

# **LatticeECP3 Family Data Sheet**

# **Data Sheet**



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# 1. Introduction

### 1.1. Features

### Higher Logic Density for Increased System Integration

- 17 K to 149 K LUTs
- 116 to 586 I/O

#### • Embedded SERDES

- 150 Mbps to 3.2 Gbps for Generic 8b10b, 10bit SERDES, and 8-bit SERDES modes
- Data Rates 230 Mbps to 3.2 Gbps per channel for all other protocols
- Up to 16 channels per device: PCI Express, SONET/SDH, Ethernet (1GbE, SGMII, XAUI), CPRI, SMPTE 3G and Serial RapidIO

#### sysDSP™

- Fully cascadable slice architecture
- 12 to 160 slices for high performance multiply and accumulate
- Powerful 54-bit ALU operations
- Time Division Multiplexing MAC Sharing
- Rounding and truncation
- Each slice supports
  - Half 36x36, two 18x18 or four 9x9 multipliers
  - Advanced 18x36 MAC and 18x18
     Multiply- Multiply-Accumulate (MMAC) operations

#### • Flexible Memory Resources

- Up to 6.85 Mbits sysMEM™ Embedded Block RAM (EBR)
- 36 K to 303 K bits distributed RAM

# • sysCLOCK Analog PLLs and DLLs

• Two DLLs and up to ten PLLs per device

#### Pre-Engineered Source Synchronous I/O

- DDR registers in I/O cells
- Dedicated read/write levelling functionality
- Dedicated gearing logic
- Source synchronous standards support
  - ADC/DAC, 7:1 LVDS, XGMII
  - High Speed ADC/DAC devices
- Dedicated DDR/DDR2/DDR3 memory with DQS support
- Optional Inter-Symbol Interference (ISI) correction on outputs

# Programmable sysI/O™ Buffer Supports Wide Range of Interfaces

- On-chip termination
- Optional equalization filter on inputs
- LVTTL and LVCMOS 33/25/18/15/12
- SSTL 33/25/18/15 I, II
- HSTL15 I and HSTL18 I, II
- PCI and Differential HSTL, SSTL
- LVDS, Bus-LVDS, LVPECL, RSDS, MLVDS

#### • Flexible Device Configuration

- Dedicated bank for configuration I/O
- SPI boot flash interface
- Dual-boot images supported
- Slave SPI
- TransFR™ I/O for simple field updates
- Soft Error Detect embedded macro

#### System Level Support

- IEEE 1149.1 and IEEE 1532 compliant
- Reveal Logic Analyzer
- ORCAstra FPGA configuration utility
- On-chip oscillator for initialization & general use
- 1.2 V core power supply

Table 1.1	LatticeFCP3™	Eamily So	loction	Guida
Table L.L.	TATTICEFU P3	Family Se	IECTION	GILLAR

Device	ECP3-17	ECP3-35	ECP3-70	ECP3-95	ECP3-150
LUTs (K)	17	33	67	92	149
sysMEM Blocks (18 Kbits)	38	72	240	240	372
Embedded Memory (Kbits)	700	1327	4420	4420	6850
Distributed RAM Bits (Kbits)	36	68	145	188	303
18 x 18 Multipliers	24	64	128	128	320
SERDES (Quad)	1	1	3	3	4
PLLs/DLLs	2/2	4/2	10 / 2	10 / 2	10 / 2
Packages and SERDES Channels,	/ I/O Combination	s			
328 csBGA (10 x 10 mm)	2 / 116	_	_	_	_
256 ftBGA (17 x 17 mm)	4 / 133	4 / 133	_	_	_
484 fpBGA (23 x 23 mm)	4 / 222	4 / 295	4 / 295	4 / 295	
672 fpBGA (27 x 27 mm)	_	4/310	8 / 380	8 / 380	8 / 380
1156 fpBGA (35 x 35 mm)	_	_	12 / 490	12 / 490	16 / 586



## 1.2. Introduction

The LatticeECP3 (EConomy Plus Third generation) family of FPGA devices is optimized to deliver high performance features such as an enhanced DSP architecture, high speed SERDES and high speed source synchronous interfaces in an economical FPGA fabric. This combination is achieved through advances in device architecture and the use of 65 nm technology making the devices suitable for high-volume, high-speed, low-cost applications.

The LatticeECP3 device family expands look-up-table (LUT) capacity to 149 K logic elements and supports up to 586 user I/O. The LatticeECP3 device family also offers up to 320 18 x 18 multipliers and a wide range of parallel I/O standards.

The LatticeECP3 FPGA fabric is optimized with high performance and low cost in mind. The LatticeECP3 devices utilize reconfigurable SRAM logic technology and provide popular building blocks such as LUT-based logic, distributed and embedded memory, Phase Locked Loops (PLLs), Delay Locked Loops (DLLs), pre-engineered source synchronous I/O support, enhanced sysDSP slices and advanced configuration support, including encryption and dual-boot capabilities.

The pre-engineered source synchronous logic implemented in the LatticeECP3 device family supports a broad range of interface standards, including DDR3, XGMII and 7:1 LVDS.

The LatticeECP3 device family also features high speed SERDES with dedicated PCS functions. High jitter tolerance and low transmit jitter allow the SERDES plus PCS blocks to be configured to support an array of popular data protocols including PCI Express, SMPTE, Ethernet (XAUI, GbE, and SGMII) and CPRI. Transmit Pre-emphasis and Receive Equalization settings make the SERDES suitable for transmission and reception over various forms of media.

The LatticeECP3 devices also provide flexible, reliable and secure configuration options, such as dual-boot capability, bitstream encryption, and TransFR field upgrade features.

The Lattice Diamond™ and ispLEVER® design software allows large complex designs to be efficiently implemented using the LatticeECP3 FPGA family. Synthesis library support for LatticeECP3 is available for popular logic synthesis tools. Diamond and ispLEVER tools use the synthesis tool output along with the constraints from its floor planning tools to place and route the design in the LatticeECP3 device. The tools extract the timing from the routing and back-annotate it into the design for timing verification.

Lattice provides many pre-engineered IP (Intellectual Property) modules for the LatticeECP3 family. By using these configurable soft core IPs as standardized blocks, designers are free to concentrate on the unique aspects of their design, increasing their productivity.



# 2. Architecture

### 2.1. Architecture Overview

Each LatticeECP3 device contains an array of logic blocks surrounded by Programmable I/O Cells (PIC). Interspersed between the rows of logic blocks are rows of sysMEM™ Embedded Block RAM (EBR) and rows of sysDSP™ Digital Signal Processing slices, as shown in Figure 2.1. The LatticeECP3-150 has four rows of DSP slices all other LatticeECP3 devices have two rows of DSP slices. In addition, the LatticeECP3 family contains SERDES Quads on the bottom of the device.

There are two kinds of logic blocks, the Programmable Functional Unit (PFU) and Programmable Functional Unit without RAM (PFF). The PFU contains the building blocks for logic, arithmetic, RAM and ROM functions. The PFF block contains building blocks for logic, arithmetic and ROM functions. Both PFU and PFF blocks are optimized for flexibility, allowing complex designs to be implemented quickly and efficiently. Logic Blocks are arranged in a twodimensional array. Only one type of block is used per row.

The LatticeECP3 devices contain one or more rows of sysMEM EBR blocks. sysMEM EBRs are large, dedicated 18Kbit fast memory blocks. Each sysMEM block can be configured in a variety of depths and widths as RAM or ROM. In addition, LatticeECP3 devices contain up to two rows of DSP slices. Each DSP slice has multipliers and adder/accumulators, which are the building blocks for complex signal processing capabilities.

The LatticeECP3 devices feature up to 16 embedded 3.2 Gbps SERDES (Serializer / Deserializer) channels. Each SERDES channel contains independent 8b/10b encoding / decoding, polarity adjust and elastic buffer logic. Each group of four SERDES channels, along with its Physical Coding Sub-layer (PCS) block, creates a quad. The functionality of the SERDES/PCS quads can be controlled by memory cells set during device configuration or by registers that are addressable during device operation. The registers in every quad can be programmed via the SERDES Client Interface (SCI). These quads (up to four) are located at the bottom of the devices.

Each PIC block encompasses two PIOs (PIO pairs) with their respective sysI/O buffers. The sysI/O buffers of the LatticeECP3 devices are arranged in seven banks, allowing the implementation of a wide variety of I/O standards. In addition, a separate I/O bank is provided for the programming interfaces. 50% of the PIO pairs on the left and right edges of the device can be configured as LVDS transmit/receive pairs. The PIC logic also includes pre-engineered support to aid in the implementation of high speed source synchronous standards such as XGMII, 7:1 LVDS, along with memory interfaces including DDR3.

The LatticeECP3 registers in PFU and sysI/O can be configured to be SET or RESET. After power up and the device is configured, it enters into user mode with these registers SET/RESET according to the configuration setting, allowing the device entering to a known state for predictable system function.

Other blocks provided include PLLs, DLLs and configuration functions. The LatticeECP3 architecture provides two Delay Locked Loops (DLLs) and up to ten Phase Locked Loops (PLLs). The PLL and DLL blocks are located at the end of the EBR/DSP rows.

The configuration block that supports features such as configuration bit-stream decryption, transparent updates and dual-boot support is located toward the center of this EBR row. Every device in the LatticeECP3 family supports a sysCONFIG™ port located in the corner between banks one and two, which allows for serial or parallel device configuration.

In addition, every device in the family has a JTAG port. This family also provides an on-chip oscillator and soft error detect capability. The LatticeECP3 devices use 1.2 V as their core voltage.



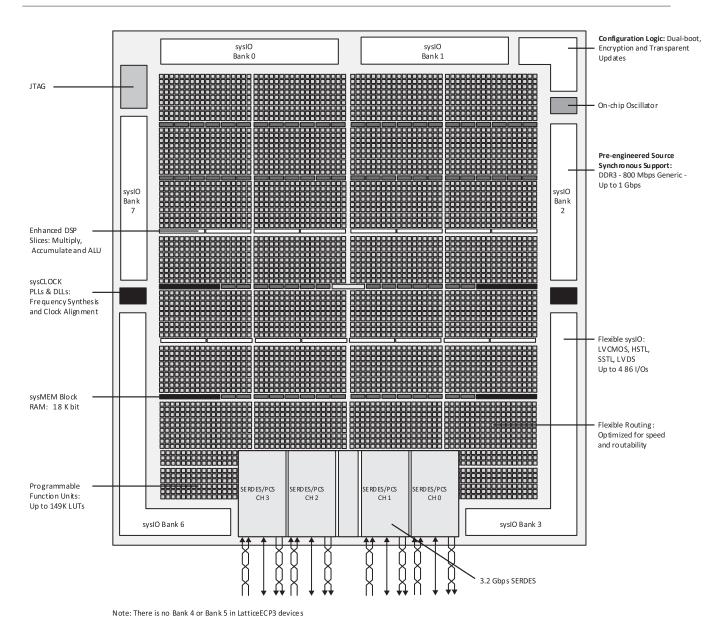


Figure 2.1. Simplified Block Diagram, LatticeECP3-35 Device (Top Level)

# 2.2. PFU Blocks

The core of the LatticeECP3 device consists of PFU blocks, which are provided in two forms, the PFU and PFF. The PFUs can be programmed to perform Logic, Arithmetic, Distributed RAM and Distributed ROM functions. PFF blocks can be programmed to perform Logic, Arithmetic and ROM functions. Except where necessary, the remainder of this data sheet will use the term PFU to refer to both PFU and PFF blocks.

Each PFU block consists of four interconnected slices numbered 0-3 as shown in Figure 2.2. Each slice contains two LUTs. All the interconnections to and from PFU blocks are from routing. There are 50 inputs and 23 outputs associated with each PFU block.



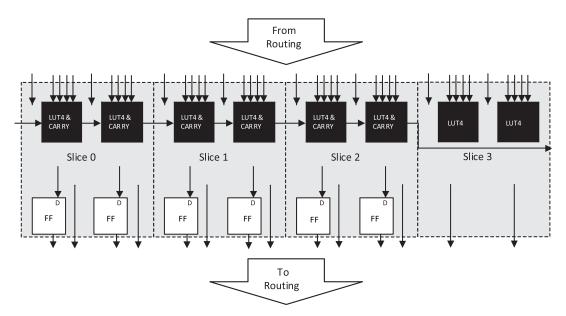


Figure 2.2. PFU Diagram

# 2.2.1. Slice

Slice 0 through Slice 2 contain two LUT4s feeding two registers, whereas Slice 3 contains two LUT4s only. For PFUs, Slice 0 through Slice 2 can be configured as distributed memory, a capability not available in the PFF. Table 2.1 shows the capability of the slices in both PFF and PFU blocks along with the operation modes they enable. In addition, each PFU contains logic that allows the LUTs to be combined to perform functions such as LUT5, LUT6, LUT7 and LUT8. There is control logic to perform set/reset functions (programmable as synchronous/ asynchronous), clock select, chip-select and wider RAM/ROM functions.

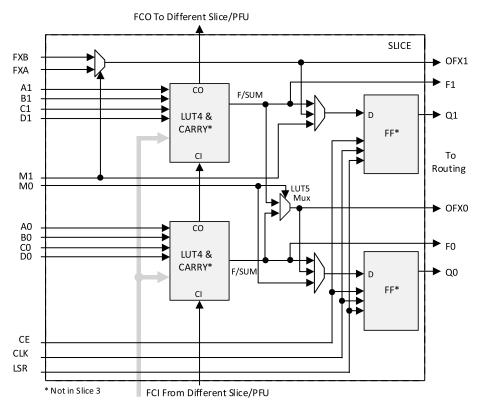
Table 2.1. Resources and Modes Available per Slice

Slice	PFU BLock		PFF Block		
	Resources	Modes	Resources	Modes	
Slice 0	2 LUT4s and 2 Registers	Logic, Ripple, RAM, ROM	2 LUT4s and 2 Registers	Logic, Ripple, ROM	
Slice 1	2 LUT4s and 2 Registers	Logic, Ripple, RAM, ROM	2 LUT4s and 2 Registers	Logic, Ripple, ROM	
Slice 2	2 LUT4s and 2 Registers	Logic, Ripple, RAM, ROM	2 LUT4s and 2 Registers	Logic, Ripple, ROM	
Slice 3	2 LUT4s	Logic, ROM	2 LUT4s	Logic, ROM	

Figure 2.3 shows an overview of the internal logic of the slice. The registers in the slice can be configured for positive/ negative and edge triggered or level sensitive clocks.

Slices 0, 1 and 2 have 14 input signals: 13 signals from routing and one from the carry-chain (from the adjacent slice or PFU). There are seven outputs: six to routing and one to carry-chain (to the adjacent PFU). Slice 3 has 10 input signals from routing and four signals to routing. Table 2.2 lists the signals associated with Slice 0 to Slice 2.





For Slices 0 and 1, memory control signals are generated from Slice 2 as follows:

WCK is CLK

WRE is from LSR  $\,$ 

 $\mbox{DI[3:2]}$  for Slice 1 and  $\mbox{DI[1:0]}$  for Slice 0 data from Slice 2

WAD [A:D] is a 4-bit address from slice 2 LUT input

Figure 2.3. Slice Diagram

**Table 2.2. Slice Signal Descriptions** 

Function	Туре	Signal Names	Description
Input	Data signal	A0, B0, C0, D0	Inputs to LUT4
Input	Data signal	A1, B1, C1, D1	Inputs to LUT4
Input	Multi-purpose	M0	Multipurpose Input
Input	Multi-purpose	M1	Multipurpose Input
Input	Control signal	CE	Clock Enable
Input	Control signal	LSR	Local Set/Reset
Input	Control signal	CLK	System Clock
Input	Inter-PFU signal	FC	Fast Carry-in <sup>1</sup>
Input	Inter-slice signal	FXA	Intermediate signal to generate LUT6 and LUT7
Input	Inter-slice signal	FXB	Intermediate signal to generate LUT6 and LUT7
Output	Data signals	F0, F1	LUT4 output register bypass signals
Output	Data signals	Q0, Q1	Register outputs
Output	Data signals	OFX0	Output of a LUT5 MUX
Output	Data signals	OFX1	Output of a LUT6, LUT7, LUT8 <sup>2</sup> MUX depending on the slice
Output	Inter-PFU signal	FCO	Slice 2 of each PFU is the fast carry chain output <sup>1</sup>

#### Notes:

- 1. See Figure 2.3 for connection details.
- 2. Requires two PFUs.

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#### 2.2.2. Modes of Operation

Each slice has up to four potential modes of operation: Logic, Ripple, RAM and ROM.

#### **Logic Mode**

In this mode, the LUTs in each slice are configured as 4-input combinatorial lookup tables. A LUT4 can have 16 possible input combinations. Any four input logic functions can be generated by programming this lookup table. Since there are two LUT4s per slice, a LUT5 can be constructed within one slice. Larger look-up tables such as LUT6, LUT7 and LUT8 can be constructed by concatenating other slices. Note LUT8 requires more than four slices.

#### **Ripple Mode**

Ripple mode supports the efficient implementation of small arithmetic functions. In ripple mode, the following functions can be implemented by each slice:

- Addition 2-bit
- Subtraction 2-bit
- Add/Subtract 2-bit using dynamic control
- Up counter 2-bit
- Down counter 2-bit
- Up/Down counter with asynchronous clear
- Up/Down counter with preload (sync)
- Ripple mode multiplier building block
- Multiplier support
- Comparator functions of A and B inputs
  - A greater-than-or-equal-to B
  - A not-equal-to B
  - A less-than-or-equal-to B

Ripple Mode includes an optional configuration that performs arithmetic using fast carry chain methods. In this configuration (also referred to as CCU2 mode) two additional signals, Carry Generate and Carry Propagate, are generated on a per slice basis to allow fast arithmetic functions to be constructed by concatenating Slices.

#### **RAM Mode**

In this mode, a 16x4-bit distributed single port RAM (SPR) can be constructed using each LUT block in Slice 0 and Slice 1 as a 16x1-bit memory. Slice 2 is used to provide memory address and control signals. A 16x2-bit pseudo dual port RAM (PDPR) memory is created by using one Slice as the read-write port and the other companion slice as the read-only port.

LatticeECP3 devices support distributed memory initialization. The Lattice design tools support the creation of a variety of different size memories. Where appropriate, the software will construct these using distributed memory primitives that represent the capabilities of the PFU. Table 2.3 shows the number of slices required to implement different distributed RAM primitives. For more information about using RAM in LatticeECP3 devices, see LatticeECP3 Memory Usage Guide (FPGA-TN-02188).

Table 2.3. Number of Slices Required to Implement Distributed RAM

	SPR 16X4	PDPR 16X4
Number of slices	3	3

Note: SPR = Single Port RAM, PDPR = Pseudo Dual Port RAM

#### **ROM Mode**

ROM mode uses the LUT logic; hence, Slices 0 through 3 can be used in ROM mode. Preloading is accomplished through the programming interface during PFU configuration.

For more information, refer to LatticeECP3 Memory Usage Guide (FPGA-TN-02188).



# 2.3. Routing

There are many resources provided in the LatticeECP3 devices to route signals individually or as busses with related control signals. The routing resources consist of switching circuitry, buffers and metal interconnect (routing) segments.

The LatticeECP3 family has an enhanced routing architecture that produces a compact design. The Diamond and ispLEVER design software tool suites take the output of the synthesis tool and places and routes the design.

# 2.4. sysCLOCK PLLs and DLLs

The sysCLOCK PLLs provide the ability to synthesize clock frequencies. The devices in the LatticeECP3 family support two to ten full-featured General Purpose PLLs.

### 2.4.1. General Purpose PLL

The architecture of the PLL is shown in Figure 2.4. A description of the PLL functionality follows.

CLKI is the reference frequency (generated either from the pin or from routing) for the PLL. CLKI feeds into the Input Clock Divider block. The CLKFB is the feedback signal (generated from CLKOP, CLKOS or from a user clock pin/logic). This signal feeds into the Feedback Divider. The Feedback Divider is used to multiply the reference frequency.

Both the input path and feedback signals enter the Phase Frequency Detect Block (PFD) which detects first for the frequency, and then the phase, of the CLKI and CLKFB are the same which then drives the Voltage Controlled Oscillator (VCO) block. In this block the difference between the input path and feedback signals is used to control the frequency and phase of the oscillator. A LOCK signal is generated by the VCO to indicate that the VCO has locked onto the input clock signal. In dynamic mode, the PLL may lose lock after a dynamic delay adjustment and not relock until the tLOCK parameter has been satisfied.

The output of the VCO then enters the CLKOP divider. The CLKOP divider allows the VCO to operate at higher frequencies than the clock output (CLKOP), thereby increasing the frequency range. The Phase/Duty Cycle/Duty Trim block adjusts the phase and duty cycle of the CLKOS signal. The phase/duty cycle setting can be pre-programmed or dynamically adjusted. A secondary divider takes the CLKOP or CLKOS signal and uses it to derive lower frequency outputs (CLKOK).

The primary output from the CLKOP divider (CLKOP) along with the outputs from the secondary dividers (CLKOK and CLKOK2) and Phase/Duty select (CLKOS) are fed to the clock distribution network.

The PLL allows two methods for adjusting the phase of signal. The first is referred to as Fine Delay Adjustment. This inserts up to 16 nominal 125 ps delays to be applied to the secondary PLL output. The number of steps may be set statically or from the FPGA logic. The second method is referred to as Coarse Phase Adjustment. This allows the phase of the rising and falling edge of the secondary PLL output to be adjusted in 22.5 degree steps. The number of steps may be set statically or from the FPGA logic.



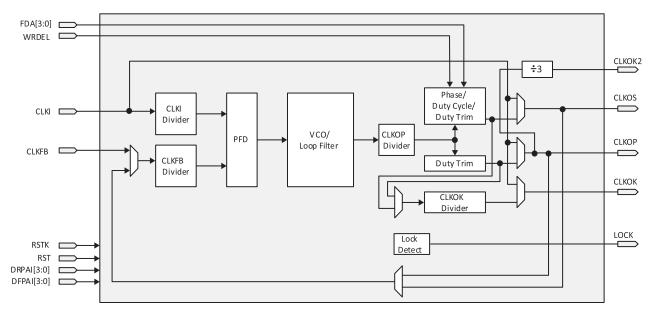


Figure 2.4. General Purpose PLL Diagram

Table 2.4 provides a description of the signals in the PLL blocks.

**Table 2.4. PLL Blocks Signal Descriptions** 

Signal	I/O	Description	
CLKI	1	Clock input from external pin or routing	
CLKFB	1	PLL feedback input from CLKOP, CLKOS, or from a user clock (pin or logic)	
RST	1	"1" to reset PLL counters, VCO, charge pumps and M-dividers	
RSTK	1	"1" to reset K-divider	
WRDEL	1	DPA Fine Delay Adjust input	
CLKOS	0	PLL output to clock tree (phase shifted/duty cycle changed)	
CLKOP	0	PLL output to clock tree (no phase shift)	
CLKOK	0	PLL output to clock tree through secondary clock divider	
CLKOK2	0	PLL output to clock tree (CLKOP divided by 3)	
LOCK	0	"1" indicates PLL LOCK to CLKI	
FDA [3:0]	1	Dynamic fine delay adjustment on CLKOS output	
DRPAI[3:0]	1	Dynamic coarse phase shift, rising edge setting	
DFPAI[3:0]	I	Dynamic coarse phase shift, falling edge setting	

### 2.4.2. Delay Locked Loops (DLL)

In addition to PLLs, the LatticeECP3 family of devices has two DLLs per device.

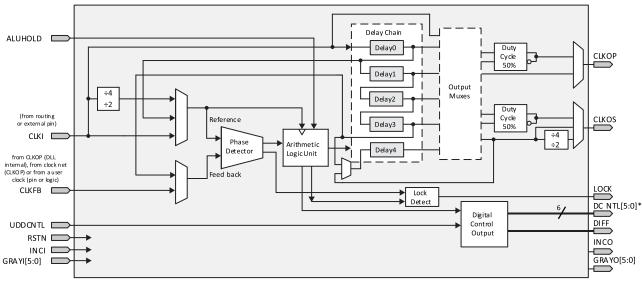
CLKI is the input frequency (generated either from the pin or routing) for the DLL. CLKI feeds into the output muxes block to bypass the DLL, directly to the DELAY CHAIN block and (directly or through divider circuit) to the reference input of the Phase Detector (PD) input mux. The reference signal for the PD can also be generated from the Delay Chain signals. The feedback input to the PD is generated from the CLKFB pin or from a tapped signal from the Delay chain.

The PD produces a binary number proportional to the phase and frequency difference between the reference and feedback signals. Based on these inputs, the ALU determines the correct digital control codes to send to the delay chain in order to better match the reference and feedback signals. This digital code from the ALU is also transmitted via the Digital Control bus (DCNTL) bus to its associated Slave Delay lines (two per DLL). The ALUHOLD input allows the user to suspend the ALU output at its current value. The UDDCNTL signal allows the user to latch the current value on the DCNTL bus.



The DLL has two clock outputs, CLKOP and CLKOS. These outputs can individually select one of the outputs from the tapped delay line. The CLKOS has optional fine delay shift and divider blocks to allow this output to be further modified, if required. The fine delay shift block allows the CLKOS output to phase shifted a further 45, 22.5 or 11.25 degrees relative to its normal position. Both the CLKOS and CLKOP outputs are available with optional duty cycle correction. Divide by two and divide by four frequencies are available at CLKOS. The LOCK output signal is asserted when the DLL is locked. Figure 2.5 shows the DLL block diagram and Table 2.5 provides a description of the DLL inputs and outputs.

The user can configure the DLL for many common functions such as time reference delay mode and clock injection removal mode. Lattice provides primitives in its design tools for these functions.



<sup>\*</sup> This signal is not user accessible. This can only be used to feed the slave delay line.

Figure 2.5. Delay Locked Loop Diagram (DLL)

Table 2.5. DLL Signals

Signal	1/0	Description	
CLKI	İ	Clock input from external pin or routing	
CLKFB	1	DLL feed input from DLL output, clock net, routing or external pin	
RSTN	1	Active low synchronous reset	
ALUHOLD	1	Active high freezes the ALU	
UDDCNTL	1	Synchronous enable signal (hold high for two cycles) from routing	
CLKOP	0	The primary clock output	
CLKOS	0	The secondary clock output with fine delay shift and/or division by 2 or by 4	
LOCK	0	Active high phase lock indicator	
INCI	1	Incremental indicator from another DLL via CIB.	
GRAYI[5:0]	1	Gray-coded digital control bus from another DLL in time reference mode.	
DIFF	0	Difference indicator when DCNTL is difference than the internal setting and update is needed.	
INCO	0	Incremental indicator to other DLLs via CIB.	
GRAYO[5:0]	0	Gray-coded digital control bus to other DLLs via CIB	

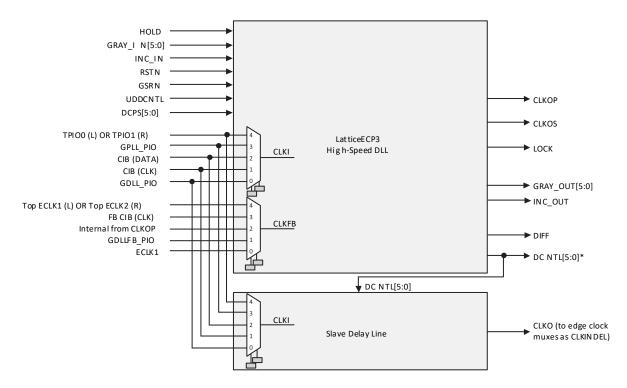
LatticeECP3 devices have two general DLLs and four Slave Delay lines, two per DLL. The DLLs are in the lowest EBR row and located adjacent to the EBR. Each DLL replaces one EBR block. One Slave Delay line is placed adjacent to the DLL and the duplicate Slave Delay line (in Figure 2.6) for the DLL is placed in the I/O ring between Banks 6 and 7 and Banks 2 and 3.

