210Yx			
$ \begin{array}{c c c c c c } \mbox{AGND} - 0.1 & \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ $	ГҮР	MAX	UNITS	
Analog Input Range         Buffer ON         AGND + 50mV           Full-Scale Input Voltage Range         (In+) - (In-)         7/P           Differential Input Impedance         Buffer OFF         0.0           Input Current         Buffer ON         0.0           Bandwidth         Fast Settling Filter         -3dB         0.0469           Sinc <sup>2</sup> Filter         -3dB         0.058         0.058           Programmable Gain Amplifier         User-Selectable Gain Range         1         0.0562           Programmable Gain Amplifier         User-Selectable Gain Range         1         0.0562           Programmable Gain Amplifier         User-Selectable Gain Range         1         0.0562           Buffer ON         Sensor Input Open Circuit         0         0           Buffer DAC Gain Error         Multiplexer channel OFF, T = +25°C         0         0           Offset DAC Range         Imput Selectable Gain Error         1         1         1           Offset DAC Gain Error         Imput Selectable Gain Error         1         1         1           System PERFORMANCE         24         1         1         1           Soliga Firler         Sinc <sup>3</sup> Filter         24         1           Nolissing Codes         Si		-		
Buffer ONAGND + 50mVFull-Scale Input Voltage Range(In+) - (In-)7/PDifferential Input ImpedanceBuffer OFF7/PInput CurrentBuffer OFF00BandwidthFast Settling Filter-3dB0.469Sinc <sup>2</sup> Filter-3dB0.262Programmable Gain AmplifierUser-Selectable Gain Range1Input CapacitanceInput Capacitance1Input Leakage CurrentMultiplexer channel OFF, T = +25°C0Burnout Current SourcesSenor Input Open Circuit0Offset DAC Range $\pm \sqrt{ReF/}$ Offset DAC Range $= \pm \sqrt{ReF/}$ $\pm \sqrt{ReF/}$ Offset DAC Gain Error1 $\pm \sqrt{ReF/}$ Offset DAC Gain Error Drift1 $= 5 \sqrt{Re}$ Resolution24ENOBSinc <sup>3</sup> Filter24ENOBSinc <sup>3</sup> Filter24Integral NonlinearityEnd Point Fit, Differential Input7Offset Error OriftEnd Point Fit, Differential Input7Offset Error OriftBefore Calibration7Offset Error OriftBefore Calibration7 </td <td></td> <td>AV<sub>DD</sub> + 0.1</td> <td>V</td>		AV <sub>DD</sub> + 0.1	V	
Differential Input Impedance     Buffer OFF     7/P       Input Current     Buffer OFF     7/P       Input Current     Buffer ON     0.0469       Bandwidth     Sinc <sup>2</sup> Filter     -3dB     0.318       Sinc <sup>3</sup> Filter     -3dB     0.262       Programmable Gain Amplifier     User-Selectable Gain Range     1       Input Capacitance     Multiplexer channel OFF, T = +25°C     0.0       Burnout Current Sources     Sensor Input Open Circuit     0       Offset DAC Range     1     1       Offset DAC Gain Error     1     1       Offset DAC Gain Error     1     1       Offset DAC Gain Error     1     1       Resolution     24     1       ENOB     Sinc <sup>3</sup> Filter     24       Indegrad Nonlinearity     End Point Fit, Differential Input     1       Offset DAC Gain Error     After Calibration     7       Offset DAC Gain Error     1     24     1       ENOB     Sinc <sup>3</sup> Filter     24     1       Offset DAC Gain Error     After Calibration     7       Offset DAC Gain Error     After Calibration     7       Offset Error     After Calibration     7       Offset Error Drift')     Before Calibration     0       System Gain Calibr		AV <sub>DD</sub> – 1.5	V	
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Fast Settling Filter-3dB0.469BandwidthSinc <sup>2</sup> Filter-3dB0.318Sinc <sup>3</sup> Filter-3dB0.262Programmable Gain AmplifierUser-Selectable Gain Range1Input CapacitanceMultiplexer channel OFF, T = +25°C0Input Leakage Current SourcesSensor Input Open Circuit0Burmout Current SourcesSensor Input Open Circuit0OFFSET DACOffset DAC Range $\pm tVREF/$ Offset DAC Gain ErrorImput Caganitation8Offset DAC Gain Error DriftImput Caganitation $\pm tVREF/$ Offset DAC Gain Error Drift1Sinc <sup>3</sup> Filter24Sinc <sup>3</sup> Filter24Colspan="2">SeeOutput NoiseSinc <sup>3</sup> Filter24Integral NonlinearityEnd Point Fit, Differential Input7Offset Drift <sup>(1)</sup> Before Calibration0Calibration7Offset Drift <sup>(1)</sup> Before Calibration7Offset Calibration Range7Offset Calibration Range7 <td colspa<="" td=""><td>PGA(5)</td><td></td><td>MΩ</td></td>	<td>PGA(5)</td> <td></td> <td>MΩ</td>	PGA(5)		MΩ
$\begin{array}{ c c c c } & A \ Bandwidth & Sinc^2 \ Filter & -3dB & 0.318 & 0.318 \\ \hline Sinc^3 \ Filter & -3dB & 0.262 \\ \hline Programmable \ Gain \ Amplifier & User-Selectable \ Gain \ Range & 1 & 0.262 \\ \hline Programmable \ Gain \ Amplifier & User-Selectable \ Gain \ Range & 1 & 0.262 \\ \hline Input \ Lagacitance & Sensor \ Input \ Open \ Circuit & 0 & 0.262 \\ \hline Burmout \ Current \ Sources & Sensor \ Input \ Open \ Circuit & 0 & 0.262 \\ \hline Burmout \ Current \ Sources & Sensor \ Input \ Open \ Circuit & 0 & 0.262 \\ \hline OffSet \ DAC \ Range & Sensor \ Input \ Open \ Circuit & 0 & 0.262 \\ \hline OffSet \ DAC \ Range & Sensor \ Input \ Open \ Circuit & 0 & 0.262 \\ \hline OffSet \ DAC \ Range & Sensor \ Input \ Open \ Circuit & 8 & 0 \\ \hline OffSet \ DAC \ Range & Sensor \ Input \ Open \ Circuit & 8 & 0 \\ \hline OffSet \ DAC \ Gain \ Error \ Drift & 0 & 0.262 \\ \hline Offset \ DAC \ Gain \ Error \ Drift & 0 & 0.262 \\ \hline SYSTEM \ PERFORMANCE & Sensor \ S$	0.5		nA	
Sinds Filter-3dB0.262Programmable Gain AmplifierUser-Selectable Gain Range1Input CapacitanceMultiplexer channel OFF, T = +25°C0Burnout Current SourcesSensor Input Open Circuit0OFFSET DACOffset DAC Range $\pm VREF/$ Offset DAC Range $= \pm VREF/$ Offset DAC Gain ErrorInput Leakage Current8Offset DAC Gain ErrorInput Current Sources $= \pm VREF/$ Offset DAC Gain Error DriftInput Current Sources $= \pm VREF/$ Offset DAC Gain Error DriftInput Current Sources $= \pm VREF/$ SYSTEM PERFORMANCEIntegral NonlinearityInput Current SourcesSeeNo Missing CodesSinc <sup>3</sup> Filter24Integral NonlinearityOffset Drift(1)Before CalibrationInput Current SourceInput Current SourceOffset Drift(1)Before CalibrationInput Current SourceInput Current SourceSystem Gain Calibration RangeAfter CalibrationInput Current SourceInput Current SourceADC Common-Mode RejectionInput Current SourceInput Current SourceInput Current SourceADC Common-Mode RejectionInput Current SourceInput SourceInput Current SourceInput Current SourceInput Current SourceInput SourceInput SourceInput Current SourceInput Current SourceInput Sou	9 • f <sub>DATA</sub>			
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Burnout Current SourcesSensor Input Open CircuitImage: Sensor Input Open CircuitOFFSET DACOffset DAC Range $\pm V_{REF/}$ Offset DAC Monotonicity8Offset DAC Gain Error1Offset DAC Gain Error Drift-SYSTEM PERFORMANCE24Resolution24ENOB24Output NoiseSeeNo Missing CodesSinc <sup>3</sup> FilterOffset Drift(1)End Point Fit, Differential InputOffset Drift(1)Before CalibrationOffset Drift(1)Before CalibrationGain Error Drift(1)Before CalibrationSystem Gain Calibration Range-50System Offset Calibration Range-50ADC Common-Mode RejectionAt DC $f_{CM} = 60Hz, f_{DATA} = 10Hz$ 1 $f_{CM} = 60Hz, f_{DATA} = 60Hz$ 1 $f_{CM} = 60Hz, f_{DATA} = 60Hz$ 1	9		pF	
OFFSET DACOffset DAC Range $1 \leq V_{REF}$ Offset DAC Monotonicity8Offset DAC Gain Error1Offset DAC Gain Error Drift1SYSTEM PERFORMANCE24Resolution24ENOB24Output NoiseSeeNo Missing CodesSinc <sup>3</sup> FilterIntegral NonlinearityEnd Point Fit, Differential InputOffset Drift(1)Before CalibrationGain Error Drift(1)Component ErrorGain Error Drift(1)Before CalibrationGain Error Dri	0.5		pА	
Offset DAC Range $\pm V_{REF}$ Offset DAC Monotonicity8Offset DAC Gain ErrorOffset DAC Gain Error DriftSYSTEM PERFORMANCEResolution24ENOBOutput NoiseSinc <sup>3</sup> FilterNo Missing CodesSinc <sup>3</sup> FilterIntegral NonlinearityEnd Point Fit, Differential InputOffset Drift(1)Before CalibrationGain ErrorAfter CalibrationOffset Calibration Range00System Offset Calibration Range-50ADC Common-Mode Rejection11 $f_{CM} = 60Hz, f_{DATA} = 10Hz$ 11 $f_{CM} = 60Hz, f_{DATA} = 60Hz$ 11	2		μΑ	
Offset DAC Monotonicity8Offset DAC Gain Error $\pm$ Offset DAC Gain Error DriftSYSTEM PERFORMANCEResolution24ENOB24Output NoiseSeeOutput NoiseSinc <sup>3</sup> Filter24Integral NonlinearityEnd Point Fit, Differential InputOffset Drift(1)Before Calibration7Offset Drift(1)Before Calibration0.Gain ErrorAfter Calibration0.Gain Error Drift(1)Before Calibration0.Goin Error Drift(1)Before Calibration0.Goin Error Drift(1)Before Calibration0.Gain Error Drift(1)Before Calibration0.Goin Error Drift(1)Before Calibration0.Goin Error Drift(1)Before Calibration0.Goin Error Drift(1)Before Calibration0.Goin Error Drift(1)<		•		
Offset DAC Gain ErrorImage: Second Seco	=/(2•PGA)		V	
Offset DAC Gain Error DriftImage: Constraint of the constr			Bits	
SYSTEM PERFORMANCEResolution24Resolution24ENOB $24$ Output NoiseSeeOutput NoiseSinc <sup>3</sup> FilterNo Missing CodesSinc <sup>3</sup> FilterIntegral NonlinearityEnd Point Fit, Differential InputOffset ErrorAfter CalibrationOffset Drift(1)Before CalibrationGain Error I <sup>(2)</sup> After CalibrationGain Error Drift(1)Before CalibrationSystem Gain Calibration Range80System Offset Calibration Range-50ADC Common-Mode RejectionAt DC $f_{CM} = 60Hz, f_{DATA} = 50Hz$ 1 $f_{CM} = 60Hz, f_{DATA} = 60Hz$ 1	±1.5		% of Range	
Resolution24ENOB $24$ ENOB $24$ Output Noise $24$ No Missing Codes $Sinc^3$ FilterNo Missing Codes $Sinc^3$ FilterIntegral NonlinearityEnd Point Fit, Differential InputOffset ErrorAfter CalibrationOffset Drift(1)Before CalibrationGain Error(2)After CalibrationGain Error Drift(1)Before CalibrationSystem Gain Calibration Range80System Offset Calibration Range-50And C Common-Mode Rejection $f_{CM} = 60Hz$ , $f_{DATA} = 10Hz$ $f_{CM} = 50Hz$ , $f_{DATA} = 60Hz$ 1 $f_{CM} = 60Hz$ , $f_{DATA} = 60Hz$ 1	1		ppm/°C	
ENOBSeeOutput NoiseSinc3 FilterNo Missing CodesSinc3 FilterNo Missing CodesSinc3 FilterIntegral NonlinearityEnd Point Fit, Differential InputOffset ErrorAfter CalibrationOffset Drift(1)Before CalibrationGain Error(2)After CalibrationGain Error Drift(1)Before CalibrationSystem Gain Calibration Range80System Offset Calibration Range-50And C Common-Mode RejectionAt DCfcM = 60Hz, fDATA = 10Hz1fcM = 60Hz, fDATA = 60Hz1fcM = 60Hz, fDATA = 60Hz1		•	_1	
Output NoiseSeeNo Missing CodesSinc <sup>3</sup> Filter24Integral NonlinearityEnd Point Fit, Differential Input7Offset ErrorAfter Calibration77Offset Drift(1)Before Calibration00Gain Error(2)After Calibration00Gain Error Drift(1)Before Calibration00System Gain Calibration Range8000System Offset Calibration Range-50100ADC Common-Mode RejectionfcM = 60Hz, fDATA = 10Hz10fcM = 60Hz, fDATA = 50Hz101fcM = 60Hz, fDATA = 60Hz1fcM = 60Hz, fDATA = 60Hz1			Bits	
No Missing CodesSinc <sup>3</sup> Filter24Integral NonlinearityEnd Point Fit, Differential Input7Offset ErrorAfter Calibration77Offset Drift(1)Before Calibration00Gain Error (2)After Calibration00Gain Error Drift(1)Before Calibration00System Gain Calibration Range8000System Offset Calibration Range-50100ADC Common-Mode RejectionfcM = 60Hz, fDATA = 10Hz11fcM = 60Hz, fDATA = 50Hz1111fcM = 60Hz, fDATA = 60Hz11	e Typical Cl	Characteristics	_1	
Integral NonlinearityEnd Point Fit, Differential InputIntegral NonlinearityOffset ErrorAfter Calibration77Offset Drift(1)Before Calibration00Gain Error(2)After Calibration00Gain Error Drift(1)Before Calibration00System Gain Calibration Range8000System Offset Calibration Range-5000After Calibration Range-50100After Calibration Range10011After Calibration Range-50100Application Common-Mode RejectionfcM = 60Hz, fDATA = 10Hz11fcM = 50Hz, fDATA = 50Hz1111fcM = 60Hz, fDATA = 60Hz1111fcM = 60Hz, fDATA = 60Hz11fcM = 60Hz1111fcM = 60Hz11fcM = 60Hz11fc	e Typical Cl	Characteristics		
Offset ErrorAfter Calibration77Offset Drift(1)Before Calibration00Gain Error(2)After Calibration00Gain Error Drift(1)Before Calibration00System Gain Calibration Range8000System Offset Calibration Range8000After Calibration Range-5000After Calibration Range100100Application Common-Mode Rejection10011fcM = 50Hz, fDATA = 10Hz1111fcM = 60Hz, fDATA = 60Hz11fcM = 60Hz, fDATA = 60Hz11			Bits	
$\begin{array}{c c} \mbox{Offset Drift}^{(1)} & \mbox{Before Calibration} & \mbox{Offset Drift}^{(1)} & \mbox{Before Calibration} & \mbox{O} & $		±0.0015	% of FSR	
$ \begin{array}{c c} Gain \ Error^{(2)} & After \ Calibration & 0. \\ \hline Gain \ Error \ Drift^{(1)} & Before \ Calibration & 0. \\ \hline System \ Gain \ Calibration \ Range & 80 \\ \hline System \ Offset \ Calibration \ Range & -50 \\ \hline ADC \ Common-Mode \ Rejection & \hline \\ \hline f_{CM} = 60Hz, \ f_{DATA} = 10Hz & 11 \\ \hline f_{CM} = 50Hz, \ f_{DATA} = 50Hz & 11 \\ \hline f_{CM} = 60Hz, \ f_{DATA} = 60Hz & 11 \\ \hline \end{array} $	7.5		ppm of FS	
Gain Error Drift(1)Before CalibrationControlSystem Gain Calibration Range80System Offset Calibration Range-50ADC Common-Mode Rejection $f_{CM} = 60Hz, f_{DATA} = 10Hz$ 100 $f_{CM} = 50Hz, f_{DATA} = 50Hz$ 11 $f_{CM} = 60Hz, f_{DATA} = 60Hz$ 11	0.1		ppm of FS/°C	
System Gain Calibration Range80System Offset Calibration Range50Application Range50ADC Common-Mode Rejection $At DC$ 100 $f_{CM} = 60Hz, f_{DATA} = 10Hz$ 1 $f_{CM} = 50Hz, f_{DATA} = 50Hz$ 1 $f_{CM} = 60Hz, f_{DATA} = 60Hz$ 1	.005		%	
System Offset Calibration Range $-50$ ADC Common-Mode RejectionAt DC1001 $f_{CM} = 60Hz$ , $f_{DATA} = 10Hz$ 11 $f_{CM} = 50Hz$ , $f_{DATA} = 50Hz$ 11 $f_{CM} = 60Hz$ , $f_{DATA} = 60Hz$ 11	0.5		ppm/°C	
ADC Common-Mode Rejection $\begin{array}{c c} At DC & 100 & 1\\ \hline f_{CM} = 60Hz, f_{DATA} = 10Hz & 1\\ \hline f_{CM} = 50Hz, f_{DATA} = 50Hz & 1\\ \hline f_{CM} = 60Hz, f_{DATA} = 60Hz & 1\\ \hline \end{array}$		120	% of FS	
ADC Common-Mode Rejection $ \begin{array}{c c} f_{CM} = 60Hz, \ f_{DATA} = 10Hz \\ \hline f_{CM} = 50Hz, \ f_{DATA} = 50Hz \\ \hline f_{CM} = 60Hz, \ f_{DATA} = 60Hz \end{array} \begin{array}{c c} 1 \\ 1 \\ 1 \\ \hline 1 \\ 1 \\ \hline 1 \\ 1 \\ 1 \\ 1 \\$		50	% of FS	
ADC Common-Mode Rejection $f_{CM} = 50Hz, f_{DATA} = 50Hz \qquad 1$ $f_{CM} = 60Hz, f_{DATA} = 60Hz \qquad 1$	115		dB	
ADC Common-Mode Rejection $f_{CM} = 50Hz, f_{DATA} = 50Hz \qquad 1$ $f_{CM} = 60Hz, f_{DATA} = 60Hz \qquad 1$	130		dB	
f <sub>CM</sub> = 60Hz, f <sub>DATA</sub> = 60Hz 1	120		dB	
	120		dB	
	100	1	dB	
Normal-Mode Rejection $f_{SIG} = 60Hz$ , $f_{DATA} = 60Hz$ 1	100		dB	
	85		dB	

<sup>(1)</sup> Calibration can minimize these errors. <sup>(2)</sup> The self-gain calibration cannot have a REF IN+ of more than  $AV_{DD}$  –1.5V with Buffer ON. To calibrate gain, turn Buffer OFF.

 $\begin{array}{l} \Delta V_{OUT} \text{ is change in digital result.} \\ \begin{array}{l} \text{(4)} & 9pF \text{ switched capacitor at f}_{SAMP} \text{ clock frequency (see Figure 14).} \\ \end{array} \\ \begin{array}{l} \text{(5)} & \text{The input impedance for PGA} = 128 \text{ is the same as that for PGA} = 64 \text{ (that is, } 7M\Omega/64).} \end{array}$ 



**ELECTRICAL CHARACTERISTICS:**  $AV_{DD} = 3V$  (continued) All specifications from  $T_{MIN}$  to  $T_{MAX}$ ,  $DV_{DD} = +2.7V$  to 5.25V,  $f_{MOD} = 15.625$ kHz, PGA = 1, Buffer ON,  $f_{DATA} = 10$ Hz, Bipolar, and  $V_{REF} = (REF IN+) - (REF IN-) = +1.25V$ , unless otherwise noted.

			MSC1210Yx			
PARAMETER		CONDITIONS	MIN	ТҮР	MAX	UNITS
VOLTAGE REFE	RENCE INPUT	•	•		•	
Reference Input	Range	REF IN+, REF IN-	AGND		AV <sub>DD</sub> <sup>(2)</sup>	V
V <sub>REF</sub>		$V_{\text{REF}} \equiv (\text{REF IN+}) - (\text{REF IN-})$	0.1	1.25	AV <sub>DD</sub>	V
		At DC		130		dB
V <sub>REF</sub> Common-	Mode Rejection	$f_{CM} = 60Hz, f_{DATA} = 60Hz$		120		dB
Input Current <sup>(4)</sup>		V <sub>REF</sub> = 1.25V		1.5		μΑ
ON-CHIP VOLTA	GE REFERENCE	•	•	·	•	
Output Voltage		VREFH = 0 at +25°C, ACLK = 1MHz	1.245	1.25	1.255	V
Power-Supply Rejection Ratio				65		dB
Short-Circuit Current Source				8		mA
Short-Circuit Current Sink				50		μΑ
Short-Circuit Duration		Sink or Source		Indefinite		
Drift				5		ppm/°C
Output Impedance		Sourcing 100µA		3		Ω
Startup Time from Power OFF		$C_{REF} = 0.1 \mu F$		8		ms
Temperature Ser	nsor Voltage	T = +25°C		115		mV
Temperature Ser	nsor Coefficient			375		μV/°C
ANALOG POWE	R-SUPPLY REQUIREMENT	S	•		•	
Analog Power-Su	upply Voltage	AV <sub>DD</sub>	2.7		3.6	V
	Analog Current (I <sub>ADC</sub> + I <sub>VREF</sub> )	PDADC = 1, ALVDIS = 1, DAB = 1		< 1		nA
		PGA = 1, Buffer OFF		200		μΑ
Analog Power-Supply Current	ADC Current	PGA = 128, Buffer OFF		500		μΑ
	(I <sub>ADC</sub> )	PGA = 1, Buffer ON		240		μΑ
		PGA = 128, Buffer ON		850		μΑ
	V <sub>REF</sub> Supply Current (I <sub>VREF</sub> )	ADC ON, , V <sub>REF</sub> = 1.25V		240		μΑ

<sup>(1)</sup> Calibration can minimize these errors.

(2) The self-gain calibration cannot have a REF IN+ of more than AV<sub>DD</sub> –1.5V with Buffer ON. To calibrate gain, turn Buffer OFF.

 $^{(3)}$   $\Delta V_{OUT}$  is change in digital result.

 $^{(4)}$  9pF switched capacitor at  $f_{\mbox{SAMP}}$  clock frequency (see Figure 14).

<sup>(5)</sup> The input impedance for PGA = 128 is the same as that for PGA = 64 (that is,  $7M\Omega/64$ ).



# **DIGITAL CHARACTERISTICS:** $DV_{DD}$ = 2.7V to 5.25V All specifications from T<sub>MIN</sub> to T<sub>MAX</sub>, f<sub>OSC</sub> = 1MHz, unless otherwise specified.

			MSC1210Yx			MSC1210Yx			
PARAMETER		CONDITIONS	MIN	ТҮР	MAX	UNITS			
DIGITAL POWER-S	UPPLY REQUIREMENTS	1	-	•	•				
Digital Power-Supply	/ Voltage	DV <sub>DD</sub>	2.7	3.0	3.6	V			
		Normal Mode, f <sub>OSC</sub> = 1MHz		1.4	1.6	mA			
		Normal Mode, f <sub>OSC</sub> = 8MHz		8	9	mA			
		Stop Mode <sup>(1)</sup>		0.5		μΑ			
Digital Power-Supply	y Current	DV <sub>DD</sub>	4.75	5.0	5.25	V			
Digital i ower-oupply current		Normal Mode, f <sub>OSC</sub> = 1MHz		2	2.2	mA			
		Normal Mode, f <sub>OSC</sub> = 8MHz		17	18	mA			
		Stop Mode <sup>(1)</sup>		0.5		μΑ			
DIGITAL INPUT/OU	TPUT (CMOS)		·		•				
	V <sub>IH</sub> (except XIN pin)		0.6 • DV <sub>DD</sub>		DV <sub>DD</sub>	V			
Logic Level	VIL (except XIN pin)		DGND	1.4       1.6         8       9         0.5       1         2       2.2         17       18         0.5       1         0.5       1         0.5       1         0.5       1         0.5       1         0.5       1         0.5       1         0.5       1         0.5       1         0.5       1         0.5       1         0.5       1         0.5       1         0.5       1         0.5       1         0.5       1         0.5       1         0.5       1         0.5       1         0.2 • DV_DD       1         0.2 • DV_DD       1         0.4       1	V				
I/O Pin Hysteresis	<u>.</u>			700		mV			
VO Pin Hysteresis Ports 0–3, Input Leakage Current, Input Mode		$V_{IH} = DV_{DD}$ or $V_{IH} = 0V$		< 1		pА			
Pins EA, XIN Input L	eakage Current					pА			
		I <sub>OL</sub> = 1mA	DGND		0.4	V			
V <sub>OL</sub> , ALE, PSEN, Ports 0–3, All Output Modes		I <sub>OL</sub> = 30mA		1.5		V			
		I <sub>OH</sub> = 1mA	DV <sub>DD</sub> - 0.4	DV <sub>DD</sub> - 0.1	DV <sub>DD</sub>	V			
V <sub>OH</sub> , ALE, PSEN, Ports 0–3, Strong Drive Output		I <sub>OH</sub> = 30mA		DV <sub>DD</sub> – 1.5		V			
Ports 0–3, Pull-Up Resistors				9		kΩ			
Pins ALE, PSEN, Pu	III-Up Resistors	Flash Programming Mode Only		9		kΩ			
Pin RST, Pull-Down	Resistor			500		kΩ			

(1) Digital Brownout Detect disabled (HCR1.2 = 1), Low Voltage Detect disabled (LVDCON.3 =1). Ports configured for CMOS output low. If in External Oscillation mode, the oscillator must be disabled.