The outputs from the DLL and Slave Delay lines are fed to the clock distribution network.

For more information, see LatticeECP3 sysCLOCK PLL/DLL Design and Usage Guide (FPGA-TN-02191).

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st This signal is not user accessible. It can only be used to feed the slave delay line.

Figure 2.6. Top-Level Block Diagram, High-Speed DLL and Slave Delay Line

## 2.4.3. PLL/DLL Cascading

LatticeECP3 devices have been designed to allow certain combinations of PLL and DLL cascading. The allowable combinations are:

- PLL to PLL supported
- PLL to DLL supported

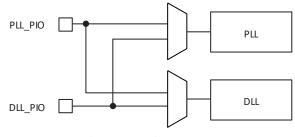
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The DLLs in the LatticeECP3 are used to shift the clock in relation to the data for source synchronous inputs. PLLs are used for frequency synthesis and clock generation for source synchronous interfaces. Cascading PLL and DLL blocks allows applications to utilize the unique benefits of both DLLs and PLLs.

For further information about the DLL, see the list of technical documentation at the end of this data sheet.

## 2.4.4. PLL/DLL PIO Input Pin Connections

All LatticeECP3 devices contains two DLLs and up to ten PLLs, arranged in quadrants. If a PLL and a DLL are next to each other, they share input pins as shown in the Figure 2.7.



Note: Not every PLL has an associated DLL.

Figure 2.7. Sharing of PIO Pins by PLLs and DLLs in LatticeECP3 Devices

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## 2.5. Clock Dividers

LatticeECP3 devices have two clock dividers, one on the left side and one on the right side of the device. These are intended to generate a slower-speed system clock from a high-speed edge clock. The block operates in a ÷2, ÷4 or ÷8 mode and maintains a known phase relationship between the divided down clock and the high-speed clock based on the release of its reset signal. The clock dividers can be fed from selected PLL/DLL outputs, the Slave Delay lines, routing or from an external clock input. The clock divider outputs serve as primary clock sources and feed into the clock distribution network. The Reset (RST) control signal resets input and asynchronously forces all outputs to low. The RELEASE signal releases outputs synchronously to the input clock. For further information on clock dividers, see LatticeECP3 sysCLOCK PLL/DLL Design and Usage Guide (FPGA-TN-02191). Figure 2.8 shows the clock divider connections.

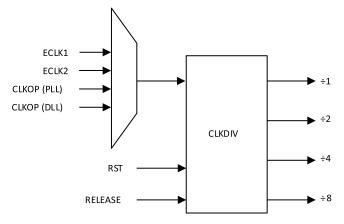


Figure 2.8. Clock Divider Connections

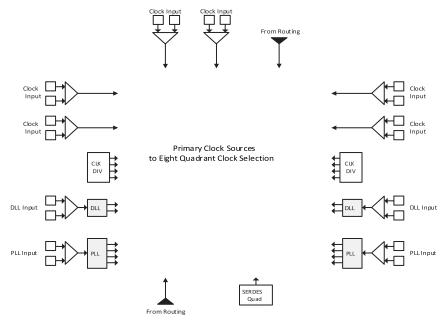
#### 2.6. Clock Distribution Network

LatticeECP3 devices have eight quadrant-based primary clocks and eight secondary clock/control sources. Two high performance edge clocks are available on the top, left, and right edges of the device to support high speed interfaces. These clock sources are selected from external I/O, the sysCLOCK PLLs, DLLs or routing. These clock sources are fed throughout the chip via a clock distribution system.

#### 2.6.1. Primary Clock Sources

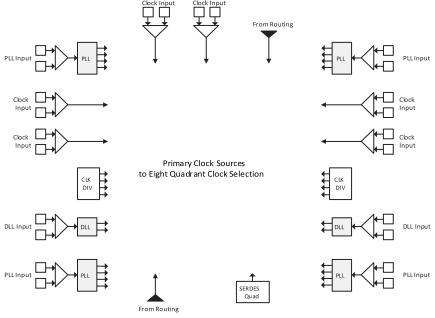
LatticeECP3 devices derive clocks from six primary source types: PLL outputs, DLL outputs, CLKDIV outputs, dedicated clock inputs, routing and SERDES Quads. LatticeECP3 devices have two to ten sysCLOCK PLLs and two DLLs, located on the left and right sides of the device. There are six dedicated clock inputs: two on the top side, two on the left side and two on the right side of the device. Figure 2.9, Figure 2.10 and Figure 2.11 show the primary clock sources for LatticeECP3 devices.





Note: Clock inputs can be configured in differential or single-ended mode.

Figure 2.9. Primary Clock Sources for LatticeECP3-17



Note: Clock in puts can be configured in differential or single-ended mode.

Figure 2.10. Primary Clock Sources for LatticeECP3-35



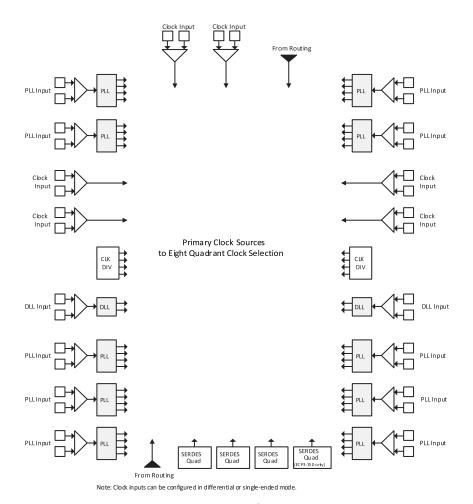


Figure 2.11. Primary Clock Sources for LatticeECP3-70, -95, -150

### 2.6.2. Primary Clock Routing

The purpose of the primary clock routing is to distribute primary clock sources to the destination quadrants of the device. A global primary clock is a primary clock that is distributed to all quadrants. The clock routing structure in LatticeECP3 devices consists of a network of eight primary clock lines (CLKO through CLK7) per quadrant. The primary clocks of each quadrant are generated from muxes located in the center of the device. All the clock sources are connected to these muxes. Figure 2.12 shows the clock routing for one quadrant. Each quadrant mux is identical. If desired, any clock can be routed globally.

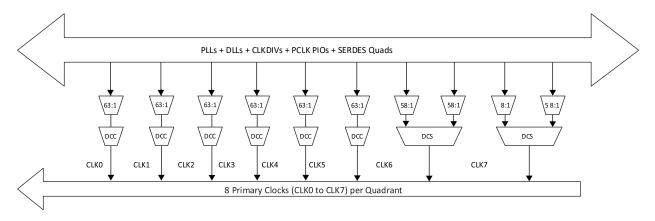


Figure 2.12. Per Quadrant Primary Clock Selection

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## 2.6.3. Dynamic Clock Control (DCC)

The DCC (Quadrant Clock Enable/Disable) feature allows internal logic control of the quadrant primary clock network. When a clock network is disabled, all the logic fed by that clock does not toggle, reducing the overall power consumption of the device.

### 2.6.4. Dynamic Clock Select (DCS)

The DCS is a smart multiplexer function available in the primary clock routing. It switches between two independent input clock sources without any glitches or runt pulses. This is achieved regardless of when the select signal is toggled. There are two DCS blocks per quadrant; in total, there are eight DCS blocks per device. The inputs to the DCS block come from the center muxes. The output of the DCS is connected to primary clocks CLK6 and CLK7 (see Figure 2.12).

Figure 2.13 shows the timing waveforms of the default DCS operating mode. The DCS block can be programmed to other modes. For more information about the DCS, see the list of technical documentation at the end of this data sheet.

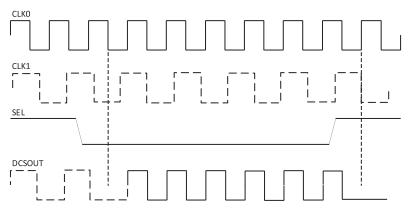


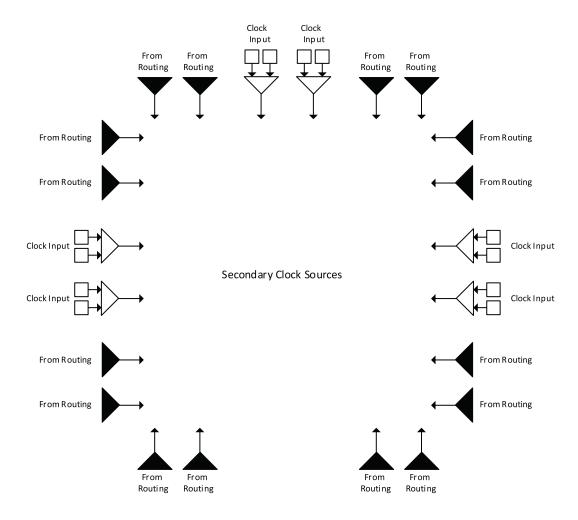
Figure 2.13. DCS Waveforms

## 2.6.5. Secondary Clock/Control Sources

LatticeECP3 devices derive eight secondary clock sources (SCO through SC7) from six dedicated clock input pads and the rest from routing. Figure 2.14 shows the secondary clock sources. All eight secondary clock sources are defined as inputs to a per-region mux SCO-SC7. SCO-SC3 are primary for control signals (CE and/or LSR), and SC4-SC7 are for the clock.

In an actual implementation, there is some overlap to maximize routability. In addition to SCO-SC3, SC7 is also an input to the control signals (LSR or CE). SCO-SC2 are also inputs to clocks along with SC4-SC7.





Note: Clock inputs can be configured in differential or single-ended mode.

Figure 2.14. Secondary Clock Sources

# 2.6.6. Secondary Clock/Control Routing

Global secondary clock is a secondary clock that is distributed to all regions. The purpose of the secondary clock routing is to distribute the secondary clock sources to the secondary clock regions. Secondary clocks in the LatticeECP3 devices are region-based resources. Certain EBR rows and special vertical routing channels bind the secondary clock regions. This special vertical routing channel aligns with either the left edge of the center DSP slice in the DSP row or the center of the DSP row. Figure 2.15 shows this special vertical routing channel and the 20 secondary clock regions for the LatticeECP3 family of devices. All devices in the LatticeECP3 family have eight secondary clock resources per region (SC0 to SC7). The same secondary clock routing can be used for control signals.

**Table 2.6. Secondary Clock Regions** 

Table Elot decondary clock hegions				
Device	Number of Secondary Clock Regions			
ECP3-17	16			
ECP3-35	16			
ECP3-70	20			
ECP3-95	20			
ECP3-150	36			

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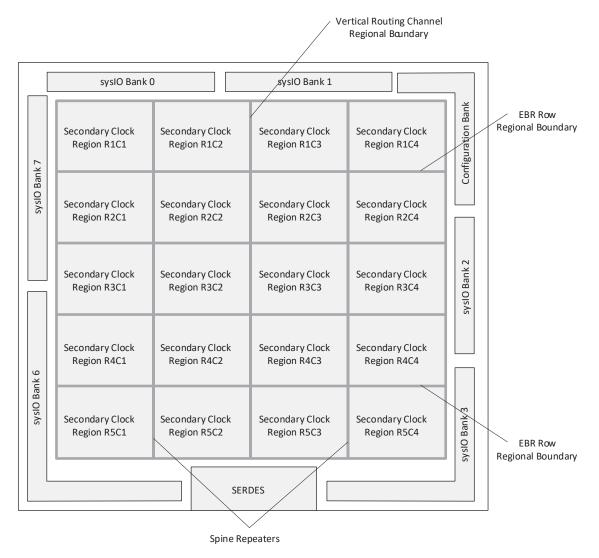


Figure 2.15. LatticeECP3-70 and LatticeECP3-95 Secondary Clock Regions

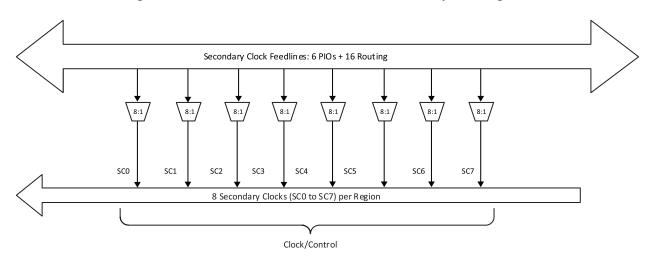


Figure 2.16. Per Region Secondary Clock Selection



#### 2.6.7. Slice Clock Selection

Figure 2.17 shows the clock selections and Figure 2.18 shows the control selections for Slice 0 through Slice 2. All the primary clocks and seven secondary clocks are routed to this clock selection mux. Other signals can be used as a clock input to the slices via routing. Slice controls are generated from the secondary clocks/controls or other signals connected via routing.

If none of the signals are selected for both clock and control then the default value of the mux output is 1. Slice 3 does not have any registers; therefore it does not have the clock or control muxes.

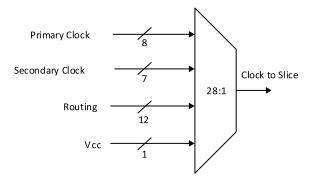


Figure 2.17. Slice0 through Slice2 Clock Selection

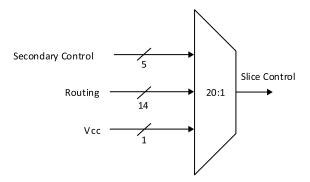


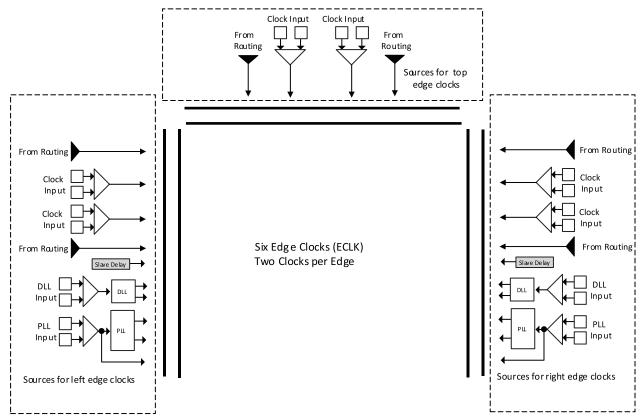
Figure 2.18. Slice0 through Slice2 Control Selection

# 2.6.8. Edge Clock Sources

Edge clock resources can be driven from a variety of sources at the same edge. Edge clock resources can be driven from adjacent edge clock PIOs, primary clock PIOs, PLLs, DLLs, Slave Delay and clock dividers as shown in Figure 2.19.

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Notes:

- 1. Clock inputs can be configured in differential or single ended mode.
- 2. The two DLLs can also drive the two top edge clocks.
- 3. The top left and top right PLL can also drive the two top edge clocks.

Figure 2.19. Edge Clock Sources

# 2.6.9. Edge Clock Routing

LatticeECP3 devices have a number of high-speed edge clocks that are intended for use with the PIOs in the implementation of high-speed interfaces. There are six edge clocks per device: two edge clocks on each of the top, left, and right edges. Different PLL and DLL outputs are routed to the two muxes on the left and right sides of the device. In addition, the CLKINDEL signal (generated from the DLL Slave Delay Line block) is routed to all the edge clock muxes on the left and right sides of the device. Figure 2.20 shows the selection muxes for these clocks.



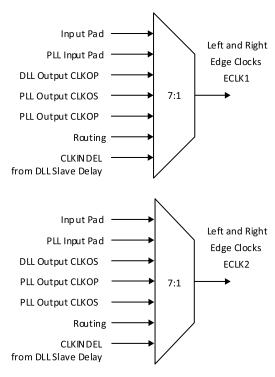


Figure 2.20. Sources of Edge Clock (Left and Right Edges)

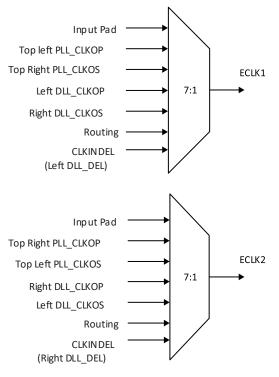


Figure 2.21. Sources of Edge Clock (Top Edge)

The edge clocks have low injection delay and low skew. They are used to clock the I/O registers and thus are ideal for creating I/O interfaces with a single clock signal and a wide data bus. They are also used for DDR Memory or Generic DDR interfaces.

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The edge clocks on the top, left, and right sides of the device can drive the secondary clocks or general routing resources of the device. The left and right side edge clocks also can drive the primary clock network through the clock dividers (CLKDIV).

# 2.7. sysMEM Memory

LatticeECP3 devices contain a number of sysMEM Embedded Block RAM (EBR). The EBR consists of an 18-Kbit RAM with memory core, dedicated input registers and output registers with separate clock and clock enable. Each EBR includes functionality to support true dual-port, pseudo dual-port, single-port RAM, ROM and FIFO buffers (via external PFUs).

## 2.7.1. sysMEM Memory Block

The sysMEM block can implement single port, dual port or pseudo dual port memories. Each block can be used in a variety of depths and widths as shown in Table 2.7. FIFOs can be implemented in sysMEM EBR blocks by implementing support logic with PFUs. The EBR block facilitates parity checking by supporting an optional parity bit for each data byte. EBR blocks provide byte-enable support for configurations with18-bit and 36-bit data widths. For more information, see LatticeECP3 Memory Usage Guide (FPGA-TN-02188).

**Table 2.7. sysMEM Block Configurations** 

Memory Mode	Configurations
Single Port	16,384 x 1
	8,192 x 2
	4,096 x 4
	2,048 x 9
	1,024 x 18
	512 x 36
	16,384 x 1
	8,192 x 2
True Dual Port	4,096 x 4
	2,048 x 9
	1,024 x 18
	16,384 x 1
	8,192 x 2
Pseudo Dual Port	4,096 x 4
Pseudo Duai Port	
	2,048 x 9
	1,024 x 18
	512 x 36

### 2.7.2. Bus Size Matching

All of the multi-port memory modes support different widths on each of the ports. The RAM bits are mapped LSB word 0 to MSB word 0, LSB word 1 to MSB word 1, and so on. Although the word size and number of words for each port varies, this mapping scheme applies to each port.

### 2.7.3. RAM Initialization and ROM Operation

If desired, the contents of the RAM can be pre-loaded during device configuration. By preloading the RAM block during the chip configuration cycle and disabling the write controls, the sysMEM block can also be utilized as a ROM.

# 2.7.4. Memory Cascading

Larger and deeper blocks of RAM can be created using EBR sysMEM Blocks. Typically, the Lattice design tools cascade memory transparently, based on specific design inputs.

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### 2.7.5. Single, Dual and Pseudo-Dual Port Modes

In all the sysMEM RAM modes the input data and address for the ports are registered at the input of the memory array. The output data of the memory is optionally registered at the output.

EBR memory supports the following forms of write behavior for single port or dual port operation:

- 1. **Normal** Data on the output appears only during a read cycle. During a write cycle, the data (at the current address) does not appear on the output. This mode is supported for all data widths.
- 2. **Write Through** A copy of the input data appears at the output of the same port during a write cycle. This mode is supported for all data widths.
- 3. **Read-Before-Write (EA devices only)** When new data is written, the old content of the address appears at the output. This mode is supported for x9, x18, and x36 data widths.

### 2.7.6. Memory Core Reset

The memory array in the EBR utilizes latches at the A and B output ports. These latches can be reset asynchronously or synchronously. RSTA and RSTB are local signals, which reset the output latches associated with Port A and Port B, respectively. The Global Reset (GSRN) signal can reset both ports. The output data latches and associated resets for both ports are as shown in Figure 2.22.

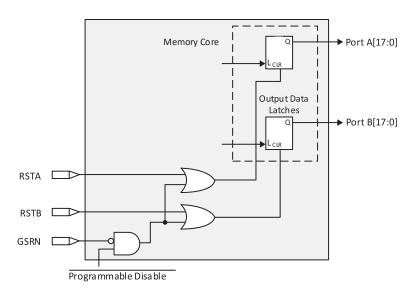


Figure 2.22. Memory Core Reset

For further information on the sysMEM EBR block, see the list of technical documentation at the end of this data sheet.

# 2.8. sysDSP™ Slice

The LatticeECP3 family provides an enhanced sysDSP architecture, making it ideally suited for low-cost, high-performance Digital Signal Processing (DSP) applications. Typical functions used in these applications are Finite Impulse Response (FIR) filters, Fast Fourier Transforms (FFT) functions, Correlators, Reed-Solomon/Turbo/Convolution encoders and decoders. These complex signal processing functions use similar building blocks such as multiply-adders and multiply-accumulators.

#### 2.8.1. sysDSP Slice Approach Compared to General DSP

Conventional general-purpose DSP chips typically contain one to four (Multiply and Accumulate) MAC units with fixed data-width multipliers; this leads to limited parallelism and limited throughput. Their throughput is increased by higher clock speeds. The LatticeECP3, on the other hand, has many DSP slices that support different data widths.

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This allows designers to use highly parallel implementations of DSP functions. Designers can optimize DSP performance vs. area by choosing appropriate levels of parallelism. Figure 2.23 compares the fully serial implementation to the mixed parallel and serial implementation.

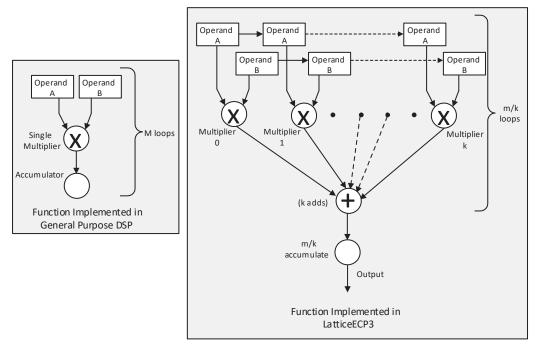


Figure 2.23. Comparison of General DSP and LatticeECP3 Approaches

# 2.9. LatticeECP3 sysDSP Slice Architecture Features

The LatticeECP3 sysDSP Slice has been significantly enhanced to provide functions needed for advanced processing applications. These enhancements provide improved flexibility and resource utilization.

The LatticeECP3 sysDSP Slice supports many functions that include the following:

- Multiply (one 18 x 36, two 18 x 18 or four 9 x 9 Multiplies per Slice)
- Multiply (36 x 36 by cascading across two sysDSP slices)
- Multiply Accumulate (up to 18 x 36 Multipliers feeding an Accumulator that can have up to 54-bit resolution)
- Two Multiplies feeding one Accumulate per cycle for increased processing with lower latency (two 18 x 18 Multiplies feed into an accumulator that can accumulate up to 52 bits)
- Flexible saturation and rounding options to satisfy a diverse set of applications situations
- Flexible cascading across DSP slices
  - Minimizes fabric use for common DSP and ALU functions
  - Enables implementation of FIR Filter or similar structures using dedicated sysDSP slice resources only
  - Provides matching pipeline registers
  - · Can be configured to continue cascading from one row of sysDSP slices to another for longer cascade chains
- Flexible and Powerful Arithmetic Logic Unit (ALU) Supports:
  - Dynamically selectable ALU OPCODE
  - Ternary arithmetic (addition/subtraction of three inputs)
  - Bit-wise two-input logic operations (AND, OR, NAND, NOR, XOR and XNOR)
  - Eight flexible and programmable ALU flags that can be used for multiple pattern detection scenarios, such as, overflow, underflow and convergent rounding, etc.
  - Flexible cascading across slices to get larger functions
- RTL Synthesis friendly synchronous reset on all registers, while still supporting asynchronous reset for legacy users
- Dynamic MUX selection to allow Time Division Multiplexing (TDM) of resources for applications that require processor-like flexibility that enables different functions for each clock cycle



For most cases, as shown in Figure 2.24, the LatticeECP3 DSP slice is backwards-compatible with the LatticeECP2™ sysDSP block, such that, legacy applications can be targeted to the LatticeECP3 sysDSP slice. The functionality of one LatticeECP2 sysDSP Block can be mapped into two adjacent LatticeECP3 sysDSP slices, as shown in Figure 2.25.

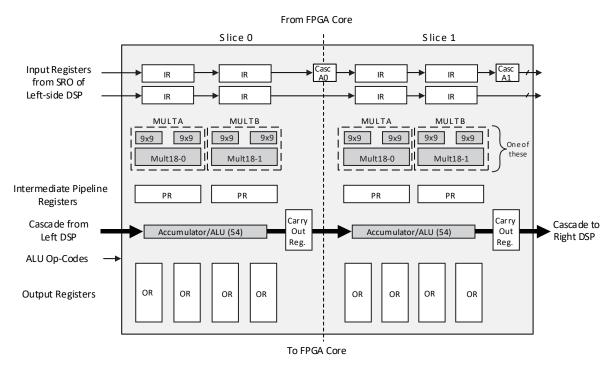
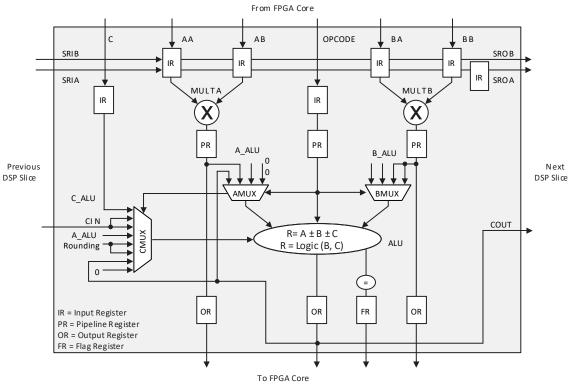


Figure 2.24. Simplified sysDSP Slice Block Diagram



Note: A\_ALU, B\_ALU and C\_ALU are internal signals generated by combining bits from AA, AB, BA BB and C inputs. See TN1182, LatticeECP3 sysDSP Usage Guide, for further information.

Figure 2.25. Detailed sysDSP Slice Diagram



The LatticeECP3 sysDSP block supports the following basic elements.

- MULT (Multiply)
- MAC (Multiply, Accumulate)
- MULTADDSUB (Multiply, Addition/Subtraction)
- MULTADDSUBSUM (Multiply, Addition/Subtraction, Summation)

Table 2.8 shows the capabilities of each of the LatticeECP3 slices versus the above functions.