### FLASH MEMORY CHARACTERISTICS: DV<sub>DD</sub> = 2.7V to 5.25V

		MSC1210Yx			
PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
Flash Memory Endurance		100,000	1,000,000		cycles
Flash Memory Data Retention		100			years
Mass and Page Erase Time	Set with FER in FTCON	10			ms
Flash Memory Write Time	Set with FWR in FTCON	30		40	μs
	$DV_{DD} = 3.0V$			10	mA
Flash Programming Current	$DV_{DD} = 5.0V$			25	mA



## AC ELECTRICAL CHARACTERISTICS<sup>(1)(2)</sup>: $DV_{DD} = 2.7V$ to 5.25V

			2.7V to 3.6V		4.75V to 5.25V			
SYMBOL	FIGURE	PARAMETER	MIN	MAX	MIN	MAX	UNITS	
System Clo	ock		•					
		External Crystal Frequency (f <sub>OSC</sub> )	1	18	1	33	MHz	
1/t <sub>CLK</sub> <sup>(4)</sup>	4	External Clock Frequency (f <sub>OSC</sub> )	0	18	0	33	MHz	
•		External Ceramic Resonator Frequency (f <sub>OSC</sub> )	1	16	1	16	MHz	
Program M	lemory							
t <sub>lhll</sub>	1	ALE Pulse Width	1.5t <sub>CLK</sub> – 5		1.5t <sub>CLK</sub> – 5		ns	
AVLL	1	Address Valid to ALE LOW	0.5t <sub>CLK</sub> – 10		0.5t <sub>CLK</sub> – 7		ns	
t <sub>llax</sub>	1	Address Hold After ALE LOW	0.5t <sub>CLK</sub>		0.5t <sub>CLK</sub>		ns	
t <sub>LLIV</sub>	1	ALE LOW to Valid Instruction In		2.5t <sub>CLK</sub> - 35		2.5t <sub>CLK</sub> - 25	ns	
t <sub>LLPL</sub>	1	ALE LOW to PSEN LOW	0.5t <sub>CLK</sub>		0.5t <sub>CLK</sub>		ns	
<sup>t</sup> PLPH	1	PSEN Pulse Width	2t <sub>CLK</sub> – 5		2t <sub>CLK</sub> – 5		ns	
t <sub>PLIV</sub>	1	PSEN LOW to Valid Instruction in		2t <sub>CLK</sub> – 40		2t <sub>CLK</sub> - 30	ns	
t <sub>PXIX</sub>	1	Input Instruction Hold After PSEN	5		-5		ns	
t <sub>PXIZ</sub>	1	Input Instruction Float After PSEN		t <sub>CLK</sub> – 5		t <sub>CLK</sub>	ns	
taviv	1	Address to Valid Instruction In		3t <sub>CLK</sub> – 40		3t <sub>CLK</sub> – 25	ns	
t <sub>PLAZ</sub>	1	PSEN LOW to Address Float		0		0	ns	
Data Memo	ory		•			•		
	_	$\overline{\text{RD}}$ Pulse Width (t <sub>MCS</sub> = 0) <sup>(5)</sup>	2t <sub>CLK</sub> – 5		2t <sub>CLK</sub> – 5		ns	
RLRH	2	RD Pulse Width $(t_{MCS} > 0)^{(5)}$	t <sub>MCS</sub> - 5		t <sub>MCS</sub> - 5		ns	
	_	WR Pulse Width $(t_{MCS} = 0)^{(5)}$	2t <sub>CLK</sub> – 5		2t <sub>CLK</sub> – 5		ns	
twlwh	3	WR Pulse Width $(t_{MCS} > 0)^{(5)}$	t <sub>MCS</sub> – 5		t <sub>MCS</sub> – 5		ns	
	2	RD LOW to Valid Data In $(t_{MCS} = 0)^{(5)}$		2t <sub>CLK</sub> - 40		2t <sub>CLK</sub> - 30	ns	
trldv	2	$\overline{RD}$ LOW to Valid Data In $(t_{MCS} > 0)^{(5)}$		t <sub>MCS</sub> – 40		t <sub>MCS</sub> – 30	ns	
t <sub>RHDX</sub>	2	Data Hold After Read	-5		-5		ns	
t <sub>RHDZ</sub>	2	Data Float After Read $(t_{MCS} = 0)^{(5)}$		t <sub>CLK</sub>		<sup>t</sup> CLK	ns	
KHUZ	2	Data Float After Read $(t_{MCS} > 0)^{(5)}$		2t <sub>CLK</sub>		2t <sub>CLK</sub>	ns	
tLLDV	2	ALE LOW to Valid Data In $(t_{MCS} = 0)^{(5)}$		2.5t <sub>CLK</sub> – 40		2.5t <sub>CLK</sub> – 25	ns	
LLDV	_	ALE LOW to Valid Data In $(t_{MCS} > 0)^{(5)}$		$t_{CLK} + t_{MCS} - 40$		t <sub>CLK</sub> + t <sub>MCS</sub> - 25	ns	
t <sub>AVDV</sub>	2	Address to Valid Data In $(t_{MCS} = 0)^{(5)}$		3t <sub>CLK</sub> – 40		3t <sub>CLK</sub> – 25	ns	
AUDV		Address to Valid Data In $(t_{MCS} > 0)^{(5)}$	0.51 5	1.5t <sub>CLK</sub> + t <sub>MCS</sub> - 40	0.51 5	1.5t <sub>CLK</sub> + t <sub>MCS</sub> - 25	ns	
tLLWL	2, 3	ALE LOW to $\overline{RD}$ or $\overline{WR}$ LOW ( $t_{MCS} = 0$ )(5)	0.5t <sub>CLK</sub> – 5	0.5t <sub>CLK</sub> + 5	0.5t <sub>CLK</sub> – 5	0.5t <sub>CLK</sub> + 5	ns	
		ALE LOW to $\overline{RD}$ or $\overline{WR}$ LOW $(t_{MCS} > 0)^{(5)}$ Address to $\overline{RD}$ or $\overline{WR}$ LOW $(t_{MCS} = 0)^{(5)}$	t <sub>CLK</sub> – 5	t <sub>CLK</sub> + 5	t <sub>CLK</sub> – 5	t <sub>CLK</sub> + 5	ns	
tavwl	2, 3	Address to RD or WR LOW $(t_{MCS} = 0)^{(3)}$ Address to RD or WR LOW $(t_{MCS} > 0)^{(5)}$	t <sub>CLK</sub> – 5		t <sub>CLK</sub> – 5 2tour 5		ns ns	
town	3	Data Valid to WR Transition	2t <sub>CLK</sub> – 5 –8		2t <sub>CLK</sub> – 5 –5		ns	
	3	Data Valid to WR Transition					ns	
	2	RD LOW to Address Float	t <sub>CLK</sub> – 8	-0.5t <sub>CLK</sub> - 5	t <sub>CLK</sub> – 5	-0.5t <sub>CLK</sub> - 5	ns	
t <sub>RLAZ</sub>	-	RD cow to Address Float RD or WR HIGH to ALE HIGH $(t_{MCS} = 0)^{(5)}$	-5	-0.5ICLK - 5	-5	-0.5ICLK - 5	ns	
twhlh	2, 3	RD or WR HIGH to ALE HIGH $(t_{MCS} = 0)^{(5)}$	-5 t <sub>CLK</sub> - 5	t <sub>CLK</sub> + 5	-5 t <sub>CLK</sub> - 5	t <sub>CLK</sub> + 5	ns	
External C	lock		ULK U	ULK V	ULK U	ULK I U	115	
	4	HIGH Time <sup>(3)</sup>	15		10		ns	
tHIGH	4	LOW Time <sup>(3)</sup>	15		10		ns	
	4	Rise Time <sup>(3)</sup>	15	5	10	5	ns	
<sup>t</sup> R	4	Fall Time <sup>(3)</sup>	-	5		5		
t <sub>F</sub> (1) Demonstra	4	Fall Time(9)		Э		5	ns	

Parameters are valid over operating temperature range, unless otherwise specified.
 Load capacitance for Port 0, ALE, and PSEN = 100pF; load capacitance for all other outputs = 80pF.
 These values are characterized but not 100% production tested.
 In the MSC1210, f<sub>OSC</sub> = f<sub>CLK</sub>, t<sub>CLK</sub> = 1/fosc = one oscillator clock period.
 t<sub>MCS</sub> is a time period related to the Stretch MOVX selection. The following table shows the value of t<sub>MCS</sub> for each stretch selection:

MD2	MD1	MD0	MOVX DURATION	t <sub>MCS</sub>
0	0	0	2 Machine Cycles	0
0	0	1	3 Machine Cycles (default)	4t <sub>CLK</sub>
0	1	0	4 Machine Cycles	8t <sub>CLK</sub>
0	1	1	5 Machine Cycles	12t <sub>CLK</sub>
1	0	0	6 Machine Cycles	16t <sub>CLK</sub>
1	0	1	7 Machine Cycles	20t <sub>CLK</sub>
1	1	0	8 Machine Cycles	24t <sub>CLK</sub>
1	1	1	9 Machine Cycles	28t <sub>CLK</sub>



#### **EXPLANATION OF THE AC SYMBOLS**

Each Timing Symbol has five characters. The first character is always 't' (= time). The other characters, depending on their positions, indicate the name of a signal or the logical status of that signal. The designators are:

- A—Address C-Clock
- D-Input Data
- H-Logic Level HIGH
- I-Instruction (program memory contents)
- L-Logic Level LOW, or ALE
- P-PSEN
- Q-Output Data

R-RD Signal t-Time V—Valid W-WR Signal X-No Longer a Valid Logic Level Z-Float Examples:

(1)  $t_{AVLL}$  = Time for address valid to ALE LOW.



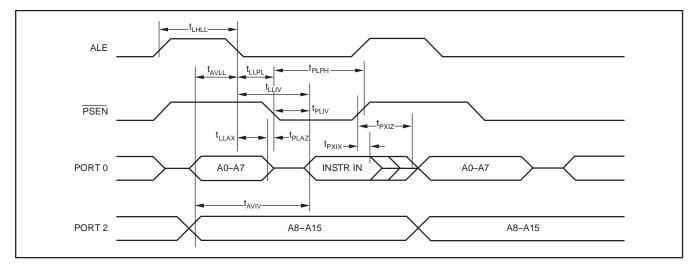
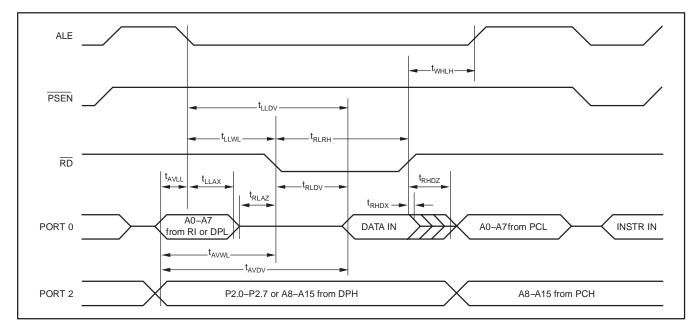


Figure 1. External Program Memory Read Cycle





# MSC1210



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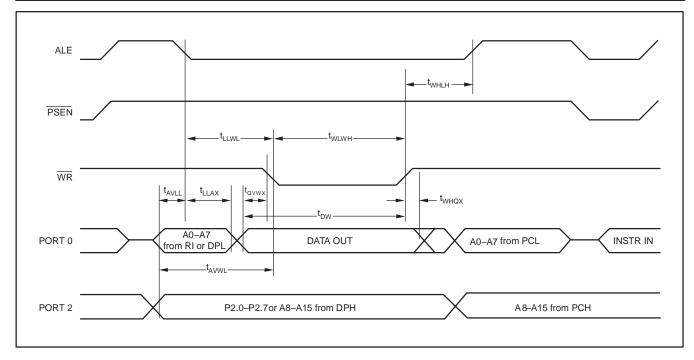


Figure 3. External Data Memory Write Cycle

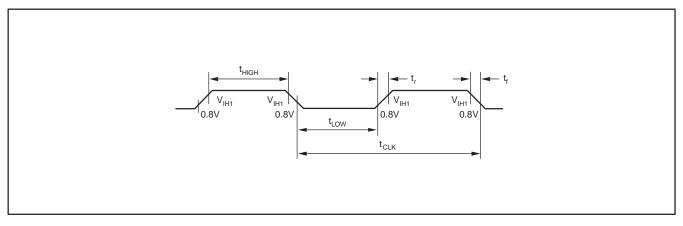
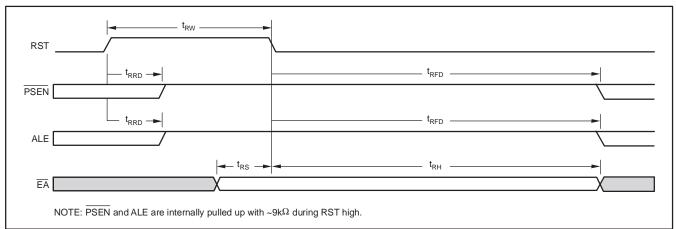


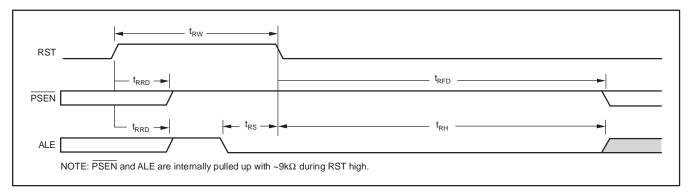
Figure 4. External Clock Drive CLK



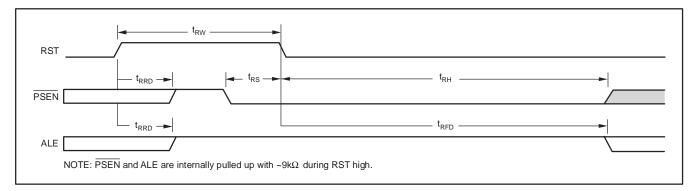
## **RESET AND POWER-ON TIMING**



#### Figure 5. Reset Timing



#### Figure 6. Parallel Flash Programming Power-On Timing (EA is ignored)

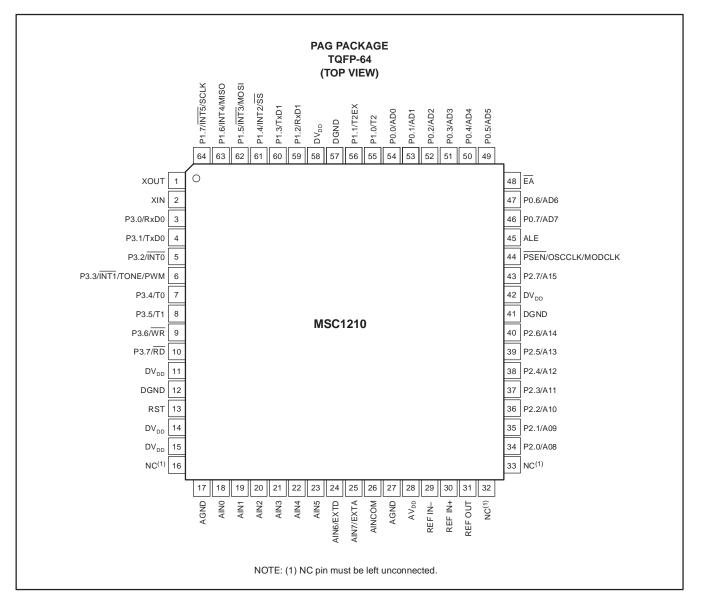


### Figure 7. Serial Flash Programming Power-On Timing (EA is ignored)

SYMBOL	PARAMETER	MIN	MAX	UNIT
t <sub>RW</sub>	RST width	2t <sub>OSC</sub>	—	ns
t <sub>RRD</sub>	RST rise to PSEN ALE internal pull HIGH	-	5	μs
t <sub>RFD</sub>	RST falling to PSEN and ALE start	-	(2 <sup>17</sup> + 512)t <sub>OSC</sub>	ns
t <sub>RS</sub>	Input signal to RST falling setup time	tosc	—	ns
<sup>t</sup> RH	RST falling to input signal hold time	(2 <sup>17</sup> + 512)t <sub>OSC</sub>	—	ns



#### PIN ASSIGNMENTS



## **PIN DESCRIPTIONS**

PIN #	NAME	DESCRIPTION						
1	XOUT		The crystal oscillator pin XOUT supports parallel resonant AT cut fundamental frequency crystals and ceramic resonators. XOUT serves as the output of the crystal amplifier.					
2	XIN			nant AT cut fundamental frequency crystals and ceramic xternal clock source instead of a crystal.				
3–10	P3.0–P3.7	Port 3 is a bidirectional I/O port. The alternate functions for Port 3 are listed below.						
		PORT 3.x	Alternate Name(s)	Alternate Use				
		P3.0	RxD0	Serial port 0 input				
		P3.1	TxD0	Serial port 0 output				
		P3.2	INTO	External interrupt 0				
		P3.3	INT1/TONE/PWM	External interrupt 1/TONE/PWM output				
		P3.4	ТО	Timer 0 external input				
		P3.5	T1	Timer 1 external input				
		P3.6	WR	External data memory write strobe				
		P3.7	RD	External data memory read strobe				
11, 14, 15, 42, 58	DVDD	Digital power sup	ply					
12, 41, 57	DGND	Digital ground						
13	RST	A HIGH on the re	set input for two t <sub>OSC</sub> periods rese	ets the device.				
16, 32, 33	NC		his pin must be left unconnected.					
17, 27	AGND	Analog ground						
18	AIN0	Analog input cha	nnel 0					
19	AIN1	Analog input cha	nnel 1					
20	AIN2	Analog input cha	nnel 2					
21	AIN3	Analog input cha	nnel 3					
22	AIN4	Analog input cha	nnel 4					
23	AIN5	Analog input cha	nnel 5					
24	AIN6, EXTD	Analog input cha	nnel 6, digital low-voltage detect inp	put, generates DLVD interrupt				
25	AIN7, EXTA	Analog input cha	nnel 7, analog low-voltage detect in	put, generates ALVD interrupt				
26	AINCOM	Analog common	for single-ended inputs or analog in	put for differential inputs				
28	AV <sub>DD</sub>		5 i 5	o disable Analog Brownout Reset function.				
29	REF IN-		e negative input (must be tied to AG					
30	REF IN+	Voltage reference	<b>o</b>					
31	REF OUT	Ū.	eference output (tie to REF IN+ for	internal VREF use)				
34–40, 43	P2.0-P2.7		• •	ons for Port 2 are listed below. Refer to P2DDR, SFR B1h–B2h.				
,		PORT 2.x	Alternate Name	Alternate Use				
		P2.0	A8	Address bit 8				
		P2.1	A9	Address bit 9				
		P2.2	A10	Address bit 10				
		P2.3	A11	Address bit 11				
		P2.4	A12	Address bit 12				
		P2.5	A13	Address bit 13				
		P2.6	A14	Address bit 14				
		P2.7	A15	Address bit 15				
			···					



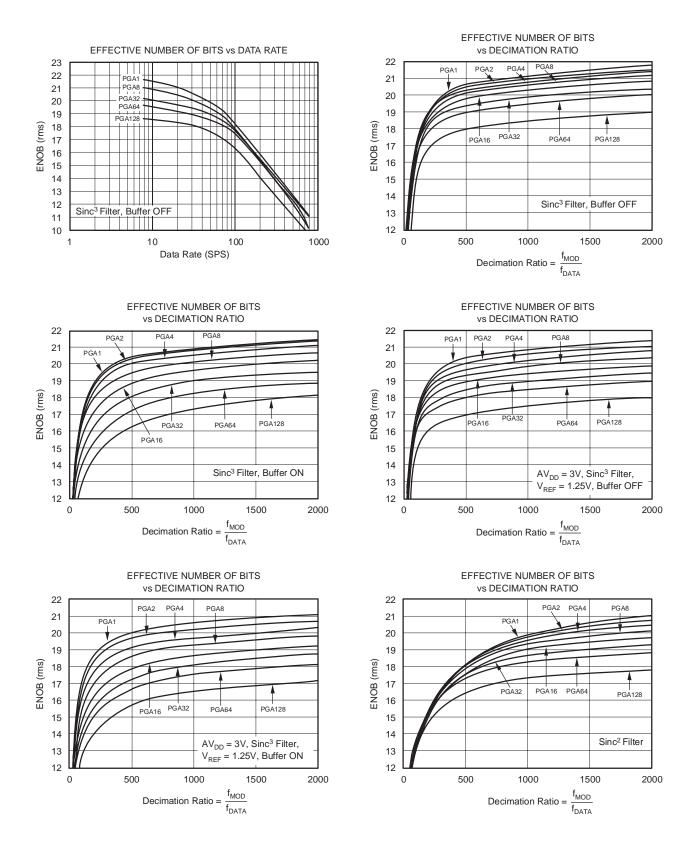
## PIN DESCRIPTIONS (continued)

PIN #	NAME	DESCRIPTION					
44	PSEN, OSCCLK, MODCLK	Program store enable. Connected to optional external memory as a chip enable. PSEN provides an active low pulse. In programming mode, PSEN is used as an input along with ALE to define serial or parallel programming mode. PSEN is held HIGH for parallel programming mode and LOW for serial programming. This pin can also be selected (when not using external memory) to output the oscillator clock, modulator clock, HIGH, or LOW. Care should be taken so that loading on this pin does not inadvertently cause the device to enter programming mode.					
		ALE	PSEN	Program Mode Selection <sup>(1)</sup>			
		NC or DV <sub>DD</sub>	NC or DV <sub>DD</sub>	Normal operation (User Application mode)			
		0	NC or DV <sub>DD</sub>	Parallel programming			
		NC or DV <sub>DD</sub>	0	Serial programming			
		0	0	Reserved			
45	ALE	at a constant rate skipped during ea define serial or pa programming. Thi	e of 1/4 the oscillator frequency, ach access to external data me arallel programming mode. ALE is pin can also be selected (wh	byte of the address during an access to external memory. ALE is emitted and can be used for external timing or clocking. One ALE pulse is mory. In programming mode, ALE is used as an input along with FSEN to is held HIGH for serial programming mode and LOW for parallel en not using external memory) to output HIGH or LOW. Care should be invertently cause the device to enter programming mode.			
48	EA			ly held LOW at the end of RESET to enable the device to fetch code ing with 0000h. No internal pull-up on this pin.			
46, 47, 49–54	P0.0-P0.7	Port 0 is a bidirectional I/O port. The alternate functions for Port 0 are listed below. Refer to P1DDR, SFR AEh-AFh					
		PORT 0.x	Alternate Name	Alternate Use			
		P0.0	AD0	Address/Data bit 0			
		P0.1	AD1	Address/Data bit 1			
		P0.2	AD2	Address/Data bit 2			
		P0.3	AD3	Address/Data bit 3			
		P0.4	AD4	Address/Data bit 4			
		P0.5	AD5	Address/Data bit 5			
		P0.6	AD6	Address/Data bit 6			
		P0.7	AD7	Address/Data bit 7			
55, 56, 59–64	P1.0-P1.7	Port 0 is a bidire	ctional I/O port. The alternate	functions for Port 0 are listed below. Refer to P1DDR, SFR AEh-AFh.			
		PORT 0.x	Alternate Name(s)	Alternate Use			
		P1.0	T2	T2 input			
		P1.1	T2EX	T2 external input			
		P1.2	RxD1	Serial port input			
		P1.3	TxD1	Serial port output			
		P1.4	INT2/SS	External Interrupt / Slave Select			
		P1.5	INT3/MOSI	External Interrupt / Master Out-Slave In			
		P1.6	INT4/MISO	External Interrupt / Master In-Slave Out			
		P1.7	INT5/SCK	External Interrupt / Serial Clock			

(1) The program mode is changed during the falling edge of the reset signal.



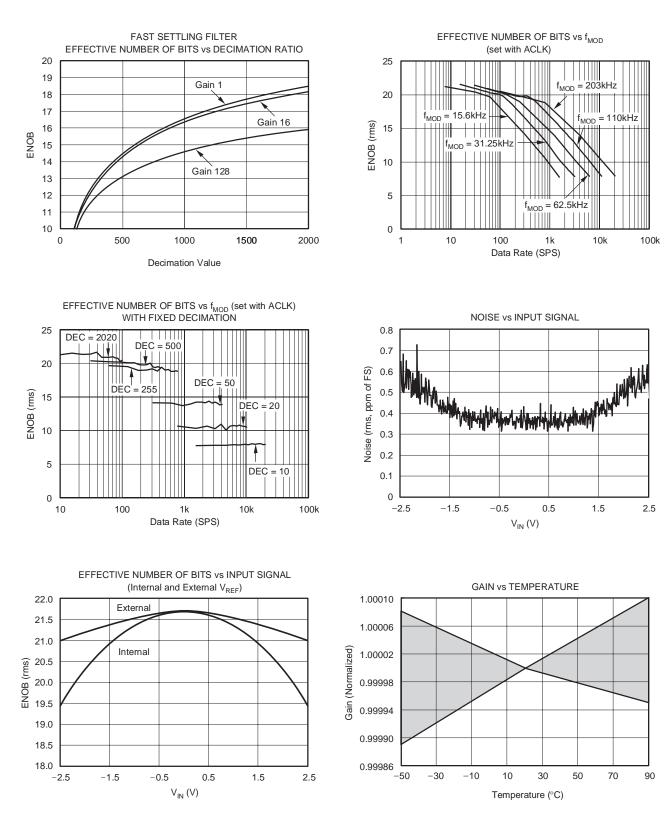
#### **TYPICAL CHARACTERISTICS**





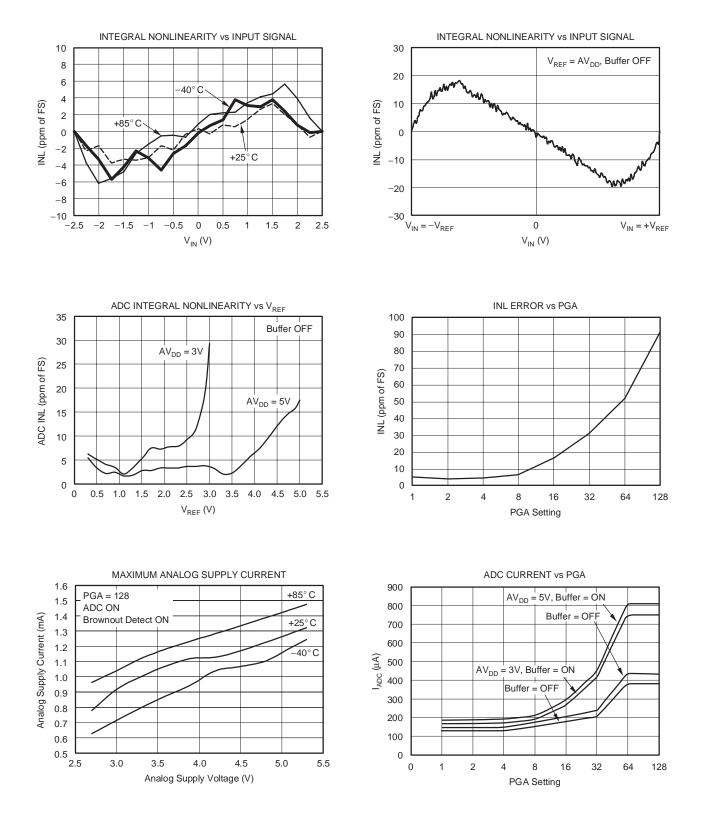
#### **TYPICAL CHARACTERISTICS (Continued)**

 $AV_{DD} = +5V$ ,  $DV_{DD} = +5V$ ,  $f_{OSC} = 8MHz$ , PGA = 1,  $f_{MOD} = 15.625Hz$ , Bipolar, Buffer ON, and  $V_{REF} = (REF IN+) - (REF IN-) = +2.5V$ , unless otherwise specified.



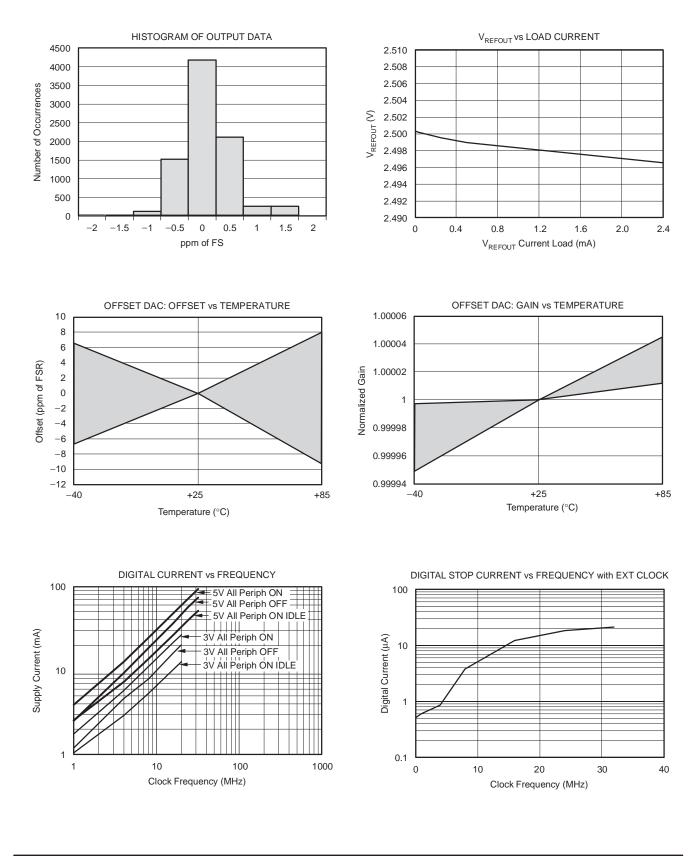


#### **TYPICAL CHARACTERISTICS (Continued)**



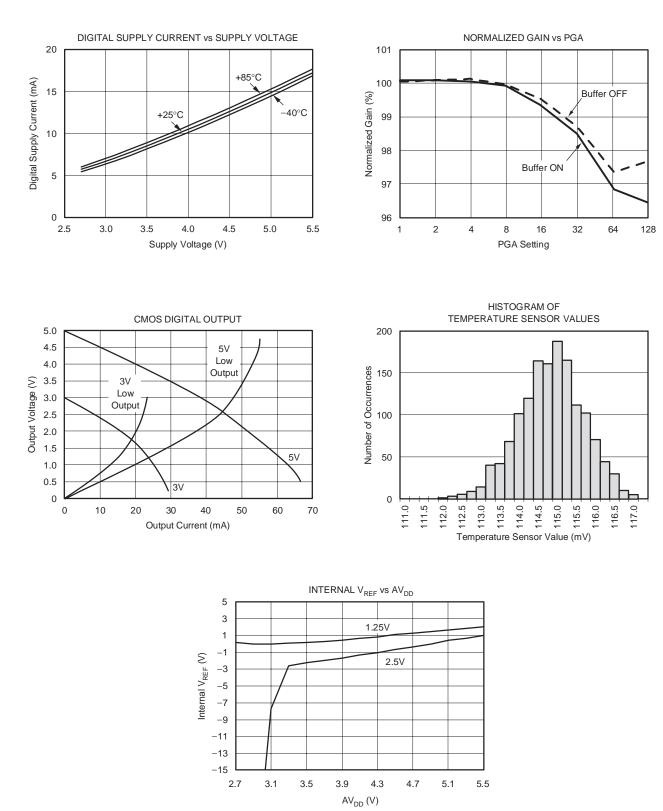


#### **TYPICAL CHARACTERISTICS (Continued)**





#### **TYPICAL CHARACTERISTICS (Continued)**



# DESCRIPTION

The MSC1210Yx is a completely integrated family of mixed-signal devices incorporating a high-resolution delta-sigma ADC, 8-channel multiplexer, burnout current sources, selectable buffered input, offset DAC (digital-to-analog converter), PGA (programmable gain amplifier), temperature sensor, voltage reference, 8-bit microcontroller, Flash Program Memory, Flash Data Memory, and Data SRAM, as shown in Figure 8.

On-chip peripherals include an additional 32-bit accumulator, an SPI-compatible serial port, dual USARTs, multiple digital input/output ports, watchdog timer, low-voltage detect, on-chip power-on reset, 16-bit PWM, and system timers, brownout reset, and three timer/counters.

The device accepts low-level differential or single-ended signals directly from a transducer. The ADC provides 24 bits of resolution and 24 bits of no-missing-code performance using a Sinc<sup>3</sup> filter with a programmable sample rate. The ADC also has a selectable filter that allows for high-resolution single-cycle conversion.

The microcontroller core is 8051 instruction set compatible. The microcontroller core is an optimized 8051 core that executes up to three times faster than the standard 8051 core, given the same clock source. That makes it possible to run the device at a lower external clock frequency and achieve the same performance at lower power than the standard 8051 core.

FRUMENTS

The MSC1210Yx allows the user to uniquely configure the Flash and SRAM memory maps to meet the needs of their application. The Flash is programmable down to 2.7V using both serial and parallel programming methods. The Flash endurance is 1 million Erase/Write cycles. In addition, 1280 bytes of RAM are incorporated on-chip.

The part has separate analog and digital supplies, which can be independently powered from 2.7V to +5.25V. At +3V operation, the power dissipation for the part is typically less than 4mW. The MSC1210Yx is packaged in a TQFP-64 package.

The MSC1210Yx is designed for high-resolution measurement applications in smart transmitters, industrial process control, weigh scales, chromatography, and portable instrumentation.

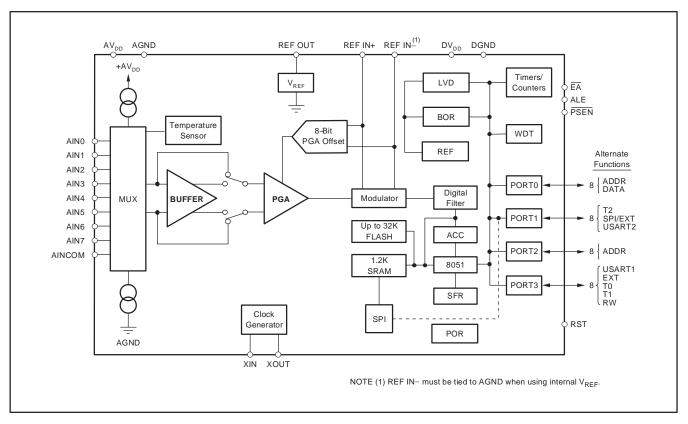


Figure 8. Block Diagram



#### **ENHANCED 8051 CORE**

All instructions in the MSC1210 family perform exactly the same functions as they would in a standard 8051. The effect on bits, flags, and registers is the same. However, the timing is different. The MSC1210 family utilizes an efficient 8051 core which results in an improved instruction execution speed of between 1.5 and 3 times faster than the original core for the same external clock speed (4 clock cycles per instruction versus 12 clock cycles per instruction, as shown in Figure 9). The internal system clock is equal to the external oscillator frequency. This translates into an effective throughput improvement of more than 2.5 times, using the same code and same external clock speed.

Therefore, a device frequency of 33MHz for the MSC1210Yx actually performs at an equivalent execution speed of 82.5MHz compared to the standard 8051 core. This allows the user to run the device at slower external clock speeds which reduces system noise and power consumption, but provides greater throughput. This performance difference can be seen in Figure 10. The timing of software loops will be faster with the MSC1210. However, the timer/counter operation of the MSC1210 may be maintained at 12 clocks per increment or optionally run at 4 clocks per increment.

The MSC1210 also provides dual data pointers (DPTRs) to speed block Data Memory moves.

Additionally, it can stretch the number of memory cycles to access external Data Memory from between two and nine instruction cycles in order to accommodate different speeds of memory or devices, as shown in Table 1. The MSC1210 provides an external memory interface with a 16-bit address bus (P0 and P2). The 16-bit address bus makes it necessary to multiplex the low address byte through the P0 port. To enhance P0 and P2 for high-speed memory access, hardware configuration control is provided to configure the ports for external memory/peripheral interface or general-purpose I/O.

Single-Byte, Single-Cycle Instruction ALE **MSC1210 Timing** PSEN AD0-AD7 PORT 2 4 Cycles CLK 12 Cycles Standard 8051 Timing ALE PSEN AD0-AD7 PORT 2 Single-Byte, Single-Cycle Instruction

Figure 10. Comparison of MSC1210 Timing to Standard 8051 Timing

Table 1. Memory Cycle Stretching. Stretching of MOVX timing as defined by MD2, MD1, and MD0 bits in CKCON register (address 8Eh).

CKCON (8Eh) MD2:MD0	INSTRUCTION CYCLES (for MOVX)	RD or WR STROBE WIDTH (SYS CLKs)	RD or WR STROBE WIDTH (μs) AT 12MHz
000	2	2	0.167
001	3 (default)	4	0.333
010	4	8	0.667
011	5	12	1.000
100	6	16	1.333
101	7	20	1.667
110	8	24	2.000
111	9	28	2.333

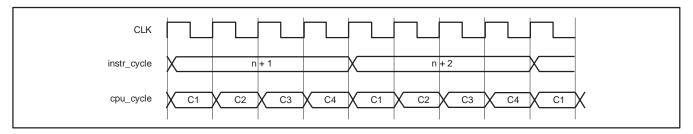


Figure 9. Instruction Timing Cycle

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Furthermore, improvements were made to peripheral features that off-load processing from the core, and the user, to further improve efficiency. For instance, 32-bit accumulation can be done through the summation register to significantly reduce the processing overhead for the multiple byte data from the ADC or other sources. This allows for 32-bit addition and shifting to be accomplished in a few instruction cycles, compared to hundreds of instruction cycles through a software implementation.

#### **Family Device Compatibility**

The hardware functionality and pin configuration across the MSC1210 family are fully compatible. To the user the only difference between family members is the memory configuration. This makes migration between family members simple. Code written for the MSC1210Y2 can be executed directly on an MSC1210Y3, MSC1210Y4, or

MSC1210Y5. This gives the user the ability to add or subtract software functions and to freely migrate between family members. Thus, the MSC1210 can become a standard device used across several application platforms.

#### **Family Development Tools**

The MSC1210 is fully compatible with the standard 8051 instruction set. This means that the user can develop software for the MSC1210 with their existing 8051 development tools. Additionally, a complete, integrated development environment is provided with each demo board, and third-party developers also provide support.

#### Power Down Modes

The MSC1210 can power down several of the on-chip peripherals and put the CPU into IDLE. For more information, see the *Idle Mode* and *Stop Mode* sections.

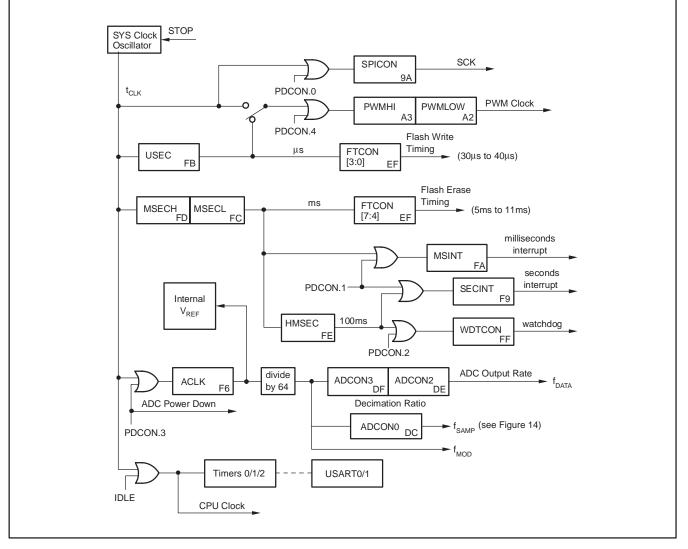


Figure 11. MSC1210 Timing Chain and Clock Control



# **OVERVIEW**

The MSC1210 ADC structure is shown in Figure 12. The figure lists the components that make up the ADC, along with the corresponding special function register (SFR) associated with each component.

#### ADC INPUT MULTIPLEXER

The input multiplexer provides for any combination of differential inputs to be selected as the input channel, as shown in Figure 13. If AINO is selected as the positive differential input channel, any other channel can be selected as the negative differential input channel. With this method,

it is possible to have up to eight fully differential input channels. It is also possible to switch the polarity of the differential input pair to negate any offset voltages.

In addition, current sources are supplied that will source or sink current to detect open or short circuits on the pins.

#### **TEMPERATURE SENSOR**

On-chip diodes provide temperature sensing capability. When the configuration register for the input MUX is set to all 1s, the diodes are connected to the input of the ADC. All other channels are open.

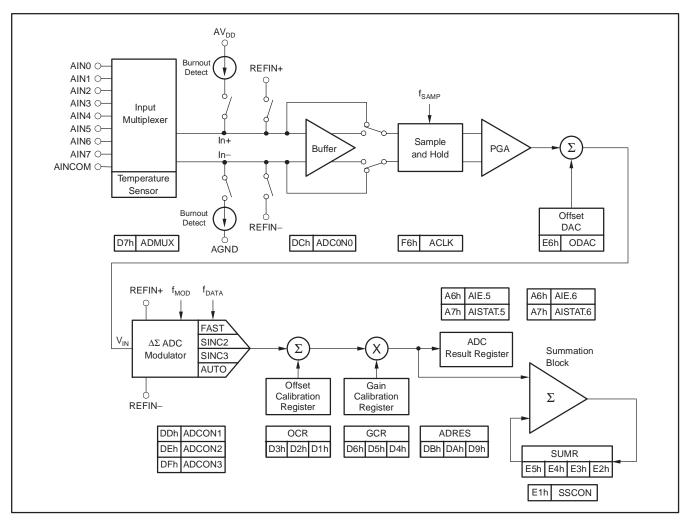


Figure 12. MSC1210 ADC Structure



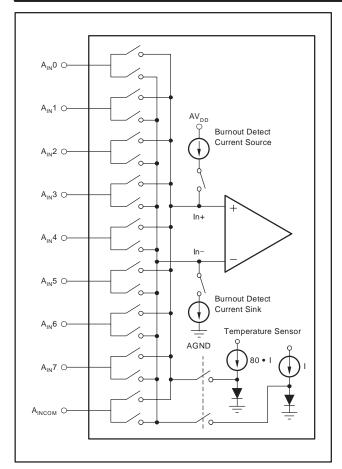


Figure 13. Input Multiplexer Configuration

#### **BURNOUT DETECT**

When the Burnout Detect (BOD) bit is set in the ADC control configuration register (ADCON0 DCh), two current sources are enabled. The current source on the positive input channel sources approximately  $2\mu$ A of current. The current source on the negative input channel sinks approximately  $2\mu$ A. This allows for the detection of an open circuit (full-scale reading) or short circuit (small differential reading) on the selected input differential pair. Buffer should be on for sensor burnout detection.

#### ADC INPUT BUFFER

The analog input impedance is always high, regardless of PGA setting (when the buffer is enabled). With the buffer enabled, the input voltage range is reduced and the analog power-supply current is higher. If the limitation of input voltage range is acceptable, then the buffer is always preferred.

The input impedance of the MSC1210 without the buffer is 7M $\Omega$ /PGA. The buffer is controlled by the state of the BUF bit in the ADC control register (ADCON0 DCh).

#### ADC ANALOG INPUT

When the buffer is not selected, the input impedance of the analog input changes with ACLK clock frequency (ACLK F6h) and gain (PGA). The relationship is:

Impedance (
$$\Omega$$
) =  $\left(\frac{1}{f_{SAMP} \cdot C_S}\right)$ 

$$A_{IN} \text{ Impedance}(\Omega) = \left(\frac{10^6}{\text{ACLK Frequency}}\right) \cdot \left(\frac{7M\Omega}{\text{PGA}}\right)$$

where ACLK frequency = 
$$\frac{f_{CLK}}{(ACLK+1)}$$

and modclk = 
$$f_{MOD} = \frac{f_{ACLK}}{64}$$
.

NOTE: The input impedance for PGA = 128 is the same as that for PGA = 64 (that is,  $\frac{7M\Omega}{64}$ ).

Figure 14 shows the basic input structure of the MSC1210.

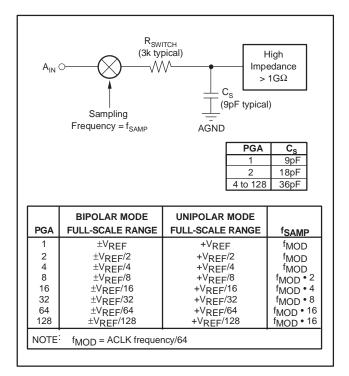


Figure 14. Analog Input Structure



#### ADC PGA

The PGA can be set to gains of 1, 2, 4, 8, 16, 32, 64, or 128. Using the PGA can actually improve the effective resolution of the ADC. For instance, with a PGA of 1 on a  $\pm 2.5V$  full-scale range, the ADC can resolve to  $1.5\mu$ V. With a PGA of 128 on a  $\pm 19$ mV full-scale range, the ADC can resolve to 75nV, as shown in Table 2.

Table 2.	ENOB	versus	PGA	(Bipolar	Mode)
----------	------	--------	-----	----------	-------

PGA SETTING	FULL-SCALE RANGE (V)	ENOB(1) AT 10HZ	RMS MEASUREMENT RESOLUTION (nV)
1	±2.5V	21.7	1468
2	±1.25	21.5	843
4	±0.625	21.4	452
8	±0.313	21.2	259
16	±0.156	20.8	171
32	±0.0781	20.4	113
64	±0.039	20	74.5
128	±0.019	19	74.5

(1) ENOB = Log<sub>2</sub>(FSR/RMS Noise) = Log<sub>2</sub>( $2^{24}$ ) - Log<sub>2</sub>( $\sigma_{CODES}$ ) = 24 - Log<sub>2</sub>( $\sigma_{CODES}$ )

#### ADC OFFSET DAC

The analog input to the PGA can be offset (in bipolar mode) by up to half the full-scale input range of the PGA by using the ODAC register (SFR E6h). The ODAC (Offset DAC) register is an 8-bit value; the MSB is the sign and the seven LSBs provide the magnitude of the offset. Since the ODAC introduces an analog (instead of digital) offset to the PGA, using the ODAC does not reduce the range of the ADC.

#### ADC MODULATOR

The modulator is a single-loop 2nd-order system. The modulator runs at a clock speed ( $f_{MOD}$ ) that is derived from the CLK using the value in the Analog Clock (ACLK) register (SFR F6h). The data rate is:

Data Rate =  $\frac{f_{MOD}}{Decimation Ratio}$ 

where 
$$f_{MOD} = \frac{f_{CLK}}{(ACLK+1) \cdot 64} = \frac{f_{ACLK}}{64}$$

and Decimation Ratio is set in [ADCON3:ADCON2].

#### ADC CALIBRATION

The offset and gain errors in the MSC1210, or the complete system, can be reduced with calibration. Calibration is controlled through the ADCON1 register (SFR DDh), bits CAL2:CAL0. Each calibration process takes seven  $t_{DATA}$  (data conversion time) periods to complete. Therefore, it takes 14  $t_{DATA}$  periods to complete both an offset and gain calibration.

For system calibration, the appropriate signal must be applied to the inputs. The system offset command requires a *zero* differential input signal. It then computes an offset value that will nullify offsets in the system. The system gain command requires a positive *full-scale* differential input signal. It then computes a value to nullify gain errors in the system. Each of these calibrations will take seven t<sub>DATA</sub> periods to complete.

Calibration should be performed after power on. It should also be done after a change in temperature, decimation ratio, buffer, Power Supply, voltage reference, or PGA. The Offset DAC wil affect offset calibration; therefore, the value of the Offset DAC should be zero until prior to performing a calibration.

At the completion of calibration, the ADC Interrupt bit goes HIGH which indicates the calibration is finished and valid data is available.

#### ADC DIGITAL FILTER

The Digital Filter can use either the Fast Settling, Sinc<sup>2</sup>, or Sinc<sup>3</sup> filter, as shown in Figure 15. In addition, the Auto mode changes the Sinc filter after the input channel or PGA is changed. When switching to a new channel or new PGA value, it will use the Fast Settling filter for the next two conversions (the first of which should be discarded). It will then use the Sinc<sup>2</sup> followed by the Sinc<sup>3</sup> filter to improve noise performance.

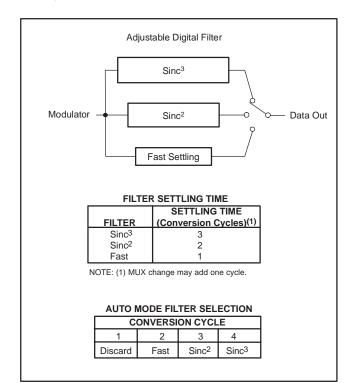


Figure 15. Filter Step Responses

This combines the low-noise advantage of the Sinc<sup>3</sup> filter with the quick response of the Fast Settling Time filter. The frequency response of each filter is shown in Figure 16.

#### **VOLTAGE REFERENCE**

The MSC1210 can use either an internal or external voltage reference. The voltage reference selection is controlled via ADC Control Register 0 (ADCON0, SFR DCh). The default power-up configuration for the voltage reference is 2.5V internal.

The internal voltage reference can be selected as either 1.25V or 2.5V. The analog power supply (AV<sub>DD</sub>) must be within the specified range for the selected internal voltage reference. The valid ranges are:  $V_{REF} = 2.5$  internal (AV<sub>DD</sub> = 3.3V to 5.25V) and  $V_{REF} = 1.25$  internal (AV<sub>DD</sub> = 2.7V to 5.25V). If the internal  $V_{REF}$  is selected, then the REFOUT pin must be connected to REFIN+, and AGND must be connected to REFIN–. The REFOUT pin should also have a 0.1µF capacitor connected to AGND, as close as possible to the pin. If the internal  $V_{REF}$  is not used, then  $V_{REF}$  should be disabled in ADCON0.

If the external voltage reference is selected, it can be used as either a single-ended input or differential input, for ratiometric measures. When using an external reference, it is important to note that the input current will increase for  $V_{\text{REF}}$  with higher PGA settings and with a higher modulator frequency. The external voltage reference can be used over the input range specified in the *Electrical Characteristics* section.

For applications requiring higher performance than that obtainable from the internal reference, use an external precision reference such as the REF50xx. The internal reference performance can be observed in the noise (and ENOB) versus input signal graphs in the Typical Characteristics section. All the rest of the ENOB plots are obtained with the inputs shorted together. By shorting the inputs, the inherent noise performance of only the ADC can be determined and displayed. When the inputs are not shorted, the extra noise comes from the reference. As can be seen in the ENOB vs Input Signal graph, the external reference adds about 0.7 bits of noise, whereas the internal reference adds about 2.3 bits of noise. This ENOB performance of 19.4 represents 21.16 bits of noise. With an LSB of 298nV, that translates to 6.3µV, or a peak-to-peak noise of almost 42µV. An external reference provides the best noise, drift, and repeatability performance for high-precision applications.



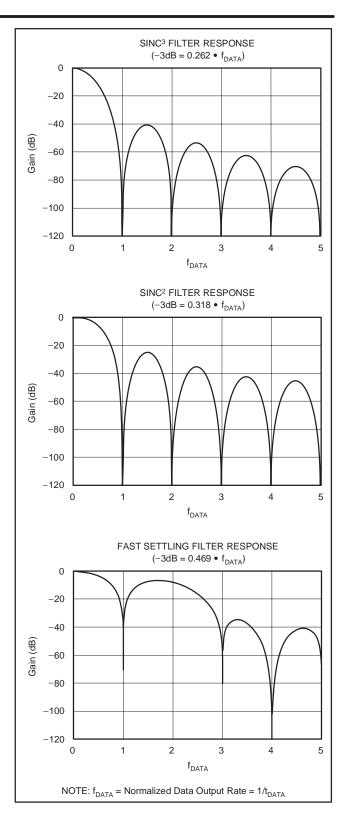


Figure 16. Filter Frequency Responses



#### RESET

The device can be reset from the following sources:

- Power-on reset
- External reset
- Software reset
- Watchdog timer reset
- Brownout reset

An external reset is accomplished by taking the RST pin high for two t<sub>OSC</sub> periods, followed by taking the RST pin low. A software reset is accomplished through the System Reset register (SRTST, 0F7h). A watchdog timer reset is enabled and controlled through Hardware Configuration Register 0 (HCR0) and the Watchdog Timer register (WDTCON, 0FFh). A brownout reset is enabled through Hardware Configuration Register 1 (HCR1). External reset, software reset, and watchdog timer reset complete after 2<sup>17</sup> clock cycles. A brownout reset completes after 2<sup>15</sup> clock cycles.

All sources of reset cause the digital pins to be pulled high from the initiation of the reset. For an external reset, taking the RST pin high stops device operation, crystal oscillation, and causes all digital pins to be pulled high from that point. Taking the RST pin low initiates the reset procedure.

A recommended external reset circuit is shown in Figure 17. The serial  $10k\Omega$  resistor is recommended for any external reset circuit configuration.

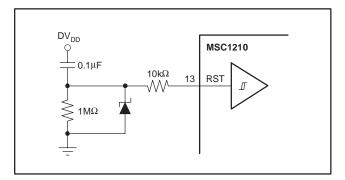


Figure 17. Typical Reset Circuit

#### POWER-ON RESET

The on-chip power-on reset (POR) circuitry releases the device from reset at approximately  $DV_{DD} = 2.0V$ . The POR accommodates power-supply ramp rates as slow as 1V/10ms. To ensure proper operation, the power supply should ramp monotonically. Note that as the device is released from reset and program execution begins, the device current consumption may increase, which may result in a power-supply voltage drop. If the power supply ramps at a slower rate, is not monotonic, or a brownout condition occurs (where the supply does not drop below the 2.0V threshold), then improper device operation may occur. The on-chip brownout reset may provide benefit in these conditions.

#### **BROWNOUT RESET**

The brownout reset (BOR) is enabled through Hardware Configuration Register 1 (HCR1). If the conditions for proper POR are not met or the device encounters a brownout condition that does not generate a POR, the BOR can be used to ensure proper device operation. The BOR will hold the state of the device when the power supply drops below the threshold level programmed in HCR1, and then generate a reset when the supply rises above the threshold level. Note that as the device is released from reset, and program execution begins, the device current consumption may increase, which may result in a power-supply voltage drop, which may initiate another brownout condition.

The BOR level should be chosen to match closely with the application. For example, with a high external clock frequency, the BOR level should match the minimum operating voltage range for the device, or improper operation may still occur.

Note that  $AV_{DD}$  must rise above 2.0V for the Analog Brownout Reset function to be disabled; otherwise, it will be enabled and hold the device in reset.

The BOR voltage is not calibrated until the end of the reset cycle; therefore, the actual BOR voltage will be approxiamtely 25% higher than the selected voltage. This can create a condition where the reset never ends (for example, when selecting a 4.5V BOR voltage for a 5V power supply).



#### **IDLE MODE**

Idle mode is entered by setting the IDLE bit in the Power Control register (PCON, 087h). In Idle mode, the CPU, Timer0, Timer1, and USARTs are stopped, but all other peripherals and digital pins remain active. The device can be returned to active mode via an active internal or external interrupt. This mode is typically used for reducing power consumption between ADC samples.

By configuring the device prior to entering Idle mode, further power reductions can be achieved (while in Idle mode). These reductions include powering down peripherals not in use in the PDCON register (0F1h).

#### STOP MODE

Stop mode is entered by setting the STOP bit in the Power Control register (PCON, 087h). In Stop mode, all internal clocks are halted. This mode has the lowest power consumption. The device can be returned to active mode only via an external or power-on reset.

By configuring the device prior to entering Stop mode, further power reductions can be achieved (while in Stop mode). These power reductions include halting the external clock into the device, configuring all digital I/O pins as open drain with low output drive, disabling the ADC buffer, disabling the internal  $V_{REF}$ , and setting PDCON to 0FFh to power down all peripherals.

In Stop mode, if the brownout reset is enabled, there is approximately  $25\mu A$  of draw from the power supply. To achieve zero current ( $\approx$  100nA) in Stop mode, disable the brownout reset via HCR1.

In Stop mode, all digital pins retain their values.

#### POWER CONSUMPTION CONSIDERATIONS

RUMENTS

The following suggestions will reduce current consumption:

- 1. Use the lowest supply voltage that will work in the application for both  $AV_{DD}$  and  $DV_{DD}$ .
- 2. Use the lowest clock frequency that will work in the application.
- 3. Use Idle mode and the system clock divider whenever possible. Note that the system clock divider also affects the ADC clock.
- 4. Avoid using 8051-compatible I/O mode on the I/O ports. The internal pull-up resistors will draw current when the outputs are low.
- 5. Use the delay line for Flash Memory control by setting the FRCM bit in the FMCON register (SFR EEh)
- 6. Power down peripherals when they are not needed. Refer to SFR PDCON, LVDCON, and ADCON0.