Table 2.8. Maximum Number of Elements in a Slice

Width of Multiply	х9	x18	x36
MULT	4	2	1/2
MAC	1	1	_
MULTADDSUB	2	1	_
MULTADDSUBSUM	1 <sup>1</sup>	1/2	_

Note: One slice can implement 1/2 9x9 m9x9addsubsum and two m9x9addsubsum with two slices.

Some options are available in the four elements. The input register in all the elements can be directly loaded or can be loaded as a shift register from previous operand registers. By selecting "dynamic operation" the following operations are possible:

- In the Add/Sub option the Accumulator can be switched between addition and subtraction on every cycle.
- The loading of operands can switch between parallel and serial operations.

For further information, refer to LatticeECP3 sysDSP Usage Guide (FPGA-TN-02193).

#### 2.9.1. MULT DSP Element

This multiplier element implements a multiply with no addition or accumulator nodes. The two operands, AA and AB, are multiplied and the result is available at the output. The user can enable the input/output and pipeline registers. Figure 2.26 shows the MULT sysDSP element.

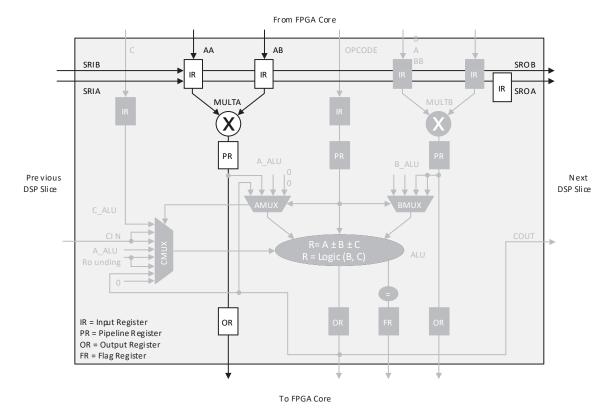


Figure 2.26. MULT sysDSP Element



#### 2.9.2. MAC DSP Element

In this case, the two operands, AA and AB, are multiplied and the result is added with the previous accumulated value. This accumulated value is available at the output. The user can enable the input and pipeline registers, but the output register is always enabled. The output register is used to store the accumulated value. The ALU is configured as the accumulator in the sysDSP slice in the LatticeECP3 family can be initialized dynamically. A registered overflow signal is also available. The overflow conditions are provided later in this document. Figure 2.27 shows the MAC sysDSP element.

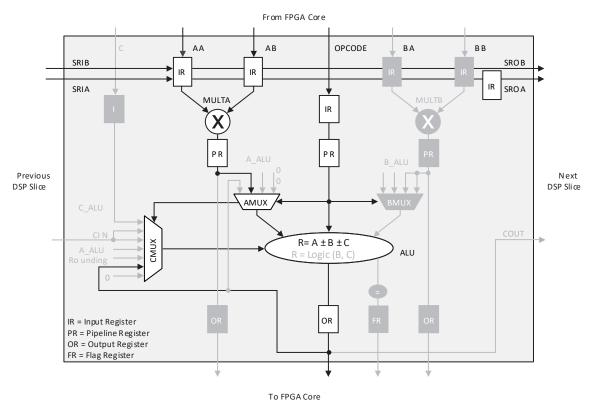


Figure 2.27. MAC DSP Element



#### 2.9.3. MMAC DSP Element

The LatticeECP3 supports a MAC with two multipliers. This is called Multiply Multiply Accumulate or MMAC. In this case, the two operands, AA and AB, are multiplied and the result is added with the previous accumulated value and with the result of the multiplier operation of operands BA and BB. This accumulated value is available at the output. The user can enable the input and pipeline registers, but the output register is always enabled. The output register is used to store the accumulated value. The ALU is configured as the accumulator in the sysDSP slice. A registered overflow signal is also available. The overflow conditions are provided later in this document. Figure 2.28 shows the MMAC sysDSP element.

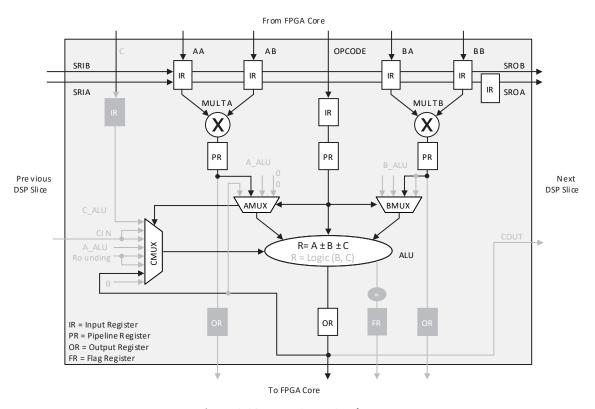


Figure 2.28. MMAC sysDSP Element



### 2.9.4. MULTADDSUB DSP Element

In this case, the operands AA and AB are multiplied and the result is added/subtracted with the result of the multiplier operation of operands BA and BB. The user can enable the input, output and pipeline registers. Figure 2.29 shows the MULTADDSUB sysDSP element.

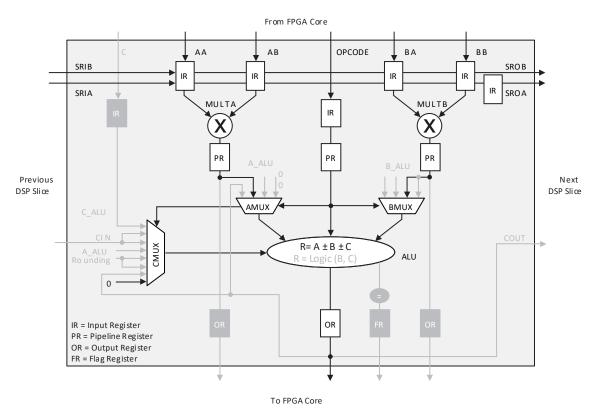


Figure 2.29. MULTADDSUB



#### 2.9.5. MULTADDSUBSUM DSP Element

In this case, the operands AA and AB are multiplied and the result is added/subtracted with the result of the multiplier operation of operands BA and BB of Slice 0. Additionally, the operands AA and AB are multiplied and the result is added/subtracted with the result of the multiplier operation of operands BA and BB of Slice 1. The results of both addition/subtractions are added by the second ALU following the slice cascade path. The user can enable the input, output and pipeline registers. Figure 2.30 and Figure 2.31 show the MULTADDSUBSUM sysDSP element.

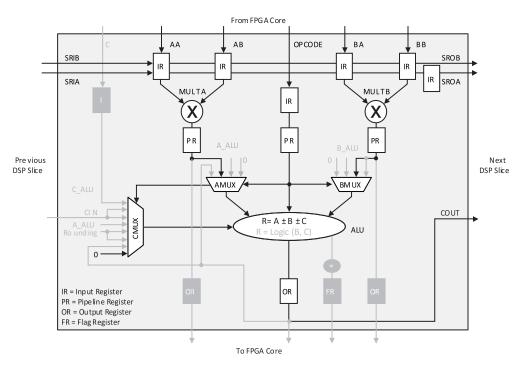


Figure 2.30. MULTADDSUBSUM Slice 0

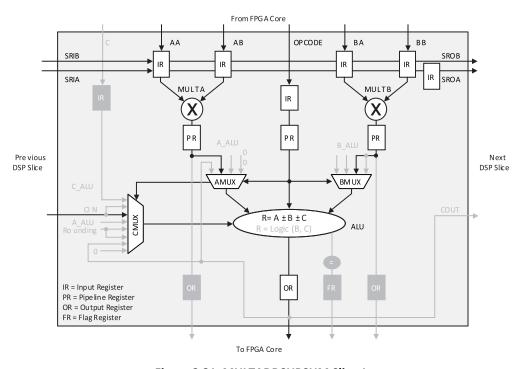


Figure 2.31. MULTADDSUBSUM Slice 1



## 2.10. Advanced sysDSP Slice Features

### 2.10.1. Cascading

The LatticeECP3 sysDSP slice has been enhanced to allow cascading. Adder trees are implemented fully in sysDSP slices, improving the performance. Cascading of slices uses the signals CIN, COUT and C Mux of the slice.

#### 2.10.2. Addition

The LatticeECP3 sysDSP slice allows for the bypassing of multipliers and cascading of adder logic. High performance adder functions are implemented without the use of LUTs. The maximum width adders that can be implemented are 54-bit.

#### 2.10.3. Rounding

The rounding operation is implemented in the ALU and is done by adding a constant followed by a truncation operation.

The rounding methods supported are:

- Rounding to zero (RTZ)
- Rounding to infinity (RTI)
- Dynamic rounding
- Random rounding
- Convergent rounding

## 2.10.4. ALU Flags

The sysDSP slice provides a number of flags from the ALU including:

- Equal to zero (EQZ)
- Equal to zero with mask (EQZM)
- Equal to one with mask (EQOM)
- Equal to pattern with mask (EQPAT)
- Equal to bit inverted pattern with mask (EQPATB)
- Accumulator Overflow (OVER)
- Accumulator Underflow (UNDER)
- Either over or under flow supporting LatticeECP2 legacy designs (OVERUNDER)

### 2.10.5. Clock, Clock Enable and Reset Resources

Global Clock, Clock Enable and Reset signals from routing are available to every sysDSP slice. From four clock sources (CLKO, CLK1, CLK2, and CLK3) one clock is selected for each input register, pipeline register and output register. Similarly Clock Enable (CE) and Reset (RST) are selected at each input register, pipeline register and output register.

#### 2.10.6. Resources Available in the LatticeECP3 Family

Table 2.9 shows the maximum number of multipliers for each member of the LatticeECP3 family. Table 2.10 shows the maximum available EBR RAM Blocks in each LatticeECP3 device. EBR blocks, together with Distributed RAM can be used to store variables locally for fast DSP operations.

Table 2.9. Maximum Number of DSP Slices in the LatticeECP3 Family

Device	DSP Slices	9x9 Multiplier 18x18 Multiplier		36x36 Multiplier
ECP3-17	12	48	24	6
ECP3-35	32	128	64	16
ECP3-70	64	256	128	32
ECP3-95	64	256	128	32
ECP3-150	160	640	320	80

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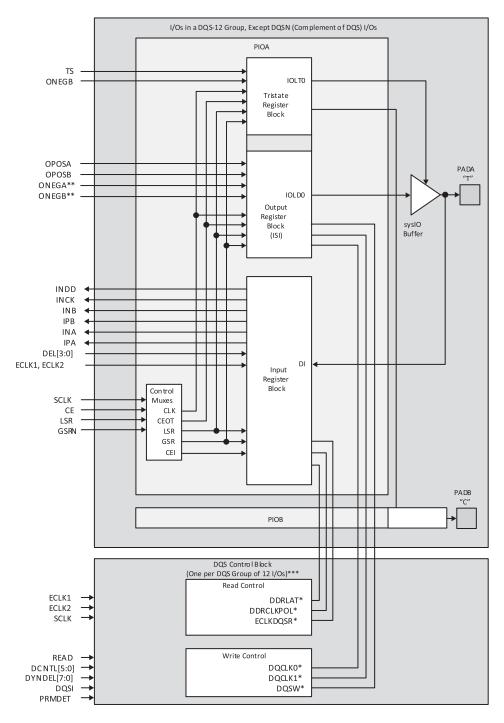
## Table 2.10. Embedded SRAM in the LatticeECP3 Family

Device	EBR SRAM Block	Total EBR SRAM (Kbits)
ECP3-17	38	700
ECP3-35	72	1327
ECP3-70	240	4420
ECP3-95	240	4420
ECP3-150	372	6850



# 3. Programmable I/O Cells (PIC)

Each PIC contains two PIOs connected to their respective sysI/O buffers as shown in Figure 3.1. The PIO Block supplies the output data (DO) and the tri-state control signal (TO) to the sysI/O buffer and receives input from the buffer. Table 3.1 provides the PIO signal list.



<sup>\*</sup> Signals are available on left/right/top edges only.

Figure 3.1. PIC Diagram

<sup>\*\*</sup> Signals are available on the left and right sides only

<sup>\*\* \*</sup> Selected PIO.



Two adjacent PIOs can be joined to provide a differential I/O pair (labeled as "T" and "C") as shown in Figure 3.1. The PAD Labels "T" and "C" distinguish the two PIOs. Approximately 50% of the PIO pairs on the left and right edges of the device can be configured as true LVDS outputs. All I/O pairs can operate as LVDS inputs.

Table 3.1. PIO Signal List

Name	Туре	Description
INDD	Input Data	Register bypassed input. This is not the same port as INCK.
IPA, INA, IPB, INB	Input Data	Ports to core for input data
OPOSA, ONEGA <sup>1</sup> , OPOSB, ONEGB <sup>1</sup>	Output Data	Output signals from core. An exception is the ONEGB port, used for tristate logic at the DQS pad.
CE	PIO Control	Clock enables for input and output block flip-flops.
SCLK	PIO Control	System Clock (PCLK) for input and output/TS blocks. Connected from clock ISB.
LSR	PIO Control	Local Set/Reset
ECLK1, ECLK2	PIO Control	Edge clock sources. Entire PIO selects one of two sources using mux.
ECLKDQSR <sup>1</sup>	Read Control	From DQS_STROBE, shifted strobe for memory interfaces only.
DDRCLKPOL <sup>1</sup>	Read Control	Ensures transfer from DQS domain to SCLK domain.
DDRLAT <sup>1</sup>	Read Control	Used to guarantee INDDRX2 gearing by selectively enabling a D-Flip-Flop in datapath.
DEL[3:0]	Read Control	Dynamic input delay control bits.
INCK	To Clock Distribution and PLL	PIO treated as clock PIO, path to distribute to primary clocks and PLL.
TS	Tristate Data	Tristate signal from core (SDR)
DQCLK0 <sup>1</sup> , DQCLK1 <sup>1</sup>	Write Control	Two clocks edges, 90 degrees out of phase, used in output gearing.
DQSW <sup>2</sup>	Write Control	Used for output and tristate logic at DQS only.
DYNDEL[7:0]	Write Control	Shifting of write clocks for specific DQS group, using 6:0 each step is approximately 25ps, 128 steps. Bit 7 is an invert (timing depends on input frequency). There is also a static control for this 8-bit setting, enabled with a memory cell.
DCNTL[6:0]	PIO Control	Original delay code from DDR DLL
DATAVALID <sup>1</sup>	Output Data	Status flag from DATAVALID logic, used to indicate when input data is captured in IOLOGIC and valid to core.
READ	For DQS_Strobe	Read signal for DDR memory interface
DQSI	For DQS_Strobe	Unshifted DQS strobe from input pad
PRMBDET	For DQS_Strobe	DQSI biased to go high when DQSI is tristate, goes to input logic block as well as core logic.
GSRN	Control from routing	Global Set/Reset

#### Notes:

- 1. Signals available on left/right/top edges only.
- Selected PIO.



#### 3.1. PIO

The PIO contains four blocks: an input register block, output register block, tristate register block and a control logic block. These blocks contain registers for operating in a variety of modes along with the necessary clock and selection logic.

## 3.1.1. Input Register Block

The input register blocks for the PIOs, in the left, right and top edges, contain delay elements and registers that can be used to condition high-speed interface signals, such as DDR memory interfaces and source synchronous interfaces, before they are passed to the device core. Figure 3.2 shows the input register block for the left, right and top edges. The input register block for the bottom edge contains one element to register the input signal and no DDR registers. The following description applies to the input register block for PIOs in the left, right and top edges only.

Input signals are fed from the sysI/O buffer to the input register block (as signal DI). If desired, the input signal can bypass the register and delay elements and be used directly as a combinatorial signal (INDD), a clock (INCK) and, in selected blocks, the input to the DQS delay block. If an input delay is desired, designers can select either a fixed delay or a dynamic delay DEL[3:0]. The delay, if selected, reduces input register hold time requirements when using a global clock.

The input block allows three modes of operation. In single data rate (SDR) the data is registered with the system clock by one of the registers in the single data rate sync register block.

In DDR mode, two registers are used to sample the data on the positive and negative edges of the modified DQS (ECLKDQSR) in the DDR Memory mode or ECLK signal when using DDR Generic mode, creating two data streams. Before entering the core, these two data streams are synchronized to the system clock to generate two data streams.

A gearbox function can be implemented in each of the input registers on the left and right sides. The gearbox function takes a double data rate signal applied to PIOA and converts it as four data streams, INA, IPA, INB and IPB. The two data streams from the first set of DDR registers are synchronized to the edge clock and then to the system clock before entering the core. Figure 2.30 provides further information on the use of the gearbox function.

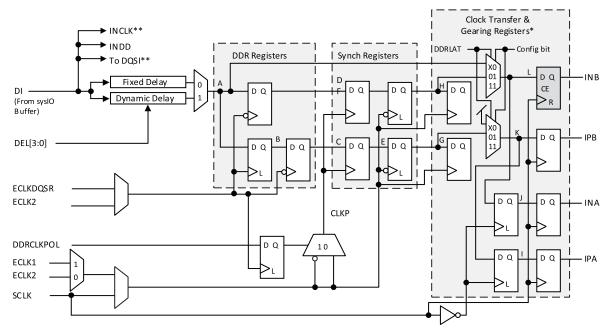
The signal DDRCLKPOL controls the polarity of the clock used in the synchronization registers. It ensures adequate timing when data is transferred to the system clock domain from the ECLKDQSR (DDR Memory Interface mode) or ECLK (DDR Generic mode). The DDRLAT signal is used to ensure the data transfer from the synchronization registers to the clock transfer and gearbox registers.

The ECLKDQSR, DDRCLKPOL and DDRLAT signals are generated in the DQS Read Control Logic Block. See Figure 3.6 for an overview of the DQS read control logic.

Further discussion about using the DQS strobe in this module is discussed in the DDR Memory section of this data sheet.

See LatticeECP3 High-Speed I/O Interface (FPGA-TN-02184) for more information on this topic.





<sup>\*</sup> Only on the left and right sides.

Note: Simplified diagram does not show CE/SET/REST details.

Figure 3.2. Input Register Block for Left, Right and Top Edges

## 3.1.2. Output Register Block

The output register block registers signals from the core of the device before they are passed to the sysI/O buffers. The blocks on the left and right PIOs contain registers for SDR and full DDR operation. The topside PIO block is the same as the left and right sides except it does not support ODDRX2 gearing of output logic. ODDRX2 gearing is used in DDR3 memory interfaces. The PIO blocks on the bottom contain the SDR registers but do not support generic DDR.

Figure 3.3 shows the Output Register Block for PIOs on the left and right edges.

In SDR mode, OPOSA feeds one of the flip-flops that then feeds the output. The flip-flop can be configured as a Dtype or latch. In DDR mode, two of the inputs are fed into registers on the positive edge of the clock. At the next clock cycle, one of the registered outputs is also latched.

A multiplexer running off the same clock is used to switch the mux between the 11 and 01 inputs that will then feed the output.

A gearbox function can be implemented in the output register block that takes four data streams: OPOSA, ONEGA, OPOSB and ONEGB. All four data inputs are registered on the positive edge of the system clock and two of them are also latched. The data is then output at a high rate using a multiplexer that runs off the DQCLKO and DQCLK1 clocks. DQCLKO and DQCLK1 are used in this case to transfer data from the system clock to the edge clock domain. These signals are generated in the DQS Write Control Logic block. See Figure 3.6 for an overview of the DQS write control logic.

See LatticeECP3 High-Speed I/O Interface (FPGA-TN-02184) for more information on this topic.

Further discussion on using the DQS strobe in this module is discussed in the DDR Memory section of this data sheet.

<sup>\*\*</sup> Selected PIO



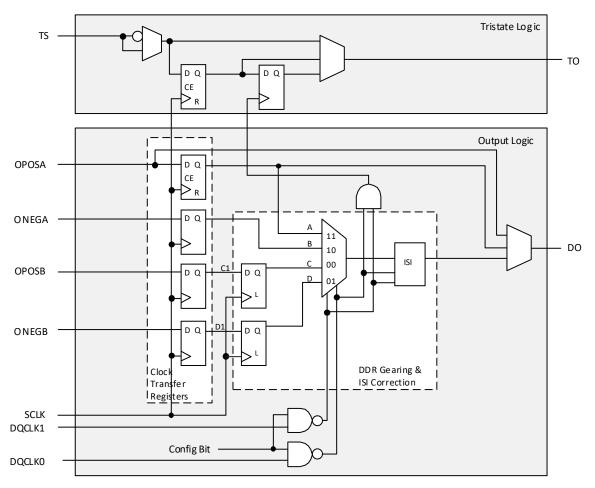


Figure 3.3. Output and Tristate Block for Left and Right Edges

#### 3.1.3. Tristate Register Block

The tristate register block registers tri-state control signals from the core of the device before they are passed to the sysl/O buffers. The block contains a register for SDR operation and an additional register for DDR operation.

In SDR and non-gearing DDR modes, TS input feeds one of the flip-flops that then feeds the output. In DDRX2 mode, the register TS input is fed into another register that is clocked using the DQCLK0 and DQCLK1 signals. The output of this register is used as a tristate control.

#### 3.1.4. ISI Calibration

The setting for Inter-Symbol Interference (ISI) cancellation occurs in the output register block. ISI correction is only available in the DDRX2 modes. ISI calibration settings exist once per output register block, so each I/O in a DQS-12 group may have a different ISI calibration setting.

The ISI block extends output signals at certain times, as a function of recent signal history. So, if the output pattern consists of a long strings of 0's to long strings of 1's, there are no delays on output signals. However, if there are quick, successive transitions from 010, the block will stretch out the binary 1. This is because the long trail of 0's will cause these symbols to interfere with the logic 1. Likewise, if there are quick, successive transitions from 101, the block will stretch out the binary 0. This block is controlled by a 3-bit delay control that can be set in the DQS control logic block. For more information about this topic, see the list of technical documentation at the end of this data sheet.

## 3.1.5. Control Logic Block

The control logic block allows the selection and modification of control signals for use in the PIO block.



## 3.2. DDR Memory Support

Certain PICs have additional circuitry to allow the implementation of high-speed source synchronous and DDR, DDR2 and DDR3 memory interfaces. The support varies by the edge of the device as detailed below.

### 3.2.1. Left and Right Edges

The left and right sides of the PIC have fully functional elements supporting DDR, DDR2, and DDR3 memory interfaces. One of every 12 PIOs supports the dedicated DQS pins with the DQS control logic block. Figure 3.4 shows the DQS bus spanning 11 I/O pins. Two of every 12 PIOs support the dedicated DQS and DQS# pins with the DQS control logic block.

#### 3.2.2. Bottom Edge

PICs on the bottom edge of the device do not support DDR memory and Generic DDR interfaces.

## 3.2.3. Top Edge

PICs on the top side are similar to the PIO elements on the left and right sides but do not support gearing on the output registers. Hence, the modes to support output/tristate DDR3 memory are removed on the top side.

The exact DQS pins are shown in a dual function in the Logic Signal Connections table in this data sheet. Additional detail is provided in the Signal Descriptions table. The DQS signal from the bus is used to strobe the DDR data from the memory into input register blocks. Interfaces on the left, right and top edges are designed for DDR memories that support 10 bits of data.

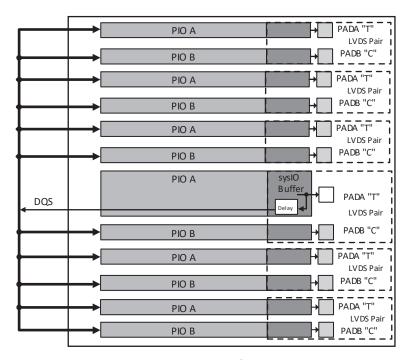


Figure 3.4. DQS Grouping on the Left, Right and Top Edges



## 3.2.4. DLL Calibrated DQS Delay Block

Source synchronous interfaces generally require the input clock to be adjusted in order to correctly capture data at the input register. For most interfaces, a PLL is used for this adjustment. However, in DDR memories the clock (referred to as DQS) is not free-running so this approach cannot be used. The DQS Delay block provides the required clock alignment for DDR memory interfaces.

The delay required for the DQS signal is generated by two dedicated DLLs (DDR DLL) on opposite side of the device. Each DLL creates DQS delays in its half of the device as shown in Figure 3.5. The DDR DLL on the left side will generate delays for all the DQS Strobe pins on Banks 0, 7 and 6 and DDR DLL on the right will generate delays for all the DQS pins on Banks 1, 2 and 3. The DDR DLL loop compensates for temperature, voltage and process variations by using the system clock and DLL feedback loop. DDR DLL communicates the required delay to the DQS delay block using a 7-bit calibration bus (DCNTL[6:0])

The DQS signal (selected PIOs only, as shown in Figure 3.4) feeds from the PAD through a DQS control logic block to a dedicated DQS routing resource. The DQS control logic block consists of DQS Read Control logic block that generates control signals for the read side and DQS Write Control logic that generates the control signals required for the write side. A more detailed DQS control diagram is shown in Figure 3.6, which shows how the DQS control blocks interact with the data paths.

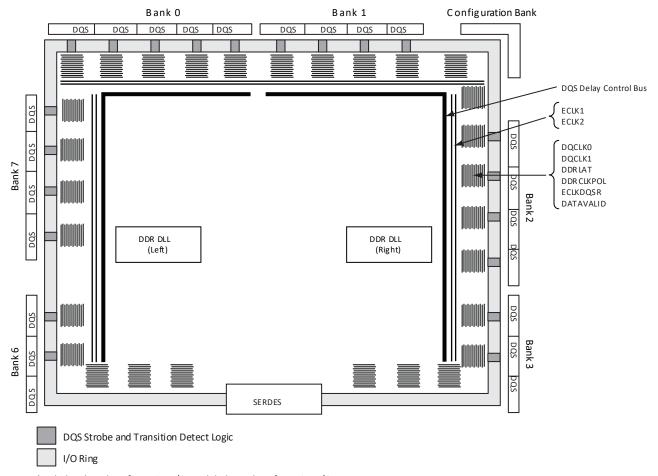
The DQS Read control logic receives the delay generated by the DDR DLL on its side and delays the incoming DQS signal by 90 degrees. This delayed ECLKDQSR is routed to 10 or 11 DQ pads covered by that DQS signal. This block also contains a polarity control logic that generates a DDRCLKPOL signal, which controls the polarity of the clock to the sync registers in the input register blocks. The DQS Read control logic also generates a DDRLAT signal that is in the input register block to transfer data from the first set of DDR register to the second set of DDR registers when using the DDRX2 gearbox mode for DDR3 memory interface.

The DQS Write control logic block generates the DQCLKO and DQCLK1 clocks used to control the output gearing in the Output register block which generates the DDR data output and the DQS output. They are also used to control the generation of the DQS output through the DQS output register block. In addition to the DCNTL [6:0] input from the DDR DLL, the DQS Write control block also uses a Dynamic Delay DYN DEL [7:0] attribute which is used to further delay the DQS to accomplish the write leveling found in DDR3 memory. Write leveling is controlled by the DDR memory controller implementation. The DYN DELAY can set 128 possible delay step settings. In addition, the most significant bit will invert the clock for a 180-degree shift of the incoming clock. This will generate the DQSW signal used to generate the DQS output in the DQS output register block.

Figure 3.5 and Figure 3.6 show how the DQS transition signals that are routed to the PIOs.

See LatticeECP3 High-Speed I/O Interface (FPGA-TN-02184) for more information on this topic.





<sup>\*</sup>Includes shared configuration I/Os and dedicated configuration I/Os.

Figure 3.5. Edge Clock, DLL Calibration and DQS Local Bus Distribution



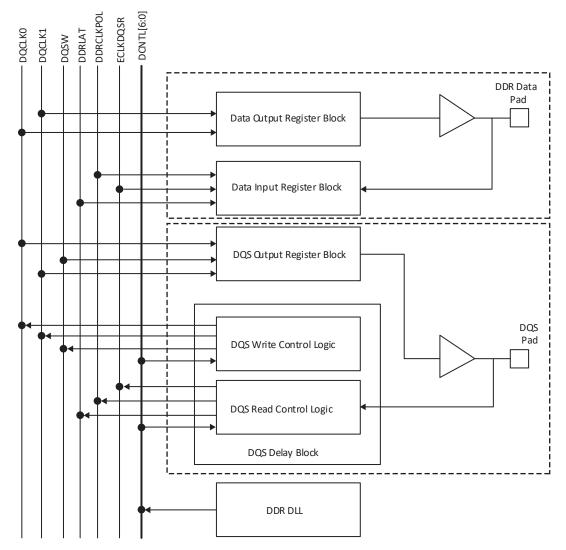


Figure 3.6. DQS Local Bus

### 3.2.5. Polarity Control Logic

In a typical DDR Memory interface design, the phase relationship between the incoming delayed DQS strobe and the internal system clock (during the READ cycle) is unknown. The LatticeECP3 family contains dedicated circuits to transfer data between these domains. A clock polarity selector is used to prevent set-up and hold violations at the domain transfer between DQS (delayed) and the system clock. This changes the edge on which the data is registered in the synchronizing registers in the input register block. This requires evaluation at the start of each READ cycle for the correct clock polarity.

Prior to the READ operation in DDR memories, DQS is in tristate (pulled by termination). The DDR memory device drives DQS low at the start of the preamble state. A dedicated circuit detects the first DQS rising edge after the preamble state. This signal is used to control the polarity of the clock to the synchronizing registers.



## 3.2.6. DDR3 Memory Support

LatticeECP3 supports the read and write leveling required for DDR3 memory interfaces.

Read leveling is supported by the use of the DDRCLKPOL and the DDRLAT signals generated in the DQS Read Control logic block. These signals dynamically control the capture of the data with respect to the DQS at the input register block.

To accomplish write leveling in DDR3, each DQS group has a slightly different delay that is set by DYN DELAY[7:0] in the DQS Write Control logic block. The DYN DELAY can set 128 possible delay step settings. In addition, the most significant bit will invert the clock for a 180-degree shift of the incoming clock.

LatticeECP3 input and output registers can also support DDR gearing that is used to receive and transmit the high speed DDR data from and to the DDR3 Memory.

LatticeECP3 supports the 1.5V SSTL I/O standard required for the DDR3 memory interface. For more information, refer to the sysIO section of this data sheet.

See LatticeECP3 High-Speed I/O (FPGA-TN-02184) Interface for more information on DDR Memory interface implementation in LatticeECP3.

# 3.3. sysI/O Buffer

Each I/O is associated with a flexible buffer referred to as a sysI/O buffer. These buffers are arranged around the periphery of the device in groups referred to as banks. The sysI/O buffers allow users to implement the wide variety of standards that are found in today's systems including LVDS, BLVDS, HSTL, SSTL Class I & II, LVCMOS, LVTTL, LVPECL, PCI.

### 3.3.1. sysI/O Buffer Banks

LatticeECP3 devices have six sysI/O buffer banks: six banks for user I/O arranged two per side. The banks on the bottom side are wraparounds of the banks on the lower right and left sides. The seventh sysI/O buffer bank (Configuration Bank) is located adjacent to Bank 2 and has dedicated/shared I/O for configuration. When a shared pin is not used for configuration it is available as a user I/O. Each bank is capable of supporting multiple I/O standards. Each sysI/O bank has its own I/O supply voltage (VCCIO). In addition, each bank, except the Configuration Bank, has voltage references, VREF1 and VREF2, which allow it to be completely independent from the others. Figure 3.7 shows the seven banks and their associated supplies.

In LatticeECP3 devices, single-ended output buffers and ratioed input buffers (LVTTL, LVCMOS and PCI) are powered using VCCIO. LVTTL, LVCMOS33, LVCMOS25 and LVCMOS12 can also be set as fixed threshold inputs independent of VCCIO.

Each bank can support up to two separate VREF voltages, VREF1 and VREF2, that set the threshold for the referenced input buffers. Some dedicated I/O pins in a bank can be configured to be a reference voltage supply pin. Each I/O is individually configurable based on the bank's supply and reference voltages.



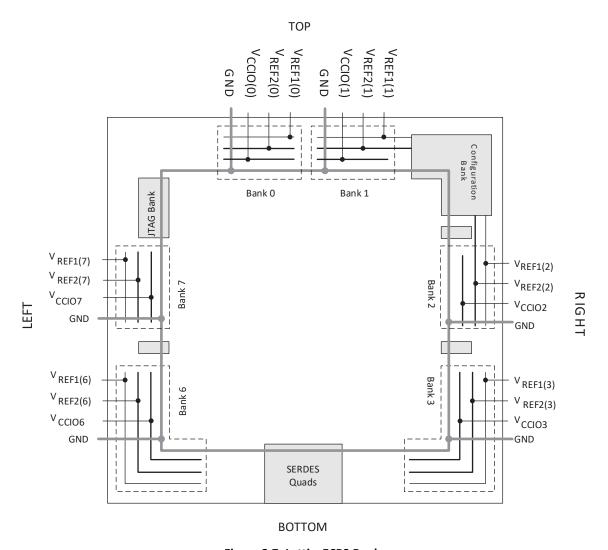


Figure 3.7. LatticeECP3 Banks

LatticeECP3 devices contain two types of sysI/O buffer pairs.

Top (Bank 0 and Bank 1) and Bottom sysIO Buffer Pairs (Single-Ended Outputs Only)

The sysl/O buffer pairs in the top banks of the device consist of two single-ended output drivers and two sets of single-ended input buffers (both ratioed and referenced). One of the referenced input buffers can also be configured as a differential input. Only the top edge buffers have a programmable PCI clamp.

The two pads in the pair are described as "true" and "comp", where the true pad is associated with the positive side of the differential input buffer and the comp (complementary) pad is associated with the negative side of the differential input buffer.

The top and bottom sides are ideal for general purpose I/O, PCI, and inputs for LVDS (LVDS outputs are only allowed on the left and right sides). The top side can be used for the DDR3 ADDR/CMD signals.

The I/O pins located on the top and bottom sides of the device (labeled PTxxA/B or PBxxA/B) are fully hot socketable. Note that the pads in Banks 3, 6 and 8 are wrapped around the corner of the device. In these banks, only the pads located on the top or bottom of the device are hot socketable. The top and bottom side pads can be identified by the Lattice Diamond tool.

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- Left and Right (Banks 2, 3, 6 and 7) sysl/O Buffer Pairs (50% Differential and 100% Single-Ended Outputs)

  The sysl/O buffer pairs in the left and right banks of the device consist of two single-ended output drivers, two sets of single-ended input buffers (both ratioed and referenced) and one differential output driver. One of the referenced input buffers can also be configured as a differential input. In these banks the two pads in the pair are described as "true" and "comp", where the true pad is associated with the positive side of the differential I/O, and the comp (complementary) pad is associated with the negative side of the differential I/O.
  - In addition, programmable on-chip input termination (parallel or differential, static or dynamic) is supported on these sides, which is required for DDR3 interface. However, there is no support for hot-socketing for the I/O pins located on the left and right side of the device as the PCI clamp is always enabled on these pins.
  - LVDS, RSDS, PPLVDS and Mini-LVDS differential output drivers are available on 50% of the buffer pairs on the left and right banks.
- Configuration Bank sysl/O Buffer Pairs (Single-Ended Outputs, Only on Shared Pins When Not Used by Configuration)
  - The sysl/O buffers in the Configuration Bank consist of ratioed single-ended output drivers and single-ended input buffers. This bank does not support PCI clamp like the other banks on the top, left, and right sides.
  - The two pads in the pair are described as "true" and "comp", where the true pad is associated with the positive side of the differential input buffer and the comp (complementary) pad is associated with the negative side of the differential input buffer.

Programmable PCI clamps are only available on the top banks. PCI clamps are used primarily on inputs and bidirectional pads to reduce ringing on the receiving end.

## 3.3.2. Typical sysI/O I/O Behavior During Power-up

The internal power-on-reset (POR) signal is deactivated when  $V_{CC}$ , VCCIO8 and  $V_{CCAUX}$  have reached satisfactory levels. After the POR signal is deactivated, the FPGA core logic becomes active. It is the user's responsibility to ensure that all other  $V_{CCIO}$  banks are active with valid input logic levels to properly control the output logic states of all the I/O banks that are critical to the application. For more information about controlling the output logic state with valid input logic levels during power-up in LatticeECP3 devices, see the list of technical documentation at the end of this data sheet.

The  $V_{CC}$  and  $V_{CCAUX}$  supply the power to the FPGA core fabric, whereas the  $V_{CCIO}$  supplies power to the I/O buffers. In order to simplify system design while providing consistent and predictable I/O behavior, it is recommended that the I/O buffers be powered-up prior to the FPGA core fabric.  $V_{CCIO}$  supplies should be powered-up before or together with the  $V_{CC}$  and  $V_{CCAUX}$  supplies.

### 3.3.3. Supported sysI/O Standards

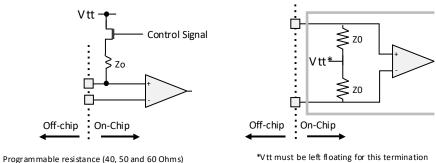
The LatticeECP3 sysI/O buffer supports both single-ended and differential standards. Single-ended standards can be further subdivided into LVCMOS, LVTTL and other standards. The buffers support the LVTTL, LVCMOS 1.2 V, 1.5 V, 1.8 V, 2.5 V and 3.3 V standards. In the LVCMOS and LVTTL modes, the buffer has individual configuration options for drive strength, slew rates, bus maintenance (weak pull-up, weak pull-down, or a bus-keeper latch) and open drain. Other single-ended standards supported include SSTL and HSTL. Differential standards supported include LVDS, BLVDS, LVPECL, MLVDS, RSDS, Mini-LVDS, PPLVDS (point-to-point LVDS), TRLVDS (Transition Reduced LVDS), differential SSTL and differential HSTL. For further information on utilizing the sysI/O buffer to support a variety of standards see LatticeECP3 sysIO Usage Guide (FPGA-TN-02194).

#### 3.3.4. On-Chip Programmable Termination

The LatticeECP3 supports a variety of programmable on-chip terminations options, including:

- Dynamically switchable Single-Ended Termination with programmable resistor values of 40, 50, or 60 Ohms. External termination to Vtt should be used for DDR2 and DDR3 memory controller implementation.
- Common mode termination of 80, 100, 120 Ohms for differential inputs





\*Vtt must be left floating for this termination

Parallel Single-Ended Input

Differential Input

Figure 3.8. On-Chip Termination

See Table 3.2 for termination options for input modes.

Table 3.2. On-Chip Termination Options for Input Modes

IO_TYPE	TERMINATE to VTT <sup>1, 2</sup>	DIFFERENTIAL TERMINATION RESISTOR <sup>1</sup>
LVDS25	þ	80, 100, 120
BLVDS25	þ	80, 100, 120
MLVDS	þ	80, 100, 120
HSTL18_I	40, 50, 60	þ
HSTL18_II	40, 50, 60	þ
HSTL18D_I	40, 50, 60	þ
HSTL18D_II	40, 50, 60	þ
HSTL15_I	40, 50, 60	þ
HSTL15D_I	40, 50, 60	þ
SSTL25_I	40, 50, 60	þ
SSTL25_II	40, 50, 60	þ
SSTL25D_I	40, 50, 60	þ
SSTL25D_II	40, 50, 60	þ
SSTL18_I	40, 50, 60	þ
SSTL18_II	40, 50, 60	þ
SSTL18D_I	40, 50, 60	þ
SSTL18D_II	40, 50, 60	þ
SSTL15	40, 50, 60	þ
SSTL15D	40, 50, 60	þ

#### Notes:

TERMINATE to VTT and DIFFRENTIAL TERMINATION RESISTOR when turned on can only have one setting per bank. Only left and right banks have this feature. Use of TERMINATE to VTT and DIFFRENTIAL TERMINATION RESISTOR are mutually exclusive in an I/O bank.

On-chip termination tolerance +/- 20%

External termination to VTT should be used when implementing DDR2 and DDR3 memory controller.

See LatticeECP3 sysIO Usage Guide (FPGA-TN-02194) for on-chip termination usage and value ranges.



#### 3.3.5. Equalization Filter

Equalization filtering is available for single-ended inputs on both true and complementary I/O, and for differential inputs on the true I/O on the left, right, and top sides. Equalization is required to compensate for the difficulty of sampling alternating logic transitions with a relatively slow slew rate. It is considered the most useful for the Input DDRX2 modes, used in DDR3 memory, LVDS, or TRLVDS signaling. Equalization filter acts as a tunable filter with settings to determine the level of correction. In the LatticeECP3 devices, there are four settings available: 0 (none), 1, 2 and 3. The default setting is 0. The equalization logic resides in the sysl/O buffers, the two bits of setting is set uniquely in each input IOLOGIC block. Therefore, each sysl/O can have a unique equalization setting within a DQS-12 group.

#### 3.3.6. Hot Socketing

LatticeECP3 devices have been carefully designed to ensure predictable behavior during power-up and powerdown. During power-up and power-down sequences, the I/O remain in tri-state until the power supply voltage is high enough to ensure reliable operation. In addition, leakage into I/O pins is controlled within specified limits. Refer to the Hot Socketing Specifications1, 2, 3 section in this data sheet.

## 3.4. SERDES and PCS (Physical Coding Sublayer)

LatticeECP3 devices feature up to 16 channels of embedded SERDES/PCS arranged in quads at the bottom of the devices supporting up to 3.2Gbps data rate. Figure 3.9 shows the position of the quad blocks for the LatticeECP3-150 devices. Table 3.3 shows the location of available SERDES Quads for all devices.

The LatticeECP3 SERDES/PCS supports a range of popular serial protocols, including:

- PCI Express 1.1
- Ethernet (XAUI, GbE 1000 Base CS/SX/LX and SGMII)
- Serial RapidIO
- SMPTE SDI (3G, HD, SD)
- CPRI
- SONET/SDH (STS-3, STS-12, STS-48)

Each quad contains four dedicated SERDES for high speed, full duplex serial data transfer. Each quad also has a PCS block that interfaces to the SERDES channels and contains protocol specific digital logic to support the standards listed above. The PCS block also contains interface logic to the FPGA fabric. All PCS logic for dedicated protocol support can also be bypassed to allow raw 8-bit or 10-bit interfaces to the FPGA fabric.

Even though the SERDES/PCS blocks are arranged in quads, multiple baud rates can be supported within a quad with the use of dedicated, per channel  $\div 1$ ,  $\div 2$  and  $\div 11$  rate dividers. Additionally, multiple quads can be arranged together to form larger data pipes.

For information on how to use the SERDES/PCS blocks to support specific protocols, as well on how to combine multiple protocols and baud rates within a device, refer to LatticeECP3 SERDES/PCS Usage Guide (FPGA-TN-02190).



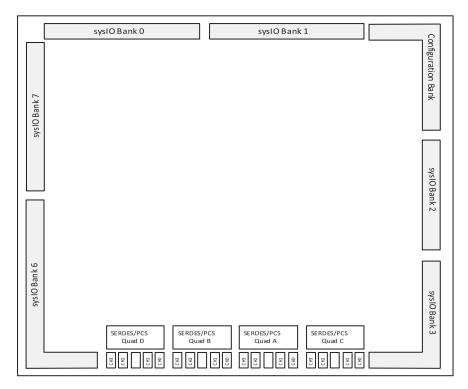


Figure 3.9. SERDES/PCS Quads (LatticeECP3-150)

Table 3.3. LatticeECP3 SERDES Standard Support

Standard	Data Rate (Mbps)	Number of General/Link Width	Encoding Style
PCI Express 1.1	2500	x1, x2, x4	8b10b
Gigabit Ethernet	1250, 2500	x1	8b10b
SGMII	1250	x1	8b10b
XAUI	3125	x4	8b10b
Serial RapidIO Type I, Serial RapidIO Type II, Serial RapidIO Type III	1250, 2500, 3125	x1, x4	8b10b
CPRI-1, CPRI-2, CPRI-3, CPRI-4	614.4, 1228.8, 2457.6, 3072.0	x1	8b10b
SD-SDI (259M, 344M)	143 <sup>1</sup> , 177 <sup>1</sup> , 270, 360, 540	x1	NRZI/Scrambled
HD-SDI (292M)	1483.5, 1485	x1	NRZI/Scrambled
3G-SDI (424M)	2967, 2970	x1	NRZI/Scrambled
SONET-STS-3 <sup>2</sup>	155.52	x1	N/A
SONET-STS-12 <sup>2</sup>	622.08	x1	N/A
SONET-STS-48 <sup>2</sup>	2488	x1	N/A

#### Notes:

- 1. For slower rates, the SERDES are bypassed and CML signals are directly connected to the FPGA routing.
- The SONET protocol is supported in 8-bit SERDES mode. See Lattice ECP3 SERDES/PCS Usage Guide (FPGA-TN-02190) for more information.

Table 3.4. Available SERDES Quads per LatticeECP3 Devices

Package	ECP3-17	ECP3-35	ECP3-70	ECP3-95	ECP3-150	
256 ftBGA	1	1	_	_	_	
328 csBGA	2 channels	_	_	_	_	
484 fpBGA	1	1	1	1	_	
672 fpBGA	_	1	2	2	2	
1156 fpBGA	_	_	3	3	4	

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#### 3.4.1. SERDES Block

A SERDES receiver channel may receive the serial differential data stream, equalize the signal, perform Clock and Data Recovery (CDR) and de-serialize the data stream before passing the 8- or 10-bit data to the PCS logic. The SERDES transmitter channel may receive the parallel 8- or 10-bit data, serialize the data and transmit the serial bit stream through the differential drivers. Figure 3.10 shows a single-channel SERDES/PCS block. Each SERDES channel provides a recovered clock and a SERDES transmit clock to the PCS block and to the FPGA core logic.

Each transmit channel, receiver channel, and SERDES PLL shares the same power supply (VCCA). The output and input buffers of each channel have their own independent power supplies (VCCOB and VCCIB).

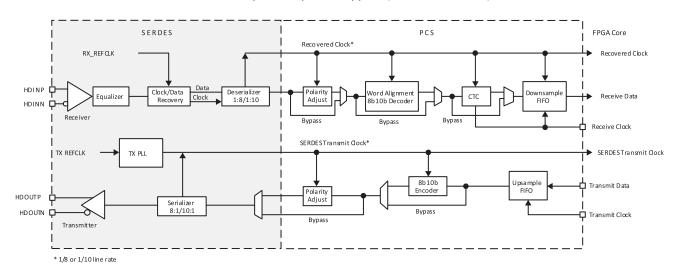


Figure 3.10. Simplified Channel Block Diagram for SERDES/PCS Block

## 3.4.2. PCS

As shown in Figure 3.10, the PCS receives the parallel digital data from the deserializer and selects the polarity, performs word alignment, decodes (8b/10b), provides Clock Tolerance Compensation and transfers the clock domain from the recovered clock to the FPGA clock via the Down Sample FIFO.

For the transmit channel, the PCS block receives the parallel data from the FPGA core, encodes it with 8b/10b, selects the polarity and passes the 8/10 bit data to the transmit SERDES channel.

The PCS also provides bypass modes that allow a direct 8-bit or 10-bit interface from the SERDES to the FPGA logic. The PCS interface to the FPGA can also be programmed to run at 1/2 speed for a 16-bit or 20-bit interface to the FPGA logic.

#### 3.4.3. SCI (SERDES Client Interface) Bus

The SERDES Client Interface (SCI) is an IP interface that allows the SERDES/PCS Quad block to be controlled by registers rather than the configuration memory cells. It is a simple register configuration interface that allows SERDES/PCS configuration without power cycling the device. The Diamond and ispLEVER design tools support all modes of the PCS. Most modes are dedicated to applications associated with a specific industry standard data protocol. Other more general purpose modes allow users to define their own operation. With these tools, the user can define the mode for each quad in a design. Popular standards such as 10Gb Ethernet, x4 PCI Express and 4x Serial RapidIO can be implemented using IP (available through Lattice), a single quad (Four SERDES channels and PCS) and some additional logic from the core.

The LatticeECP3 family also supports a wide range of primary and secondary protocols. Within the same quad, the LatticeECP3 family can support mixed protocols with semi-independent clocking as long as the required clock frequencies are integer x1, x2, or x11 multiples of each other. Table 3.5 lists the allowable combination of primary and secondary protocol combinations.



## 3.5. Flexible Quad SERDES Architecture

The LatticeECP3 family SERDES architecture is a quad-based architecture. For most SERDES settings and standards, the whole quad (consisting of four SERDES) is treated as a unit. This helps in silicon area savings, better utilization and overall lower cost.

However, for some specific standards, the LatticeECP3 quad architecture provides flexibility; more than one standard can be supported within the same quad.

Table 3.5 shows the standards can be mixed and matched within the same quad. In general, the SERDES standards whose nominal data rates are either the same or a defined subset of each other, can be supported within the same quad. In Table 3.5, the Primary Protocol column refers to the standard that determines the reference clock and PLL settings. The Secondary Protocol column shows the other standard that can be supported within the same quad.

Furthermore, Table 3.5 also implies that more than two standards in the same quad can be supported, as long as they conform to the data rate and reference clock requirements. For example, a quad may contain PCI Express 1.1, SGMII, Serial RapidIO Type I and Serial RapidIO Type II, all in the same quad.

Table 3.5. LatticeECP3 Primary and Secondary Protocol Support

Primary Protocol	Secondary Protocol
PCI Express 1.1	SGMII
PCI Express 1.1	Gigabit Ethernet
PCI Express 1.1	Serial RapidIO Type I
PCI Express 1.1	Serial RapidIO Type II
Serial RapidIO Type I	SGMII
Serial RapidIO Type I	Gigabit Ethernet
Serial RapidIO Type II	SGMII
Serial RapidIO Type II	Gigabit Ethernet
Serial RapidIO Type II	Serial RapidIO Type I
CPRI-3	CPRI-2 and CPRI-1
3G-SDI	HD-SDI and SD-SDI

There are some restrictions to be aware of when using spread spectrum. When a quad shares a PCI Express x1 channel with a non-PCI Express channel, ensure that the reference clock for the quad is compatible with all protocols within the quad. For example, a PCI Express spread spectrum reference clock is not compatible with most Gigabit Ethernet applications because of tight CTC ppm requirements.

While the LatticeECP3 architecture will allow the mixing of a PCI Express channel and a Gigabit Ethernet, Serial RapidIO or SGMII channel within the same quad, using a PCI Express spread spectrum clocking as the transmit reference clock will cause a violation of the Gigabit Ethernet, Serial RapidIO and SGMII transmit jitter specifications.

For further information on SERDES, see LatticeECP3 SERDES/PCS Usage Guide (FPGA-TN-02190).

# 3.6. IEEE 1149.1-Compliant Boundary Scan Testability

All LatticeECP3 devices have boundary scan cells that are accessed through an IEEE 1149.1 compliant Test Access Port (TAP). This allows functional testing of the circuit board on which the device is mounted through a serial scan path that can access all critical logic nodes. Internal registers are linked internally, allowing test data to be shifted in and loaded directly onto test nodes, or test data to be captured and shifted out for verification. The test access port consists of dedicated I/O: TDI, TDO, TCK and TMS. The test access port has its own supply voltage V<sub>CCJ</sub> and can operate with LVCMOS3.3, 2.5, 1.8, 1.5 and 1.2 standards.

For more information, see LatticeECP3 sysCONFIG Usage Guide (FPGA-TN-02192).



# 3.7. Device Configuration

All LatticeECP3 devices contain two ports that can be used for device configuration. The Test Access Port (TAP), which supports bit-wide configuration, and the sysCONFIG port, support dual-byte, byte and serial configuration. The TAP supports both the IEEE Standard 1149.1 Boundary Scan specification and the IEEE Standard 1532 In-System Configuration specification. The sysCONFIG port includes seven I/O used as dedicated pins with the remaining pins used as dual-use pins. See LatticeECP3 sysCONFIG Usage Guide (FPGA-TN-02192) for more information about using the dual-use pins as general purpose I/O.

There are various ways to configure a LatticeECP3 device:

- JTAG
- 2. Standard Serial Peripheral Interface (SPI and SPIm modes) interface to boot PROM memory
- 3. System microprocessor to drive a x8 CPU port (PCM mode)
- 4. System microprocessor to drive a serial slave SPI port (SSPI mode)
- 5. Generic byte wide flash with a MachXO™ device, providing control and addressing

On power-up, the FPGA SRAM is ready to be configured using the selected sysCONFIG port. Once a configuration port is selected, it will remain active throughout that configuration cycle. The IEEE 1149.1 port can be activated any time after power-up by sending the appropriate command through the TAP port.

LatticeECP3 devices also support the Slave SPI Interface. In this mode, the FPGA behaves like a SPI Flash device (slave mode) with the SPI port of the FPGA to perform read-write operations.

#### 3.7.1. Enhanced Configuration Options

LatticeECP3 devices have enhanced configuration features such as: decryption support, TransFR™ I/O and dualboot image support.

#### 1. TransFR (Transparent Field Reconfiguration)

TransFR I/O (TFR) is a unique Lattice technology that allows users to update their logic in the field without interrupting system operation using a single ispVM command. TransFR I/O allows I/O states to be frozen during device configuration. This allows the device to be field updated with a minimum of system disruption and downtime. See Minimizing System Interruption During Configuration Using TransFR Technology (FPGA-TN-02198) for details.

#### 2. Dual-Boot Image Support

Dual-boot images are supported for applications requiring reliable remote updates of configuration data for the system FPGA. After the system is running with a basic configuration, a new boot image can be downloaded remotely and stored in a separate location in the configuration storage device. Any time after the update the LatticeECP3 can be re-booted from this new configuration file. If there is a problem, such as corrupt data during download or incorrect version number with this new boot image, the LatticeECP3 device can revert back to the original backup golden configuration and try again. This all can be done without power cycling the system. For more information, see LatticeECP3 sysCONFIG Usage Guide (FPGA-TN-02192).