#### **MEMORY MAP**

The MSC1210 contains on-chip SFR, Flash Memory, Scratchpad SRAM Memory, Boot ROM, and SRAM. THe SFR registers are primarily used for control and status. The standard 8051 features and additional peripheral features of the MSC1210 are controlled through the SFR. Reading from an undefined SFR and writing to undefined SFR registers is not recommended, and will have indeterminate effects.

Flash Memory is used for both Program Memory and Data Memory. The user has the ability to select the partition size of Program and Data Memories. The partition size is set through hardware configuration bits, which are programmed through either the parallel or serial programming methods. Both Program and Data Flash Memories are erasable and writable (programmable) in User Application mode (UAM). However, program execution can only occur from Program Memory. As an added precaution, a lock feature can be activated through the hardware configuration bits, which disables erase and writes to 4kB of Program Flash Memory or the entire Program Flash Memory in UAM.

The MSC1210 includes 1kB of SRAM on-chip. SRAM starts at address 0 and is accessed through the MOVX instruction. This SRAM can also be located to start at 8400h and can be accessed as both Program and Data Memory.



#### FLASH MEMORY

The page size for Flash memory is 128 bytes. The respective page must be erased before it can be written to, regardless of whether it is mapped to Program or Data Memory space. The MSC1210 uses a memory addressing scheme that separates Program Memory (FLASH/ROM) from Data Memory (FLASH/RAM). Each area is 64kB beginning at address 0000h and ending at FFFFh, as shown in Figure 18. The program and data segments can overlap since they are accessed in different ways. Program Memory is fetched by the microcontroller automatically. There is one instruction (MOVC) that is used to explicitly read the program area. This is commonly used to read lookup tables.

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The Data Memory area is accessed explicitly using the MOVX instruction. This instruction provides multiple ways of specifying the target address. It is used to access the 64kB of Data Memory. The address and data range of devices with on-chip Program and Data Memory overlap the 64kB memory space. When on-chip memory is enabled, accessing memory in the on-chip range will cause the device to access internal memory. Memory accesses beyond the internal range will be addressed externally via Ports 0 and 2.

The MSC1210 has two Hardware Configuration registers (HCR0 and HCR1) that are programmable only during Flash Memory Programming mode.

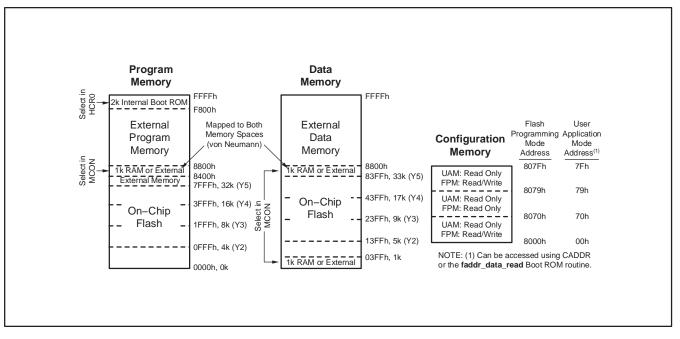


Figure 18. Memory Map

The MSC1210 allows the user to partition the Flash Memory between Program Memory and Data Memory. For instance, the MSC1210Y5 contains 32kB of Flash Memory on-chip. Through the HW configuration registers, the user can define the partition between Program Memory (PM) and Data Memory (DM), as shown in Table 3 and Table 4. The MSC1210 family offers four memory configurations, as shown.

HCR0	MSC1	1210Y2 MSC1210Y3		210Y3	MSC1210Y4		MSC1210Y5	
DFSEL	PM	DM	PM	DM	PM	DM	PM	DM
000	0kB	4kB	0kB	8kB	0kB	16kB	0kB	32kB
001	0kB	4kB	0kB	8kB	0kB	16kB	0kB	32kB
010	0kB	4kB	0kB	8kB	0kB	16kB	16kB	16kB
011	0kB	4kB	0kB	8kB	8kB	8kB	24kB	8kB
100	0kB	4kB	4kB	4kB	12kB	4kB	28kB	4kB
101	2kB	2kB	6kB	2kB	14kB	2kB	30kB	2kB
110	3kB	1kB	7kB	1kB	15kB	1kB	31kB	1kB
111 (default)	4kB	0kB	8kB	0kB	16kB	0kB	32kB	0kB
<b>NOTE:</b> When a 0kB program memory configuration is selected, program execution is external.								

#### Table 3. MSC1210 Flash Partitioning

HCR0	MSC1	210Y2	MSC1	210Y3	MSC1	210Y4	MSC1210Y5	
DFSEL	PM	DM	PM	DM	PM	DM	PM	DM
000	0000	0400- 13FF	0000	0400- 23FF	0000	0400- 43FF	0000	0400- 83FF
001	0000	0400- 13FF	0000	0400- 23FF	0000	0400- 43FF	0000	0400- 83FF
010	0000	0400- 13FF	0000	0400- 23FF	0000	0400- 43FF	0000- 3FFF	0400- 43FF
011	0000	0400- 13FF	0000	0400- 23FF	0000- 1FFF	0400- 23FF	0000- 5FFF	0400- 23FF
100	0000	0400- 13FF	0000- 0FFF	0400- 13FF	0000- 2FFF	0400- 13FF	0000- 6FFF	0400- 13FF
101	0000- 07FF	0400- 0BFF	0000- 17FF	0400- 0BFF	0000- 37FF	0400- 0BFF	0000- 77FF	0400- 0BFF
110	0000- 0BFF	0400- 07FF	0000- 1BFF	0400- 07FF	0000- 3BFF	0400- 07FF	0000- 7BFF	0400- 07FF
111 (default)	0000- 0FFF	0000	0000- 1FFF	0000	0000- 3FFF	0000	0000- 7FFF	0000
	NOTE: Program memory accesses above the highest listed address will access external program memory.							

#### Table 4. MSC1210 Flash Memory Partitioning

It is important to note that the Flash Memory is readable and writable by the user through the MOVX instruction when configured as either Program or Data Memory (via the MXWS bit in the MWS, SFR 8Fh). This means that the user may partition the device for maximum Flash Program Memory size (no Flash Data Memory) and use Flash Program Memory as Flash Data Memory. This may lead to undesirable behavior if the PC points to an area of Flash Program Memory that is being used for data storage. Therefore, it is recommended to use Flash partitioning when Flash Memory is used for data storage. Flash partitioning prohibits execution of code from Data Flash Memory. Additionally, the Program Memory erase/write can be disabled through hardware configuration bits (HCR0), while still providing access (read/write/erase) to Data Flash Memory.

The effect of memory mapping on Program and Data Memory is straightforward. The Program Memory is decreased in size from the top of internal Program Memory. Therefore, if the MSC1210Y5 is partitioned with 31kB of Flash Program Memory and 1kB of Flash Data Memory, external Program Memory execution will begin at 7C00h (versus 8000h for 32kB). The Flash Data Memory is added on top of the SRAM memory. Therefore, access to Data Memory (through MOVX) will access SRAM for addresses 0000h–03FFh and access Flash Memory for addresses 0400h–07FFh.

#### **Data Memory**

The MSC1210 can address 64kB of Data Memory. The MOVX instruction is used to access the Data SRAM Memory. This includes 1,024 bytes of on-chip Data SRAM Memory. The data bus values do not appear on Port 0 (during data bus timing) for internal memory access.

The MSC1210 also has on-chip Flash Data Memory which is readable and writable (depending on Memory Write Select register) during normal operation (full  $V_{DD}$  range). This memory is mapped into the external Data Memory space directly above the SRAM.

The MOVX instruction is used to write the flash memory. Flash memory must be erased before it can be written. Flash memory is erased in 128 byte pages.





#### CONFIGURATION MEMORY

The MSC1210 Configuration Memory consists of 128 bytes. In UAM, all Configuration Memory is readable using the **faddr\_data\_read** Boot ROM routine, and the CADDR and CDATA registers. In UAM, however, none of the Configuration Memory is writable.

In serial or parallel programming mode, all Configuration Memory is readable. Most locations are also writable, except for addresses 8070h through 8079h, which are read-only.

The two hardware configuration registers reside in configuration memory at 807Eh (HCR1) and 807Fh (HCR0).

Figure 19 shows the configuration memory mapping for programming mode and UAM. Note that reading/writing configuration memory in Flash Programming mode (FPM) requires 16-bit addressing, whereas reading configuration memory in User Application mode (UAM) requires only 8-bit addressing.

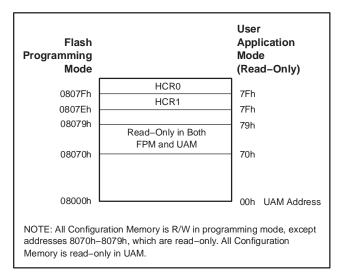


Figure 19. Configuration Memory Map

#### **REGISTER MAP**

The Register Map is illustrated in Figure 20. It is entirely separate from the Program and Data Memory areas mentioned before. A separate class of instructions is used to access the registers. There are 256 potential register locations. In practice, the MSC1210 has 256 bytes of Scratchpad RAM and up to 128 SFRs. This is possible, since the upper 128 Scratchpad RAM locations can only

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be accessed indirectly. Thus, a direct reference to one of the upper 128 locations must be an SFR access. Direct RAM is reached at locations 0 to 7Fh (0 to 127).

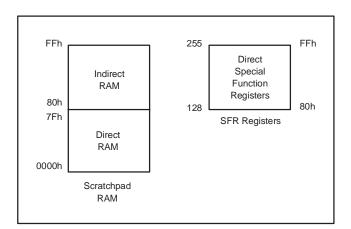


Figure 20. Register Map

SFRs are accessed directly between 80h and FFh (128 to 255). The RAM locations between 128 and 255 can be reached through an indirect reference to those locations. Scratchpad RAM is available for general-purpose data storage. It is commonly used in place of off-chip RAM when the total data contents are small. When off-chip RAM is needed, the Scratchpad area will still provide the fastest general-purpose access. Within the 256 bytes of RAM, there are several special-purpose areas.

#### **Bit Addressable Locations**

In addition to direct register access, some individual bits are also accessible. These are individually addressable bits in both the RAM and SFR area. In the Scratchpad RAM area, registers 20h to 2Fh are bit addressable. This provides 128 ( $16 \times 8$ ) individual bits available to software. A bit access is distinguished from a full-register access by the type of instruction. In the SFR area, any register location ending in a 0 or 8 is bit addressable. Figure 21 shows details of the on-chip RAM addressing including the locations of individual RAM bits.

#### **Working Registers**

As part of the lower 128 bytes of RAM, there are four banks of Working Registers, as shown in Figure 21. The Working Registers are general-purpose RAM locations that can be addressed in a special way. They are designated R0

through R7. Since there are four banks, the currently selected bank will be used by any instruction using R0—R7. This allows software to change context by simply switching banks. This is controlled via the Program Status Word register (PSW; 0D0h) in the SFR area described below. Registers R0 and R1 also allow their contents to be used for indirect addressing of the upper 128 bytes of RAM. Thus, an instruction can designate the value stored in R0 (for example) to address the upper RAM. The 16 bytes immediately above the R0—R7 registers are bit addressable; any of the 128 bits in this area can be directly accessed using bit addressable instructions.

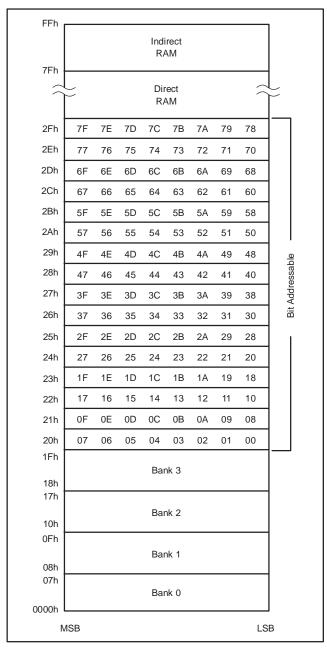


Figure 21. Scratchpad Register Addressing



#### Stack

Another use of the Scratchpad area is for the programmer's stack. This area is selected using the Stack Pointer (SP; 81h) SFR. Whenever a call or interrupt is invoked, the return address is placed on the Stack. It also is available to the programmer for variables, etc., since the Stack can be moved and there is no fixed location within the RAM designated as Stack. The Stack Pointer will default to 07h on reset. The user can then move it as needed. A convenient location would be the upper RAM area (> 7Fh) since this is only available indirectly. The SP will point to the last used value. Therefore, the next value placed on the Stack is put at SP + 1. Each PUSH or CALL will increment the SP by the appropriate value. Each POP or RET will decrement as well.

#### Program Memory

After reset, the CPU begins execution from Program Memory location 0000h. The selection of where Program Memory execution begins is made by tying the  $\overline{EA}$  pin to DV<sub>DD</sub> for internal access, or DGND for external access. When  $\overline{EA}$  is tied to DV<sub>DD</sub>, any PC fetches outside the internal Program Memory address occur from external memory. If  $\overline{EA}$  is tied to DGND, then all PC fetches address external memory. The standard internal Program Memory size for MSC1210 family members is shown in Table 5. If enabled the Boot ROM will appear from address F800h to FFFFh.

Table 5. MSC1210 Maximum Internal Pr	ogram
Memory Sizes	•

PRODUCT	STANDARD INTERNAL PROGRAM MEMORY SIZE (BYTES)
MSC1210Y5	32k
MSC1210Y4	16k
MSC1210Y3	8k
MSC1210Y2	4k

#### **Boot ROM**

There is a 2kB Boot ROM that controls operation during serial or parallel programming. The Boot ROM routines can be accessed during the user mode if it is enabled. The Boot ROM routines are listed in Table 6. When enabled, the Boot ROM routines will be located at memory addresses F800h–FFFFh during user mode. In program mode the Boot ROM is located in the first 2kB of Program Memory. For additional information, refer to Application Note SBAA085, *MSC1210 ROM Routines*, available for download from the TI web site (www.ti.com).

Table 6. M	ISC1210 Boot	ROM Rout	ines
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ADDRESS	ROUTINE	C DECLARATIONS	DESCRIPTION
FFD5	put_string	<pre>void put_string (char code *string);</pre>	Output string
FFD7	page_erase	char page_erase (int faddr, char fdata, char fdm);	Erase flash page
FFD9	write_flash	Assembly only; DPTR = address, R5 = data	Fast flash write
FFDB	write_flash_chk	char write_flash_chk (int faddr, char fdata, char fdm);	Write flash byte, verify
FFDD	write_flash_byte	char write_flash_byte (int faddr, char fdata, char fdm);	Write flash byte
FFDF	faddr_data_read	char faddr_data_read (char faddr);	Read HW config byte from addr
FFE1	data_x_c_read	char data_x_c_read (int faddr, char fdm);	Read xdata or code byte
FFE3	tx_byte	void tx_byte (char);	Send byte to USART0
FFE5	tx_hex	void tx_hex (char);	Send hex value to USART0
FFE7	putok	void putok (void);	Send "OK" to USART0
FFE9	rx_byte	char rx_byte (void);	Read byte from USART0
FFEB	rx_byte_echo	char rx_byte_echo (void);	Read and echo byte on USART0
FFED	rx_hex_echo	int rx_hex_echo (void);	Read and echo hex on USART0
FFEF	rx_hex_int_echo	int rx_hex_int_echo (void);	Read int as hex and echo: USART0
FFF1	rx_hex_rev_echo	int rx_hex_rev_echo (void);	Read int reversed as hex and echo: USART0
FFF3	autobaud	void autobaud (void);	Set baud rate with received CR
FFF5	putspace4	void putspace4 (void);	Output 4 spaces to USART0
FFF7	putspace3	void putspace3 (void);	Output 3 spaces to USART0
FFF9	putspace2	void putspace2 (void);	Output 2 spaces to USART0
FFFB	putspace1	void putspace1 (void);	Output 1 space to USART0
FFFB	putcr	void putcr (void);	Output CR, LF to USART0
F979	cmd_parse	void cmd_parser (void);	See SBAA076
FD37	monitor_isr	void monitor_isr () interrupt 6	Push registers and call cmd_parser



#### ACCESSING EXTERNAL MEMORY

If external memory is used, P0 and P2 can be configured as address and data lines. If external memory is not used, P0 and P2 can be configured as general-purpose I/O lines through the Hardware Configuration Register.

To enable access to external memory, bits 0 and 1 of the HCR1 register must be set to 0. When these bits are enabled all memory addresses for both internal and external memory will appear on ports 0 and 2. During the data portion of the cycle for internal memory, Port 0 will be zero for security purposes.

Accesses to external memory are of two types: accesses to external Program Memory and accesses to external Data Memory. Accesses to external Program Memory use signal PSEN (program store enable) as the read strobe. Accesses to external Data Memory use RD or WR (alternate functions of P3.7 and P3.6) to strobe the memory.

External Program Memory and external Data Memory may be combined if desired by applying the RD and PSEN signals to the inputs of an AND gate and using the output of the gate as the read strobe to the external Program/Data Memory.

Program fetches from external Program Memory always use a 16-bit address. Accesses to external Data Memory can use either a 16-bit address (MOVX @DPTR) or an 8-bit address (MOVX @R<sub>I</sub>).

If Port 2 is selected for external memory use (HCR1, bit 0), it cannot be used as general-purpose I/O. This bit (or Bit 1 of HCR1) also forces bits P3.6 and P3.7 to be used for WR and RD instead of I/O. Port 2, P3.6, and P3.7 should all be written to '1.'

If an 8-bit address is being used (MOVX  $@R_1$ ), the contents of the MPAGE (92h) SFR remain at the Port 2 pins throughout the external memory cycle. This will facilitate paging.

In any case, the low byte of the address is time-multiplexed with the data byte on Port 0. The ADDR/DATA signals use CMOS drivers in the Port 0, Port 2,  $\overline{WR}$ , and  $\overline{RD}$  output buffers. Thus, in this application the Port 0 pins are not open-drain outputs, and do not require external pull-ups for high-speed access. Signal ALE (Address Latch Enable) should be used to capture the address byte into an external latch. The address byte is valid at the negative transition of ALE. Then, in a write cycle, the data byte to be written appears on Port 0 just before  $\overline{WR}$  is activated, and remains there until after  $\overline{WR}$  is deactivated. In a read cycle, the incoming byte is accepted at Port 0 just before the read strobe is deactivated.

The functions of Port 0 and Port 2 are selected in Hardware Configuration Register 1. This can only be changed during the Flash Program mode. There is no conflict in the use of these registers; they will either be used as general-purpose I/O or for external memory access. The default state is for Port 0 and Port 2 to be used as general-purpose I/O. If an external memory access is attempted when they are configured as general-purpose I/O, the values of Port 0 and Port 2 will not be affected.

External Program Memory is accessed under two conditions:

- 1. Whenever signal EA is LOW during reset, then all future accesses are external; or
- 2. Whenever the Program Counter (PC) contains a number that is outside of the internal Program Memory address range, if the ports are enabled.

If Port 0 and Port 2 are selected for external memory, all 8 bits of Port 0 and Port 2, as well as P3.6 and P3.7, are dedicated to an output function and may not be used for general-purpose I/O. During external program fetches, Port 2 outputs the high byte of the PC.

#### **Programming Flash Memory**

There are four sections of Flash Memory for programming:

- 1. 128 configuration bytes.
- 2. Reset sector (4kB) (not to be confused with the 2kB Boot ROM).
- 3. Program Memory.
- 4. Data Memory.

#### Flash Programming Mode

There are two programming modes: parallel and serial. The programming mode is selected by the state of the ALE and  $\overrightarrow{PSEN}$  signals during power-on reset. Serial programming mode is selected with  $\overrightarrow{PSEN} = 0$  and  $\overrightarrow{ALE} = 1$ . Parallel programming mode is selected with  $\overrightarrow{PSEN} = 1$  and  $\overrightarrow{ALE} = 0$  (see Figure 22). If they are both HIGH, the MSC1210 will operate in normal user mode. Both signals LOW is a reserved mode and is not defined. Programming mode is exited with a reset (BOR, WDT, software, or POR) and the normal mode selected.



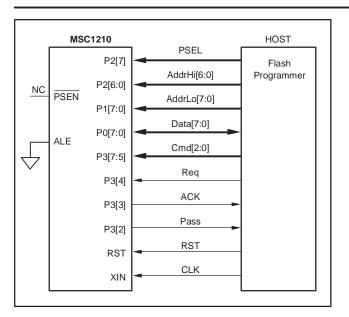


Figure 22. Parallel Programming Configuration

The MSC1210 is shipped with Flash Memory erased (all 1s). Parallel programming methods typically involve a third-party programmer. Serial programming methods typically involve in-system programming. UAM allows Flash Program and Data Memory programming. The actual code for Flash programming cannot execute from Flash. That code must execute from the Boot ROM, internal (von Neumann) RAM or external memory.

Figure 23 shows the serial programming conection.

Serial programming mode works through USART0, and has special protocols, which are discussed at length in Application Note SBAA076, *Programming the MSC1210*, available for download at www.ti.com. The serial programming mode works at a maximum baud rate determined by f<sub>OSC</sub>.

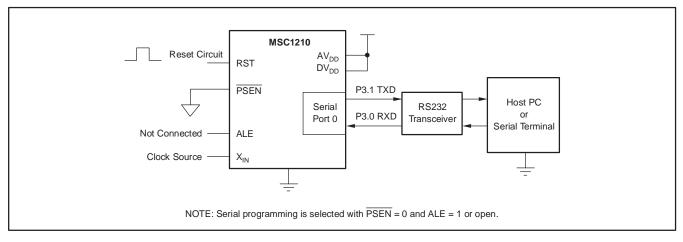


Figure 23. Serial Programming Connection



#### **INTERRUPTS**

The MSC1210 uses a three-priority interrupt system. As shown in Table 7, each interrupt source has an independent priority bit, flag, interrupt vector, and enable (except that nine interrupts share the Auxiliary Interrupt [AI] at the highest priority). In addition, interrupts can be globally enabled or disabled. The interrupt structure is compatible with the original 8051 family. All of the standard interrupts are available.

#### HARDWARE CONFIGURATION MEMORY

The 128 configuration bytes can only be written during the program mode. The bytes are accessed through SFR registers CADDR (SFR 93h) and CDATA (SFR 94h). Two of the configuration bytes control Flash partitioning and system control. If the security bit is set, these bits can not be changed except with a Mass Erase command that erases all of the Flash Memory including the 128 configuration bytes.

	INTERRUPT					PRIORITY
INTERRUPT/EVENT	ADDR	NUM	PRIORITY	FLAG	ENABLE	CONTROL
DV <sub>DD</sub> Low Voltage/HW Breakpoint	33h	6	HIGH	EDLVB (AIE.0) <sup>(1)</sup> EBP (BPCON.0) <sup>(1)</sup>	EDLVB (AIE.0) <sup>(1)</sup> EBP (BPCON.0) <sup>(1)</sup>	N/A
AV <sub>DD</sub> Low Voltage	33h	6	0	EALV (AIE.1) <sup>(1)</sup>	EALV (AIE.1) <sup>(1)</sup>	N/A
SPI Receive	33h	6	0	ESPIR (AIE.2) <sup>(1)</sup>	ESPIR (AIE.2) <sup>(1)</sup>	N/A
SPI Transmit	33h	6	0	ESPIT (AIE.3)(1)	ESPIT (AIE.3) <sup>(1)</sup>	N/A
Milliseconds Timer	33h	6	0	EMSEC (AIE.4)(1)	EMSEC (AIE.4) <sup>(1)</sup>	N/A
ADC	33h	6	0	EADC (AIE.5) <sup>(1)</sup>	EADC (AIE.5) <sup>(1)</sup>	N/A
Summation Register	33h	6	0	ESUM (AIE.6)(1)	ESUM (AIE.6) <sup>(1)</sup>	N/A
Seconds Timer	33h	6	0	ESEC (AIE.7) <sup>(1)</sup>	ESEC (AIE.7) <sup>(1)</sup>	N/A
External Interrupt 0	03h	0	1	IE0 (TCON.1) <sup>(2)</sup>	EX0 (IE.0) <sup>(4)</sup>	PX0 (IP.0)
Timer 0 Overflow	0Bh	1	2	TF0 (TCON.5)(3)	ET1 (IE.1) <sup>(4)</sup>	PT0 (IP.1)
External Interrupt 1	13h	2	3	IE1 (TCON.3) <sup>(2)</sup>	EX1 (IE.2) <sup>(4)</sup>	PX1 (IP.2)
Timer 1 Overflow	0Bh	3	4	TF1 (TCON.7) <sup>(3)</sup>	ET1 (IE.3) <sup>(4)</sup>	PT1 (IP.3)
Serial Port 0	23h	4	5	RI_0 (SCON0.0) TI_0 (SCON0.1)	ES0 (IE.4) <sup>(4)</sup>	PS0 (IP.4)
Timer 2 Overflow	2Bh	5	6	TF2 (T2CON.7)	ET2 (IE.5) <sup>(4)</sup>	PT2 (IP.5)
Serial Port 1	3Bh	7	7	RI_1 (SCON1.0) TI_1 (SCON1.1)	ES1 (IE.6) <sup>(4)</sup>	PS1 (IP.6)
External Interrupt 2	43h	8	8	IE2 (EXIF.4)	EX2 (EIE.0) <sup>(4)</sup>	PX2 (EIP.0)
External Interrupt 3	4Bh	9	9	IE3 (EXIF.5)	EX3 (EIE.1) <sup>(4)</sup>	PX3 (EIP.1)
External Interrupt 4	53h	10	10	IE4 (EXIF.6)	EX4 (EIE.2) <sup>(4)</sup>	PX4 (EIP.2)
External Interrupt 5	5Bh	11	11	IE5 (EXIF.7)	EX5 (EIE.3) <sup>(4)</sup>	PX5 (EIP.3)
Watchdog	63h	12	12 LOW	WDTI (EICON.3)	EWDI (EIE.4) <sup>(4)</sup>	PWDI (EIP.4)

#### Table 7. Interrupt Summary

(1) These interrupts set the AI flag (EICON.4) and are enabled by EAI (EICON.5).

(2) If edge-triggered, cleared automatically by hardware when the service routine is vectored to. If level-triggered, the flag follows the state of the pin.

(3) Cleared automatically by hardware when interrupt vector occurs.

(4) Globally enabled by  $\overline{EA}$  (IE.7).



#### Hardware Configuration Register 0 (HCR0)—Accessed Using SFR Registers CADDR and CDATA.

	bit 7	bit 6	bit 5	bit 4	bit 3	bit 2	bit 1	bit 0
CADDR 7Fh	EPMA	PML	RSL	EBR	EWDR	DFSEL2	DFSEL1	DFSEL0

**NOTE:** HCR0 is programmable only in Flash Programming mode, but can be read in User Application mode using the CADDR and CDATA SFRs or the **faddr\_data\_read** Boot ROM routine.

#### EPMA Enable Programming Memory Access (Security Bit).

bit 7 0: After reset in programming modes, Flash Memory can only be accessed in UAM until a mass erase is done. 1: Fully Accessible (default)

#### PML Program Memory Lock (PML has Priority Over RSL).

- bit 60: Enable all Flash Programming modes in program mode, can be written in UAM.1: Enable read-only for program mode; cannot be written in UAM (default).
- RSL Reset Sector Lock. The reset sector can be used to provide another method of Flash Memory programming. This bit 5 will allow Program Memory updates without changing the jumpers for in-circuit code updates or program development. The code in this boot sector would then provide the monitor and programming routines with the ability to jump into the main Flash code when programming is finished.
  - 0: Enable Reset Sector Writing
  - 1: Enable Read-Only Mode for Reset Sector (4kB) (default)
- **EBR** Enable Boot ROM. Boot ROM is 2kB of code located in ROM, not to be confused with the 4kB Boot Sector located bit 4 in Flash Memory.
  - 0: Disable Internal Boot ROM
    - 1: Enable Internal Boot ROM (default)

#### EWDR Enable Watchdog Reset.

- bit 3 0: Disable Watchdog Reset
  - 1: Enable Watchdog Reset (default)

#### DFSEL Data Flash Memory Size (see Table 3 and Table 4).

- bits 2–0 000: Reserved
  - 001: 32kB, 16kB, 8kB, or 4kB Data Flash Memory
  - 010: 16kB, 8kB, or 4kB Data Flash Memory
  - 011: 8kB or 4kB Data Flash Memory
  - 100: 4kB Data Flash Memory
  - 101: 2kB Data Flash Memory
  - 110: 1kB Data Flash Memory
  - 111: No Data Flash Memory (default)



#### Hardware Configuration Register 1 (HCR1)

	bit 7	bit 6	bit 5	bit 4	bit 3	bit 2	bit 1	bit 0
CADDR 7Eh	DBLSEL1	DBLSEL0	ABLSEL1	ABLSEL0	DAB	DDB	EGP0	EGP23

**NOTE:** HCR1 is programmable only in Flash Programming mode, but can be read in User Application mode using the CADDR and CDATA SFRs or the **faddr\_data\_read** Boot ROM routine.

#### DBLSEL **Digital Brownout Level Select** bits 7-6 00: 4.5V 01: 4.2V 10: 2.7V 11: 2.5V (default) ABLSEL **Analog Brownout Level Select** bits 5-4 00: 4.5V 01: 4.2V 10: 2.7V 11: 2.5V (default) DAB **Disable Analog Power-Supply Brownout Reset** bit 3 0: Enable Analog Brownout Reset 1: Disable Analog Brownout Reset (default) (will not disable unless AV<sub>DD</sub> > 2.0V) DDB **Disable Digital Power-Supply Brownout Reset** 0: Enable Digital Brownout Reset bit 2 1: Disable Digital Brownout Reset (default) EGP0 Enable General-Purpose I/O for Port 0 0: Port 0 is Used for External Memory, P3.6 and P3.7 Used for WR and RD. bit 1 1: Port 0 is Used as General-Purpose I/O (default) EGP23 Enable General-Purpose I/O for Ports 2 and 3 0: Port 2 is Used for External Memory, P3.6 and P3.7. Used for WR and RD. bit 0 1: Port 2 and Port3 are Used as General-Purpose I/O (default)

#### **Configuration Memory Programming**

Certain key functions such as Brownout Reset and Watchdog Timer are controlled by the hardware configuration bits. These bits are nonvolatile and can only be changed through serial and parallel programming. Other peripheral control and status functions, such as ADC configuration, timer setup, and Flash control, are controlled through the SFRs.



#### SFR Definitions (Boldface definitions indicate that the register is unique to the MSC1210Yx)

ADDRESS	REGISTER	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0	RESET VALUES
80h	P0	P0.7	P0.6	P0.5	P0.4	P0.3	P0.2	P0.1	P0.0	FFh
81h	SP									07h
82h	DPL0									00h
83h	DPH0									00h
84h	DPL1									00h
85h	DPH1									00h
86h	DPS	0	0	0	0	0	0	0	SEL	00h
87h	PCON	SMOD	0	1	1	GF1	GF0	STOP	IDLE	30h
88h	TCON	TF1	TR1	TF0	TR0	IE1	IT1	IE0	IT0	00h
89h	TMOD			ner 1 –––––				ner 0		00h
		GATE	C/T	M1	MO	GATE	C/T	M1	MO	-
8Ah	TL0	-			-	-				00h
8Bh	TL1									00h
8Ch	TH0									00h
8Dh	TH1									00h
8Eh	CKCON	0	0	T2M	T1M	том	MD2	MD1	MD0	01h
8Fh	MWS	0	0	0	0	0	0	0	MXWS	00h
90h	P1	P1.7 INT5/SCK	P1.6 INT4/MISO	P1.5 INT3/MOSI	P1.4 INT2/SS	P1.3 TXD1	P1.2 RXD1	P1.1 T2EX	P1.0 T2	FFh
91h	EXIF	IE5	IE4	IE3	IE2	1	0	0	0	08h
92h	MPAGE									00h
93h	CADDR									00h
94h	CDATA									00h
95h	MCON	BPSEL	0	0					RAMMAP	00h
96h										
97h										
98h	SCON0	SM0_0	SM1_0	SM2_0	REN_0	TB8_0	RB8_0	TI_0	RI_0	00h
99h	SBUF0									00h
9Ah	SPICON	SCK2	SCK1	SCK0	0	ORDER	MSTR	СРНА	CPOL	00h
9Bh	SPIDATA									00h
9Dh	SPITCON			CLK_EN	DRV_DLY	DRV_EN				00h
A0h	P2	P2.7	P2.6	P2.5	P2.4	P2.3	P2.2	P2.1	P2.0	FFh
A1h	PWMCON			PPOL	PWMSEL	SPDSEL	TPCNTL2	TPCNTL1	TPCNTL0	00h
A2h	PWMLOW TONELOW	PWM7 TDIV7	PWM6 TDIV6	PWM5 TDIV5	PWM4 TDIV4	PWM3 TDIV3	PWM2 TDIV2	PWM1 TDIV1	PWM0 TDIV0	00h
A3h	PWMHI TONEHI	PWM15 TDIV15	PWM14 TDIV14	PWM13 TDIV13	PWM12 TDIV12	PWM11 TDIV11	PWM10 TDIV10	PWM9 TDIV9	PWM8 TDIV8	00h
A4h										
A5h	PAI	0	0	0	0	PAI3	PAI2	PAI1	PAI0	00h
A6h	AIE	ESEC	ESUM	EADC	EMSEC	ESPIT	ESPIR	EALV	EDLVB	00h
A7h	AISTAT	SEC	SUM	ADC	MSEC	SPIT	SPIR	ALVD	DLVD	00h
A8h	IE	EA	ES1	ET2	ES0	ET1	EX1	ET0	EX0	00h
A9h	BPCON	BP	0	0	0	0	0	PMSEL	EBP	00h
AAh	BPL									00h
ABh	BPH									00h
ACh	PODDRL	P03H	P03L	P02H	P02L	P01H	P01L	P00H	P00L	00h
ADh	P0DDRH	P07H	P07L	P06H	P06L	P05H	P05L	P04H	P04L	00h



SFR Definitions (continued) (Boldface definitions indicate that the register is unique to the MSC1210Yx)

ADDRESS	REGISTER	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0	RESET VALUES
AEh	P1DDRL	P13H	P13L	P12H	P12L	P11H	P11L	P10H	P10L	00h
AFh	P1DDRH	P17H	P17L	P16H	P16L	P15H	P15L	P14H	P14L	00h
B0h	P3	<u>P3.</u> 7 RD	<u>P3.</u> 6 WR	P3.5 T1	P3.4 T0	<u>P3.3</u> INT1	<u>P3.2</u> INT0	P3.1 TXD0	P3.0 RXD0	FFh
B1h	P2DDRL	P23H	P23L	P22H	P22L	P21H	P21L	P20H	P20L	00h
B2h	P2DDRH	P27H	P27L	P26H	P26L	P25H	P25L	P24H	P24L	00h
B3h	P3DDRL	P33H	P33L	P32H	P32L	P31H	P31L	P30H	P30L	00h
B4h	P3DDRH	P37H	P37L	P36H	P36L	P35H	P35L	P34H	P34L	00h
B5h										
B6h										
B7h										
B8h	IP	1	PS1	PT2	PS0	PT1	PX1	PT0	PX0	80h
B9h										
BAh										
BBh										
BCh						1		1		1
BDh						1				
BEh										
BFh										
C0h	SCON1	SM0_1	SM1_1	SM2_1	REN_1	TB8_1	RB8_1	TI_1	RI_1	00h
C1h	SBUF1									00h
C2h										
C3h										
C4h										
C5h										
C6h	EWU						EWUWDT	EWUEX1	EWUEX0	00h
C7h										
C8h	T2CON	TF2	EXF2	RCLK	TCLK	EXEN2	TR2	C/T2	CP/RL2	00h
C9h										
CAh	RCAP2L									00h
CBh	RCAP2H									00h
CCh	TL2									00h
CDh	TH2									00h
CEh										
CFh										
D0h	PSW	CY	AC	F0	RS1	RS0	OV	F1	Р	00h
D1h	OCL								LSB	00h
D2h	осм					1		1		00h
D3h	осн	MSB				1	1	1		00h
D4h	GCL					1	1	1	LSB	5Ah
D5h	GCM					1				ECh
D6h	GCH	MSB								5Fh
D7h	ADMUX	INP3	INP2	INP1	INP0	INN3	INN2	INN1	INN0	01h
D8h	EICON	SMOD1	1	EAI	AI	WDTI	0	0	0	40h
D9h	ADRESL								LSB	00h
DAh	ADRESM									00h



#### **SFR Definitions (continued)** (**Boldface** definitions indicate that the register is unique to the MSC1210Yx)

ADDRESS	REGISTER	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0	RESET VALUES
DBh	ADRESH	MSB								00h
DCh	ADCON0	—	BOD	EVREF	VREFH	EBUF	PGA2	PGA1	PGA0	30h
DDh	ADCON1	—	POL	SM1	SM0	-	CAL2	CAL1	CAL0	0000_0000b
DEh	ADCON2	DR7	DR6	DR5	DR4	DR3	DR2	DR1	DR0	1Bh
DFh	ADCON3	0	0	0	0	0	DR10	DR9	DR8	06h
E0h	ACC									00h
E1h	SSCON	SSCON1	SSCON0	SCNT2	SCNT1	SCNT0	SHF2	SHF1	SHF0	00h
E2h	SUMR0									00h
E3h	SUMR1									00h
E4h	SUMR2									00h
E5h	SUMR3									00h
E6h	ODAC									00h
E7h	LVDCON	ALVDIS	ALVD2	ALVD1	ALVD0	DLVDIS	DLVD2	DLVD1	DLVD0	00h
E8h	EIE	1	1	1	EWDI	EX5	EX4	EX3	EX2	E0h
E9h	HWPC0	0	0	0	0	0	0	MEMO	RY SIZE	0000_00xxb
EAh	HWPC1	0	0	0	0	0	0	0	0	00h
EBh	HDWVER									xxh
ECh	Reserved									00h
EDh	Reserved									00h
EEh	FMCON	0	PGERA	0	FRCM	0	BUSY	1	0	02h
EFh	FTCON	FER3	FER2	FER1	FER0	FWR3	FWR2	FWR1	FWR0	A5h
F0h	В	B.7	B.6	B.5	B.4	B.3	B.2	B.1	B.0	00h
F1h	PDCON	0	0	0	PDPWM	PDADC	PDWDT	PDST	PDSPI	1Fh
F2h	PASEL	0	0	PSEN2	PSEN1	PSEN0	0	ALE1	ALE0	00h
F3h										
F4h										
F5h										
F6h	ACLK	0	FREQ6	FREQ5	FREQ4	FREQ3	FREQ2	FREQ1	FREQ0	03h
F7h	SRST	0	0	0	0	0	0	0	RSTREQ	00h
F8h	EIP	1	1	1	PWDI	PX5	PX4	PX3	PX2	E0h
F9h	SECINT	WRT	SECINT6	SECINT5	SECINT4	SECINT3	SECINT2	SECINT1	SECINT0	7Fh
FAh	MSINT	WRT	MSINT6	MSINT5	MSINT4	MSINT3	MSINT2	MSINT1	MSINT0	7Fh
FBh	USEC	0	0	0	FREQ4	FREQ3	FREQ2	FREQ1	FREQ0	03h
FCh	MSECL									9Fh
FDh	MSECH									0Fh
FEh	HMSEC									63h
FFh	WDTCON	EWDT	DWDT	RWDT	WDCNT4	WDCNT3	WDCNT2	WDCNT1	WDCNT0	00h



Table 8. Special Function	n Register	Cross	Reference
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SFR	ADDRESS	FUNCTIONS	CPU	INTERRUPTS	PORTS	SERIAL COMM.	POWER AND CLOCKS	TIMER COUNTERS	PWM	FLASH MEMORY	ADC
P0	80h	Port 0			X						
SP	81h	Stack Pointer	Х								
DPL0	82h	Data Pointer Low 0	X								
DPH0	83h	Data Pointer High 0	X								+
DPL1	84h	Data Pointer Low 1	Х								
DPH1	85h	Data Pointer High 1	X								+
DPS	86h	Data Pointer Select	X								
PCON	87h	Power Control					Х				+
TCON	88h	Timer/Counter Control				Х		Х			+
TMOD	89h	Timer Mode Control				X		X			
TL0	8Ah	Timer0 LSB				~		X			
TL1	8Bh	Timer1 LSB						X			
THO	8Ch	Timer0 MSB						X			+
TH1	8Dh	Timer1 MSB						X			
CKCON	8Eh	Clock Control				Х	Х	X			
MWS	8Fh	Memory Write Select				^	^	^		X	+
P1	90h	Port 1			Х					^	
EXIF	9011 91h			Х	^						
MPAGE	91h 92h	External Interrupt Flag	X	^							
CADDR		Memory Page	^							X	
CDATA	93h	Configuration Address									<u> </u>
MCON	94h	Configuration Data	V							Х	<u> </u>
	95h	Memory Control	X								<u> </u>
SCON0	98h	Serial Port 0 Control				X		Х			
SBUF0	99h	Serial Data Buffer 0				Х					<u> </u>
SPICON	9Ah	SPI Control				X					
I2CCON		I <sup>2</sup> C Control				Х					<u> </u>
SPIDATA	9Bh	SPI Data				Х					<u> </u>
I2CDATA		I <sup>2</sup> C Data				Х					<u> </u>
SPITCON	9Dh	SPI Transmit Control				Х					<u> </u>
I2CSTAT		I <sup>2</sup> C Status				Х					<u> </u>
P2	A0h	Port 2			Х						<u> </u>
PWMCON	A1h	PWM Control					Х		Х		<u> </u>
PWMLOW	A2h	PWM Low Byte							Х		<u> </u>
TONELOW		Tone Low Byte							Х		
PWMHI	A3h	PWM HIgh Byte							Х		
TONEHI		Tone Low Byte							Х		
PAI	A5h	Pending Auxiliary Interrupt		Х	Х	Х	Х	Х			Х
AIE	A6h	Auxiliary Interrupt Enable		Х	Х	Х	Х	Х			Х
AISTAT	A7h	Auxiliary Interrupt Status		Х	Х	Х	Х	Х			Х
IE	A8h	Interrupt Enable		Х							
BPCON	A9h	Breakpoint Control	Х						Х		
BPL	AAh	Breakpoint Low Address	Х						Х		
BPH	ABh	Breakpoint High Address	Х						Х		
P0DDRL	ACh	Port 0 Data Direction Low			Х						
PODDRH	ADh	Port 0 Data Direction High			Х						
P1DDRL	AEh	Port 1 Data Direction Low			Х						
P1DDRH	AFh	Port 1 Data Direction High			Х						
P3	B0h	Port 3			Х						



050						SERIAL	POWER AND	TIMER		FLASH	
SFR	ADDRESS	FUNCTIONS	CPU	INTERRUPTS	PORTS	COMM.	CLOCKS	COUNTERS	PWM	MEMORY	ADC
P2DDRL	B1h	Port 2 Data Direction Low	_		Х						
P2DDRH	B2h	Port 2 Data Direction High			Х						<u> </u>
P3DDRL	B3h	Port 3 Data Direction Low			Х						
P3DDRH	B4h	Port 3 Data Direction High			Х						
IP	B8h	Interrupt Priority		Х							
SCON1	C0h	Serial Port 1 Control				Х		Х			
SBUF1	C1h	Serial Data Buffer 1				Х					
EWU	C6h	Enable Wake Up		Х			Х				
T2CON	C8h	Timer 2 Control		Х				Х			
RCAP2L	CAh	Timer 2 Capture LSB		х				Х			
RCAP2H	CBh	Timer 2 Capture MSB		Х				Х			
TL2	CCh	Timer 2 LSB						Х			
TH2	CDh	Timer 2 MSB						Х			
PSW	D0h	Program Status Word	Х								
OCL	D1h	ADC Offset Calibration Low Byte									Х
OCM	D2h	ADC Offset Calibration Mid Byte									Х
OCH	D3h	ADC Offset Calibration High Byte									Х
GCL	D4h	ADC Gain Calibration Low Byte									Х
GCM	D5h	ADC Gain Calibration Mid Byte									Х
GCH	D6h	ADC Gain Calibration High Byte									Х
ADMUX	D7h	ADC Input Multiplexer									Х
EICON	D8h	Enable Interrupt Control		х		Х	Х				Х
ADRESL	D9h	ADC Results Low Byte									Х
ADRESM	DAh	ADC Results Middle Byte									Х
ADRESH	DBh	ADC Results High Byte									Х
ADCON0	DCh	ADC Control 0									Х
ADCON1	DDh	ADC Control 1									X
ADCON2	DEh	ADC Control 2									X
ADCON3	DFh	ADC Control 3									X
ACC	E0h	Accumulator	X								~
SSCON	E1h	Summation/Shifter Control	X								Х
SUMR0	E2h	Summation 0	X								X
SUMR1	E3h	Summation 1	X								X
SUMR2	E4h	Summation 2	X								X
SUMR3	E5h	Summation 3	X								X
ODAC	E6h	Offset DAC	^								X
							v				^
LVDCON	E7h	Low Voltage Detect Control		v			Х				<u> </u>
EIE	E8h	Extended Interrupt Enable		Х							<u> </u>
HWPC0	E9h	Hardware Product Code 0	X								<u> </u>
HWPC1	EAh	Hardware Product Code 1	X								<u> </u>
HWVER	EBh	Hardware Version	Х								<u> </u>
FMCON	EEh	Flash Memory Control	_							X	<u> </u>
FTCON	EFh	Flash Memory Timing Control								Х	

## Table 8. Special Function Register Cross Reference (continued)



## Table 8. Special Function Register Cross Reference (continued)

SFR	ADDRESS	FUNCTIONS	CPU	INTERRUPTS	PORTS	SERIAL COMM.	POWER AND CLOCKS	TIMER COUNTERS	PWM	FLASH MEMORY	ADC
В	F0h	Second Accumulator	Х								
PDCON	F1h	Power Down Control				Х	Х	Х			Х
PASEL	F2h	PSEN/ALE Select			Х		Х				
ACLK	F6h	Analog Clock					Х				Х
SRST	F7h	System Reset	Х				Х				
EIP	F8h	Extended Interrupt Priority		Х							
SECINT	F9h	Seconds Timer Interrupt		Х			Х				
MSINT	FAh	Milliseconds Timer Interrupt		Х			Х				
USEC	FBh	One Microsecond TImer					Х		Х	Х	
MSECL	FCh	One Millisecond TImer Low Byte					Х			Х	
MSECH	FDh	One Millisecond Timer High Byte					Х			Х	
HMSEC	FEh	One Hundred Millisecond TImer					Х				
WDTCON	FFh	Watchdog Timer	Х				Х				
HCR0	3Fh	Hardware Configuration Reg. 0								Х	
HCR1	3Eh	Hardware Configuration Reg. 1					Х				



#### Port 0 (P0)

	7	6	5	4	3	2	1	0	Reset Value
SFR 80h	P0.7	P0.6	P0.5	P0.4	P0.3	P0.2	P0.1	P0.0	FFh

P0.7-0 Port 0. This port functions as a multiplexed address/data bus during external memory access, and as a generalbits 7-0 purpose I/O port when external memory access is not needed. During external memory cycles, this port will contain the LSB of the address when ALE is HIGH, and Data when ALE is LOW. When used as a general-purpose I/O, this port drive is selected by P0DDRL and P0DDRH (ACh, ADh). Whether Port 0 is used as general-purpose I/O or for external memory access is determined by the Flash Configuration Register (HCR1.1)

#### Stack Pointer (SP)

	7	6	5	4	3	2	1	0	Reset Value
SFR 81h	SP.7	SP.6	SP.5	SP.4	SP.3	SP.2	SP.1	SP.0	07h

**SP.7–0** Stack Pointer. The stack pointer identifies the location where the stack will begin. The stack pointer is incremented bits 7–0 before every PUSH or CALL operation and decremented after each POP or RET/RETI. This register defaults to 07h after reset.

#### Data Pointer Low 0 (DPL0)

	7	6	5	4	3	2	1	0	Reset Value
SFR 82h	DPL0.7	DPL0.6	DPL0.5	DPL0.4	DPL0.3	DPL0.2	DPL0.1	DPL0.0	00h

**DPL0.7–0 Data Pointer Low 0.** This register is the low byte of the standard 8051 16-bit data pointer. DPL0 and DPH0 are used to point to non-scratchpad data RAM. The current data pointer is selected by DPS (SFR 86h).

#### Data Pointer High 0 (DPH0)

	7	6	5	4	3	2	1	0	Reset Value
SFR 83h	DPH0.7	DPH0.6	DPH0.5	DPH0.4	DPH0.3	DPH0.2	DPH0.1	DPH0.0	00h

**DPH0.7–0 Data Pointer High 0.** This register is the high byte of the standard 8051 16-bit data pointer. DPL0 and DPH0 are used to point to non-scratchpad data RAM. The current data pointer is selected by DPS (SFR 86h).

#### Data Pointer Low 1 (DPL1)

	7	6	5	4	3	2	1	0	Reset Value
SFR 84h	DPL1.7	DPL1.6	DPL1.5	DPL1.4	DPL1.3	DPL1.2	DPL1.1	DPL1.0	00h

**DPL1.7-0 Data Pointer Low 1.** This register is the low byte of the auxiliary 16-bit data pointer. When the SEL bit (DPS.0, bits 7–0 SFR 86h) is set, DPL1 and DPH1 are used in place of DPL0 and DPH0 during DPTR operations.



#### Data Pointer High 1 (DPH1)

	7	6	5	4	3	2	1	0	Reset Value
SFR 85h	DPH1.7	DPH1.6	DPH1.5	DPH1.4	DPH1.3	DPH1.2	DPH1.1	DPH1.0	00h

**DPH1.7–0 Data Pointer High.** This register is the high byte of the auxiliary 16-bit data pointer. When the SEL bit (DPS.0, SFR 86h) is set, DPL1 and DPH1 are used in place of DPL0 and DPH0 during DPTR operations.

#### Data Pointer Select (DPS)

	7	6	5	4	3	2	1	0	Reset Value
SFR 86h	0	0	0	0	0	0	0	SEL	00h

SEL Data Pointer Select. This bit selects the active data pointer.

0: Instructions that use the DPTR will use DPL0 and DPH0.

1: Instructions that use the DPTR will use DPL1 and DPH1.

#### **Power Control (PCON)**

bit 0

	7	6	5	4	3	2	1	0	Reset Value					
SFR 87h	SMOD	0	1	1	GF1	GF0	STOP	IDLE	30h					
SMOD bit 7	<b>Serial Port (</b> 0: Serial Por 1: Serial Por	t 0 baud rate	will be a sta	ndard baud	rate.	Ũ		erial Port 0. when using Ti	mer 1.					
<b>GF1</b> bit 3	General-Pur	General-Purpose User Flag 1. This is a general-purpose flag for software control.												
<b>GF0</b> bit 2	General-Pur	rpose User I	<b>Flag 0.</b> This	is a general-	purpose flag	for software	control.							
STOP bit 1	•	<b>Stop Mode Select.</b> Setting this bit will halt the oscillator and block external clocks. This bit will always read as a 0. All DACs and digital pins keep their respective output values. Exit with RESET.												
IDLE bit 0	<b>Idle Mode Select.</b> Setting this bit will freeze the CPU, Timer 0, 1, and 2, and the USARTs; other peripherals remain active. All DACs and digital pins keep their respective output values. This bit will always be read as a 0. Exit with AI (A6h) and EWU (C6h) interrupts. The internal reference remains unchanged.													



## Timer/Counter Control (TCON)

	7	6	5	4	3	2	1	0	Reset Value
SFR 88h	TF1	TR1	TF0	TR0	IE1	IT1	IE0	IT0	00h
TF1 bit 7	This bit can t routine. 0: No Timer	be cleared by	software and as been dete	d is automatio					current mode. errupt service
TR1 bit 6	<b>Timer 1 Run</b> in TH1, TL1. 0: Timer is h 1: Timer is e	alted.	s bit enables/	disables the o	operation of T	imer 1. Halting	g this timer w	ill preserve the	e current count
<b>TF0</b> bit 5	This bit can t routine. 0: No Timer		software and as been dete	d is automatio					current mode. errupt service
<b>TR0</b> bit 4	<b>Timer 0 Run</b> count in THC 0: Timer is h 1: Timer is e	), TL0. alted.	is bit enable	s/disables th	e operation c	of Timer 0. Ha	alting this tim	ner will preserv	ve the current
IE1 bit 3		et until cleare	ed in so <u>ftwa</u> r	e or the star					Γ1 = 1, this bit 0, this bit will
IT1 bit 2	0: INT1 is lev			ects whether	the INT1 pin	will detect e	dge or level	triggered inte	rrupts.
IE0 bit 3		et until cleare	ed in softwar	e or the star					Γ0 = 1, this bit 0, this bit will
<b>IT0</b> bit 2	0: INTO is lev			cts whether	the INT0 pin	will detect e	dge or level	triggered inter	rrupts.



#### Timer Mode Control (TMOD)

	7	6	5	4	3	3 2 1 0				
		TIME	ER 1		TIMER 0				Reset Value	
SFR 89h	GATE	C/T	M1	MO	GATE	C/T	M1	MO	00h	

#### GATE Timer 1 Gate Control. This bit enables/disables the ability of Timer 1 to increment.

0: Timer 1 will clock when TR1 = 1, regardless of the state of pin  $\overline{INT1}$ .

1: Timer 1 will clock only when TR1 = 1 and pin  $\overline{INT1}$  = 1.

#### C/T Timer 1 Counter/Timer Select.

bit 6

bit 7

0: Timer is incremented by internal clocks.

1: Timer is incremented by pulses on T1 pin when TR1 (TCON.6, SFR 88h) is 1.

#### M1, M0 Timer 1 Mode Select. These bits select the operating mode of Timer 1.

bits 5-4

M1	MO	MODE
0	0	Mode 0: 8-bit counter with 5-bit prescale.
0	1	Mode 1: 16 bits.
1	0	Mode 2: 8-bit counter with auto reload.
1	1	Mode 3: Timer 1 is halted, but holds its count.

GATE Timer 0 Gate Control. This bit enables/disables the ability of Timer 0 t	0 to increment.
---	-----------------

- bit 3 0: Timer 0 will clock when TR0 = 1, regardless of the state of pin INT0 (software control).
  - 1: Timer 0 will clock only when TR0 = 1 and pin  $\overline{INT0}$  = 1 (hardware control).

#### C/T Timer 0 Counter/Timer Select.

bit 2 0: Timer is incremented by internal clocks.

1: Timer is incremented by pulses on pin T0 when TR0 (TCON.4, SFR 88h) is 1.