## 3.7.2. Soft Error Detect (SED) Support

LatticeECP3 devices have dedicated logic to perform Cycle Redundancy Code (CRC) checks. During configuration, the configuration data bitstream can be checked with the CRC logic block. In addition, the LatticeECP3 device can also be programmed to utilize a Soft Error Detect (SED) mode that checks for soft errors in configuration SRAM. The SED operation can be run in the background during user mode. If a soft error occurs, during user mode (normal operation) the device can be programmed to generate an error signal.

For further information on SED support, see LatticeECP3 Soft Error Detection (SED) Usage Guide (FPGA-TN-02207).

#### 3.7.3. External Resistor

LatticeECP3 devices require a single external,  $10 \text{ kOhm} \pm 1\%$  value between the XRES pin and ground. Device configuration will not be completed if this resistor is missing. There is no boundary scan register on the external resistor pad.

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## 3.7.4. On-Chip Oscillator

Every LatticeECP3 device has an internal CMOS oscillator which is used to derive a Master Clock (MCCLK) for configuration. The oscillator and the MCCLK run continuously and are available to user logic after configuration is completed. The software default value of the MCCLK is nominally 2.5 MHz. Table 3.6 lists all the available MCCLK frequencies. When a different Master Clock is selected during the design process, the following sequence takes place:

- 1. Device powers up with a nominal Master Clock frequency of 3.1 MHz.
- 2. During configuration, users select a different master clock frequency.
- 3. The Master Clock frequency changes to the selected frequency once the clock configuration bits are received.
- 4. If the user does not select a master clock frequency, then the configuration bitstream defaults to the MCCLK frequency of 2.5 MHz.

This internal 130 MHz +/- 15% CMOS oscillator is available to the user by routing it as an input clock to the clock tree. For further information on the use of this oscillator for configuration or user mode, see LatticeECP3 sysCONFIG Usage Guide (FPGA-TN-02192).

Table 3.6. Selectable Master Clock (MCCLK) Frequencies During Configuration (Nominal)

MCCLK (MHz)	MCCLK (MHz)
	10
2.51	13
4.3	15 <sup>2</sup>
5.4	20
6.9	26
8.1	33 <sup>3</sup>
9.2	

#### Notes:

- Software default MCCLK frequency. Hardware default is 3.1 MHz.
- 2. Maximum MCCLK with encryption enabled.
- 3. Maximum MCCLK without encryption.

# 3.8. Density Shifting

The LatticeECP3 family is designed to ensure that different density devices in the same family and in the same package have the same pinout. Furthermore, the architecture ensures a high success rate when performing design migration from lower density devices to higher density devices. In many cases, it is also possible to shift a lower utilization design targeted for a high-density device to a lower density device. However, the exact details of the final resource utilization will impact the likelihood of success in each case. An example is that some user I/O may become No Connects in smaller devices in the same package. Refer to the LatticeECP3 Pin Migration Tables and Diamond software for specific restrictions and limitations.



# 4. DC and Switching Characteristics

# 4.1. Absolute Maximum Ratings<sup>1, 2, 3</sup>

Supply Voltage  $V_{CC}$  0.5 V to 1.32 V Supply Voltage  $V_{CCAUX}$  0.5 V to 3.75 V Supply Voltage  $V_{CCI}$  0.5 V to 3.75 V Output Supply Voltage  $V_{CCIO}$  0.5 V to 3.75 V Input or I/O Tristate Voltage Applied<sup>4</sup> 0.5 V to 3.75 V Storage Temperature (Ambient) 65 V to 150 °C Junction Temperature (T<sub>J</sub>) +125 °C

- 1. Stress above those listed under the "Absolute Maximum Ratings" may cause permanent damage to the device. Functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.
- 2. Compliance with the Lattice Thermal Management document is required.
- 3. All voltages referenced to GND.
- 4. Overshoot and undershoot of -2 V to (VIHMAX + 2) volts is permitted for a duration of <20 ns.

# 4.2. Recommended Operating Conditions<sup>1</sup>

Symbol	Parameter	Min.	Max.	Units
V <sub>CC</sub> <sup>2</sup>	Core Supply Voltage	1.14	1.26	V
V <sub>CCAUX</sub> <sup>2,4</sup>	Auxiliary Supply Voltage, Terminating Resistor Switching Power Supply (SERDES)	3.135	3.465	٧
V <sub>CCPLL</sub>	PLL Supply Voltage	3.135	3.465	V
V <sub>CCIO</sub> <sup>2,3</sup>	I/O Driver Supply Voltage	1.14	3.465	٧
V <sub>CCJ</sub> <sup>2</sup>	Supply Voltage for IEEE 1149.1 Test Access Port	1.14	3.465	٧
V <sub>REF1</sub> <sup>7</sup> and V <sub>REF2</sub>	Input Reference Voltage	0.5	1.7	٧
V <sub>TT5</sub>	Termination Voltage	0.5	1.3125	٧
t <sub>JCOM</sub>	Junction Temperature, Commercial Operation	0	85	°C
t <sub>JIND</sub>	Junction Temperature, Industrial Operation	-40	100	°C
SERDES External Power	Supply6			
V <sub>CCIB</sub>	Input Buffer Power Supply (1.2 V)	1.14	1.26	V
	Input Buffer Power Supply (1.5 V)	1.425	1.575	V
V <sub>CCOB</sub>	Output Buffer Power Supply (1.2 V)	1.14	1.26	V
	Output Buffer Power Supply (1.5 V)	1.425	1.575	V
V <sub>CCA</sub>	Transmit, Receive, PLL and Reference Clock Buffer Power Supply	1.14	1.26	V

#### Notes:

- For correct operation, all supplies except V<sub>REF</sub> and V<sub>TT</sub> must be held in their valid operation range. This is true independent of feature usage.
- 2. If  $V_{CCIO}$  or  $V_{CCI}$  is set to 1.2 V, they must be connected to the same power supply as  $V_{CC}$ . If  $V_{CCIO}$  or  $V_{CCI}$  is set to 3.3 V, they must be connected to the same power supply as  $V_{CCAUX}$ .
- 3. See recommended voltages by I/O standard in subsequent table.
- 4. V<sub>CCAUX</sub> ramp rate must not exceed 30 mV/μs during power-up when transitioning between 0 V and 3.3 V.
- 5. If not used,  $V_{TT}$  should be left floating.
- See LatticeECP3 SERDES/PCS Usage Guide (FPGA-TN-02190) for information on board considerations for SERDES power supplies.
- 7. When implementing generic DDR interface using the DQS circuit (GDDRX1/2\_RX.DQS.Centered/Aligned) and LVDS inputs, it is recommended the VREF1 port of that bank be tied to VCCIO. This is required for the correct startup of the DQSBUF circuit. See LatticeECP3 High-Speed I/O Interface (TN1180) for more information.

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# 4.3. Hot Socketing Specifications<sup>1, 2, 3</sup>

Symbol	Parameter	Condition	Min.	Тур.	Max.	Units
IDK_HS⁴	Input or I/O Leakage Current	$0 \le V_{IN} \le V_{IH}$ (Max.)	1	1	+/-1	mA
IDK <sup>5</sup>	Input or I/O Leakage Current	$0 \le V_{IN} < V_{CCIO}$	-	-	+/-1	mA
		$V_{CCIO} \le V_{IN} \le V_{CCIO} + 0.5V$	1	18	1	mA

#### Notes:

- 1.  $V_{CC}$ ,  $V_{CCAUX}$  and  $V_{CCIO}$  should rise/fall monotonically.
- 2.  $I_{DK}$  is additive to  $I_{PU}$ ,  $I_{PD}$  or  $I_{BH}$ .
- 3. LVCMOS and LVTTL only.
- 4. Applicable to general purpose I/O pins located on the top and bottom sides of the device.
- 5. Applicable to general purpose I/O pins located on the left and right sides of the device.

# 4.4. Hot Socketing Requirements

Description	Min.	Тур.	Max.	Units
Input current per SERDES I/O pin when device is powered down and inputs	-	-	8	mA
driven.				

#### Notes:

- Assumes the device is powered down, all supplies grounded, both P and N inputs driven by CML driver with maximum allowed VCCOB (1.575 V), 8b10b data, internal AC coupling.
- Each P and N input must have less than the specified maximum input current. For a 16-channel device, the total input current would be 8 mA\*16 channels \*2 input pins per channel = 256 mA

## 4.5. ESD Performance

Refer to the LatticeECP3 Product Family Qualification Summary for complete qualification data, including ESD performance.



## 4.6. DC Electrical Characteristics

**Over Recommended Operating Conditions** 

Symbol	Parameter	Condition	Min.	Тур.	Max.	Units
I <sub>IL</sub> , I <sub>IH</sub> <sup>1, 4</sup>	Input or I/O Low Leakage	$0 \le V_{IN} \le (V_{CCIO} - 0.2 V)$	_	_	10	μΑ
I <sub>IH</sub> 1, 3	Input or I/O High Leakage	$(V_{CCIO} - 0.2 \text{ V}) < V_{IN} \le 3.6 \text{ V}$	_	_	150	μΑ
I <sub>PU</sub>	I/O Active Pull-up Current	$0 \le V_{IN} \le 0.7 V_{CCIO}$	-30	_	-210	μΑ
I <sub>PD</sub>	I/O Active Pull-down Current	$V_{IL}$ (MAX) $\leq V_{IN} \leq V_{CCIO}$	30	_	210	μΑ
I <sub>BHLS</sub>	Bus Hold Low Sustaining Current	V <sub>IN</sub> = V <sub>IL</sub> (MAX)	30	_	_	μА
I <sub>вннѕ</sub>	Bus Hold High Sustaining Current	V <sub>IN</sub> = 0.7 V <sub>CCIO</sub>	-30	_	_	μА
I <sub>BHLO</sub>	Bus Hold Low Overdrive Current	$0 \le V_{IN} \le V_{CCIO}$	_	_	210	μА
I <sub>BHHO</sub>	Bus Hold High Overdrive Current	$0 \le V_{IN} \le V_{CCIO}$	_	_	-210	μА
$V_{BHT}$	Bus Hold Trip Points	$0 \le V_{IN} \le V_{IH}$ (MAX)	V <sub>IL</sub> (MAX)	_	V <sub>IH</sub> (MIN)	V
C1	I/O Capacitance <sup>2</sup>	$V_{CCIO} = 3.3 \text{ V}, 2.5 \text{ V}, 1.8 \text{ V}, 1.5 \text{ V},$ $1.2 \text{ V}, V_{CC} = 1.2 \text{ V}, V_{IO} = 0 \text{ to } V_{IH}$ (MAX)	_	5	8	pf
C2	Dedicated Input Capacitance <sup>2</sup>	$V_{CCIO} = 3.3 \text{ V}, 2.5 \text{ V}, 1.8 \text{ V}, 1.5 \text{ V},$ $1.2 \text{ V}, V_{CC} = 1.2 \text{ V}, V_{IO} = 0 \text{ to } V_{IH}$ (MAX)	_	5	7	pf

#### Notes:

- 1. Input or I/O leakage current is measured with the pin configured as an input or as an I/O with the output driver tri-stated. It is not measured with the output driver active. Bus maintenance circuits are disabled.
- 2.  $T_A$  25 oC, f = 1.0 MHz.
- 3. Applicable to general purpose I/O in top and bottom banks.
- 4. When used as  $V_{REF}$ , maximum leakage= 25  $\mu$ A.



# 4.7. LatticeECP3 Supply Current (Standby)

**Over Recommended Operating Conditions** 

Symbol	Parameter	Device	Тур	Units	
			-6L, -7L, -8L	-6, -7, -8	
I <sub>CC</sub>	Core Power Supply Current	ECP-17EA	29.8	49.4	mA
		ECP3-35EA	53.7	89.4	mA
		ECP3-70EA	137.3	230.7	mA
		ECP3-95EA	137.3	230.7	mA
		ECP3-150EA	219.5	370.9	mA
I <sub>CCAUX</sub>	Auxiliary Power Supply Current	ECP-17EA	18.3	19.4	mA
		ECP3-35EA	19.6	23.1	mA
		ECP3-70EA	26.5	32.4	mA
		ECP3-95EA	26.5	32.4	mA
		ECP3-150EA	37.0	45.7	mA
I <sub>CCPLL</sub>	PLL Power Supply Current (Per PLL)	ECP-17EA	0.0	0.0	mA
		ECP3-35EA	0.1	0.1	mA
		ECP3-70EA	0.1	0.1	mA
		ECP3-95EA	0.1	0.1	mA
		ECP3-150EA	0.1	0.1	mA
I <sub>CCIO</sub>	Bank Power Supply Current (Per Bank)	ECP-17EA	1.3	1.4	mA
		ECP3-35EA	1.3	1.4	mA
		ECP3-70EA	1.4	1.5	mA
		ECP3-95EA	1.4	1.5	mA
		ECP3-150EA	1.4	1.5	mA
I <sub>CCJ</sub>	JTAG Power Supply Current	All Devices	2.5	2.5	mA
I <sub>CCA</sub>	Transmit, Receive, PLL and Reference Clock Buffer	ECP-17EA	6.1	6.1	mA
	Power Supply	ECP3-35EA	6.1	6.1	mA
		ECP3-70EA	18.3	18.3	mA
		ECP3-95EA	18.3	18.3	mA
		ECP3-150EA	24.4	24.4	mA

#### Notes:

- For further information on supply current, see the list of technical documentation at the end of this data sheet.
- $\bullet \qquad \text{Assumes all outputs are tristated, all inputs are configured as LVCMOS and held at the $V_{\text{CCIO}}$ or GND.}$
- Frequency 0 MHz.
- Pattern represents a "blank" configuration data file. A "blank" pattern configures the part to the following conditions:
  - All outputs are tri-stated, all inputs are held at either V<sub>CCIO</sub>, or GND.
  - All clock inputs are at 0 MHz.
  - No pull-ups on I/O.
- $T_J = 85$  °C, power supplies at nominal voltage.
- To determine the LatticeECP3 peak start-up current data, use the Power Calculator tool.



# 4.8. SERDES Power Supply Requirements

**Over Recommended Operating Conditions** 

Symbol	Description	Тур.	Max.	Units
Standby (Power I	Down)			
I <sub>CCA-SB</sub>	V <sub>CCA</sub> current (per channel)	3	5	mA
I <sub>CCIB-SB</sub>	Input buffer current (per channel)	_	_	mA
I <sub>CCOB-SB</sub>	Output buffer current (per channel)	_	_	mA
Operating (Data	Rate = 3.2 Gbps)			
I <sub>CCA-OP</sub>	V <sub>CCA</sub> current (per channel)	68	77	mA
I <sub>CCIB-OP</sub>	Input buffer current (per channel)	5	7	mA
I <sub>CCOB-OP</sub>	Output buffer current (per channel)	19	25	mA
Operating (Data	Rate = 2.5 Gbps)			
I <sub>CCA-OP</sub>	V <sub>CCA</sub> current (per channel)	66	76	mA
I <sub>CCIB-OP</sub>	Input buffer current (per channel)	4	5	mA
I <sub>CCOB-OP</sub>	Output buffer current (per channel)	15	18	mA
Operating (Data	Rate = 1.25 Gbps)			
I <sub>CCA-OP</sub>	VCCA current (per channel)	62	72	mA
I <sub>CCIB-OP</sub>	Input buffer current (per channel)	4	5	mA
I <sub>CCOB-OP</sub>	Output buffer current (per channel)	15	18	mA
Operating (Data	Rate = 250 Mbps)			
I <sub>CCA-OP</sub>	V <sub>CCA</sub> current (per channel)	55	65	mA
I <sub>CCIB-OP</sub>	Input buffer current (per channel)	4	5	mA
I <sub>CCOB-OP</sub>	Output buffer current (per channel)	14	17	mA
Operating (Data	Rate = 150 Mbps)			•
I <sub>CCA-OP</sub>	V <sub>CCA</sub> current (per channel)	55	65	mA
I <sub>CCIB-OP</sub>	Input buffer current (per channel)	4	5	mA
I <sub>CCOB-OP</sub>	Output buffer current (per channel)	14	17	mA

#### Notes:

- Equalization enabled, pre-emphasis disabled.
- One quarter of the total quad power (includes contribution from common circuits, all channels in the quad operating, preemphasis disabled, equalization enabled).
- Pre-emphasis adds 20 mA to ICCA-OP data.



# 4.9. sysI/O Recommended Operating Conditions

Standard		v <sub>ccio</sub>			V <sub>REF</sub> (V)	
	Min.	Тур.	Max.	Min.	Тур.	Max.
LVCMOS33 <sup>2</sup>	3.135	3.3	3.465	_	_	_
LVCMOS33D	3.135	3.3	3.465	_	_	_
LVCMOS25 <sup>2</sup>	2.375	2.5	2.625	_	_	_
LVCMOS18	1.71	1.8	1.89	_	_	_
LVCMOS15	1.425	1.5	1.575	_	_	_
LVCMOS12 <sup>2</sup>	1.14	1.2	1.26	_	_	_
LVTTL33 <sup>2</sup>	3.135	3.3	3.465	_	_	_
PCI33	3.135	3.3	3.465	_	_	_
SSTL15 <sup>3</sup>	1.43	1.5	1.57	0.68	0.75	0.9
SSTL18_I, II <sup>2</sup>	1.71	1.8	1.89	0.833	0.9	0.969
SSTL25_I, II <sup>2</sup>	2.375	2.5	2.625	1.15	1.25	1.35
SSTL33_I, II <sup>2</sup>	3.135	3.3	3.465	1.3	1.5	1.7
HSTL15_I <sup>2</sup>	1.425	1.5	1.575	0.68	0.75	0.9
HSTL18_I, II <sup>2</sup>	1.71	1.8	1.89	0.816	0.9	1.08
LVDS25 <sup>2</sup>	2.375	2.5	2.625	_	_	_
LVDS25E	2.375	2.5	2.625	_	_	_
MLVDS <sup>1</sup>	2.375	2.5	2.625	_	_	_
LVPECL33 <sup>1, 2</sup>	3.135	3.3	3.465	_	_	_
Mini LVDS	2.375	2.5	2.625	_	_	_
BLVDS25 <sup>1, 2</sup>	2.375	2.5	2.625	_	_	_
RSDS <sup>2</sup>	2.375	2.5	2.625	_	_	_
RSDSE <sup>1, 2</sup>	2.375	2.5	2.625	_	_	_
TRLVDS	3.14	3.3	3.47	_	_	_
PPLVDS	3.14/2.25	3.3/2.5	3.47/2.75	_	_	_
SSTL15D <sup>3</sup>	1.43	1.5	1.57	_	_	_
SSTL18D_I <sup>2, 3</sup> , II <sup>2, 3</sup>	1.71	1.8	1.89	_	_	_
SSTL25D_ I <sup>2</sup> , II <sup>2</sup>	2.375	2.5	2.625	_	_	_
SSTL33D_ I <sup>2</sup> , II <sup>2</sup>	3.135	3.3	3.465	_	_	_
HSTL15D_ I <sup>2</sup>	1.425	1.5	1.575	_	_	_
HSTL18D_ I <sup>2</sup> , II <sup>2</sup>	1.71	1.8	1.89	_	_	_

#### Notes:

- 1. Inputs on chip. Outputs are implemented with the addition of external resistors.
- 2. For input voltage compatibility, see LatticeECP3 sysIO Usage Guide (FPGA-TN-02194).
- 3. VREF is required when using Differential SSTL to interface to DDR memory.



# 4.10. sysI/O Single-Ended DC Electrical Characteristics<sup>2</sup>

Input/Output	\	/ <sub>IL</sub>	V <sub>IH</sub>		V <sub>OL</sub> Max.	V <sub>OH</sub> Min.	I <sub>OL</sub> <sup>1</sup> (mA)	I <sub>OH</sub> <sup>1</sup> (mA)
Standard	Min. (V)	Max. (V)	Min. (V)	Max. (V)	(V)	(V)		
LVCMOS33	-0.3	0.8	2.0	3.6	0.4	V <sub>CCIO</sub> - 0.4	20, 16, 12, 8, 4	-20, -16, - 12, -8, -4
					0.2	V <sub>CCIO</sub> - 0.2	0.1	-0.1
LVCMOS25	-0.3	0.7	1.7	3.6	0.4	V <sub>CCIO</sub> - 0.4	20, 16, 12, 8, 4	-20, -16, - 12, -8, -4
					0.2	V <sub>CCIO</sub> - 0.2	0.1	-0.1
LVCMOS18	-0.3	0.35 V <sub>CCIO</sub>	0.65 V <sub>CCIO</sub>	3.6	0.4	V <sub>CCIO</sub> - 0.4	16, 12, 8, 4	-16, -12, -8, -4
					0.2	V <sub>CCIO</sub> - 0.2	0.1	-0.1
LVCMOS15	-0.3	0.35	0.65	3.6	0.4	V <sub>CCIO</sub> - 0.4	8, 4	-8, -4
		V <sub>CCIO</sub>	V <sub>CCIO</sub>		0.2	V <sub>CCIO</sub> - 0.2	0.1	-0.1
LVCMOS12	-0.3	0.35	0.65	3.6	0.4	V <sub>CCIO</sub> - 0.4	6, 2	-6, -2
		$V_{CC}$	$V_{CC}$		0.2	V <sub>CCIO</sub> - 0.2	0.1	-0.1
LVTTL33	-0.3	0.8	2.0	3.6	0.4	V <sub>CCIO</sub> - 0.4	20, 16, 12, 8, 4	-20, -16, - 12, -8, -4
					0.2	V <sub>CCIO</sub> - 0.2	0.1	-0.1
PCI33	-0.3	0.3 V <sub>CCIO</sub>	0.5 V <sub>CCIO</sub>	3.6	0.1 V <sub>CCIO</sub>	0.9 V <sub>CCIO</sub>	1.5	-0.5
SSTL18_I	-0.3	V <sub>REF</sub> - 0.125	V <sub>REF</sub> + 0.125	3.6	0.4	V <sub>CCIO</sub> - 0.4	6.7	-6.7
SSTL18_II (DDR2 Memory)	-0.3	V <sub>REF</sub> - 0.125	V <sub>REF</sub> + 0.125	3.6	0.28	V <sub>CCIO</sub> - 0.28	8 11	-8 -11
SSTL2_I	-0.3	V <sub>REF</sub> -	V <sub>REF</sub> +	3.6	0.54	V <sub>CCIO</sub> - 0.62	7.6	-7.6
55.1 <u>11</u> .	0.0	0.18	0.18			1000 0.02	12	-12
SSTL2_II (DDR	-0.3	V <sub>REF</sub> -	V <sub>REF</sub> +	3.6	0.35	V <sub>CCIO</sub> - 0.43	15.2	-15.2
Memory)		0.18	0.18			00.0	20	-20
SSTL3_I	-0.3	V <sub>REF</sub> - 0.2	V <sub>REF</sub> + 0.2	3.6	0.7	V <sub>CCIO</sub> - 1.1	8	-8
SSTL3_II	-0.3	V <sub>REF</sub> - 0.2	V <sub>REF</sub> + 0.2	3.6	0.5	V <sub>CCIO</sub> - 0.9	16	-16
SSTL15 (DDR3	-0.3	V <sub>REF</sub> -	V <sub>REF</sub> +	3.6	0.3	V <sub>CCIO</sub> - 0.3	7.5	-7.5
Memory)		0.1	0.1			V <sub>CCIO</sub> * 0.8	9	-9
HSTL15_I	-0.3	V <sub>REF</sub> -	V <sub>REF</sub> +	3.6	0.4	V <sub>CCIO</sub> -	4	-4
		0.1	0.1			0.4	8	-8
HSTL18_I	-0.3	V <sub>REF</sub> -	V <sub>REF</sub> +	3.6	0.4	V <sub>CCIO</sub> -	8	-8
		0.1	0.1			0.4	12	-12
HSTL18_II	-0.3	V <sub>REF</sub> - 0.1	V <sub>REF</sub> + 0.1	3.6	0.4	V <sub>CCIO</sub> - 0.4	16	-16

#### Notes:

- 1. For electromigration, the average DC current drawn by I/O pads between two consecutive VCCIO or GND pad connections, or between the last VCCIO or GND in an I/O bank and the end of an I/O bank, as shown in the Logic Signal Connections table (also shown as I/O grouping) shall not exceed n \* 8 mA, where n is the number of I/O pads between the two consecutive bank VCCIO or GND connections or between the last VCCIO and GND in a bank and the end of a bank. I/O Grouping can be found in the Data Sheet Pin Tables, which can also be generated from the Lattice Diamond software.
- 2. For I/Os with mixed voltage support,  $V_{OH}$  follows respective sysl/O bank  $V_{CCIO}$  supply voltage, and  $V_{IL}$  /  $V_{IH}$  follows the I/O signaling standard.



# 4.11. sysI/O Differential Electrical Characteristics

#### LVDS25

**Over Recommended Operating Conditions** 

Parameter	Description	Test Conditions	Min.	Тур.	Max.	Units
V <sub>INP</sub> *, V <sub>INM</sub> *	Input Voltage	_	0	_	2.4	V
V <sub>CM</sub> *	Input Common Mode Voltage	Half the Sum of the Two Inputs	0.05	_	2.35	V
$V_{THD}$	Differential Input Threshold	Difference Between the Two Inputs	+/-100	_	_	mV
I <sub>IN</sub>	Input Current	Power On or Power Off	_	_	+/-10	μΑ
V <sub>OH</sub>	Output High Voltage for V <sub>OP</sub> or V <sub>OM</sub>	RT = 100 Ω	_	1.38	1.60	V
V <sub>OL</sub>	Output Low Voltage for V <sub>OP</sub> or V <sub>OM</sub>	RT = 100 Ω	0.9 V	1.03	_	V
V <sub>OD</sub>	Output Voltage Differential	$(V_{OP} - V_{OM})$ , RT = 100 $\Omega$	250	350	450	mV
$\Delta V_{OD}$	Change in V <sub>OD</sub> Between High and Low	_	_	_	50	mV
Vos	Output Voltage Offset	$(V_{OP} + V_{OM})/2$ , RT = 100 $\Omega$	1.125	1.20	1.375	V
ΔV <sub>OS</sub>	Change in V <sub>OS</sub> Between H and L	_	_	_	50	mV
I <sub>SAB</sub>	Output Short Circuit Current	V <sub>OD</sub> = 0 V Driver Outputs Shorted to Each Other	_	_	12	mA

<sup>\*</sup>Note: On the left and right sides of the device, this specification is valid only for  $V_{CCIO} = 2.5 \text{ V}$  or 3.3 V.

#### 4.11.1. Differential HSTL and SSTL

Differential HSTL and SSTL outputs are implemented as a pair of complementary single-ended outputs. All allowable single-ended output classes (class I and class II) are supported in this mode.

#### 4.11.2. LVDS25E

The top and bottom sides of LatticeECP3 devices support LVDS outputs via emulated complementary LVCMOS outputs in conjunction with a parallel resistor across the driver outputs. The scheme shown in Figure 4.1 is one possible solution for point-to-point signals.