M1, M0 Timer 0 Mode Select. These bits select the operating mode of Timer 0.

#### bits 1–0

M1	MO	MODE
0	0	Mode 0: 8-bit counter with 5-bit prescale.
0	1	Mode 1: 16 bits.
1	0	Mode 2: 8-bit counter with auto reload.
1	1	Mode 3: Two 8-bit counters.

#### Timer 0 LSB (TL0)

	7	6	5	4	3	2	1	0	Reset Value
SFR 8Ah	TL0.7	TL0.6	TL0.5	TL0.4	TL0.3	TL0.2	TL0.1	TL0.0	00h

## **TL0.7–0 Timer 0 LSB.** This register contains the least significant byte of Timer 0.

bits 7-0

#### Timer 1 LSB (TL1)

	7	6	5	4	3	2	1	0	Reset Value
SFR 8Bh	TL1.7	TL1.6	TL1.5	TL1.4	TL1.3	TL1.2	TL1.1	TL1.0	00h

**TL1.7–0 Timer 1 LSB.** This register contains the least significant byte of Timer 1.

bits 7–0



#### Timer 0 MSB (TH0)

	7	6	5	4	3	2	1	0	Reset Value
SFR 8Ch	TH0.7	TH0.6	TH0.5	TH0.4	TH0.3	TH0.2	TH0.1	TH0.0	00h

**TH0.7–0 Timer 0 MSB.** This register contains the most significant byte of Timer 0.

bits 7-0

#### Timer 1 MSB (TH1)

	7	6	5	4	3	2	1	0	Reset Value
SFR 8Dh	TH1.7	TH1.6	TH1.5	TH1.4	TH1.3	TH1.2	TH1.1	TH1.0	00h

**TH1.7–0 Timer 1 MSB.** This register contains the most significant byte of Timer 1.

bits 7-0

#### **Clock Control (CKCON)**

		7	6	5	4	3	2	1	0	Reset Value
SFR	R 8Eh	0	0	T2M	T1M	TOM	MD2	MD1	MD0	01h

T2MTimer 2 Clock Select. This bit controls the division of the system clock that drives Timer 2. This bit has no effect when<br/>the timer is in baud rate generator or clock output mode. Clearing this bit to 0 maintains 8051 compatibility. This bit<br/>has no effect on instruction cycle timing.

0: Timer 2 uses a divide-by-12 of the crystal frequency.

1: Timer 2 uses a divide-by-4 of the crystal frequency.

- **T1M Timer 1 Clock Select.** This bit controls the division of the system clock that drives Timer 1. Clearing this bit to 0 maintains 8051 compatibility. This bit has no effect on instruction cycle timing.
  - 0: Timer 1 uses a divide-by-12 of the crystal frequency.

1: Timer 1 uses a divide-by-4 of the crystal frequency.

- **TOM** Timer 0 Clock Select. This bit controls the division of the system clock that drives Timer 0. Clearing this bit to 0 maintains 8051 compatibility. This bit has no effect on instruction cycle timing.
  - 0: Timer 0 uses a divide-by-12 of the crystal frequency.

1: Timer 0 uses a divide-by-4 of the crystal frequency.

MD2, MD1, MD0 bits 2–0 Stretch MOVX Select 2–0. These bits select the time by which external MOVX cycles are to be stretched. This allows slower memory or peripherals to be accessed without using ports or manual software intervention. The width of the RD or WR strobe will be stretched by the specified interval, which will be transparent to the software except for the increased time to execute the MOVX instruction. All internal MOVX instructions on devices containing MOVX SRAM are performed at the 2 instruction cycle rate.

MD2	MD1	MD0	STRETCH VALUE	MOVX DURATION	RD or WR STROBE WIDTH (SYS CLKs)	RD or WR STROBE WIDTH (μs) at 12MHz
0	0	0	0	2 Instruction Cycles	2	0.167
0	0	1	1	3 Instruction Cycles (default) <sup>(1)</sup>	4	0.333
0	1	0	2	4 Instruction Cycles	8	0.667
0	1	1	3	5 Instruction Cycles	12	1.000
1	0	0	4	6 Instruction Cycles	16	1.333
1	0	1	5	7 Instruction Cycles	20	1.667
1	1	0	6	8 Instruction Cycles	24	2.000
1	1	1	7	9 Instruction Cycles	28	2.333

(1) For applications without external memory, no extra cycle is needed. To increase speed, set MD2, MD1, and MD0 to '000'.



#### Memory Write Select (MWS)

		7	6	5	4	3	2	1	0	Reset Value
ſ	SFR 8Fh	0	0	0	0	0	0	0	MXWS	00h

#### MXWS MOVX Write Select. This allows writing to the internal Flash program memory.

0: No writes are allowed to the internal Flash program memory.

1: Writing is allowed to the internal Flash program memory, unless PML (HCR0) or RSL (HCR0) are on.

#### Port 1 (P1)

bit 0

	7	6	5	4	3	2	1	0	Reset Value
SFR 90h	P1.7 INT5/SCK	P1.6 INT4/MISO	P1.5 INT3/MOSI	P1.4 INT2/SS	P1.3 TXD1	P1.2 RXD1	P1.1 T2EX	P1.0 T2	FFh

**P1.7-0 General-Purpose I/O Port 1.** This register functions as a general-purpose I/O port. In addition, all the pins have an alternative function listed below. Each of the functions is controlled by several other SFRs. The associated Port 1 latch bit must contain a logic '1' before the pin can be used in its alternate function capacity. To use the alternate function, set the appropriate mode in P1DDRL (SFR AEh), P1DDRH (SFR AFh).

**INT5/SCK** External Interrupt 5. A falling edge on this pin will cause an external interrupt 5 if enabled.

- bit 7 SPI Clock. The master clock for SPI data transfers.
- **INT4/MISO** External Interrupt 4. A rising edge on this pin will cause an external interrupt 4 if enabled.
- bit 6 Master In Slave Out. For SPI data transfers, this pin receives data for the master and transmits data from the slave.
- **INT3/MOSI** External Interrupt 3. A falling edge on this pin will cause an external interrupt 3 if enabled. bit 5 Master Out Slave In. For SPI data transfers, this pin transmits master data and receives slave data.
- **INT2/SS** External Interrupt 2. A rising edge on this pin will cause an external interrupt 2 if enabled.
- bit 4 Slave Select. During SPI operation, this pin provides the select signal for the slave device but does not control the output drive of MISO.
- **TXD1** Serial Port 1 Transmit. This pin transmits the serial Port 1 data in serial port modes 1, 2, 3, and emits the synchrobit 3 nizing clock in serial port mode 0.
- **RXD1** Serial Port 1 Receive. This pin receives the serial Port 1 data in serial port modes 1, 2, 3, and is a bidirectional data transfer pin in serial port mode 0.
- T2EX
   Timer 2 Capture/Reload Trigger. A 1 to 0 transition on this pin will cause the value in the T2 registers to be

   bit 1
   transferred into the capture registers, if enabled by EXEN2 (T2CON.3, SFR C8h). When in auto-reload mode, a 1 to 0 transition on this pin will reload the Timer 2 registers with the value in RCAP2L and RCAP2H if enabled by EXEN2 (T2CON.3, SFR C8h).

   (T2CON.3, SFR C8h).
- T2 Timer 2 External Input. A 1 to 0 transition on this pin will cause Timer 2 to increment.

bit 0

#### **External Interrupt Flag (EXIF)**

	7	6	5	4	3	2	1	0	Reset Value
SFR 91h	IE5	IE4	IE3	IE2	1	0	0	0	08h

IE5 External Interrupt 5 Flag. This bit will be set when a falling edge is detected on INT5. This bit must be cleared manually by software. Setting this bit in software will cause an interrupt if enabled.

**IE4 External Interrupt 4 Flag.** This bit will be set when a rising edge is detected on INT4. This bit must be cleared bit 6 manually by software. Setting this bit in software will cause an interrupt if enabled.

**IE3 External Interrupt 3 Flag.** This bit will be set when a falling edge is detected on INT3. This bit must be cleared bit 5 manually by software. Setting this bit in software will cause an interrupt if enabled.

**IE2 External Interrupt 2 Flag.** This bit will be set when a rising edge is detected on INT2. This bit must be cleared bit 4 manually by software. Setting this bit in software will cause an interrupt if enabled.

#### Memory Page (MPAGE)

	7	6	5	4	3	2	1	0	Reset Value
SFR 92h									00h

 MPAGE
 The 8051 uses Port 2 for the upper 8 bits of the external data memory access by MOVX A, @Ri and MOVX @Ri, A

 bits 7-0
 instructions. The MSC1210 uses register MPAGE instead of Port 2. To access external data memory using the MOVX A, @Ri and MOVX @Ri, A instructions, the user should preload the upper byte of the address into MPAGE (versus preloading into P2 for the standard 8051).

#### Configuration Address Register (CADDR) (write-only)

	7	6	5	4	3	2	1	0	Reset Value
SFR 93h									00h

**CADDR Configuration Address Register.** This register supplies the address for reading bytes in the 128 bytes of Flash bits 7–0 Configuration memory. This is a write-only register.

CAUTION: If this register is written to while executing from Flash Memory, the CDATA register will be incorrect. The faddr\_data\_read routine in the Boot ROM can be used for this purpose.

#### Configuration Data Register (CDATA) (read-only)

	7	6	5	4	3	2	1	0	Reset Value
SFR 94h									00h

**CDATA Configuration Data Register.** This register will contain the data in the 128 bytes of Flash Configuration memory that are located at the last written address in the CADDR register. This is a read-only register.

bit 0



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#### Memory Control (MCON)

	7	6	5	4	3	2	1	0	Reset Value
SFR 95h	BPSEL	0	0	_	—	—	_	RAMMAP	00h

#### BPSEL Breakpoint Address Selection

bit 7 Write: Select one of two Breakpoint registers: 0 or 1.

0: Select breakpoint register 0.

1: Select breakpoint register 1.

Read: Provides the Breakpoint register that created the last interrupt: 0 or 1.

#### RAMMAP Memory Map 1kB extended SRAM.

- 0: Address is: 0000h—03FFh (default) (Data Memory)
  - 1: Address is 8400h—87FFh (Data and Program Memory)

#### Serial Port 0 Control (SCON0)

	7	6	5	4	3	2	1	0	Reset Value
SFR 98h	SM0_0	SM1_0	SM2_0	REN_0	TB8_0	RB8_0	TI_0	RI_0	00h

**SM0–2** Serial Port 0 Mode. These bits control the mode of serial Port 0. Modes 1, 2, and 3 have 1 start and 1 stop bit in addition to the 8 or 9 data bits.

MODE	SM0	SM1	SM2	FUNCTION	LENGTH	PERIOD
0	0	0	0	Synchronous	8 bits	12 p <sub>CLK</sub> <sup>(1)</sup>
0	0	0	1	Synchronous	8 bits	<sup>4</sup> PCLK <sup>(1)</sup>
1(2)	0	1	0	Asynchronous	10 bits	Timer 1 or 2 Baud Rate Equation
1(2)	0	1	1	Valid Stop Required <sup>(3)</sup>	10 bits	Timer 1 Baud Rate Equation
2	1	0	0	Asynchronous	11 bits	$64  \text{p}_{\text{CLK}}^{(1)}  (\text{SMOD} = 0)$
2	1	0	1	Asynchronous with Multiprocessor Communication <sup>(4)</sup>	11 bits	$32 p_{CLK}^{(1)} (SMOD = 1) 64 p_{CLK}^{(1)} (SMOD = 0) 32 p_{CLK}^{(1)} (SMOD = 1)$
3(2)	1	1	0	Asynchronous	11 bits	Timer 1 or 2 Baud Rate Equation
3(2)	1	1	1	Asynchronous with Multiprocessor Communication <sup>(4)</sup>	11 bits	Timer 1 or 2 Baud Rate Equation

(4) For modes 1 and 3, the selection of Timer 1 or 2 for baud rate is specified via the T2CON (C8h) r (3) RI 0 will only be activated when a valid STOP is received.

(4) RI\_0 will only be activated when a valid ST
 (4) RI\_0 will not be activated if bit 9 = 0.

**REN\_0 Receive Enable.** This bit enables/disables the serial Port 0 received shift register.

bit 4 0: Serial Port 0 reception disabled.

1: Serial Port 0 received enabled (modes 1, 2, and 3). Initiate synchronous reception (mode 0).

**TB8\_0** 9th Transmission Bit State. This bit defines the state of the 9th transmission bit in serial Port 0 modes 2 and 3.

- bit 3
- **RB8\_0 9th Received Bit State.** This bit identifies the state of the 9th reception bit of received data in serial Port 0 modes 2 and 3. In serial port mode 1, when SM2\_0 = 0, RB8\_0 is the state of the stop bit. RB8\_0 is not used in mode 0.
- TI\_0Transmitter Interrupt Flag. This bit indicates that data in the serial Port 0 buffer has been completely shifted out. In serial<br/>port mode 0, TI\_0 is set at the end of the 8th data bit. In all other modes, this bit is set at the end of the last data bit.<br/>This bit must be manually cleared by software.
- RI\_0 Receiver Interrupt Flag. This bit indicates that a byte of data has been received in the serial Port 0 buffer. In serial bit 0 port mode 0, RI\_0 is set at the end of the 8th bit. In serial port mode 1, RI\_0 is set after the last sample of the incoming stop bit subject to the state of SM2\_0. In modes 2 and 3, RI\_0 is set after the last sample of RB8\_0. This bit must be manually cleared by software.

#### Serial Data Buffer 0 (SBUF0)

Γ		7	6	5	4	3	2	1	0	Reset Value
Γ	SFR 99h									00h

**SBUF0** Serial Data Buffer 0. Data for Serial Port 0 is read from or written to this location. The serial transmit and receive buffers are separate registers, but both are addressed at this location.

# SPI Control (SPICON). Any change resets the SPI interface, counters, and pointers. PDCON controls which is enabled.

	7	6	5	4	3	2	1	0	Reset Value
SFR 9Ah	SCK2	SCK1	SCK0	0	ORDER	MSTR	CPHA	CPOL	00h

**SCK Selection.** Selection of t<sub>CLK</sub> divider for generation of SCK in Master mode.

bits 7-5

SCK2	SCK1	SCK0	SCK PERIOD
0	0	0	tCLK/2
0	0	1	<sup>t</sup> CLK <sup>/4</sup>
0	1	0	tCLK/8
0	1	1	t <sub>CLK</sub> /16
1	0	0	tCLK/32
1	0	1	tCLK/64
1	1	0	t <sub>CLK</sub> /128
1	1	1	t <sub>CLK</sub> /256

#### ORDER Set Bit Order for Transmit and Receive.

- bit 3 0: Most Significant Bits First
  - 1: Least Significant Bits First

#### MSTR SPI Master Mode.

- bit 2 0: Slave Mode
  - 1: Master Mode

#### CPHA Serial Clock Phase Control.

bit 1 0: Valid data starting from half SCK period before the first edge of SCK 1: Valid data starting from the first edge of SCK

#### CPOL Serial Clock Polarity.

bit 0 0: SCK idle at logic LOW 1: SCK idle at logic HIGH

#### SPI Data Register (SPIDATA)

	7	6	5	4	3	2	1	0	Reset Value
SFR 9Bh									00h

**SPIDATA** SPI Data Register. Data for SPI is read from or written to this location. The SPI transmit and receive buffers are separate registers, but both are addressed at this location. Read to clear the receive interrupt and write to clear the transmit interrupt.



#### SPI Transmit Control Register (SPITCON)

	7	6	5	4	3	2	1	0	Reset Value
SFR 9Dh			CLK_EN	DRV_DLY	DRV_EN				00h

#### CLK\_EN SCK Driver Enable.

- bit 5 0: Disable SCK Driver (Master Mode) 1: Enable SCK Driver (Master Mode)
- DRV\_DLY
   Drive Delay. (Refer to DRV\_EN bit)

   bit 4
   0: Drive output immediately

   1: Drive output after current byte transfer

#### DRV\_EN Drive Enable.

bit 3

DRV_DLY	DRV_EN	MOSI or MISO OUTPUT CONTROL
0	0	Tristate immediately
0	1	Drive immediately
1	0	Tristate after the current byte transfer
1	1	Drive after the current byte transfer

#### Port 2 (P2)

	7	6	5	4	3	2	1	0	Reset Value
SFR A0h									FFh

 P2
 Port 2. This port functions as an address bus during external memory access, and as a general-purpose I/O port.

 bits 7-0
 During external memory cycles, this port will contain the MSB of the address. Whether Port 2 is used as general-purpose I/O or for external memory access is determined by the Flash Configuration Register (HCR1.0).

#### **PWM Control (PWMCON)**

	7	6	5	4	3	2	1	0	Reset Value
SFR A1h	—	—	PPOL	PWMSEL	SPDSEL	TPCNTL2	TPCNTL1	TPCNTL0	00h

**PPOL Period Polarity.** Specifies the starting level of the PWM pulse.

bit 5 0: ON Period. PWM Duty register programs the ON period.

1: OFF Period. PWM Duty register programs the OFF period.

PWMSEL PWM Register Select. Select which 16-bit register is accessed by PWMLOW/PWMHIGH.

- bit 4 0: Period (must be 0 for TONE mode) 1: Duty
- SPDSEL Speed Select. bit 3 0: 1MHz (the USEC Clock) 1: SYSCLK

\_ \_ \_

#### TPCNTL Tone Generator/Pulse Width Modulation Control.

bits 2–0

TPCNTL.2	TPCNTL.1	TPCNTL.0	MODE
0	0	0	Disable (default)
0	0	1	PWM
0	1	1	TONE—Square
1	1	1	TONE—Staircase

#### Tone Low (TONELOW)/PWM Low (PWMLOW)

	7	6	5	4	3	2	1	0	Reset Value
SFR A2h	TDIV7 PWM7	TDIV6 PWM6	TDIV5 PWM5	TDIV4 PWM4	TDIV3 PWM3	TDIV2 PWM2	TDIV1 PWM1	TDIV0 PWM0	00h

**TDIV7-0 Tone Divisor.** The low order bits that define the half-time period. For staircase mode the output is high impedance for the last 1/4 of this period.

**PWMLOW** Pulse Width Modulator Low Bits. These 8 bits are the least significant 8 bits of the PWM register. bits 7–0

#### Tone High (TONEHI)/PWM High (PWMHI)

	7	6	5	4	3	2	1	0	Reset Value
SFR A3h	TDIV15 PWM15	TDIV14 PWM14	TDIV13 PWM13	TDIV12 PWM12	TDIV11 PWM11	TDIV10 PWM10	TDIV9 PWM9	TDIV8 PWM8	00h

**TDIV15-8 Tone Divisor.** The high order bits that define the half time period. For staircase mode the output is high impedance for the last 1/4 of this period.

**PWMHI** Pulse Width Modulator High Bits. These 8 bits are the high order bits of the PWM register. bits 7–0

#### Pending Auxiliary Interrupt (PAI)

	7	6	5	4	3	2	1	0	Reset Value
SFR A5h	—		—		PAI3	PAI2	PAI1	PAI0	00h

PAI Pending Auxiliary Interrupt Register. The results of this register can be used as an index to vector to the appropriate interrupt routine. All of these interrupts vector through address 0033h.

PAI3	PAI2	PAI1	PAI0	AUXILIARY INTERRUPT STATUS
0	0	0	0	No Pending Auxiliary IRQ
0	0	0	1	Digital Low Voltage IRQ Pending
0	0	1	0	Analog Low Voltage IRQ Pending
0	0	1	1	SPI Receive IRQ Pending.
0	1	0	0	SPI Transmit IRQ Pending.
0	1	0	1	One Millisecond System Timer IRQ Pending.
0	1	1	0	Analog-to-Digital Conversion IRQ Pending.
0	1	1	1	Accumulator IRQ Pending.
1	0	0	0	One Second System Timer IRQ Pending.



#### Auxiliary Interrupt Enable (AIE)

	7	6	5	4	3	2	1	0	Reset Value
SFR A6h	ESEC	ESUM	EADC	EMSEC	ESPIT	ESPIR	EALV	EDLVB	00h

Interrupts are enabled by EICON.4 (SFR D8H). The other interrupts are controlled by the IE and EIE registers.

# ESECEnable Seconds Timer Interrupt (lowest priority auxiliary interrupt).bit 7Write: Set mask bit for this interrupt 0 = masked, 1 = enabled.<br/>Read: Current value of Seconds Timer Interrupt before masking.ESUMEnable Summation Interrupt.<br/>Write: Set mask bit for this interrupt 0 = masked, 1 = enabled.<br/>Read: Current value of Summation Interrupt before masking.

#### EADC Enable ADC Interrupt.

bit 5 Write: Set mask bit for this interrupt 0 = masked, 1 = enabled. Read: Current value of **ADC Interrupt** before masking.

#### EMSEC Enable Millisecond System Timer Interrupt.

bit 4 Write: Set mask bit for this interrupt 0 = masked, 1 = enabled. Read: Current value of **Millisecond System Timer Interrupt** before masking.

#### ESPIT Enable SPI Transmit Interrupt.

bit 3 Write: Set mask bit for this interrupt 0 = masked, 1 = enabled. Read: Current value of **SPI Transmit Interrupt** before masking.

#### ESPIR Enable SPI Receive Interrupt.

bit 2 Write: Set mask bit for this interrupt 0 = masked, 1 = enabled. Read: Current value of **SPI Receive Interrupt** before masking.

#### EALV Enable Analog Low Voltage Interrupt.

- bit 1 Write: Set mask bit for this interrupt 0 = masked, 1 = enabled. Read: Current value of **Analog Low Voltage Interrupt** before masking.
- EDLVB Enable Digital Low Voltage or Breakpoint Interrupt (highest priority auxiliary interrupt).
- bit 0 Write: Set mask bit for this interrupt 0 = masked, 1 = enabled. Read: Current value of **Digital Low Voltage** or **Breakpoint Interrupt** before masking.

## Auxiliary Interrupt Status Register (AISTAT)

**J**ij

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	7	6	5	4	3	2	1	0	Reset Value	
SFR A7h	SEC	SUM	ADC	MSEC	SPIT	SPIR	ALVD	DLVD	00h	
EC	Second Sys	tem Timer Inte	errupt Status	s Flag (lowe	st priority	AI).				
it 7	Second System Timer Interrupt Status Flag (lowest priority AI). 0: SEC interrupt inactive or masked.									
	1: SEC Interrupt active. (It is set inactive by reading the SECINT register.)									
UM	Summation	Summation Register Interrupt Status Flag.								
it 6		D: SUM interrupt inactive or masked.								
		rupt active. (It i		by reading	the lowest b	yte of the S	ummation re	egister.)		
DC	ADC Interru	pt Status Flag	I.							
it 5	0: ADC interro	upt inactive or r	masked (If act	tive, it is set i	nactive by re	ading the lov	west byte of	the Data Ou	tput Registe	
	1: ADC interr	<ul><li>0: ADC interrupt inactive or masked (If active, it is set inactive by reading the lowest byte of the Data Output Register).</li><li>1: ADC interrupt active. (If active, no new data will be written to the Data Output Register.)</li></ul>								
ISEC	Millisecond	Millisecond System Timer Interrupt Status Flag.								
it 4		Millisecond System Timer Interrupt Status Flag. 0: MSEC interrupt inactive or masked.								
				ve by reading	n the MSINT	register)				
	1: MSEC interrupt active. (It is set inactive by reading the MSINT register.)									
PIT	SPI Transmit Interrupt Status Flag.									
it 3		nit interrupt ina			unition of a state of					
	1: SPI transm	nit interrupt act	ive. (It is set	inactive by w	riting to the	SPIDATA re	egister.)			
PIR	SPI Receive	Interrupt Stat	tus Flag.							
it 2	0: SPI receive	e interrupt inac	ctive or mask	ed.						
	1: SPI receive	e interrupt activ	ve. (It is set i	nactive by re	ading from	the SPIDAT/	A register.)			
LVD	Analog Low	Voltage Deter	ct Interrupt S	Status Flag.						
it 1	0: ALVD inter	rupt inactive o	r masked.							
	1: ALVD inter	rupt active. (In	terrupt stays	active until 1	he AV <sub>DD</sub> vol	tage exceed	ls the thresh	old.)		
DLVD	-	Voltage Detec	-	oint Interrup	t Status Fla	ag (highest	priority AI)			
it O		rrupt inactive o								
	1: DLVD inter cleared.)	rrupt active. (Ir	nterrupt stays	active until	the DV <sub>DD</sub> vo	ltage excee	eds the three	shold or the	Breakpoint	

#### Interrupt Enable (IE)

	7	6	5	4	3	2	1	0	Reset Value	
SFR A8h	EA	ES1	ET2	ES0	ET1	EX1	ET0	EX0	00h	
EA	Global Inter	rupt Enable. Th	nis bit contro	ls the globa	l masking of	all interrupt	s except the	use in AIE (S	SFR A6h).	
oit 7		<ul> <li>Global Interrupt Enable. This bit controls the global masking of all interrupts except those in AIE (SFR A6h).</li> <li>Disable interrupt sources. This bit overrides individual interrupt mask settings for this register.</li> <li>Enable all individual interrupt masks. Individual interrupts in this register will occur if enabled.</li> </ul>								
		•					0	•		
ES1	Enable Seria	Enable Serial Port 1 Interrupt. This bit controls the masking of the serial Port 1 interrupt.								
bit 6		Disable all serial Port 1 Interrupt. This bit controls the masking of the serial Port 1 Interrupt.								
		errupt requests	•	y the RI_1 (	SCON1.0, S	SFR C0h) or	TI_1 (SCO	N1.1, SFR (	C0h) flags.	
ET2	Enable Time	er 2 Interrupt. T	This bit cont	rols the ma	sking of th	e Timer 2 ir	nterrupt.			
bit 5		I Timer 2 interru			J					
		: Enable interrupt requests generated by the TF2 flag (T2CON.7, SFR C8h).								
ES0	Enable Seria	Enable Serial port 0 interrupt. This bit controls the masking of the serial Port 0 interrupt.								
bit 4	0: Disable all serial Port 0 interrupts. 1: Enable interrupt requests generated by the RI_0 (SCON0.0, SFR 98h) or TI_0 (SCON0.1, SFR 98h) flags.									
	1: Enable inte	errupt requests	generated b	y the RI_0 (	SCON0.0, S	SFR 98h) or	TI_0 (SCON	N0.1, SFR 9	8h) flags.	
ET1	Enable Time	er 1 Interrupt. T	This bit cont	rols the ma	sking of th	e Timer 1 ir	nterrupt.			
bit 3	0: Disable Ti	mer 1 interrupt.								
	1: Enable inte	errupt requests	generated b	y the TF1 fl	ag (TCON.7	, SFR 88h).				
EX1	Enable Exte	rnal Interrupt 1	I. This bit co	ontrols the	masking of	external in	terrupt 1.			
bit 2	0: Disable ex	ternal interrupt	1.							
	1: Enable inter	1: Enable interrupt requests generated by the INT1 pin.								
ET0	Enable Time	Enable Timer 0 Interrupt. This bit controls the masking of the Timer 0 interrupt.								
bit 1	0: Disable all	0: Disable all Timer 0 interrupts.								
	1: Enable inte	1: Enable interrupt requests generated by the TF0 flag (TCON.5, SFR 88h).								
EX0		rnal Interrupt (		ontrols the	masking of	external in	terrupt 0.			
bit 0		ternal interrupt								
	1: Enable inte	errupt requests	generated b	y the INT0	oin.					

#### **Breakpoint Control (BPCON)**

	7	6	5	4	3	2	1	0	Reset Value
SFR A9h	BP	0	0	0	0	0	PMSEL	EBP	00h

Writing to register sets the breakpoint condition specified by MCON, BPL, and BPH.

- BP Breakpoint Interrupt. This bit indicates that a break condition has been recognized by a hardware breakpoint register(s).
   bit 7 Read: Status of Breakpoint Interrupt. Will indicate a breakpoint match for any of the breakpoint registers.
  - Write: 0: No effect.

1: Clear Breakpoint 1 for breakpoint register selected by MCON (SFR 95h).

 
 PMSEL
 Program Memory Select. Write this bit to select memory for address breakpoints of register selected in MCON (SFR 95h).

0: Break on address in data memory.

1: Break on address in program memory.

**EBP Enable Breakpoint.** This bit enables this breakpoint register. Address of breakpoint register selected by

bit 0 MCON (SFR 95h).

0: Breakpoint disabled.

1: Breakpoint enabled.

#### Breakpoint Low (BPL) Address for BP Register Selected in MCON (95h)

	7	6	5	4	3	2	1	0	Reset Value
SFR AAh	BPL.7	BPL.6	BPL.5	BPL.4	BPL.3	BPL.2	BPL.1	BPL.0	00h

BPL.7–0 Breakpoint Low Address. The low 8 bits of the 16-bit breakpoint address.

bits 7–0

#### Breakpoint High Address (BPH) Address for BP Register Selected in MCON (95h)

	7	6	5	4	3	2	1	0	Reset Value
SFR ABh	BPH.7	BPH.6	BPH.5	BPH.4	BPH.3	BPH.2	BPH.1	BPH.0	00h

**BPH.7–0** Breakpoint High Address. The high 8 bits of the 16-bit breakpoint address.

bits 7–0



#### Port 0 Data Direction Low Register (P0DDRL)

	7	6	5	4	3	2	1	0	Reset Value
SFR ACh	P03H	P03L	P02H	P02L	P01H	P01L	P00H	P00L	00h

#### P0.3 Port 0 Bit 3 Control.

bits 7-6

P03H	P03L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

#### P0.2 Port 0 Bit 2 Control.

bits 5-4

P02H	P02L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

#### P0.1 Port 0 Bit 1 Control.

bits 3-2

P01H	P01L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

#### P0.0 Port 0 Bit 0 Control.

bits 1-0

P00H	P00L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

NOTE: Port 0 also controlled by EA and Memory Access Control HCR1.1.



#### Port 0 Data Direction High Register (P0DDRH)

	7	6	5	4	3	2	1	0	Reset Value
SFR ADh	P07H	P07L	P06H	P06L	P05H	P05L	P04H	P04L	00h

#### P0.7 Port 0 Bit 7 Control.

bits 7-6

P07H	P07L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

#### P0.6 Port 0 Bit 6 Control.

bits 5-4

F	P06H	P06L	
	0	0	Standard 8051 (Pull-Up)
	0	1	CMOS Output
	1	0	Open Drain Output
	1	1	Input

#### P0.5 Port 0 Bit 5 Control.

bits 3-2

P05H	P05L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

#### P0.4 Port 0 Bit 4 Control.

bits 1-0

P04H	P04L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

NOTE: Port 0 also controlled by EA and Memory Access Control HCR1.1.



#### Port 1 Data Direction Low Register (P1DDRL)

	7	6	5	4	3	2	1	0	Reset Value
SFR AEh	P13H	P13L	P12H	P12L	P11H	P11L	P10H	P10L	00h

#### P1.3 Port 1 Bit 3 Control.

bits 7-6

P13H	P13L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

#### P1.2 Port 1 Bit 2 Control.

bits 5-4

P12H	P12L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

#### P1.1 Port 1 Bit 1 Control.

bits 3–2

P11H	P11L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

Standard 8051 (Pull-Up) CMOS Output

Open Drain Output

Input

#### P1.0 Port 1 Bit 0 Control.

P10H

0

0 1

1

bits 1-0

P10L

0

1

0

1



## Port 1 Data Direction High Register (P1DDRH)

	7	6	5	4	3	2	1	0	Reset Value
SFR AFh	P17H	P17L	P16H	P16L	P15H	P15L	P14H	P14L	00h

#### P1.7 Port 1 Bit 7 Control.

bits 7-6

P17H	P17L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

#### P1.6 Port 1 Bit 6 Control.

bits 5-4

P16H	P16L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

#### P1.5 Port 1 Bit 5 Control.

bits 3-2

P15H	P15L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

#### P1.4 Port 1 Bit 4 Control.

bits 1-0

P14H	P14L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

#### Port 3 (P3)

bit 5

bit 3

bit 2

	7	6	5	4	3	2	1	0	Reset Value
SFR B0h	P3.7 RD	P3.6 WR	P3.5 T1	P3.4 T0	P3.3 INT1	P3.2 INT0	P3.1 TXD0	P3.0 RXD0	FFh

**P3.7–0 General-Purpose I/O Port 3.** This register functions as a general-purpose I/O port. In addition, all the pins have an alternative function listed below. Each of the functions is controlled by several other SFRs. The associated Port 3 latch bit must contain a logic '1' before the pin can be used in its alternate function capacity.

- WRExternal Data Memory Write Strobe. This pin provides an active low write strobe to an external memory device.bit 6If Port 0 or Port 2 is selected for external memory in the HCR1 register, this function will be enabled even if a '1' is<br/>not written to this latch bit. When external memory is selected, the settings of P3DRRH are ignored.
- T1 Timer/Counter 1 External Input. A 1 to 0 transition on this pin will increment Timer 1.
- T0 Timer/Counter 0 External Input. A 1 to 0 transition
- **T0 Timer/Counter 0 External Input.** A 1 to 0 transition on this pin will increment Timer 0. bit 4
- **INT1 External Interrupt 1.** A falling edge/low level on this pin will cause an external interrupt 1 if enabled.
- **INTO External Interrupt 0.** A falling edge/low level on this pin will cause an external interrupt 0 if enabled.
- **TXD0** Serial Port 0 Transmit. This pin transmits the serial Port 0 data in serial port modes 1, 2, 3, and emits the synchronizing clock in serial port mode 0.

**RXD0** Serial Port 0 Receive. This pin receives the serial Port 0 data in serial port modes 1, 2, 3, and is a bidirectional data bit 0 transfer pin in serial port mode 0.

RDExternal Data Memory Read Strobe. This pin provides an active low read strobe to an external memory device.bit 7If Port 0 or Port 2 is selected for external memory in the HCR1 register, this function will be enabled even if a '1' is<br/>not written to this latch bit. When external memory is selected, the settings of P3DRRH are ignored.



#### Port 2 Data Direction Low Register (P2DDRL)

	7	6	5	4	3	2	1	0	Reset Value
SFR B1h	P23H	P23L	P22H	P22L	P21H	P21L	P20H	P20L	00h

#### P2.3 Port 2 Bit 3 Control.

bits 7-6

P23H	P23L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

#### P2.2 Port 2 Bit 2 Control.

bits 5-4

	P22H	P22L	
I	0	0	Standard 8051 (Pull-Up)
	0	1	CMOS Output
	1	0	Open Drain Output
	1	1	Input

#### P2.1 Port 2 Bit 1 Control.

bits 3-2

P21H	P21L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

#### P2.0 Port 2 Bit 0 Control.

bits 1-0

P20H	P20L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

NOTE: Port 2 also controlled by EA and Memory Access Control HCR1.1.



#### Port 2 Data Direction High Register (P2DDRH)

	7	6	5	4	3	2	1	0	Reset Value
SFR B2h	P27H	P27L	P26H	P26L	P25H	P25L	P24H	P24L	00h

#### P2.7 Port 2 Bit 7 Control.

bits 7-6

P27H	P27L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

#### P2.6 Port 2 Bit 6 Control.

bits 5-4

P26H	P26L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

#### P2.5 Port 2 Bit 5 Control.

bits 3-2

P25H	P25L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

#### P2.4 Port 2 Bit 4 Control.

bits 1-0

## Port 2 Bit 4 Control.

P24H	P24L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

NOTE: Port 2 also controlled by EA and Memory Access Control HCR1.1.



## Port 3 Data Direction Low Register (P3DDRL)

	7	6	5	4	3	2	1	0	Reset Value
SFR B3h	P33H	P33L	P32H	P32L	P31H	P31L	P30H	P30L	00h

#### P3.3 Port 3 Bit 3 Control.

bits 7-6

P33H	P33L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

#### P3.2 Port 3 Bit 2 Control.

bits 5-4

P32H	P32L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

#### P3.1 Port 3 Bit 1 Control.

bits 3-2

P31H	P31L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

#### P3.0 Port 3 Bit 0 Control.

bits 1-0

P30H	P30L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input



#### Port 3 Data Direction High Register (P3DDRH)

	7	6	5	4	3	2	1	0	Reset Value
SFR B4h	P37H	P37L	P36H	P36L	P35H	P35L	P34H	P34L	00h

#### P3.7 Port 3 Bit 7 Control.

bits 7-6

P37H	P37L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

NOTE: Port 3.7 also controlled by EA and Memory Access Control HCR1.1.

#### P3.6 Port 3 Bit 6 Control.

bits 5-4

P36H	P36L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

NOTE: Port 3.6 also controlled by EA and Memory Access Control HCR1.1.

#### P3.5 Port 3 Bit 5 Control.

bits 3-2

P35H	P35L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input

#### P3.4 Port 3 Bit 4 Control.

bits 1-0

P34H	P34L	
0	0	Standard 8051 (Pull-Up)
0	1	CMOS Output
1	0	Open Drain Output
1	1	Input



## Interrupt Priority (IP)

	7	6	5	4	3	2	1	0	Reset Value				
SFR B8h	1	PS1	PT2	PS0	PT1	PX1	PT0	PX0	80h				
PS1	Serial Port 1	Interrupt. Th	is bit contro	ls the priority	of the seria	l Port 1 interi	upt.						
bit 6	0 = Serial Port 1 priority is determined by the natural priority order.												
	1 = Serial Port 1 is a high priority interrupt.												
PT2	Timer 2 Inter	<b>rupt.</b> This bit	controls the	e priority of the	e Timer 2 in	terrupt.							
bit 5	0 = Timer 2 p	riority is deter	mined by th	ne natural pric	ority order.								
	1 = Timer 2 priority is a high priority interrupt.												
PS0	Serial Port 0	Interrupt. Th	is bit contro	ls the priority	of the seria	l Port 0 interi	upt.						
bit 4	0 = Serial Po	rt 0 priority is	determined	by the natura	al priority orc	ler.	-						
	1 = Serial Po	rt 0 is a high p	priority interi	upt.									
PT1	Timer 1 Inter	<b>rupt.</b> This bit	controls the	e priority of the	e Timer 1 in	terrupt.							
bit 3	0 = Timer 1 p	riority is deter	mined by th	ne natural pric	ority order.								
	1 = Timer 1 p	riority is a hig	h priority int	errupt.									
PX1	External Inte	errupt 1. This	bit controls	the priority of	external int	errupt 1.							
bit 2		nterrupt 1 pric			natural pric	rity order.							
	1 = External i	nterrupt 1 is a	high priorit	y interrupt.									
РТ0	Timer 0 Inter	<b>rupt.</b> This bit	controls the	e priority of the	e Timer 0 in	terrupt.							
bit 1	0 = Timer 0 p	0 = Timer 0 priority is determined by the natural priority order.											
	1 = Timer 0 p	priority is a hig	h priority int	errupt.									
PX0	External Inte	errupt 0. This	bit controls	the priority of	external int	errupt 0.							
bit 0		nterrupt 0 prio			natural pric	rity order.							
	1 = External i	nterrupt 0 is a	ı high priorit	y interrupt.									



#### Serial Port 1 Control (SCON1)

	7	6	5	4	3	2	1	0	Reset Value
SFR C0h	SM0_1	SM1_1	SM2_1	REN_1	TB8_1	RB8_1	TI_1	RI_1	00h

**SM0–2** Serial Port 1 Mode. These bits control the mode of serial Port 1. Modes 1, 2, and 3 have 1 start and 1 stop bit in addition to the 8 or 9 data bits.

MODE	SM0	SM1	SM2	FUNCTION	LENGTH	PERIOD
0	0	0	0	Synchronous	8 bits	12 p <sub>CLK</sub> <sup>(1)</sup>
0	0	0	1	Synchronous	8 bits	4 PCLK <sup>(1)</sup>
1	0	1	0	Asynchronous	10 bits	Timer 1 Baud Rate Equation
1	0	1	1	Valid Stop Required <sup>(2)</sup>	10 bits	Timer 1 or Baud Rate Equation
2	1	0	0	Asynchronous	11 bits	$64 \text{ p}_{CLK}^{(1)} (SMOD = 0)$ $32 \text{ p}_{CLK}^{(1)} (SMOD = 1)$
2	1	0	1	Asynchronous with Multiprocessor Communication <sup>(3)</sup>	11 bits	$64 \text{ p}_{\text{CLK}}^{(1)} (\text{SMOD} = 0)$ $32 \text{ p}_{\text{CLK}}^{(1)} (\text{SMOD} = 1)$
3	1	1	0	Asynchronous	11 bits	Timer 1 Baud Rate Equation
3	1	1	1	Asynchronous with Multiprocessor Communication <sup>(3)</sup>	11 bits	Timer 1 Baud Rate Equation

(3) RI\_0 will not be activated if bit 9 = 0.

**REN\_1 Receive Enable.** This bit enables/disables the serial Port 1 received shift register.

bit 4 0 = Serial Port 1 reception disabled.

1 = Serial Port 1 received enabled (modes 1, 2, and 3). Initiate synchronous reception (mode 0).

## **TB8\_1 9th Transmission Bit State.** This bit defines the state of the 9th transmission bit in serial Port 1 modes 2 and 3. bit 3

**RB8\_19th Received Bit State.** This bit identifies the state of the 9th reception bit of received data in serial Port 1 modesbit 22 and 3. In serial port mode 1, when SM2\_1 = 0, RB8\_1 is the state of the stop bit. RB8\_1 is not used in mode 0.

TI\_1Transmitter Interrupt Flag. This bit indicates that data in the serial Port 1 buffer has been completely shifted out.bit 1In serial port mode 0, TI\_1 is set at the end of the 8th data bit. In all other modes, this bit is set at the end of the last data bit. This bit must be cleared by software to transmit the next byte.

RI\_1 Receiver Interrupt Flag. This bit indicates that a byte of data has been received in the serial Port 1 buffer. In serial bit 0 port mode 0, RI\_1 is set at the end of the 8th bit. In serial port mode 1, RI\_1 is set after the last sample of the incoming stop bit subject to the state of SM2\_1. In modes 2 and 3, RI\_1 is set after the last sample of RB8\_1. This bit must be cleared by software to receive the next byte.

bit 0

#### Serial Data Buffer 1 (SBUF1)

	7	6	5	4	3	2	1	0	Reset Value
SFR C1h									00h

**SBUF1.7–0** Serial Data Buffer 1. Data for serial Port 1 is read from or written to this location. The serial transmit and receive buffers are separate registers, but both are addressed at this location.

#### Enable Wake Up (EWU) Waking Up from IDLE Mode

		7	6	5	4	3	2	1	0	Reset Value
SFR C6h	-	_	_	_	_	_	EWUWDT	EWUEX1	EWUEX0	00h

Auxiliary interrupts will wake up from IDLE. They are enabled with EAI (EICON.5, SFR D8h).

#### **EWUWDT** Enable Wake Up Watchdog Timer. Wake using watchdog timer interrupt.

bit 2 0 = Don't wake up on watchdog timer interrupt.

1 = Wake up on watchdog timer interrupt.

**EWUEX1** Enable Wake Up External 1. Wake using external interrupt source 1.

bit 1 0 = Don't wake up on external interrupt source 1.

1 = Wake up on external interrupt source 1.

#### **EWUEX0** Enable Wake Up External 0. Wake using external interrupt source 0.

0 = Don't wake up on external interrupt source 0.

1 = Wake up on external interrupt source 0.



#### Timer 2 Control (T2CON)

		7	6	5	4	3	2	1	0	Reset Value	
SFR C8h		TF2	EXF2	RCLK	TCLK	EXEN2	TR2	C/T2	CP/RL2	00h	
<b>TF2</b> bit 7 <b>EXF2</b> bit 6	TF2 w Timer (T2CC	ill only be <b>2 Exterr</b> DN.3) bit.	e set if RCLk nal Flag. A no	and TCLK a egative trans	are both clea ition on the T	red to 0. Writ 2EX pin (P1.	ing a 1 to TF 1) will cause	2 forces a Ti this flag to be	mer 2 interru e set based o	by software. opt if enabled. on the EXEN2 it in software	
RCLK bit 5	0 = Tir 1 = Tir Setting	mer 1 ov mer 2 ov	erflow is use erflow is use will force Ti	ed to determi ed to determi	ne receiver l ne receiver l	Port 0 timeb baud rate for baud rate for beration mod	USART0. USART0.	-		des 1 or 3. le by 2 of the	
TCLK bit 4	<ul> <li>Transmit Clock Flag. This bit determines the serial Port 0 timebase when transmitting data in serial modes 1 or 3.</li> <li>0 = Timer 1 overflow is used to determine transmitter baud rate for USART0.</li> <li>1 = Timer 2 overflow is used to determine transmitter baud rate for USART0.</li> <li>Setting this bit will force Timer 2 into baud rate generation mode. The timer will operate from a divide by 2 of the external clock.</li> </ul>										
EXEN2 bit 3	baud r 0 = Tir	ates for mer 2 wil	the serial po I ignore all e	rt. xternal even	its at T2EX.			·		ot generating	
<b>TR2</b> bit 2	<ul> <li>1 = Timer 2 will capture or reload a value if a negative transition is detected on the T2EX pin.</li> <li>Timer 2 Run Control. This bit enables/disables the operation of Timer 2. Halting this timer will preserve the curren count in TH2, TL2.</li> <li>0 = Timer 2 is halted.</li> <li>1 = Timer 2 is enabled.</li> </ul>										
<b>C/T2</b> bit 1	<ul> <li>Counter/Timer Select. This bit determines whether Timer 2 will function as a timer or counter. Independent of this bit, Timer 2 runs at 2 clocks per tick when used in baud rate generator mode.</li> <li>0 = Timer 2 functions as a timer. The speed of Timer 2 is determined by the T2M bit (CKCON.5).</li> <li>1 = Timer 2 will count negative transitions on the T2 pin (P1.0).</li> </ul>										
CP/RL2 bit 0	or TCL	_K is set,	this bit will ds will occur	not function when Timer	and the time 2 overflows	r the capture r will functior or a falling e	n in an auto-i edge is detec	reload mode ted on T2E	following ea		

1 = Timer 2 captures will occur when a falling edge is detected on T2EX if EXEN2 = 1.

### Timer 2 Capture LSB (RCAP2L)

	7	6	5	4	3	2	1	0	Reset Value
SFR CAh									00h

RCAP2LTimer 2 Capture LSB. This register is used to capture the TL2 value when Timer 2 is configured in capture mode.bits 7-0RCAP2L is also used as the LSB of a 16-bit reload value when Timer 2 is configured in auto-reload mode.

## Timer 2 Capture MSB (RCAP2H)

	7	6	5	4	3	2	1	0	Reset Value
SFR CBh									00h

**RCAP2H** Timer 2 Capture MSB. This register is used to capture the TH2 value when Timer 2 is configured in capture mode. RCAP2H is also used as the MSB of a 16-bit reload value when Timer 2 is configured in auto-reload mode.

#### Timer 2 LSB (TL2)

	7	6	5	4	3	2	1	0	Reset Value
SFR CCh									00h

TL2 Timer 2 LSB. This register contains the least significant byte of Timer 2.

bits 7-0

#### Timer 2 MSB (TH2)

	7	6	5	4	3	2	1	0	Reset Value
SFR CDh									00h

**TH2 Timer 2 MSB.** This register contains the most significant byte of Timer 2.

bits 7-0

#### **Program Status Word (PSW)**

		7	6	5	4	3	2	1	0	Reset Value
Γ	SFR D0h	CY	AC	F0	RS1	RS0	OV	F1	Р	00h

**CY Carry Flag.** This bit is set when the last arithmetic operation resulted in a carry (during addition) or a borrow (during bit 7 subtraction). Otherwise, it is cleared to 0 by all arithmetic operations.

AC Auxiliary Carry Flag. This bit is set to 1 if the last arithmetic operation resulted in a carry into (during addition), or a borrow (during subtraction) from the high-order nibble. Otherwise, it is cleared to 0 by all arithmetic operations.

F0 User Flag 0. This is a bit-addressable, general-purpose flag for software control.

bit 5

**RS1, RS0** Register Bank Select 1–0. These bits select which register bank is addressed during register accesses.

bits 4–3

RS1	RS0	REGISTER BANK	ADDRESS
0	0	0	00h – 07h
0	1	1	08h – 0Fh
1	0	2	10h – 17h
1	1	3	18h – 1Fh

**OV Overflow Flag.** This bit is set to 1 if the last arithmetic operation resulted in a carry (addition), borrow (subtraction), bit 2 or overflow (multiply or divide). Otherwise it is cleared to 0 by all arithmetic operations.

F1 User Flag 1. This is a bit-addressable, general-purpose flag for software control.

bit 1

P Parity Flag. This bit is set to 1 if the modulo-2 sum of the 8 bits of the accumulator is 1 (odd parity); and cleared to 0 on even parity.

#### ADC Offset Calibration Register Low Byte (OCL)

	7	6	5	4	3	2	1	0	Reset Value
SFR D1h								LSB	00h

**OCL ADC Offset Calibration Register Low Byte.** This is the low byte of the 24-bit word that contains the ADC offset calibration. A value that is written to this location will set the ADC offset calibration value.

#### ADC Offset Calibration Register Middle Byte (OCM)

	7	6	5	4	3	2	1	0	Reset Value
SFR D2h									00h

**OCM** ADC Offset Calibration Register Middle Byte. This is the middle byte of the 24-bit word that contains the ADC offset calibration. A value that is written to this location will set the ADC offset calibration value.

#### ADC Offset Calibration Register High Byte (OCH)

	7	6	5	4	3	2	1	0	Reset Value
SFR D3h	MSB								00h

**OCH ADC Offset Calibration Register High Byte.** This is the high byte of the 24-bit word that contains the ADC offset calibration. A value that is written to this location will set the ADC offset calibration value.

#### ADC Gain Calibration Register Low Byte (GCL)

	7	6	5	4	3	2	1	0	Reset Value
SFR D4h								LSB	5Ah

**GCL** ADC Gain Calibration Register Low Byte. This is the low byte of the 24-bit word that contains the ADC gain calibration. A value that is written to this location will set the ADC gain calibration value.

#### ADC Gain Calibration Register Middle Byte (GCM)

	7	6	5	4	3	2	1	0	Reset Value
SFR D5h									ECh

**GCM** ADC Gain Calibration Register Middle Byte. This is the middle byte of the 24-bit word that contains the ADC gain calibration. A value that is written to this location will set the ADC gain calibration value.

#### ADC Gain Calibration Register High Byte (GCH)

	7	6	5	4	3	2	1	0	Reset Value
SFR D6h	MSB								5Fh

**GCH** ADC Gain Calibration Register High Byte. This is the high byte of the 24-bit word that contains the ADC gain calibration. A value that is written to this location will set the ADC gain calibration value.



## ADC Multiplexer Register (ADMUX)

	7	6	5	4	3	2	1	0	Reset Value
SFR D7h	INP3	INP2	INP1	INP0	INN3	INN2	INN1	INN0	01h

#### **INP3-0** Input Multiplexer Positive Channel. This selects the positive signal input.

bits 7-4

INP3	INP2	INP1	INP0	POSITIVE INPUT
0	0	0	0	AIN0 (default)
0	0	0	1	AIN1
0	0	1	0	AIN2
0	0	1	1	AIN3
0	1	0	0	AIN4
0	1	0	1	AIN5
0	1	1	0	AIN6
0	1	1	1	AIN7
1	0	0	0	AINCOM
1	1	1	1	Temperature Sensor (requires ADMUX = FFh)

#### INN3-0 Input Multiplexer Negative Channel. This selects the negative signal input.

bits 3-0

INN3	INN2	INN1	INN0	NEGATIVE INPUT
0	0	0	0	AINO
0	0	0	1	AIN1 (default)
0	0	1	0	AIN2
0	0	1	1	AIN3
0	1	0	0	AIN4
0	1	0	1	AIN5
0	1	1	0	AIN6
0	1	1	1	AIN7
1	0	0	0	AINCOM
1	1	1	1	Temperature Sensor (requires ADMUX = FFh)



#### Enable Interrupt Control (EICON)

	7	6	5	4	3	2	1	0	Reset Value
SFR D8h	SMOD1	1	EAI	AI	WDTI	0	0	0	40h
SMOD1 bit 7	<b>Serial Port 1 M</b> 0 = Standard b 1 = Double bac	aud rate for	Port 1 (defa		aud rate for	Port 1 will be	e doubled.		

- EAIEnable Auxiliary Interrupt. The Auxiliary Interrupt accesses nine different interrupts which are masked and<br/>identified by SFR registers PAI (SFR A5h), AIE (SFR A6h), and AISTAT (SFR A7h).
  - 0 = Auxiliary Interrupt disabled (default).
  - 1 = Auxiliary Interrupt enabled.
- Al Auxiliary Interrupt Flag. Al must be cleared by software before exiting the interrupt service routine, after the source of the interrupt is cleared. Otherwise, the interrupt occurs again. Setting Al in software generates an Auxiliary Interrupt, if enabled.
  - 0 = No Auxiliary Interrupt detected (default).
  - 1 = Auxiliary Interrupt detected.
- WDTIWatchdog Timer Interrupt Flag. WDTI must be cleared by software before exiting the interrupt service routine.bit 3Otherwise, the interrupt will occur again. Setting WDTI in software generates a watchdog time interrupt, if enabled.<br/>The Watchdog timer can generate an interrupt or reset. The interrupt is available only if the reset action is disabled<br/>in HCR0.
  - 0 = No Watchdog Timer Interrupt detected (default).

1 = Watchdog Timer Interrupt detected.

#### ADC Results Register Low Byte (ADRESL)

	7	6	5	4	3	2	1	0	Reset Value
SFR D9h								LSB	00h

ADRESL bits 7–0 The ADC Results Low Byte. This is the low byte of the 24-bit word that contains the ADC converter results. Reading from this register clears the ADC interrupt. However, AI in EICON (SFR D8h) must also be cleared.

#### ADC Results Register Middle Byte (ADRESM)

	7	6	5	4	3	2	1	0	Reset Value
SFR DAh									00h

**ADRESM** The ADC Results Middle Byte. This is the middle byte of the 24-bit word that contains the ADC converter results. bits 7–0

#### ADC Results Register High Byte (ADRESH)

	7	6	5	4	3	2	1	0	Reset Value
SFR DBh	MSB								00h

**ADRESH** The ADC Results High Byte. This is the high byte of the 24-bit word that contains the ADC converter results. bits 7–0



#### ADC Control Register 0 (ADCON0)

	7	6	5	4	3	2	1	0	Reset Value
SFR DCh		BOD	EVREF	VREFH	EBUF	PGA2	PGA1	PGA0	30h

#### **BOD** bit 6 Burnout Detect. When enabled this connects a positive current source to the positive channel and a negative current source to the negative channel. If the channel is open circuit then the ADC results will be full-scale. 0 = Burnout Current Sources Off (default).

1 = Burnout Current Sources On.