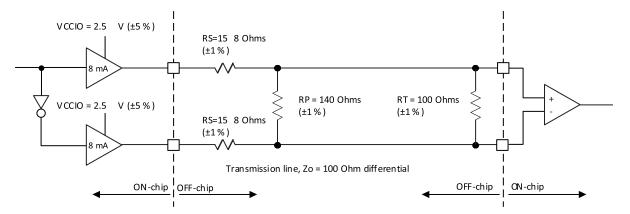


Figure 4.1. LVDS25E Output Termination Example

Table 4.1. LVDS25E DC Conditions

Parameter	Description	Typical	Units
V <sub>CCIO</sub>	Output Driver Supply (+/-5%)	2.50	V
Z <sub>OUT</sub>	Driver Impedance	20	Ω
R <sub>S</sub>	Driver Series Resistor (+/-1%)	158	Ω
R <sub>P</sub>	Driver Parallel Resistor (+/–1%)	140	Ω
R <sub>T</sub>	Receiver Termination (+/–1%)	100	Ω
V <sub>OH</sub>	Output High Voltage	1.43	V

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V <sub>OL</sub>	Output Low Voltage	1.07	V
V <sub>OD</sub>	Output Differential Voltage	0.35	V
V <sub>CM</sub>	Output Common Mode Voltage	1.25	V
Z <sub>BACK</sub>	Back Impedance	100.5	Ω
I <sub>DC</sub>	DC Output Current	6.03	mA

#### 4.11.3. LVCMOS33D

All I/O banks support emulated differential I/O using the LVCMOS33D I/O type. This option, along with the external resistor network, provides the system designer the flexibility to place differential outputs on an I/O bank with 3.3 V  $V_{\text{CCIO}}$ . The default drive current for LVCMOS33D output is 12 mA with the option to change the device strength to 4 mA, 8 mA, 16 mA or 20 mA. Follow the LVCMOS33 specifications for the DC characteristics of the LVCMOS33D.



#### 4.11.4. BLVDS25

The LatticeECP3 devices support the BLVDS standard. This standard is emulated using complementary LVCMOS outputs in conjunction with a parallel external resistor across the driver outputs. BLVDS is intended for use when multi-drop and bi-directional multi-point differential signaling is required. The scheme shown in Figure 4.2 is one possible solution for bi-directional multi-point differential signals.

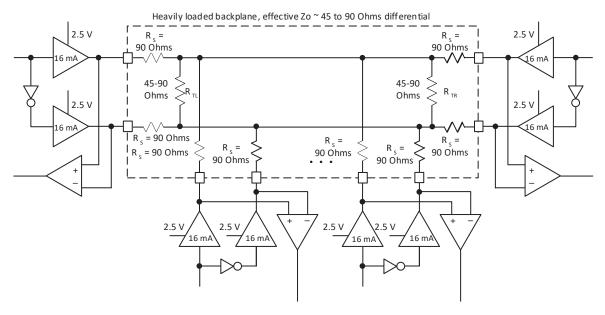


Figure 4.2. BLVDS25 Multi-point Output Example

**Table 4.2. BLVDS25 DC Conditions** 

**Over Recommended Operating Conditions** 

Parameter	Description	Туј	Typical		
		Zo = 45Ω	Zo = 90Ω		
V <sub>CCIO</sub>	Output Driver Supply (+/- 5%)	2.50	2.50	V	
Z <sub>OUT</sub>	Driver Impedance	10.00	10.00	Ω	
Rs	Driver Series Resistor (+/- 1%)	90.00	90.00	Ω	
R <sub>TL</sub>	Driver Parallel Resistor (+/- 1%)	45.00	90.00	Ω	
$R_{TR}$	Receiver Termination (+/- 1%)	45.00	90.00	Ω	
V <sub>OH</sub>	Output High Voltage	1.38	1.48	V	
V <sub>OL</sub>	Output Low Voltage	1.12	1.02	V	
V <sub>OD</sub>	Output Differential Voltage	0.25	0.46	V	
V <sub>CM</sub>	Output Common Mode Voltage	1.25	1.25	V	
I <sub>DC</sub>	DC Output Current	11.24	10.20	mA	

Note: For input buffer, see the LVDS table.



#### 4.12. LVPECL33

The LatticeECP3 devices support the differential LVPECL standard. This standard is emulated using complementary LVCMOS outputs in conjunction with a parallel resistor across the driver outputs. The LVPECL input standard is supported by the LVDS differential input buffer. The scheme shown in Figure 4.3 is one possible solution for point-to-point signals.

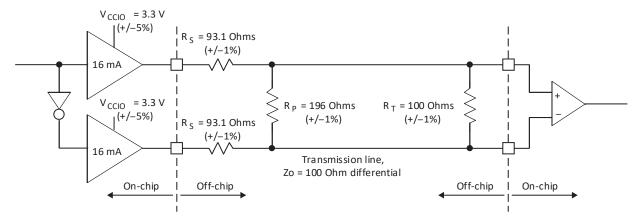


Figure 4.3. Differential LVPECL33

**Table 4.3. LVPECL33 DC Conditions** 

**Over Recommended Operating Conditions** 

Parameter	Description	Typical	Units
V <sub>CCIO</sub>	Output Driver Supply (+/–5%)	3.30	V
Z <sub>OUT</sub>	Driver Impedance	10	Ω
R <sub>S</sub>	Driver Series Resistor (+/-1%)	93	Ω
R <sub>P</sub>	Driver Parallel Resistor (+/-1%)	196	Ω
R <sub>T</sub>	Receiver Termination (+/-1%)	100	Ω
V <sub>OH</sub>	Output High Voltage	2.05	V
V <sub>OL</sub>	Output Low Voltage	1.25	V
V <sub>OD</sub>	Output Differential Voltage	0.80	V
V <sub>CM</sub>	Output Common Mode Voltage	1.65	V
Z <sub>BACK</sub>	Back Impedance	100.5	Ω
I <sub>DC</sub>	DC Output Current	12.11	mA

Note: For input buffer, see the LVDS table.



#### 4.13. RSDS25E

The LatticeECP3 devices support differential RSDS and RSDSE standards. This standard is emulated using complementary LVCMOS outputs in conjunction with a parallel resistor across the driver outputs. The RSDS input standard is supported by the LVDS differential input buffer. The scheme shown in Figure 4.4 is one possible solution for RSDS standard implementation. Resistor values in Figure 4.4 are industry standard values for 1% resistors.

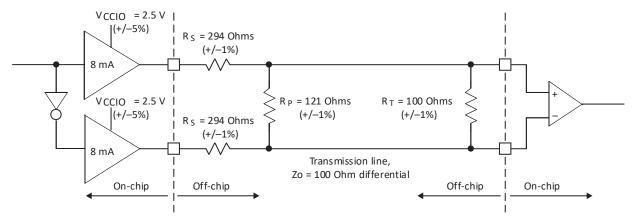


Figure 4.4. RSDS25E (Reduced Swing Differential Signaling)

#### **Table 4.4. RSDS25E DC Conditions**

**Over Recommended Operating Conditions** 

Parameter	Description	Typical	Units
V <sub>CCIO</sub>	Output Driver Supply (+/–5%)	2.50	V
Z <sub>OUT</sub>	Driver Impedance	20	Ω
$R_S$	Driver Series Resistor (+/-1%)	294	Ω
$R_P$	Driver Parallel Resistor (+/-1%)	121	Ω
R <sub>T</sub>	Receiver Termination (+/-1%)	100	Ω
V <sub>OH</sub>	Output High Voltage	1.35	V
V <sub>OL</sub>	Output Low Voltage	1.15	V
V <sub>OD</sub>	Output Differential Voltage	0.20	V
V <sub>CM</sub>	Output Common Mode Voltage	1.25	V
Z <sub>BACK</sub>	Back Impedance	101.5	Ω
I <sub>DC</sub>	DC Output Current	3.66	mA

Note: For input buffer, see the LVDS table.



#### 4.13.1. MLVDS25

The LatticeECP3 devices support the differential MLVDS standard. This standard is emulated using complementary LVCMOS outputs in conjunction with a parallel resistor across the driver outputs. The MLVDS input standard is supported by the LVDS differential input buffer. The scheme shown in Figure 4.5 is one possible solution for MLVDS standard implementation. Resistor values in Figure 4.5 are industry standard values for 1% resistors.

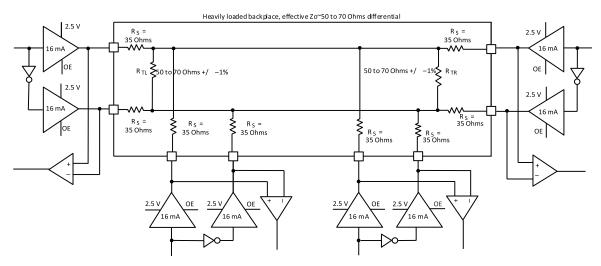


Figure 4.5. MLVDS25 (Multipoint Low Voltage Differential Signaling)

**Table 4.5. MLVDS25 DC Conditions** 

Parameter	Description	Typical	Typical		
		Ζο=50Ω	Ζο=70Ω		
V <sub>CCIO</sub>	Output Driver Supply (+/–5%)	2.50	2.50	V	
Z <sub>OUT</sub>	Driver Impedance	10.00	10.00	Ω	
RS	Driver Series Resistor (+/-1%)	35.00	35.00	Ω	
R <sub>TL</sub>	Driver Parallel Resistor (+/–1%)	50.00	70.00	Ω	
R <sub>TR</sub>	Receiver Termination (+/-1%)	50.00	70.00	Ω	
V <sub>OH</sub>	Output High Voltage	1.52	1.60	V	
V <sub>OL</sub>	Output Low Voltage	0.98	0.90	V	
V <sub>OD</sub>	Output Differential Voltage	0.54	0.70	V	
V <sub>CM</sub>	Output Common Mode Voltage	1.25	1.25	V	
IDC	DC Output Current	21.74	20.00	mA	

**Note**: For input buffer, see the LVDS table.



# 4.14. Typical Building Block Function Performance

# 4.14.1. Pin-to-Pin Performance (LVCMOS25 12 mA Drive)<sup>1, 2</sup>

Function	-8 Timing	Units			
Basic Functions					
16-bit Decoder	4.7	ns			
32-bit Decoder	4.7	ns			
64-bit Decoder	5.7	ns			
4:1 MUX	4.1	ns			
8:1 MUX	4.3	ns			
16:1 MUX	4.7	ns			
32:1 MUX	4.8	ns			

#### Notes:

- These functions were generated using the ispLEVER design tool. Exact performance may vary with device and tool version. The
  tool uses internal parameters that have been characterized but are not tested on every device.
- 2. Commercial timing numbers are shown. Industrial numbers are typically slower and can be extracted from the Diamond or ispLEVER software.

# 4.14.2. Register-to-Register Performance<sup>1, 2, 3</sup>

Function	–8 Timing	Units
Basic Functions	·	
16-bit Decoder	500	MHz
32-bit Decoder	500	MHz
64-bit Decoder	500	MHz
4:1 MUX	500	MHz
8:1 MUX	500	MHz
16:1 MUX	500	MHz
32:1 MUX	445	MHz
8-bit adder	500	MHz
16-bit adder	500	MHz
64-bit adder	305	MHz
16-bit counter	500	MHz
32-bit counter	460	MHz
64-bit counter	320	MHz
64-bit accumulator	315	MHz
Embedded Memory Functions		
512x36 Single Port RAM, EBR Output Registers	340	MHz
1024x18 True-Dual Port RAM (Write Through or Normal, EBR Output Registers)	340	MHz
1024x18 True-Dual Port RAM (Read-Before-Write, EBR Output Registers	130	MHz
1024x18 True-Dual Port RAM (Write Through or Normal, PLC Output Registers)	245	MHz
Distributed Memory Functions		
16x4 Pseudo-Dual Port RAM (One PFU)	500	MHz
32x4 Pseudo-Dual Port RAM	500	MHz
64x8 Pseudo-Dual Port RAM	400	MHz
DSP Function	·	
18x18 Multiplier (All Registers)	400	MHz
9x9 Multiplier (All Registers)	400	MHz
36x36 Multiply (All Registers)	260	MHz



## 4.14.3. Register-to-Register Performance<sup>1, 2, 3</sup>

Function	-8 Timing	Units
18x18 Multiply/Accumulate (Input and Output Registers)	200	MHz
18x18 Multiply-Add/Sub (All Registers)	400	MHz

#### Notes:

- 1. These timing numbers were generated using ispLEVER tool. Exact performance may vary with device and tool version. The tool uses internal parameters that have been characterized but are not tested on every device.
- 2. Commercial timing numbers are shown. Industrial numbers are typically slower and can be extracted from the Diamond or ispLEVER software.
- 3. For details on -9 speed grade devices, contact your Lattice Sales Representative.

## 4.15. Derating Timing Tables

Logic timing provided in the following sections of this data sheet and the Diamond and ispLEVER design tools are worst case numbers in the operating range. Actual delays at nominal temperature and voltage for best case process, can be much better than the values given in the tables. The Diamond and ispLEVER design tools can provide logic timing numbers at a particular temperature and voltage.



# **4.16.** LatticeECP3 External Switching Characteristics <sup>1, 2, 3, 13</sup> Over Recommended Commercial Operating Conditions

Parameter	Description	Device	-	-8	_	-7	_	-6	Units
			Min.	Max.	Min.	Max.	Min.	Max.	
Clocks									
Primary Clock <sup>6</sup>									
f <sub>MAX_PRI</sub>	Frequency for Primary Clock Tree	ECP3-150EA	-	500	_	420	_	375	MHz
t <sub>w_PRI</sub>	Clock Pulse Width for Primary Clock	ECP3-150EA	0.8	_	0.9	_	1.0	_	ns
t <sub>skew_pri</sub>	Primary Clock Skew Within a Device	ECP3-150EA	I	300	-	330	_	360	ps
t <sub>skew_prib</sub>	Primary Clock Skew Within a Bank	ECP3-150EA	-	250	_	280	_	300	ps
f <sub>MAX_PRI</sub>	Frequency for Primary Clock Tree	ECP3-70EA/95EA	_	500	_	420	_	375	MHz
t <sub>w_PRI</sub>	Pulse Width for Primary Clock	ECP3-70EA/95EA	0.8	_	0.9	_	1.0	_	ns
t <sub>skew_pri</sub>	Primary Clock Skew Within a Device	ECP3-70EA/95EA	-	360	_	370	_	380	ps
t <sub>skew_prib</sub>	Primary Clock Skew Within a Bank	ECP3-70EA/95EA	-	310	_	320	_	330	ps
f <sub>MAX_PRI</sub>	Frequency for Primary Clock Tree	ECP3-35EA	_	500	_	420	_	375	MHz
t <sub>W_PRI</sub>	Pulse Width for Primary Clock	ECP3-35EA	0. 8	_	0.9	_	1.0	_	ns
t <sub>skew_pri</sub>	Primary Clock Skew Within a Device	ECP3-35EA	I	300		330	_	360	ps
t <sub>SKEW_PRIB</sub>	Primary Clock Skew Within a Bank	ECP3-35EA	-	250	_	280	_	300	ps
f <sub>MAX_PRI</sub>	Frequency for Primary Clock Tree	ECP3-17EA	_	500	_	420	_	375	MHz
tw_pri	Pulse Width for Primary Clock	ECP3-17EA	0.8	_	0.9	_	1.0	_	ns
t <sub>skew_pri</sub>	Primary Clock Skew Within a Device	ECP3-17EA	_	310	_	340	_	370	ps
t <sub>skew_prib</sub>	Primary Clock Skew Within a Bank	ECP3-17EA	_	220	_	230	_	240	ps



Parameter	Description	Device	-	-8	-	-7	-	-6	Units
			Min.	Max.	Min.	Max.	Min.	Max.	
Edge Clock <sup>6</sup>	-	1		•		1	1	1	I.
f <sub>MAX_EDGE</sub>	Frequency for Edge Clock	ECP3-150EA	_	500	_	420	_	375	MHz
t <sub>W_EDGE</sub>	Clock Pulse Width for Edge Clock	ECP3-150EA	0.9	_	1.0	_	1.2	_	ns
t <sub>skew_edge_dqs</sub>	Edge Clock Skew Within an Edge of the Device	ECP3-150EA	ı	200	_	210	_	220	ps
f <sub>MAX_EDGE</sub>	Frequency for Edge Clock	ECP3-70EA/95EA	_	500	_	420	_	375	MHz
t <sub>W_EDGE</sub>	Clock Pulse Width for Edge Clock	ECP3-70EA/95EA	0. 9	_	1.0	_	1.2	_	ns
t <sub>skew_edge_dqs</sub>	Edge Clock Skew Within an Edge of the Device	ECP3-70EA/95EA	-	200	_	210	_	220	ps
f <sub>MAX_EDGE</sub>	Frequency for Edge Clock	ECP3-35EA	_	500	_	420	_	375	MHz
t <sub>W_EDGE</sub>	Clock Pulse Width for Edge Clock	ECP3-35EA	0. 9	_	1.0	_	1.2	_	ns
t <sub>skew_edge_dqs</sub>	Edge Clock Skew Within an Edge of the Device	ECP3-35EA	-	200	_	210	_	220	ps
f <sub>MAX_EDGE</sub>	Frequency for Edge Clock	ECP3-17EA	_	500	_	420	_	375	MHz
$t_{W\_{EDGE}}$	Clock Pulse Width for Edge Clock	ECP3-17EA	0. 9	_	1.0	_	1.2	_	ns
t <sub>SKEW_EDGE_DQS</sub>	Edge Clock Skew Within an Edge of the Device	ECP3-17EA	_	200	_	210	_	220	ps
Generic SDR	•								
	Parameters Using Ded		rimary Clo	1	ut PLL <sup>2</sup>				
t <sub>co</sub>	Clock to Output - PIO Output Register	ECP3-150EA	_	3.9	_	4.3	_	4.7	ns
t <sub>SU</sub>	Clock to Data Setup - PIO Input Register	ECP3-150EA	0.0	_	0.0	_	0.0	_	ns
t <sub>H</sub>	Clock to Data Hold - PIO Input Register	ECP3-150EA	1.5	_	1.7	_	2.0	_	ns
t <sub>su_del</sub>	Clock to Data Setup - PIO Input Register with Data Input Delay	ECP3-150EA	1.3	_	1.5	_	1.7	_	ns
t <sub>H_DEL</sub>	Clock to Data Hold - PIO Input Register with Input Data Delay	ECP3-150EA	0.0	_	0.0	_	0.0	_	ns

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Parameter	Description	Device	_	-8	_	7	-6		Units
			Min.	Max.	Min.	Max.	Min.	Max.	
f <sub>MAX_IO</sub>	Clock Frequency of I/O and PFU Register	ECP3-150EA	_	500	-	420	_	375	MHz
t <sub>co</sub>	Clock to Output - PIO Output Register	ECP3-70EA/95EA	_	3.8	_	4.2	_	4.6	ns
t <sub>SU</sub>	Clock to Data Setup - PIO Input Register	ECP3-70EA/95EA	0.0	_	0.0	_	0.0	_	ns
t <sub>H</sub>	Clock to Data Hold - PIO Input Register	ECP3-70EA/95EA	1.4	_	1.6	_	1.8	_	ns
t <sub>su_del</sub>	Clock to Data Setup - PIO Input Register with Data Input Delay	ECP3-70EA/95EA	1.3	_	1.5	_	1.7	_	ns
t <sub>h_del</sub>	Clock to Data Hold - PIO Input Register with Input Data Delay	ECP3-70EA/95EA	0.0	_	0.0	_	0.0	_	ns
f <sub>MAX_IO</sub>	Clock Frequency of I/O and PFU Register	ECP3-70EA/95EA	ı	500	_	420	_	375	MHz
t <sub>co</sub>	Clock to Output - PIO Output Register	ECP3-35EA	I	3.7	-	4.1	_	4.5	ns
t <sub>su</sub>	Clock to Data Setup - PIO Input Register	ECP3-35EA	0.0	_	0.0	_	0.0	_	ns
t <sub>H</sub>	Clock to Data Hold - PIO Input Register	ECP3-35EA	1.2	_	1.4	_	1.6	_	ns
t <sub>SU_DEL</sub>	Clock to Data Setup - PIO Input Register with Data Input Delay	ECP3-35EA	1.3	_	1.4	-	1.5	_	ns
t <sub>H_DEL</sub>	Clock to Data Hold - PIO Input Register with Input Data Delay	ECP3-35EA	0.0	_	0.0	-	0.0	_	ns
f <sub>MAX_IO</sub>	Clock Frequency of I/O and PFU Register	ECP3-35EA	_	500	_	420	_	375	MHz
t <sub>co</sub>	Clock to Output - PIO Output Register	ECP3-17EA	-	3.5	_	3.9	_	4.3	ns
t <sub>SU</sub>	Clock to Data Setup - PIO Input Register	ECP3-17EA	0.0	_	0.0	_	0.0	_	ns
t <sub>H</sub>	Clock to Data Hold - PIO Input Register	ECP3-17EA	1.3	_	1.5	-	1.6	_	ns



Parameter	Description	Device	_	-8	_	-7	-6		Units
			Min.	Max.	Min.	Max.	Min.	Max.	
t <sub>su_del</sub>	Clock to Data Setup - PIO Input Register with Data Input Delay	ECP3-17EA	1.3	_	1.4	_	1.5	_	ns
t <sub>H_DEL</sub>	Clock to Data Hold - PIO Input Register with Input Data Delay	ECP3-17EA	0.0	_	0.0	_	0.0	_	ns
f <sub>MAX_IO</sub>	Clock Frequency of I/O and PFU Register	ECP3-17EA	_	500	_	420	_	375	MHz
General I/O Pi	n Parameters Using Dec	dicated Clock Input P	rimary Clo	ock with P	LL with Cl	ock Injecti	on Remov	al Setting	2
t <sub>COPLL</sub>	Clock to Output - PIO Output Register	ECP3-150EA	_	3.3	_	3.6	_	39	ns
t <sub>SUPLL</sub>	Clock to Data Setup - PIO Input Register	ECP3-150EA	0.7	_	0.8	_	0.9	_	ns
t <sub>HPLL</sub>	Clock to Data Hold - PIO Input Register	ECP3-150EA	0.8	_	0.9	_	1.0	_	ns
t <sub>SU_DELPLL</sub>	Clock to Data Setup - PIO Input Register with Data Input Delay	ECP3-150EA	1.6	_	1.8	_	2.0	_	ns
t <sub>H_DELPLL</sub>	Clock to Data Hold - PIO Input Register with Input Data Delay	ECP3-150EA	_	0.0	_	0.0	_	0.0	ns
tcopll	Clock to Output - PIO Output Register	ECP3-70EA/95EA	_	3.3	_	3.5	_	3.8	ns
t <sub>SUPLL</sub>	Clock to Data Setup - PIO Input Register	ECP3-70EA/95EA	0.7	_	0.8	_	0.9	_	ns
t <sub>HPLL</sub>	Clock to Data Hold - PIO Input Register	ECP3-70EA/95EA	0.7	_	0.7	_	0.8	_	ns
t <sub>su_delpll</sub>	Clock to Data Setup - PIO Input Register with Data Input Delay	ECP3-70EA/95EA	1.6	_	1.8	_	2.0	_	ns
t <sub>H_DELPLL</sub>	Clock to Data Hold - PIO Input Register with Input Data Delay	ECP3-70EA/95EA	0.0	_	0.0	_	0.0	_	ns
t <sub>COPLL</sub>	Clock to Output - PIO Output Register	ECP3-35EA	_	3.2	_	3.4	_	3.6	ns
t <sub>SUPLL</sub>	Clock to Data Setup - PIO Input Register	ECP3-35EA	0.6	_	0.7	_	0.8	_	ns



Description		-8		-7		-6		Units
		Min.	Max.	Min.	Max.	Min.	Max.	
Clock to Data Hold - PIO Input Register	ECP3-35EA	0.3	-	0.3	_	0.4	_	ns
Clock to Data Setup - PIO Input Register with Data Input Delay	ECP3-35EA	1.6	_	1.7	-	1.8	-	ns
Clock to Data Hold - PIO Input Register with Input Data Delay	ECP3-35EA	0.0	_	0.0	-	0.0	_	ns
Clock to Output - PIO Output Register	ECP3-17EA	_	3.0	_	3.3	_	3.5	ns
Clock to Data Setup - PIO Input Register	ECP3-17EA	0.6	_	0.7	_	0.8	_	ns
Clock to Data Hold - PIO Input Register	ECP3-17EA	0.3	_	0.3	_	0.4	_	ns
Clock to Data Setup - PIO Input Register with Data Input Delay	ECP3-17EA	1.6	_	1.7	_	1.8	_	ns
Clock to Data Hold - PIO Input Register with Input Data Delay	ECP3-17EA	0.0	_	0.0	_	0.0	_	ns
	1	•	•			•		I.
nputs with Clock and	Data (>10 Bits Wide	e) Centered	d at Pin (G	DDRX1_R	K.SCLK.Cei	ntered) Us	ing PCLK F	Pin for
Data Setup Before CLK	All ECP3EA Devices	480	_	480	_	480	_	ps
Data Hold After CLK	All ECP3EA Devices	480	_	480	_	480	_	ps
DDRX1 Clock Frequency	All ECP3EA Devices	_	250	_	250	_	250	MHz
nputs with Clock and	Data (>10 Bits Wide	e) Aligned a	at Pin (GD	DRX1_RX.	SCLK.PLL.	Aligned) U	sing PLLCL	KIN Pin
and Top Sides and Clo	ock Left and Right S	ides						
Data Setup Before CLK	All ECP3EA Devices	_	0.225	_	0.225	_	0.225	UI
Data Hold After CLK	All ECP3EA Devices	0.775	_	0.775	_	0.775	_	UI
DDRX1 Clock	All ECP3EA	_	250	_	250	_	250	MHz
	Hold - PIO Input Register  Clock to Data Setup - PIO Input Register with Data Input Delay  Clock to Data Hold - PIO Input Register with Input Data Delay  Clock to Output - PIO Output Register  Clock to Data Setup - PIO Input Register with Data Input Delay  Clock to Data Hold - PIO Input Register with Data Input Delay  Clock to Data Hold - PIO Input Register with Input Data Delay  Data Setup Before CLK  Data Hold After CLK  Data Setup Before CLK  Data Setup Before CLK  Data Setup Before CLK  Data Hold After CLK	Hold - PIO Input Register  Clock to Data Setup - PIO Input Register with Data Input Delay  Clock to Data Hold - PIO Input Register with Input Data Delay  Clock to Output - PIO Output Register Clock to Data Setup - PIO Input Register  Clock to Data Hold - PIO Input Register  Clock to Data Setup - PIO Input Register  Clock to Data Hold - PIO Input Register with Data Input Delay  Clock to Data Setup - PIO Input Register with Data Input Delay  Clock to Data Setup - PIO Input Register with Data Input Delay  Clock to Data Hold - PIO Input Register with Input Data Delay  Data Setup Before CLK  Data Hold After CLK  DDRX1 Clock Frequency  Devices  DDRX1 Clock and Data (>10 Bits Wide  All ECP3EA Devices  DDRX1 Clock Frequency  Devices  Data Hold After CLK  Data Setup Before CLK  Devices  Data Setup Before CLK  Devices  All ECP3EA Devices  Data Hold After CLK  Data Setup Before CLK  Devices  All ECP3EA Devices  All ECP3EA Devices  All ECP3EA Devices  Data Hold After CLK  Data Hold After CLK  Devices  All ECP3EA Devices  All ECP3EA Devices  All ECP3EA Devices  All ECP3EA Devices  Data Hold After CLK  Devices  All ECP3EA Devices  All ECP3EA Devices	Clock to Data Hold - PIO Input Register  Clock to Data Setup - PIO Input Register with Data Input Delay  Clock to Data Hold - PIO Input Register with Data Input Delay  Clock to Data Hold - PIO Input Register with Input Data Delay  Clock to Output - PIO Output Register  Clock to Data Setup - PIO Input Register  Clock to Data Setup - PIO Input Register  Clock to Data Hold - PIO Input Register  Clock to Data Setup - PIO Input Register  Clock to Data Setup - PIO Input Register  Clock to Data Setup - PIO Input Register with Data Input Delay  Clock to Data Hold - PIO Input Register with Data Input Delay  Clock to Data Hold - PIO Input Register with Input Data Delay  Clock to Data Hold - PIO Input Register with Input Data Delay  All ECP3EA  Devices  DATA Hold After CLK Devices  DATA Hold After CLK DEVICES  DATA Hold After CLK Devices  DATA Setup Before CLK DEVICES  DATA Hold After All ECP3EA Devices  DATA Setup Before CLK Devices  DATA Setup Before CLK Devices  DATA Hold After All ECP3EA Devices  DATA Setup Before CLK Devices  DATA Setup Before CLK Devices  DATA Hold After All ECP3EA Devices  DATA Setup Before CLK Devices  DATA Hold After CLK Devices  DATA Hold After CLK Devices  DATA Hold After All ECP3EA Devices  DATA Hold After CLK Devices	Clock to Data Hold - PIO Input Register  Clock to Data Setup - PIO Input Register with Data Input Delay  Clock to Data Hold - PIO Input Register with Input Data Delay  Clock to Data Clock to Output Register with Input Data Delay  Clock to Output Register  Clock to Data Setup - PIO Input Register  Clock to Data Setup - PIO Input Register  Clock to Data Hold - PIO Input Register  Clock to Data Setup - PIO Input Register with Data Input Delay  Clock to Data Hold - PIO Input Register with Data Input Delay  Clock to Data Hold - PIO Input Register with Input Data Delay  Clock to Data Hold - PIO Input Register with Input Data Delay  Clock to Data Hold - PIO Input Register with Input Data Delay  Clock to Data Hold - PIO Input Register with Input Data Delay  Clock to Data Hold - PIO Input Register with Input Data Delay  Clock to Data Hold - PIO Input Register with Input Data Delay  Clock to Data Hold - PIO Input Register with Input Data Delay  Clock to Data Hold - PIO Input Register with Input Data Delay  Clock to Data Hold - PIO Input Register with Input Data Delay  Clock to Data Hold - PIO Input Register with Input Data Delay  Clock to Data Hold - PIO Input Register with Input Data Delay  Clock to Data Hold - PIO Input Register  CLK Data Hold After CLK Data Setup Before CLK Data Setup Before CLK Data Hold After CLK Data Hold	Clock to Data Hold - PIO Input Register  Clock to Data Setup - PIO Input Register with Data Input Data Delay  Clock to Data Hold - PIO Input Register with Data Input Data Delay  Clock to Data Hold - PIO Input Register with Input Data Delay  Clock to Output - PIO Output Register  Clock to Data Setup - PIO Input Register  Clock to Data Setup Data Setup Data PIO Input Register  Clock to Data Hold - PIO Input Register  Clock to Data Hold - PIO Input Register  Clock to Data Hold - PIO Input Register  Clock to Data Setup - PIO Input Register  Clock to Data Hold - PIO Input Register with Data Input Delay  Clock to Data Hold - PIO Input Register with Data Input Delay  Clock to Data Hold - PIO Input Register with Input Data Delay  Clock to Data Hold - PIO Input Register with Input Data Delay  Clock to Data Hold After All ECP3EA ABO - 480  Data Setup Data Hold After CLK Devices  Data Hold After All ECP3EA - 250 - Engular All ECP3EA ABO ABORX1_RX  Data Setup Devices  Data Setup Devices  Data Setup All ECP3EA - 250 - Engular All ECP3EA ABO ABORX1_RX  Data Setup Devices  Data Setup Before CLK Devices  Data Setup All ECP3EA - 250 - Enguls Setup Before CLK Devices  Data Setup Before CLK Devices  Data Setup All ECP3EA - 0.225 - Enguls Hold After CLK Devices  Data Setup Before CLK Devices  Data Hold After All ECP3EA - 0.775 - 0.775  Data Hold After All ECP3EA ABO - 7.75 - 0.775	Clock to Data	Clock to Data	Clock to Data



Parameter	Description	Device	_	8	_	7	-6		Units
			Min.	Max.	Min.	Max.	Min.	Max.	
Generic DDRX1 Clock Input	Inputs with Clock and	Data (>10 Bits Wide)	Aligned a	t Pin (GDI	DRX1_RX.	SCLK.Align	ned) Using	DLL - CLKI	N Pin fo
Data Left, Righ	t and Top Sides and Clo	ock Left and Right Sid	es						
t <sub>DVACLKGDDR</sub>	Data Setup Before CLK	All ECP3EA Devices	_	0.225	_	0.225	_	0.225	UI
t <sub>DVECLKGDDR</sub>	Data Hold After CLK	All ECP3EA Devices	0.775	_	0.775	_	0.775	_	UI
f <sub>MAX_GDDR</sub>	DDRX1 Clock Frequency	All ECP3EA Devices	_	250	_	250	_	250	MHz
Generic DDRX1 Clock Input	Inputs with Clock and	Data (<10 Bits Wide)	Centered	at Pin (GI	DDRX1_R	(.DQS.Cen	tered) Usi	ng DQS Pi	n for
t <sub>SUGDDR</sub>	Data Setup After CLK	All ECP3EA Devices	535	_	535	_	535	_	ps
t <sub>HOGDDR</sub>	Data Hold After CLK	All ECP3EA Devices	535	1	535	1	535	1	ps
f <sub>MAX_GDDR</sub>	DDRX1 Clock Frequency	All ECP3EA Devices	_	250	ı	250	_	250	MHz
Generic DDRX1 Input	Inputs with Clock and	Data (<10bits wide)	Aligned at	Pin (GDD	RX1_RX.D	QS.Aligne	d) Using D	QS Pin fo	Clock
Data and Clock	Left and Right Sides								
t <sub>DVACLKGDDR</sub>	Data Setup Before CLK	All ECP3EA Devices	_	0.225	-	0.225	_	0.225	UI
t <sub>DVECLKGDDR</sub>	Data Hold After CLK	All ECP3EA Devices	0.775	_	0.775	_	0.775	_	UI
f <sub>MAX_GDDR</sub>	DDRX1 Clock Frequency	All ECP3EA Devices	_	250	_	250	_	250	MHz
Generic DDRX2 Clock Input	Inputs with Clock and	Data (>10 Bits Wide)	Centered	l at Pin (GI	DDRX2_R	(.ECLK.Ce	ntered) Us	ing PCLK P	in for
Left and Right	Sides								
t <sub>SUGDDR</sub>	Data Setup Before CLK	ECP3-150EA	321	_	403	_	471	_	ps
t <sub>HOGDDR</sub>	Data Hold After CLK	ECP3-150EA	321	_	403	_	471	_	ps
f <sub>MAX_GDDR</sub>	DDRX2 Clock Frequency	ECP3-150EA	_	405	-	325	_	280	MHz
t <sub>SUGDDR</sub>	Data Setup Before CLK	ECP3-70EA/95EA	321	_	403	1	535	1	
									ps
t <sub>HOGDDR</sub>	Data Hold After CLK	ECP3-70EA/95EA	321	_	403	_	535	-	ps
	Data Hold After	ECP3-70EA/95EA  ECP3-70EA/95EA	321	— 405	403 —	325	535		ps
f <sub>MAX_GDDR</sub>	Data Hold After CLK DDRX2 Clock								ps
thogddr  fmax_gddr  tsugddr  thogddr	Data Hold After CLK  DDRX2 Clock Frequency Data Setup	ECP3-70EA/95EA	_		_	325	_		ps MHz
f <sub>MAX_GDDR</sub> t <sub>SUGDDR</sub> t <sub>HOGDDR</sub>	Data Hold After CLK  DDRX2 Clock Frequency  Data Setup Before CLK  Data Hold After	ECP3-70EA/95EA ECP3-35EA	335		- 425	325	- 535		ps MHz ps
f <sub>MAX_GDDR</sub>	Data Hold After CLK  DDRX2 Clock Frequency  Data Setup Before CLK  Data Hold After CLK  DDRX2 Clock	ECP3-70EA/95EA  ECP3-35EA  ECP3-35EA	335	405 — —	- 425	325	- 535	250 — —	ps MHz ps ps

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Parameter	Description	Device	-	-8	_	7	_	6	Units
			Min.	Max.	Min.	Max.	Min.	Max.	
f <sub>MAX_GDDR</sub>	DDRX2 Clock Frequency	ECP3-17EA	_	405	_	325	_	250	MHz
Generic DDRX2	Inputs with Clock and	Data (>10 Bits Wide)	Aligned a	at Pin (GD	DRX2_RX.	ECLK.Align	ned)		
Left and Right	Side Using DLLCLKIN Pi	n for Clock Input							
t <sub>dvaclkgddr</sub>	Data Setup Before CLK	ECP3-150EA	-	0.225	_	0.225	_	0.225	UI
t <sub>DVECLKGDDR</sub>	Data Hold After CLK	ECP3-150EA	0.775	_	0.775	_	0.775	1	UI
$f_{MAX\_GDDR}$	DDRX2 Clock Frequency	ECP3-150EA	1	460	_	385	_	345	MHz
t <sub>DVACLKGDDR</sub>	Data Setup Before CLK	ECP3-70EA/95EA	_	0.225	_	0.225	_	0.225	UI
t <sub>DVECLKGDDR</sub>	Data Hold After CLK	ECP3-70EA/95EA	0.775	_	0.775	_	0.775	1	UI
f <sub>MAX_GDDR</sub>	DDRX2 Clock Frequency	ECP3-70EA/95EA	-	460	_	385	_	311	MHz
tdvaclkgddr	Data Setup Before CLK	ECP3-35EA	_	0.210	_	0.210	_	0.210	UI
t <sub>DVECLKGDDR</sub>	Data Hold After CLK	ECP3-35EA	0.790	_	0.790	_	0.790	_	UI
f <sub>MAX_GDDR</sub>	DDRX2 Clock Frequency	ECP3-35EA	_	460	_	385	_	311	MHz
t <sub>DVACLKGDDR</sub>	Data Setup Before CLK	ECP3-17EA	_	0.210	_	0.210	_	0.210	UI
toveclkgddr	Data Hold After CLK	ECP3-17EA	0.790	_	0.790	_	0.790	_	UI
f <sub>MAX_GDDR</sub>	DDRX2 Clock Frequency	ECP3-17EA	_	460	_	385	_	311	MHz
<b>Top Side Using</b>	PCLK Pin for Clock Inp	ut							
t <sub>DVACLKGDDR</sub>	Data Setup Before CLK	ECP3-150EA	_	0.225	_	0.225	_	0.225	UI
t <sub>DVECLKGDDR</sub>	Data Hold After CLK	ECP3-150EA	0.775	_	0.775	_	0.775	1	UI
$f_{\text{MAX\_GDDR}}$	DDRX2 Clock Frequency	ECP3-150EA	1	235	_	170	_	130	MHz
t <sub>DVACLKGDDR</sub>	Data Setup Before CLK	ECP3-70EA/95EA	_	0.225	_	0.225	_	0.225	UI
t <sub>DVECLKGDDR</sub>	Data Hold After CLK	ECP3-70EA/95EA	0.775	_	0.775	_	0.775	_	UI
f <sub>MAX_GDDR</sub>	DDRX2 Clock Frequency	ECP3-70EA/95EA	_	235	_	170	_	130	MHz
t <sub>DVACLKGDDR</sub>	Data Setup Before CLK	ECP3-35EA	_	0.210	_	0.210	_	0.210	UI
t <sub>DVECLKGDDR</sub>	Data Hold After CLK	ECP3-35EA	0.790	_	0.790	_	0.790	_	UI
f <sub>MAX_GDDR</sub>	DDRX2 Clock Frequency	ECP3-35EA	_	235	_	170	_	130	MHz
t <sub>DVACLKGDDR</sub>	Data Setup Before CLK	ECP3-17EA	_	0.210	_	0.210	_	0.210	UI
t <sub>DVECLKGDDR</sub>	Data Hold After CLK	ECP3-17EA	0.790	_	0.790	_	0.790	_	UI



Parameter	Description	Device	-	-8	-	-7	-	-6	Units
			Min.	Max.	Min.	Max.	Min.	Max.	
f <sub>MAX_GDDR</sub>	DDRX2 Clock Frequency	ECP3-17EA	_	235	_	170	_	130	MHz
Generic DDRX2	Inputs with Clock and	Data (>10bits wide)	are Aligne	ed at Pin (	GDDRX2_F	RX.ECLK.A	ligned) (N	o CLKDIV)	
Left and Right	Sides Using DLLCLKPIN	for Clock Input							
t <sub>DVACLKGDDR</sub>	Data Setup Before CLK	ECP3-150EA	_	0.225	_	0.225	_	0.225	UI
t <sub>DVECLKGDDR</sub>	Data Hold After CLK	ECP3-150EA	0.775	_	0.775	_	0.775	_	UI
f <sub>MAX_GDDR</sub>	DDRX2 Clock Frequency	ECP3-150EA	_	460	_	385	_	345	MHz
t <sub>DVACLKGDDR</sub>	Data Setup Before CLK	ECP3-70EA/95EA	-	0.225	_	0.225	_	0.225	UI
t <sub>DVECLKGDDR</sub>	Data Hold After CLK	ECP3-70EA/95EA	0.775	_	0.775	_	0.775	_	UI
f <sub>MAX_GDDR</sub>	DDRX2 Clock Frequency	ECP3-70EA/95EA	1	460	_	385	_	311	MHz
t <sub>DVACLKGDDR</sub>	Data Setup Before CLK	ECP3-35EA	_	0.210	_	0.210	_	0.210	UI
t <sub>DVECLKGDDR</sub>	Data Hold After CLK	ECP3-35EA	0.790	_	0.790	_	0.790	_	UI
f <sub>MAX_GDDR</sub>	DDRX2 Clock Frequency	ECP3-35EA	_	460	_	385	_	311	MHz
t <sub>dvaclkgddr</sub>	Data Setup Before CLK (Left and Right Sides)	ECP3-17EA	_	0.210	_	0.210	_	0.210	UI
t <sub>DVECLKGDDR</sub>	Data Hold After CLK	ECP3-17EA	0.790	_	0.790	_	0.790	_	UI
f <sub>MAX_GDDR</sub>	DDRX2 Clock Frequency	ECP3-17EA	_	460	_	385	_	311	MHz
Top Side Using	PCLK Pin for Clock Inp	ut							
t <sub>DVACLKGDDR</sub>	Data Setup Before CLK	ECP3-150EA	_	0.225	_	0.225	_	0.225	UI
t <sub>DVECLKGDDR</sub>	Data Hold After CLK	ECP3-150EA	0.775	_	0.775	_	0.775	_	UI
$f_{MAX\_GDDR}$	DDRX2 Clock Frequency	ECP3-150EA	_	235	_	170	_	130	MHz
t <sub>DVACLKGDDR</sub>	Data Setup Before CLK	ECP3-70EA/95EA	_	0.225	_	0.225	_	0.225	UI
t <sub>DVECLKGDDR</sub>	Data Hold After CLK	ECP3-70EA/95EA	0.775	_	0.775	_	0.775	_	UI
f <sub>MAX_GDDR</sub>	DDRX2 Clock Frequency	ECP3-70EA/95EA	-	235	_	170	_	130	MH
t <sub>DVACLKGDDR</sub>	Data Setup Before CLK	ECP3-35EA	-	0.210	_	0.210	_	0.210	UI
t <sub>DVECLKGDDR</sub>	Data Hold After CLK	ECP3-35EA	0.790	_	0.790	_	0.790	_	UI
f <sub>MAX_GDDR</sub>	DDRX2 Clock Frequency	ECP3-35EA	_	235	_	170	_	130	MHz
t <sub>DVACLKGDDR</sub>	Data Setup Before CLK	ECP3-17EA	_	0.210	_	0.210	_	0.210	UI



Parameter	Description	Device	-	8	_	·7	-6		Units
			Min.	Max.	Min.	Max.	Min.	Max.	
t <sub>DVECLKGDDR</sub>	Data Hold After CLK	ECP3-17EA	0.790	_	0.790	_	0.790	_	UI
f <sub>MAX_GDDR</sub>	DDRX2 Clock Frequency	ECP3-17EA	_	235	_	170	_	130	MHz
Generic DDRX2 Clock Input	Inputs with Clock and	Data (<10 Bits Wide)	Centered	at Pin (G	DDRX2_R)	(.DQS.Cen	itered) Usi	ing DQS Pi	n for
Left and Right	Sides								
t <sub>SUGDDR</sub>	Data Setup Before CLK	All ECP3EA Devices	330	_	330	_	352	_	ps
t <sub>HOGDDR</sub>	Data Hold After CLK	All ECP3EA Devices	330	-	330	_	352	_	ps
f <sub>MAX_GDDR</sub>	DDRX2 Clock Frequency	All ECP3EA Devices	_	400	_	400	_	375	MHz
	Inputs with Clock and	Data (<10 Bits Wide)	Aligned a	t Pin (GD	DRX2_RX.	DQS.Align	ed) Using	DQS Pin fo	or Clock
Input Left and Right:	Sides								
t <sub>DVACLKGDDR</sub>	Data Setup Before CLK	All ECP3EA Devices	_	0.225	_	0.225	_	0.225	UI
t <sub>DVECLKGDDR</sub>	Data Hold After CLK	All ECP3EA Devices	0.775	-	0.775	_	0.775	_	UI
f <sub>MAX_GDDR</sub>	DDRX2 Clock Frequency	All ECP3EA Devices	_	400	_	400	_	375	MHz
Generic DDRX1	Output with Clock and	Data (>10 Bits Wide	e) Centere	d at Pin (G	DDRX1_T	X.SCLK.Ce	ntered) <sup>10</sup>		
t <sub>DVBGDDR</sub>	Data Valid Before CLK	ECP3-150EA	670	_	670	_	670	_	ps
t <sub>DVAGDDR</sub>	Data Valid After CLK	ECP3-150EA	670	-	670	_	670	_	ps
f <sub>MAX_GDDR</sub>	DDRX1 Clock Frequency	ECP3-150EA	_	250	_	250	_	250	MHz
t <sub>DVBGDDR</sub>	Data Valid Before CLK	ECP3-70EA/95EA	666	-	665	_	664	_	ps
t <sub>DVAGDDR</sub>	Data Valid After CLK	ECP3-70EA/95EA	666	_	665	_	664	_	ps
f <sub>MAX_GDDR</sub>	DDRX1 Clock Frequency	ECP3-70EA/95EA	-	250	_	250	_	250	MHz
t <sub>DVBGDDR</sub>	Data Valid Before CLK	ECP3-35EA	683	_	688	_	690	_	ps
t <sub>DVAGDDR</sub>	Data Valid After CLK	ECP3-35EA	683	_	688	_	690	_	ps
f <sub>MAX_GDDR</sub>	DDRX1 Clock Frequency	ECP3-35EA	_	250	_	250	_	250	MHz
t <sub>DVBGDDR</sub>	Data Valid Before CLK	ECP3-17EA	683	_	688	_	690	_	ps
t <sub>DVAGDDR</sub>	Data Valid After CLK	ECP3-17EA	683	_	688	_	690	_	ps
f <sub>MAX_GDDR</sub>	DDRX1 Clock Frequency	ECP3-17EA	_	250	_	250	_	250	MHz



Parameter	Description	Device	_	-8	_	-7	-6		Units
	,		Min.	Max.	Min.	Max.	Min.	Max.	
Generic DDRX1	Output with Clock and	I Data Aligned at Pin	(GDDRX1	TX.SCLK.	Aligned)10	)	1	1	I
t <sub>DIBGDDR</sub>	Data Invalid Before Clock	ECP3-150EA	_	335		338	_	341	ps
t <sub>DIAGDDR</sub>	Data Invalid After Clock	ECP3-150EA	_	335	_	338	_	341	ps
f <sub>MAX_GDDR</sub>	DDRX1 Clock Frequency	ECP3-150EA	-	250	_	250	_	250	MHz
t <sub>DIBGDDR</sub>	Data Invalid Before Clock	ECP3-70EA/95EA	-	339	_	343	_	347	ps
t <sub>DIAGDDR</sub>	Data Invalid After Clock	ECP3-70EA/95EA	-	339	_	343	_	347	ps
f <sub>MAX_GDDR</sub>	DDRX1 Clock Frequency	ECP3-70EA/95EA	_	250	_	250	_	250	MHz
t <sub>DIBGDDR</sub>	Data Invalid Before Clock	ECP3-35EA	_	322	_	320	_	321	ps
t <sub>DIAGDDR</sub>	Data Invalid After Clock	ECP3-35EA	_	322	_	320	_	321	ps
f <sub>MAX_GDDR</sub>	DDRX1 Clock Frequency	ECP3-35EA	-	250	_	250	_	250	MHz
t <sub>DIBGDDR</sub>	Data Invalid Before Clock	ECP3-17EA	_	322	_	320	_	321	ps
t <sub>DIAGDDR</sub>	Data Invalid After Clock	ECP3-17EA	_	322	_	320	_	321	ps
f <sub>MAX_GDDR</sub>	DDRX1 Clock Frequency	ECP3-17EA	_	250	_	250	_	250	MHz
Generic DDRX1	Output with Clock and	Data (<10 Bits Wide	e) Centere	d at Pin (	GDDRX1_T	X.DQS.Ce	ntered) <sup>10</sup>	•	
Left and Right	Sides								
t <sub>DVBGDDR</sub>	Data Valid Before CLK	ECP3-150EA	670	_	670	_	670	_	ps
t <sub>DVAGDDR</sub>	Data Valid After CLK	ECP3-150EA	670	_	670	_	670	_	ps
f <sub>MAX_GDDR</sub>	DDRX1 Clock Frequency	ECP3-150EA	_	250	_	250	_	250	MHz
t <sub>DVBGDDR</sub>	Data Valid Before CLK	ECP3-70EA/95EA	657	_	652	_	650	_	ps
t <sub>DVAGDDR</sub>	Data Valid After CLK	ECP3-70EA/95EA	657	_	652	_	650	_	ps
f <sub>MAX_GDDR</sub>	DDRX1 Clock Frequency	ECP3-70EA/95EA	ı	250	_	250	_	250	MHz
t <sub>DVBGDDR</sub>	Data Valid Before CLK	ECP3-35EA	670	_	675	_	676	_	ps
t <sub>DVAGDDR</sub>	Data Valid After CLK	ECP3-35EA	670	_	675	_	676	_	ps
f <sub>MAX_GDDR</sub>	DDRX1 Clock Frequency	ECP3-35EA	_	250	_	250	_	250	MHz
t <sub>DVBGDDR</sub>	Data Valid Before CLK	ECP3-17EA	670	_	670	_	670	_	ps
t <sub>DVAGDDR</sub>	Data Valid After CLK	ECP3-17EA	670	_	670	_	670	_	ps
f <sub>MAX_GDDR</sub>	DDRX1 Clock Frequency	ECP3-17EA	_	250	_	250	_	250	MHz



Parameter	Description	Device	-	-8	-	-7	_	-6	Units
			Min.	Max.	Min.	Max.	Min.	Max.	
Generic DDRX2	2 Output with Clock and	Data (>10 Bits Wide	e) Aligned	at Pin (GE	DRX2_TX	.Aligned)	'	•	II.
Left and Right	Sides	·		<u> </u>					
t <sub>DIBGDDR</sub>	Data Invalid Before Clock	All ECP3EA Devices	_	200	_	210	_	220	ps
t <sub>DIAGDDR</sub>	Data Invalid After Clock	All ECP3EA Devices	_	200	_	210	_	220	ps
f <sub>MAX_GDDR</sub>	DDRX2 Clock Frequency	All ECP3EA Devices	_	500	_	420	_	375	MHz
Generic DDRX2	2 Output with Clock and	Data (<10 Bits Wide	e) Centere	d at Pin U	sing DQSE	LL (GDDR	X2_TX.DQ	SDLL.Cent	ered) <sup>11</sup>
Left and Right		•	<u> </u>						<u> </u>
<sup>t</sup> DVBGDDR	Data Valid Before CLK	All ECP3EA Devices	400	_	400	_	431	_	ps
<sup>t</sup> DVAGDDR	Data Valid After CLK	All ECP3EA Devices	400	_	400	_	432	_	ps
fMAX_GDDR	DDRX2 Clock Frequency	All ECP3EA Devices	_	400	_	400	_	375	MHz
Generic DDRX2	Output with Clock and	Data (>10 Bits Wide	e) Centere	d at Pin U	sing PLL (0	GDDRX2_T	X.PLL.Cen	tered) <sup>10</sup>	l.
Left and Right					·	_			
t <sub>DVBGDDR</sub>	Data Valid Before CLK	All ECP3EA Devices	285	_	370	_	431	_	ps
t <sub>DVAGDDR</sub>	Data Valid After CLK	All ECP3EA Devices	285	_	370	_	432	_	ps
f <sub>MAX_GDDR</sub>	DDRX2 Clock Frequency	All ECP3EA Devices	_	500	_	420	_	375	MHz
Memory Interf	ace	l	1	l		1	l		ı
	Pin Parameters (Input	Data are Strobe Edg	e Aligned,	Output S	trobe Edge	e is Data C	entered)4		
t <sub>DVADQ</sub>	Data Valid After DQS (DDR Read)	All ECP3 Devices	_	0.225	_	0.225	_	0.225	UI
t <sub>DVEDQ</sub>	Data Hold After DQS (DDR Read)	All ECP3 Devices	0.64	_	0.64	_	0.64	_	UI
t <sub>DQVBS</sub>	Data Valid Before DQS	All ECP3 Devices	0.25	_	0.25	_	0.25	_	UI
t <sub>DQVAS</sub>	Data Valid After DQS	All ECP3 Devices	0.25	_	0.25	_	0.25	_	UI
f <sub>MAX_DDR</sub>	DDR Clock Frequency	All ECP3 Devices	95	200	95	200	95	166	MHz
f <sub>MAX_DDR2</sub>	DDR2 clock frequency	All ECP3 Devices	125	266	125	200	125	166	MHz
DDR3 (Using Pl	LL for SCLK) I/O Pin Para	ameters							
t <sub>DVADQ</sub>	Data Valid After DQS (DDR Read)	All ECP3 Devices	_	0.225	_	0.225	_	0.225	UI
t <sub>DVEDQ</sub>	Data Hold After DQS (DDR Read)	All ECP3 Devices	0.64	_	0.64	_	0.64	_	UI
t <sub>DQVBS</sub>	Data Valid Before DQS	All ECP3 Devices	0.25	_	0.25	_	0.25	_	UI
t <sub>DQVAS</sub>	Data Valid After DQS	All ECP3 Devices	0.25	_	0.25	_	0.25	_	UI
f <sub>MAX_DDR3</sub>	DDR3 clock frequency	All ECP3 Devices	300	400	266	333	266	300	MHz



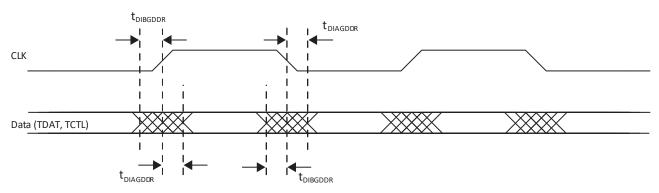
Parameter	Description	Device	_	-8		-7	_	6	Units
			Min.	Max.	Min.	Max.	Min.	Max.	
DDR3 Clock Tim	ing								
t <sub>CH</sub> (avg) <sup>9</sup>	Average High Pulse Width	All ECP3 Devices	0.47	0.53	0.47	0.53	0.47	0.53	UI
t <sub>CL</sub> (avg) <sup>9</sup>	Average Low Pulse Width	All ECP3 Devices	0.47	0.53	0.47	0.53	0.47	0.53	UI
t <sub>JIT</sub> (per, lck) <sup>9</sup>	Output Clock Period Jitter During DLL Locking Period	All ECP3 Devices	-90	90	-90	90	-90	90	ps
t <sub>JIT</sub> (cc, lck) <sup>9</sup>	Output Cycle-to- Cycle Period Jitter During DLL Locking Period	All ECP3 Devices	_	180	_	180	_	180	ps

#### Notes:

- Commercial timing numbers are shown. Industrial numbers are typically slower and can be extracted from the Diamond or ispLEVER software.
- 2. General I/O timing numbers based on LVCMOS 2.5, 12 mA, Fast Slew Rate, Opf load.
- 3. Generic DDR timing numbers based on LVDS I/O.
- 4. DDR timing numbers based on SSTL25. DDR2 timing numbers based on SSTL18.
- 5. DDR3 timing numbers based on SSTL15.
- 6. Uses LVDS I/O standard.
- 7. The current version of software does not support per bank skew numbers; this will be supported in a future release.
- 8. Maximum clock frequencies are tested under best case conditions. System performance may vary upon the user environment.
- 9. Using settings generated by IPexpress.
- 10. These numbers are generated using best case PLL located in the center of the device.
- 11. Uses SSTL25 Class II Differential I/O Standard.
- 12. All numbers are generated with ispLEVER 8.1 software.
- 13. For details on -9 speed grade devices, contact your Lattice Sales Representative.