# **EVREF** Enable Internal Voltage Reference. If the internal voltage reference is not used, it should be turned off to save power and reduce noise.

0 = Internal Voltage Reference Off.

1 = Internal Voltage Reference On (default). NOTE: REFIN- must be connected to AGND, and REFOUT to REFIN+.

#### VREFH Voltage Reference High Select. The internal voltage reference can be selected to be 2.5V or 1.25V.

bit 4 0 = REFOUT is 1.25V.

1 = REFOUT is 2.5V (default).

# **EBUF** Enable Buffer. Enable the input buffer to provide higher input impedance but limits the input voltage range and dissipates more power.

0 = Buffer disabled (default).

1 = Buffer enabled.

#### PGA2-0 Programmable Gain Amplifier. Sets the gain for the PGA from 1 to 128.

bits 2–0

PGA2	PGA1	PGA0	GAIN
0	0	0	1 (default)
0	0	1	2
0	1	0	4
0	1	1	8
1	0	0	16
1	0	1	32
1	1	0	64
1	1	1	128



#### ADC Control Register 1 (ADCON1)

	7	6	5	4	3	2	1	0	Reset Value
SFR DDh	—	POL	SM1	SM0	—	CAL2	CAL1	CAL0	0000 0000b

**POL Polarity.** Polarity of the ADC result and Summation register.

bit 6 0 = Bipolar.

1 = Unipolar. The LSB size is 1/2 the size of bipolar (twice the resolution).

POL	ANALOG INPUT	DIGITAL OUTPUT		
	+FSR	7FFFFh		
0	ZERO	000000h		
	-FSR	800000h		
	+FSR	FFFFFh		
1	ZERO	000000h		
	–FSR	000000h		

SM1–0 Settling Mode. Selects the type of filter or auto select which defines the digital filter settling characteristics.

bits 5-4

SM1	SM0	SETTLING MODE
0	0	Auto
0	1	Fast Settling Filter
1	0	Sinc <sup>2</sup> Filter
1	1	Sinc <sup>3</sup> Filter

#### CAL2–0 Calibration Mode Control Bits.

bits 2–0 Writing to these bits starts ADC calibration.

CAL2	CAL1	CAL0	CALIBRATION MODE
0	0	0	No Calibration (default)
0	0	1	Self-Calibration, Offset and Gain
0	1	0	Self-Calibration, Offset only
0	1	1	Self-Calibration, Gain only
1	0	0	System Calibration, Offset only (requires external connection)
1	0	1	System Calibration, Gain only (requires external connection)
1	1	0	Reserved
1	1	1	Reserved

NOTE: Read Value—000b.

#### ADC Control Register 2 (ADCON2)

	7	6	5	4	3	2	1	0	Reset Value
SFR DEh	DR7	DR6	DR5	DR4	DR3	DR2	DR1	DR0	1Bh

#### DR7–0 Decimation Ratio LSB.

bits 7-0

#### **ADC Control Register 3 (ADCON3)**

	7	6	5	4	3	2	1	0	Reset Value
SFR DFh	—	-			-	DR10	DR9	DR8	06h

**DR10-8** Decimation Ratio Most Significant 3 Bits. The output data rate =  $f_{CLK}/[(ACLK + 1) \cdot 64 \cdot Decimation Ratio]$ . bits 2–0



#### Accumulator (A or ACC)

	7	6	5	4	3	2	1	0	Reset Value
SFR E0h	ACC.7	ACC.6	ACC.5	ACC.4	ACC.3	ACC.2	ACC.1	ACC.0	00h

ACC.7–0 Accumulator. This register serves as the accumulator for arithmetic and logic operations.

bits 7-0

#### Summation/Shifter Control (SSCON)

	7	6	5	4	3	2	1	0	Reset Value
SFR E1h	SSCON1	SSCON0	SCNT2	SCNT1	SCNT0	SHF2	SHF1	SHF0	00h

The Summation register is powered down when the ADC is powered down. If all zeroes are written to this register the 32-bit SUMR3–0 registers will be cleared. The Summation registers will do sign extend if Bipolar is selected in ADCON1.

#### SSCON1-0 Summation/Shift Count.

bits 7-6

SOURCE	SSCON1	SSCON0	MODE
CPU	0	0	Values written to the SUM registers are accumulated when the SUMR0 value is written (sum/shift ignored)
ADC	0	1	Summation register Enabled. Source is ADC, summation count is working.
CPU	1	0	Shift Enabled. Summation register is shifted by SHF Count bits. It takes four system clocks to execute.
ADC	1	1	Accumulate and Shift Enable. Values are accumulated for SUM Count times and then shifted by SHF Count.

**SCNT2-0** Summation Count. When the summation is complete an interrupt will be generated unless masked. Reading the SUMR0 register clears the interrupt.

SCNT2	SCNT1	SCNT0	SUMMATION COUNT
0	0	0	2
0	0	1	4
0	1	0	8
0	1	1	16
1	0	0	32
1	0	1	64
1	1	0	128
1	1	1	256

#### SHF2–0 Shift Count.

bits 2-0

SHF2	SHF1	SHF0	SHIFT	DIVIDE
0	0	0	1	2
0	0	1	2	4
0	1	0	3	8
0	1	1	4	16
1	0	0	5	32
1	0	1	6	64
1	1	0	7	128
1	1	1	8	256



#### Summation Register 0 (SUMR0)

	7	6	5	4	3	2	1	0	Reset Value
SFR E2h								LSB	00h

SUMR0Summation Register 0. This is the least significant byte of the 32-bit summation register or bits 0 to 7.bits 7-0Write: Will cause values in SUMR3-0 to be added to the summation register.

Read: Will clear the Summation Count Interrupt. AI in EICON (SFR D8h) must also be cleared.

#### Summation Register 1 (SUMR1)

	7	6	5	4	3	2	1	0	Reset Value
SFR E3h									00h

SUMR1 Summation Register 1. These are bits 8–15 of the 32-bit summation register.

bits 7-0

#### Summation Register 2 (SUMR2)

	7	6	5	4	3	2	1	0	Reset Value
SFR E4h									00h

#### SUMR2 Summation Register 2. These are bits 16–23 of the 32-bit summation register.

bits 7-0

#### Summation Register 3 (SUMR3)

	7	6	5	4	3	2	1	0	Reset Value
SFR E5h	MSB								00h

**SUMR3** Summation Register 3. This is the most significant byte of the 32-bit summation register or bits 24–31. bits 7–0

#### **Offset DAC Register (ODAC)**

	7	6	5	4	3	2	1	0	Reset Value
SFR E6h									00h

ODACOffset DAC Register. This register will shift the input by up to half of the ADC full-scale input range. The offset DACbits 7-0value is summed with the ADC input prior to conversion. Writing 00h or 80h to ODAC turns off the offset DAC.bit 7Offset DAC Sign bit.

0 = Positive

1 = Negative

bit 6–0 Offset = 
$$\frac{-V_{REF}}{2 \cdot PGA} \cdot \left(\frac{ODAC[6:0]}{127}\right) \cdot (-1)^{bit7}$$

**NOTE:** ODAC cannot be used to offset the input so that the buffer can be used for AGND signals. Offset DAC should be cleared before offset calibration, since the offset DAC output is applied directly to the ADC input.



#### Low Voltage Detect Control (LVDCON)

	7	6	5	4	3	2	1	0	Reset Value
SFR E7h	ALVDIS	ALVD2	ALVD1	ALVD0	DLVDIS	DLVD2	DLVD1	DLVD0	00h

**NOTE:** By default, both analog and digital low-voltage detections are enabled, which causes approximately 25µA of current consumption from the power supply. To minimize this power consumption, both low-voltage detections should be disabled before entering Stop mode.

#### ALVDIS Analog Low Voltage Detect Disable.

0 = Enable Detection of Low Analog Supply Voltage

1 = Disable Detection of Low Analog Supply Voltage

#### ALVD2–0 Analog Voltage Detection Level.

bits 6-4

bit 7

ALVD2	ALVD1	ALVD0	VOLTAGE LEVEL
0	0	0	AV <sub>DD</sub> 2.7V (default)
0	0	1	AV <sub>DD</sub> 3.0V
0	1	0	AV <sub>DD</sub> 3.3V
0	1	1	AV <sub>DD</sub> 4.0V
1	0	0	AV <sub>DD</sub> 4.2V
1	0	1	AV <sub>DD</sub> 4.5V
1	1	0	AV <sub>DD</sub> 4.7V
1	1	1	External Voltage AIN7 compared to 1.2V

#### DLVDIS Digital Low Voltage Detect Disable.

bit 3 0 = Enable Detection of Low Digital Supply Voltage 1 = Disable Detection of Low Digital Supply Voltage

#### DLVD2–0 Digital Voltage Detection Level.

bits 2-0

#### Digital Voltage Deteotion Level

DLVD2	DLVD1	DLVD0	VOLTAGE LEVEL
0	0	0	DV <sub>DD</sub> 2.7V (default)
0	0	1	DV <sub>DD</sub> 3.0V
0	1	0	DV <sub>DD</sub> 3.3V
0	1	1	DV <sub>DD</sub> 4.0V
1	0	0	DV <sub>DD</sub> 4.2V
1	0	1	DV <sub>DD</sub> 4.5V
1	1	0	DV <sub>DD</sub> 4.7V
1	1	1	External Voltage AIN6 compared to
			1.2V



#### **Extended Interrupt Enable (EIE)**

	7	6	5	4	3	2	1	0	Reset Value
SFR E8h	1	1	1	EWDI	EX5	EX4	EX3	EX2	E0h
EWDI	Enable Watch	ndog Interru	pt. This bit	enables/disal	oles the wat	chdog interru	pt. The Wat	chdog time	r is enabled b
oit 4	(SFR FFh) and					0		•	
	0 = Disable the	e Watchdog	Interrupt						
	1 = Enable Inte	errupt Reque	est Generate	ed by the Wa	tchdog Time	er			
	_								
EX5	External Inter	-		enables/disat	oles externa	l interrupt 5.			
bit 3	0 = Disable External Interrupt 5								
1 = Enable External Interrupt 5									
EX4	External Inter	External Interrupt 4 Enable. This bit enables/disables external interrupt 4.							
bit 2	0 = Disable Ex	-							
	1 = Enable Ex		•						
EX3	External Inter	-		enables/disat	oles externa	l interrupt 3.			
bit 1	0 = Disable Ex	ternal Interro	upt 3						
	1 = Enable Ex	ternal Interru	ipt 3						
EX2	External Inter	rupt 2 Enab	le. This bit e	enables/disat	oles externa	l interrupt 2.			
bit 0	0 = Disable Ex	-							
	1 = Enable Ex		•						
			ipiz						

## Hardware Product Code Register 0 (HWPC0) (read-only)

	7	6	5	4	3	2	1	0	Reset Value
SFR E9h	0	0	0	0	0	0	MEMOR	RY SIZE	0000_00xxb

#### HWPC1.7-0 Hardware Product Code LSB. Read-only.

bits 7-0

	IORY ZE	MODEL	FLASH MEMORY
0	0	MSC1210Y2	4kB
0	1	MSC1210Y3	8kB
1	0	MSC1210Y4	16kB
1	1	MSC1210Y5	32kB

## Hardware Product Code Register 1 (HWPC1) (read-only)

	7	6	5	4	3	2	1	0	Reset Value
SFR EAh	0	0	0	0	0	0	0	0	00h

# HWPC1.7-0 Hardware Product Code MSB. Read-only.



#### Hardware Version Register (HDWVER)

	7	6	5	4	3	2	1	0	Reset Value
SFR EBh									

#### Flash Memory Control (FMCON)

	7	6	5	4	3	2	1	0	Reset Value
SFR EEh	0	PGERA	0	FRCM	0	BUSY	1	0	02h

#### PGERA Page Erase.

bit 6 0 = MOVX to Flash will perform a byte write operation

1 = MOVX to Flash will perform a page erase operation

FRCM	Frequency Control Mode.
bit 4	0 = Bypass (default)
	1 = Use Delay Line. Saves power when reading Flash (recommended)
BUSY	Write/Erase BUSY Signal.
bit 2	0 = Idle or Available

1 = Busy

#### Flash Memory Timing Control Register (FTCON)

	7	6	5	4	3	2	1	0	Reset Value
SFR EFh	FER3	FER2	FER1	FER0	FWR3	FWR2	FWR1	FWR0	A5h

Refer to Flash Timing Characteristics.

FER3-0	<b>Set Erase.</b> Flash Erase Time = $(1 + FER) \bullet (MSEC + 1) \bullet t_{CLK}$ .
bits 7–4	A minimum of 10ms is needed for industrial temperature range.

A minimum of 4ms is needed for commercial temperature range.

**FWR3-0** Set Write. Flash Write Time =  $(1 + FWR) \cdot (USEC + 1) \cdot 5 \cdot t_{CLK}$ .

bits 3-0 Write time should be  $30-40\mu$ s.

#### B Register (B)

	7	6	5	4	3	2	1	0	Reset Value
SFR F0h	B.7	B.6	B.5	B.4	B.3	B.2	B.1	B.0	00h

**B.7–0 B Register.** This register serves as a second accumulator for certain arithmetic operations. bits 7–0



#### **Power-Down Control Register (PDCON)**

	7	6	5	4	3	2	1	0	Reset Value
SFR F1h	0	0	0	PDPWM	PDADC	PDWDT	PDST	PDSPI	1Fh

Turning peripheral modules off puts the MSC1210 in the lowest power mode.

PDPWM bit 4	Pulse Width Module Control. 0 = PWM On 1 = PWM Power Down
PDADC bit 3	ADC Control. 0 = ADC On 1 = ADC, V <sub>REF</sub> , and summation registers are powered down.
PDWDT bit 2	Watchdog Timer Control. 0 = Watchdog Timer On 1 = Watchdog Timer Power Down
<b>PDST</b> bit 1	<b>System Timer Control.</b> 0 = System Timer On 1 = System Timer Power Down
<b>PDSPI</b> bit 0	<b>SPI System Control.</b> 0 = SPI System On 1 = SPI System Power Down

#### **PSEN/ALE Select (PASEL)**

	7	6	5	4	3	2	1	0	Reset Value
SFR F2h	0	0	PSEN2	PSEN1	PSEN0	0	ALE1	ALE0	00h

### PSEN2–0 PSEN Mode Select.

bits 5–3

PSEN2	PSEN1	PSEN0	
0	0	х	PSEN
0	1	х	CLK
1	0	х	ADC MODCLK
1	1	0	LOW
1	1	1	HIGH

#### ALE1–0 ALE Mode Select.

bits 1-0

ALE1	ALE0	
0	х	ALE
1	0	LOW
1	1	HIGH

**NOTE:** For power-saving purposes, it is recommended that the PSEN and ALE pins be set to low or high mode when external memory is not used.



#### Analog Clock (ACLK)

	7	6	5	4	3	2	1	0	Reset Value
SFR F6h	0	FREQ6	FREQ5	FREQ4	FREQ3	FREQ2	FREQ1	FREQ0	03h

**FREQ6–0** Clock Frequency – 1. This value + 1 divides the system clock to create the ADC clock.

bit 6–0

$$f_{ACLK} = \frac{f_{CLK}}{FREQ + 1} = \frac{f_{CLK}}{ACLK + 1}$$
$$f_{MOD} = \frac{f_{CLK}}{(ACLK + 1) \cdot 64}$$

Output Data Rate =  $\frac{T_{MOD}}{Decimation}$ 

#### System Reset Register (SRST)

	7	6	5	4	3	2	1	0	Reset Value
SFR F7h	0	0	0	0	0	0	0	RSTREQ	00h

**RSTREQ** Reset Request. Setting this bit to 1 and then clearing to 0 will generate a system reset. bit 0

#### **Extended Interrupt Priority (EIP)**

	7	6	5	4	3	2	1	0	Reset Value
SFR F8h	1	1	1	PWDI	PX5	PX4	PX3	PX2	E0h

<b>PWDI</b> bit 4	<ul> <li>Watchdog Interrupt Priority. This bit controls the priority of the watchdog interrupt.</li> <li>0 = The watchdog interrupt is low priority.</li> <li>1 = The watchdog interrupt is high priority.</li> </ul>
<b>PX5</b> bit 3	<ul> <li>External Interrupt 5 Priority. This bit controls the priority of external interrupt 5.</li> <li>0 = External interrupt 5 is low priority.</li> <li>1 = External interrupt 5 is high priority.</li> </ul>
PX4 bit 2	<ul> <li>External Interrupt 4 Priority. This bit controls the priority of external interrupt 4.</li> <li>0 = External interrupt 4 is low priority.</li> <li>1 = External interrupt 4 is high priority.</li> </ul>
<b>PX3</b> bit 1	<ul> <li>External Interrupt 3 Priority. This bit controls the priority of external interrupt 3.</li> <li>0 = External interrupt 3 is low priority.</li> <li>1 = External interrupt 3 is high priority.</li> </ul>
<b>PX2</b> bit 0	<ul> <li>External Interrupt 2 Priority. This bit controls the priority of external interrupt 2.</li> <li>0 = External interrupt 2 is low priority.</li> <li>1 = External interrupt 2 is high priority.</li> </ul>



#### Seconds Timer Interrupt (SECINT)

	7	6	5	4	3	2	1	0	Reset Value
SFR F9h	WRT	SECINT6	SECINT5	SECINT4	SECINT3	SECINT2	SECINT1	SECINT0	7Fh

This system clock is divided by the value of the 16-bit register MSECH:MSECL. Then, the 1ms timer tick is divided by the register HMSEC that provides the 100ms signal used by this seconds timer. Therefore, the seconds timer can generate an interrupt that occurs from 100ms to 12.8 seconds. Reading this register clears the Seconds Interrupt. This Interrupt can be monitored in the AIE register.

WRT Write Control. Determines whether to write the value immediately or wait until the current count is finished. bit 7 Read = 0.

0 = Delay Write Operation. The SEC value is loaded when the current count expires.

1 = Write Immediately. The counter is loaded once the CPU completes the write operation.

**SECINT6-0** Seconds Count. Normal operation uses 100ms as the clock interval, and would equal: (SEC + 1)/10 seconds. bits 6–0 Seconds Interrupt =  $(1 + SEC) \cdot (HMSEC + 1) \cdot (MSEC + 1) \cdot t_{CLK}$ 

#### Milliseconds Interrupt (MSINT)

	7	6	5	4	3	2	1	0	Reset Value
SFR FAh	WRT	MSINT6	MSINT5	MSINT4	MSINT3	MSINT2	MSINT1	MSINT0	7Fh

The clock used for this timer is the 1ms clock, which results from dividing the system clock by the values in registers MSECH:MSECL. Reading this register clears the milliseconds interrupt. AI in EICON (SFR D8h) must also be cleared.

WRT Write Control. Determines whether to write the value immediately or wait until the current count is finished.bit 7 Read = 0.

0 = Delay Write Operation. The MSINT value is loaded when the current count expires.

1 = Write Immediately. The MSINT counter is loaded once the CPU completes the write operation.

MSINT6-0 Seconds Count. Normal operation would use 1ms as the clock interval.

bits 6–0 MS Interrupt Interval = (1 + MSINT) • (MSEC + 1) • t<sub>CLK</sub>

#### One Microsecond Register (USEC)

	7	6	5	4	3	2	1	0	Reset Value
SFR FBh	0	0	0	FREQ4	FREQ3	FREQ2	FREQ1	FREQ0	03h

FREQ4–0 Clock Frequency – 1. This value + 1 divides the system clock to create a 1µs clock.

bits 4–0 USEC = CLK/(FREQ + 1). This clock is used to set Flash write time. See FTCON (SFR EFh).

#### One Millisecond Low Register (MSECL)

	7	6 5		4	3	2	1	0	Reset Value	
SFR FCh	MSECL7	MSECL6	MSECL5	MSECL4	MSECL3	MSECL2	MSECL1	MSECL0	9Fh	

MSECL7-0 One Millisecond Low. This value in combination with the next register is used to create a 1ms clock.

bits 7–0 1ms = (MSECH • 256 + MSECL + 1) • t<sub>CLK</sub>. This clock is used to set Flash erase time. See FTCON (SFR EFh).



#### **One Millisecond High Register (MSECH)**

	7	6	5	4	3	2	1	0	Reset Value
SFR FDh	MSECH7	MSECH6	MSECH5	MSECH4	MSECH3	MSECH2	MSECH1	MSECH0	0Fh

**MSECH7-0** One Millisecond High. This value in combination with the previous register is used to create a 1ms clock. bits 7–0  $1ms = (MSECH \cdot 256 + MSECL + 1) \cdot t_{CLK}$ 

#### **One Hundred Millisecond Register (HMSEC)**

	7	6	5	5 4		2	1	0	Reset Value	
SFR FEh	HMSEC7	HMSEC6	HMSEC5	HMSEC4	HMSEC3	HMSEC2	HMSEC1	HMSEC0	63h	

HMSEC7-0 One Hundred Millisecond. This clock divides the 1ms clock to create a 100ms clock.

bits 7–0 100ms = (MSECH • 256 + MSECL + 1) • (HMSEC + 1) • t<sub>CLK</sub>

#### Watchdog Timer Register (WDTCON)

ĺ		7	6	5	4	3	2	1	0	Reset Value
	SFR FFh	EWDT	DWDT	RWDT	WDCNT4	WDCNT3	WDCNT2	WDCNT1	WDCNT0	00h

#### EWDT Enable Watchdog (R/W).

bit 7 Write 1/Write 0 sequence sets the Watchdog Enable Counting bit.

#### DWDT Disable Watchdog (R/W).

bit 6 Write 1/Write 0 sequence clears the Watchdog Enable Counting bit.

#### RWDT Reset Watchdog (R/W).

bit 5 Write 1/Write 0 sequence restarts the Watchdog Counter.

#### WDCNT4-0 Watchdog Count (R/W).

bits 4–0 Watchdog expires in (WDCNT + 1) • HMSEC to (WDCNT + 2) • HMSEC, if the sequence is not asserted. There is an uncertainty of 1 count.

## **Revision History**

DATE	REV	PAGE	SECTION	DESCRIPTION
1/08	J	70	Serial Port Mode 1	Deleted note (2) from SM0-2 table.
10/07	I	26	Voltage Reference	Added paragraph to end of section.

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.



10-Dec-2020

## PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty		Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking	Samples
	(1)		Drawing		QLy	(2)	(6)	(3)		(4/5)	
MSC1210Y2PAGR	ACTIVE	TQFP	PAG	64	1500	RoHS & Green	NIPDAU	Level-4-260C-72 HR		MSC1210Y2	Samples
MSC1210Y2PAGT	ACTIVE	TQFP	PAG	64	250	RoHS & Green	NIPDAU	Level-4-260C-72 HR		MSC1210Y2	Samples
MSC1210Y3PAGR	ACTIVE	TQFP	PAG	64	1500	RoHS & Green	NIPDAU	Level-4-260C-72 HR		MSC1210Y3	Samples
MSC1210Y3PAGT	ACTIVE	TQFP	PAG	64	250	RoHS & Green	NIPDAU	Level-4-260C-72 HR		MSC1210Y3	Samples
MSC1210Y4PAGR	ACTIVE	TQFP	PAG	64	1500	RoHS & Green	NIPDAU	Level-4-260C-72 HR		MSC1210Y4	Samples
MSC1210Y4PAGT	ACTIVE	TQFP	PAG	64	250	RoHS & Green	NIPDAU	Level-4-260C-72 HR		MSC1210Y4	Samples
MSC1210Y5PAGR	ACTIVE	TQFP	PAG	64	1500	RoHS & Green	NIPDAU	Level-4-260C-72 HR		MSC1210Y5	Samples
MSC1210Y5PAGRG4	ACTIVE	TQFP	PAG	64	1500	RoHS & Green	NIPDAU	Level-4-260C-72 HR		MSC1210Y5	Samples
MSC1210Y5PAGT	ACTIVE	TQFP	PAG	64	250	RoHS & Green	NIPDAU	Level-4-260C-72 HR		MSC1210Y5	Samples

<sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

<sup>(2)</sup> RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

<sup>(3)</sup> MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

<sup>(4)</sup> There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.



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## PACKAGE OPTION ADDENDUM

10-Dec-2020

<sup>(5)</sup> Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

<sup>(6)</sup> Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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# PACKAGE MATERIALS INFORMATION

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## TAPE AND REEL INFORMATION





## QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



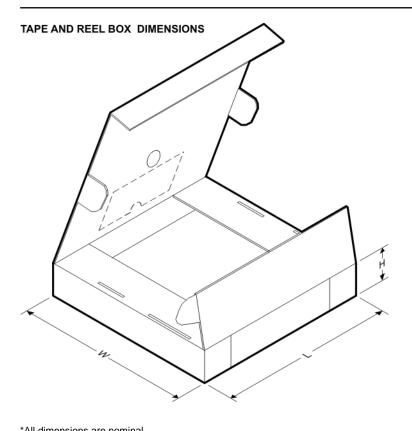
*All dimensions are nominal												
Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
MSC1210Y2PAGR	TQFP	PAG	64	1500	330.0	24.4	13.0	13.0	1.5	16.0	24.0	Q2
MSC1210Y2PAGT	TQFP	PAG	64	250	180.0	24.4	13.0	13.0	1.5	16.0	24.0	Q2
MSC1210Y3PAGR	TQFP	PAG	64	1500	330.0	24.4	13.0	13.0	1.5	16.0	24.0	Q2
MSC1210Y3PAGT	TQFP	PAG	64	250	180.0	24.4	13.0	13.0	1.5	16.0	24.0	Q2
MSC1210Y4PAGR	TQFP	PAG	64	1500	330.0	24.4	13.0	13.0	1.5	16.0	24.0	Q2
MSC1210Y4PAGT	TQFP	PAG	64	250	180.0	24.4	13.0	13.0	1.5	16.0	24.0	Q2
MSC1210Y5PAGR	TQFP	PAG	64	1500	330.0	24.4	13.0	13.0	1.5	16.0	24.0	Q2
MSC1210Y5PAGT	TQFP	PAG	64	250	180.0	24.4	13.0	13.0	1.5	16.0	24.0	Q2

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# PACKAGE MATERIALS INFORMATION

14-Feb-2019



*All dimensions are nominal							
Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
MSC1210Y2PAGR	TQFP	PAG	64	1500	350.0	350.0	43.0
MSC1210Y2PAGT	TQFP	PAG	64	250	213.0	191.0	55.0
MSC1210Y3PAGR	TQFP	PAG	64	1500	350.0	350.0	43.0
MSC1210Y3PAGT	TQFP	PAG	64	250	213.0	191.0	55.0
MSC1210Y4PAGR	TQFP	PAG	64	1500	350.0	350.0	43.0
MSC1210Y4PAGT	TQFP	PAG	64	250	213.0	191.0	55.0
MSC1210Y5PAGR	TQFP	PAG	64	1500	350.0	350.0	43.0
MSC1210Y5PAGT	TQFP	PAG	64	250	213.0	191.0	55.0

## **MECHANICAL DATA**

MTQF006A - JANUARY 1995 - REVISED DECEMBER 1996

#### PAG (S-PQFP-G64)

PLASTIC QUAD FLATPACK



NOTES: A. All linear dimensions are in millimeters.

B. This drawing is subject to change without notice.

C. Falls within JEDEC MS-026



# LAND PATTERN DATA



A. All linear dimensions are in millimeters.B. This drawing is subject to change without no

- B. This drawing is subject to change without notice.
   C. Laser cutting apertures with trapezoidal walls and also rounding corrections.
  - C. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric metal load solder paste. Refer to IPC-7525 for other stencil recommendations.
  - D. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



# LAND PATTERN DATA



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