#### **Transmit Parameters**



#### Receive Parameters

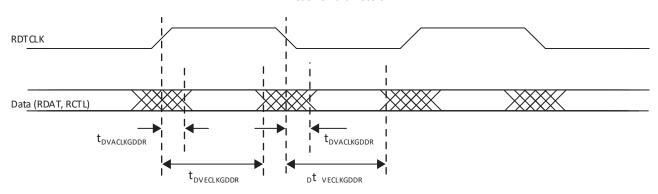


Figure 4.6. Generic DDRX1/DDRX2 (With Clock and Data Edges Aligned)

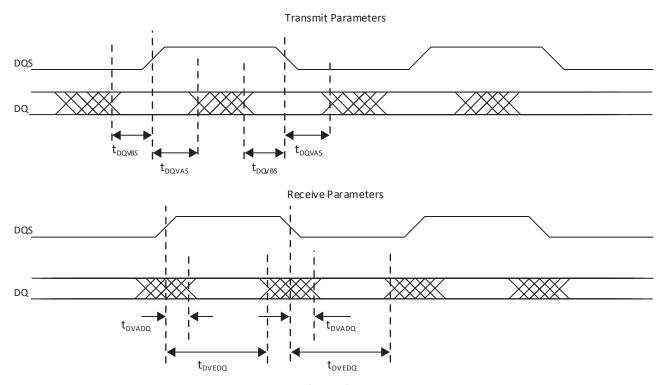


Figure 4.7. DDR/DDR2/DDR3 Parameters

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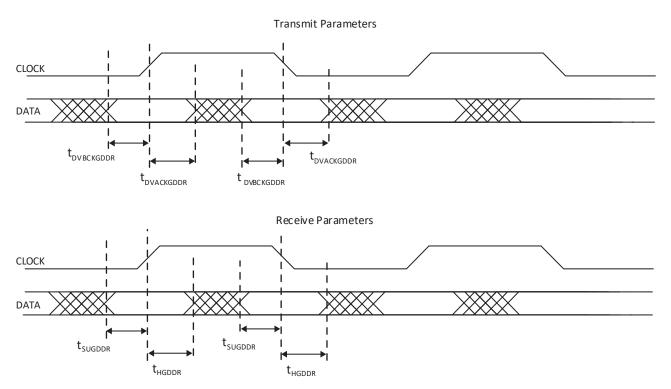


Figure 4.8. Generic DDRX1/DDRX2 (With Clock Center on Data Window)



## 4.17. LatticeECP3 Internal Switching Characteristics 1, 2, 5

**Over Recommended Commercial Operating Conditions** 

Parameter	Description	-	-8	-	-7	-6		Units.
		Min.	Max.	Min.	Max.	Min.	Max.	
PFU/PFF Log	gic Mode Timing	1		1	l .	1	l .	ı
t <sub>LUT4_PFU</sub>	LUT4 delay (A to D inputs to F output)	_	0.147	_	0.163	_	0.179	ns
t <sub>LUT6_PFU</sub>	LUT6 delay (A to D inputs to OFX output)	_	0.281	_	0.335	_	0.379	ns
t <sub>LSR_PFU</sub>	Set/Reset to output of PFU (Asynchronous)	_	0.593	_	0.674	_	0.756	ns
t <sub>LSRREC_PFU</sub>	Asynchronous Set/Reset recovery time for PFU Logic	_	0.298	_	0.345	_	0.391	ns
t <sub>SUM_PFU</sub>	Clock to Mux (M0,M1) Input Setup Time	0.134	_	0.144	_	0.153	_	ns
t <sub>HM_PFU</sub>	Clock to Mux (M0,M1) Input Hold Time	-0.097	_	-0.103	_	-0.109	_	ns
t <sub>SUD_PFU</sub>	Clock to D input setup time	0.061	_	0.068	_	0.075	_	ns
t <sub>HD_PFU</sub>	Clock to D input hold time	0.019	_	0.013	_	0.015	_	ns
t <sub>CK2Q_PFU</sub>	Clock to Q delay, (D-type Register Configuration)	_	0.243	_	0.273	_	0.303	ns
PFU Dual Po	ort Memory Mode Timing	•		•		•	•	
t <sub>CORAM_PFU</sub>	Clock to Output (F Port)	_	0.710	_	0.803	_	0.897	ns
t <sub>SUDATA_PFU</sub>	Data Setup Time	-0.137	_	-0.155	_	-0.174	_	ns
t <sub>HDATA_PFU</sub>	Data Hold Time	0.188	_	0.217	_	0.246	_	ns
t <sub>SUADDR PFU</sub>	Address Setup Time	-0.227	_	-0.257	_	-0.286	_	ns
t <sub>HADDR_PFU</sub>	Address Hold Time	0.240	_	0.275	_	0.310	_	ns
t <sub>SUWREN PFU</sub>	Write/Read Enable Setup Time	-0.055	_	-0.055	_	-0.063	_	ns
t <sub>HWREN PFU</sub>	Write/Read Enable Hold Time	0.059	_	0.059	_	0.071	_	ns
PIC Timing		•		•	<b>'</b>	1		II.
PIO Input/O	Output Buffer Timing							
t <sub>IN PIO</sub>	Input Buffer Delay (LVCMOS25)	_	0.423	_	0.466	_	0.508	ns
t <sub>out PIO</sub>	Output Buffer Delay (LVCMOS25)	_	1.241	_	1.301	_	1.361	ns
IOLOGIC Inp	out/Output Timing	1		1	L		L	I.
t <sub>SUI_PIO</sub>	Input Register Setup Time (Data Before Clock)	0.956	_	1.124	_	1.293	_	ns
t <sub>HI_PIO</sub>	Input Register Hold Time (Data after Clock)	0.225	-	0.184	_	0.240	_	ns
t <sub>COO_PIO</sub>	Output Register Clock to Output Delay4	_	1.09	_	1.16	_	1.23	ns
t <sub>SUCE_PIO</sub>	Input Register Clock Enable Setup Time	0.220	_	0.185	_	0.150	_	ns
t <sub>HCE_PIO</sub>	Input Register Clock Enable Hold Time	-0.085	_	-0.072	_	-0.058	_	ns
t <sub>SULSR_PIO</sub>	Set/Reset Setup Time	0.117	_	0.103	_	0.088	_	ns
t <sub>HLSR PIO</sub>	Set/Reset Hold Time	-0.107	1	-0.094	_	-0.081	_	ns



Parameter	Description	_	8	_	7	-	-6	Units.
		Min.	Max.	Min.	Max.	Min.	Max.	
EBR Timing	g					•		•
t <sub>CO_EBR</sub>	Clock (Read) to output from Address or Data	_	2.78	_	2.89	_	2.99	ns
t <sub>COO_EBR</sub>	Clock (Write) to output from EBR output Register	1	0.31	_	0.32	_	0.33	ns
t <sub>SUDATA_EBR</sub>	Setup Data to EBR Memory	-0.218	_	-0.227	1	-0.237	_	ns
t <sub>HDATA_EBR</sub>	Hold Data to EBR Memory	0.249	_	0.257	_	0.265	_	ns
t <sub>SUADDR_EBR</sub>	Setup Address to EBR Memory	-0.071	_	-0.070	1	-0.068	_	ns
t <sub>HADDR_EBR</sub>	Hold Address to EBR Memory	0.118	_	0.098	1	0.077	_	ns
t <sub>SUWREN_EBR</sub>	Setup Write/Read Enable to EBR Memory	-0.107	_	-0.106	I	-0.106	1	ns
t <sub>HWREN_EBR</sub>	Hold Write/Read Enable to EBR Memory	0.141	_	0.145	I	0.149	1	ns
t <sub>SUCE_EBR</sub>	Clock Enable Setup Time to EBR Output Register	0.087	_	0.096	-	0.104	_	ns
t <sub>HCE_EBR</sub>	Clock Enable Hold Time to EBR Output Register	-0.066	_	-0.080	_	-0.094	_	ns
t <sub>SUBE_EBR</sub>	Byte Enable Set-Up Time to EBR Output Register	-0.071	_	-0.070	_	-0.068	_	ns
t <sub>HBE_EBR</sub>	Byte Enable Hold Time to EBR Output Register	0.118	_	0.098	_	0.077	_	ns
DSP Block Ti	iming <sup>3</sup>					•		•
t <sub>SUI DSP</sub>	Input Register Setup Time	0.32	_	0.36	_	0.39	_	ns
t <sub>HI_DSP</sub>	Input Register Hold Time	-0.17	_	-0.19	_	-0.21	_	ns
t <sub>SUP_DSP</sub>	Pipeline Register Setup Time	2.23	_	2.30	_	2.37	_	ns
t <sub>HP_DSP</sub>	Pipeline Register Hold Time	-1.02	_	-1.09	_	-1.15	_	ns
t <sub>SUO_DSP</sub>	Output Register Setup Time	3.09	_	3.22	_	3.34	_	ns
t <sub>HO_DSP</sub>	Output Register Hold Time	-1.67	_	-1.76	_	-1.84	_	ns
t <sub>COI_DSP</sub>	Input Register Clock to Output Time	_	3.05	_	3.35	_	3.73	ns
t <sub>COP_DSP</sub>	Pipeline Register Clock to Output Time	_	1.30		1.47	_	1.64	ns
t <sub>COO_DSP</sub>	Output Register Clock to Output Time	_	0.58	_	0.60	_	0.62	ns
t <sub>SUOPT_DSP</sub>	Opcode Register Setup Time	0.31	_	0.35	1	0.39	_	ns
t <sub>HOPT_DSP</sub>	Opcode Register Hold Time	-0.20	_	-0.24	_	-0.27	_	ns
t <sub>SUDATA_DSP</sub>	Cascade_data through ALU to Output Register Setup Time	1.69	_	1.94	_	2.14	_	ns
t <sub>HPDATA_DSP</sub>	Cascade_data through ALU to Output Register Hold Time	-0.58	_	-0.80	_	-0.97	_	ns

#### Notes:

- 1. Internal parameters are characterized but not tested on every device.
- 2. Commercial timing numbers are shown. Industrial timing numbers are typically slower and can be extracted from the Diamond or ispLEVER software.
- 3. DSP slice is configured in Multiply Add/Sub 18 x 18 mode.
- 4. The output register is in Flip-flop mode.
- 5. For details on –9 speed grade devices, contact your Lattice Sales Representative.

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FPGA-DS-02074-3.2

## 4.18. Timing Diagrams

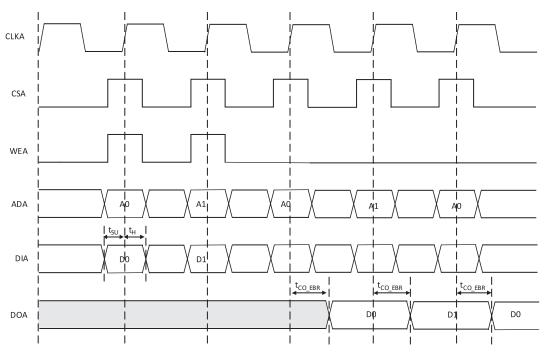


Figure 4.9. Read/Write Mode (Normal)

**Note**: Input data and address are registered at the positive edge of the clock and output data appears after the positive edge of the clock.

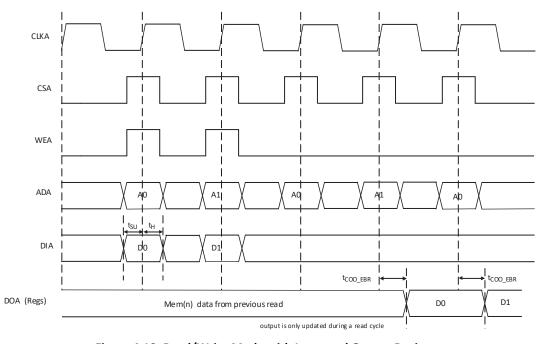


Figure 4.10. Read/Write Mode with Input and Output Registers

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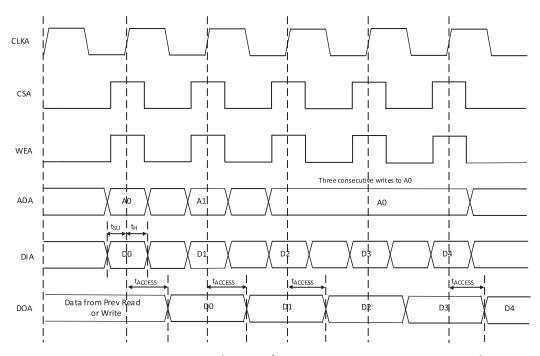


Figure 4.11. Write Through (SP Read/Write on Port A, Input Registers Only)

**Note**: Input data and address are registered at the positive edge of the clock and output data appears after the positive edge of the clock.



# **4.19.** LatticeECP3 Family Timing Adders<sup>1, 2, 3, 4, 5, 7</sup> Over Recommended Commercial Operating Conditions

	ded Commercial Operating Conditions	0	7	•	l lm!+-
Buffer Type	Description	-8	<b>-7</b>	-6	Units
Input Adjusters			1		I
LVDS25E	LVDS, Emulated, V <sub>CCIO</sub> = 2.5 V	0.03	-0.01	-0.03	ns
LVDS25	LVDS, V <sub>CCIO</sub> = 2.5 V	0.03	0.00	-0.04	ns
BLVDS25	BLVDS, Emulated, V <sub>CCIO</sub> = 2.5 V	0.03	0.00	-0.04	ns
MLVDS25	MLVDS, Emulated, V <sub>CCIO</sub> = 2.5 V	0.03	0.00	-0.04	ns
RSDS25	RSDS, $V_{CCIO} = 2.5 \text{ V}$	0.03	-0.01	-0.03	ns
PPLVDS	Point-to-Point LVDS	0.03	-0.01	-0.03	ns
TRLVDS	Transition-Reduced LVDS	0.03	0.00	-0.04	ns
Mini MLVDS	Mini LVDS	0.03	-0.01	-0.03	ns
LVPECL33	LVPECL, Emulated, V <sub>CCIO</sub> = 3.3 V	0.17	0.23	0.28	ns
HSTL18_I	HSTL_18 class I, V <sub>CCIO</sub> = 1.8 V	0.20	0.17	0.13	ns
HSTL18_II	HSTL_18 class II, V <sub>CCIO</sub> = 1.8 V	0.20	0.17	0.13	ns
HSTL18D_I	Differential HSTL 18 class I	0.20	0.17	0.13	ns
HSTL18D_II	Differential HSTL 18 class II	0.20	0.17	0.13	ns
HSTL15_I	HSTL_15 class I, V <sub>CCIO</sub> = 1.5 V	0.10	0.12	0.13	ns
HSTL15D_I	Differential HSTL 15 class I	0.10	0.12	0.13	ns
SSTL33_I	SSTL_3 class I, V <sub>CCIO</sub> = 3.3 V	0.17	0.23	0.28	ns
SSTL33 II	SSTL_3 class II, V <sub>CCIO</sub> = 3.3 V	0.17	0.23	0.28	ns
SSTL33D_I	Differential SSTL_3 class I	0.17	0.23	0.28	ns
SSTL33D_II	Differential SSTL_3 class II	0.17	0.23	0.28	ns
SSTL25_I	SSTL_2 class I, V <sub>CCIO</sub> = 2.5 V	0.12	0.14	0.16	ns
SSTL25_II	SSTL_2 class II, V <sub>CCIO</sub> = 2.5 V	0.12	0.14	0.16	ns
SSTL25D_I	Differential SSTL_2 class I	0.12	0.14	0.16	ns
SSTL25D II	Differential SSTL_2 class II	0.12	0.14	0.16	ns
SSTL18_I	SSTL_18 class I, V <sub>CCIO</sub> = 1.8 V	0.08	0.06	0.04	ns
SSTL18_II	SSTL_18 class II, V <sub>CCIO</sub> = 1.8 V	0.08	0.06	0.04	ns
SSTL18D I	Differential SSTL_18 class I	0.08	0.06	0.04	ns
SSTL18D_I	Differential SSTL_18 class II	0.08	0.06	0.04	ns
SSTL15D_II	SSTL_15, V <sub>CCIO</sub> = 1.5 V	0.087	0.059	0.032	ns
	Differential SSTL 15	0.087	1	0.032	
SSTL15D LVTTL33	_	0.087	0.059	0.032	ns
LVCMOS33	LVTTL, V <sub>CCIO</sub> = 3.3 V LVCMOS, V <sub>CCIO</sub> = 3.3 V	0.07	0.07	0.07	ns
LVCMOS25		0.07	0.07		ns
	LVCMOS, V <sub>CCIO</sub> = 2.5 V			0.00	ns
LVCMOS18	LVCMOS, V <sub>CCIO</sub> = 1.8 V	-0.13	-0.13	-0.13	ns
LVCMOS15	LVCMOS, V <sub>CCIO</sub> = 1.5 V	-0.07	-0.07	-0.07	ns
LVCMOS12	LVCMOS, V <sub>CCIO</sub> = 1.2 V	-0.20	-0.19	-0.19	ns
PCI33	PCI, V <sub>CCIO</sub> = 3.3 V	0.07	0.07	0.07	ns
Output Adjusters					
LVDS25E	LVDS, Emulated, V <sub>CCIO</sub> = 2.5 V	1.02	1.14	1.26	ns
LVDS25	LVDS, V <sub>CCIO</sub> = 2.5 V	-0.11	-0.07	-0.03	ns
BLVDS25	BLVDS, Emulated, V <sub>CCIO</sub> = 2.5 V	1.01	1.13	1.25	ns
MLVDS25	MLVDS, Emulated, V <sub>CCIO</sub> = 2.5 V	1.01	1.13	1.25	ns
RSDS25	RSDS, $V_{CCIO} = 2.5 \text{ V}$	-0.07	-0.04	-0.01	ns
PPLVDS	Point-to-Point LVDS, True LVDS, VCCIO = 2.5 V or 3.3 V	-0.22	-0.19	-0.16	ns



Buffer Type	Description	-8	-7	-6	Units
LVPECL33	LVPECL, Emulated, V <sub>CCIO</sub> = 3.3 V	0.67	0.76	0.86	ns
HSTL18_I	HSTL_18 class I 8mA drive, V <sub>CCIO</sub> = 1.8 V	1.20	1.34	1.47	ns
HSTL18_II	HSTL_18 class II, V <sub>CCIO</sub> = 1.8 V	0.89	1.00	1.11	ns
HSTL18D_I	Differential HSTL 18 class I 8 mA drive	1.20	1.34	1.47	ns
HSTL18D_II	Differential HSTL 18 class II	0.89	1.00	1.11	ns
HSTL15_I	HSTL_15 class I 4 mA drive, V <sub>CCIO</sub> = 1.5 V	1.67	1.83	1.99	ns
HSTL15D_I	Differential HSTL 15 class I 4 mA drive	1.67	1.83	1.99	ns
SSTL33_I	SSTL_3 class I, V <sub>CCIO</sub> = 3.3 V	1.12	1.17	1.21	ns
SSTL33_II	SSTL_3 class II, V <sub>CCIO</sub> = 3.3 V	1.08	1.12	1.15	ns
SSTL33D_I	Differential SSTL_3 class I	1.12	1.17	1.21	ns
SSTL33D_II	Differential SSTL_3 class II	1.08	1.12	1.15	ns
SSTL25_I	SSTL_2 class I 8 mA drive, V <sub>CCIO</sub> = 2.5 V	1.06	1.19	1.31	ns
SSTL25_II	SSTL_2 class II 16 mA drive, V <sub>CCIO</sub> = 2.5 V	1.04	1.17	1.31	ns
SSTL25D_I	Differential SSTL_2 class I 8 mA drive	1.06	1.19	1.31	ns
SSTL25D_II	Differential SSTL_2 class II 16 mA drive	1.04	1.17	1.31	ns
SSTL18_I	SSTL_1.8 class I, V <sub>CCIO</sub> = 1.8 V	0.70	0.84	0.97	ns
SSTL18_II	SSTL_1.8 class II 8 mA drive, V <sub>CCIO</sub> = 1.8 V	0.70	0.84	0.97	ns
SSTL18D_I	Differential SSTL_1.8 class I	0.70	0.84	0.97	ns
SSTL18D_II	Differential SSTL_1.8 class II 8 mA drive	0.70	0.84	0.97	ns
SSTL15	SSTL_1.5, V <sub>CCIO</sub> = 1.5 V	1.22	1.35	1.48	ns
SSTL15D	Differential SSTL_15	1.22	1.35	1.48	ns
LVTTL33_4mA	LVTTL 4 mA drive, V <sub>CCIO</sub> = 3.3V	0.25	0.24	0.23	ns
LVTTL33 8mA	LVTTL 8 mA drive, V <sub>CCIO</sub> = 3.3V	-0.06	-0.06	-0.07	ns
LVTTL33 12mA	LVTTL 12 mA drive, V <sub>CCIO</sub> = 3.3V	-0.01	-0.02	-0.02	ns
LVTTL33_16mA	LVTTL 16 mA drive, V <sub>CCIO</sub> = 3.3V	-0.07	-0.07	-0.08	ns
LVTTL33_20mA	LVTTL 20 mA drive, V <sub>CCIO</sub> = 3.3V	-0.12	-0.13	-0.14	ns
LVCMOS33_4mA	LVCMOS 3.3 4 mA drive, fast slew rate	0.25	0.24	0.23	ns
LVCMOS33_8mA	LVCMOS 3.3 8 mA drive, fast slew rate	-0.06	-0.06	-0.07	ns
LVCMOS33_12mA	LVCMOS 3.3 12 mA drive, fast slew rate	-0.01	-0.02	-0.02	ns
LVCMOS33_16mA	LVCMOS 3.3 16 mA drive, fast slew rate	-0.07	-0.07	-0.08	ns
LVCMOS33_20mA	LVCMOS 3.3 20 mA drive, fast slew rate	-0.12	-0.13	-0.14	ns
LVCMOS25_4mA	LVCMOS 2.5 4 mA drive, fast slew rate	0.12	0.10	0.09	ns
LVCMOS25_8mA	LVCMOS 2.5 8 mA drive, fast slew rate	-0.05	-0.06	-0.07	ns
LVCMOS25_12mA	LVCMOS 2.5 12 mA drive, fast slew rate	0.00	0.00	0.00	ns
LVCMOS25_16mA	LVCMOS 2.5 16 mA drive, fast slew rate	-0.12	-0.13	-0.14	ns
LVCMOS25_20mA	LVCMOS 2.5 20 mA drive, fast slew rate	-0.12	-0.13	-0.14	ns
LVCMOS18_4mA	LVCMOS 1.8 4 mA drive, fast slew rate	0.11	0.12	0.14	ns
LVCMOS18_8mA	LVCMOS 1.8 8 mA drive, fast slew rate	0.11	0.12	0.14	ns
LVCMOS18_12mA	LVCMOS 1.8 12 mA drive, fast slew rate	-0.04	-0.03	-0.03	ns
LVCMOS18_16mA	LVCMOS 1.8 16 mA drive, fast slew rate	-0.04	-0.03	-0.03	ns
LVCMOS15_4mA	LVCMOS 1.5 4 mA drive, fast slew rate	0.21	0.25	0.29	ns
LVCMOS15_8mA	LVCMOS 1.5 8 mA drive, fast slew rate	0.05	0.07	0.09	ns
LVCMOS12_2mA	LVCMOS 1.2 2 mA drive, fast slew rate	0.43	0.51	0.59	ns
LVCMOS12_6mA	LVCMOS 1.2 6 mA drive, fast slew rate	0.23	0.28	0.33	ns
LVCMOS33_4mA	LVCMOS 3.3 4 mA drive, slow slew rate	1.44	1.58	1.72	ns
LVCMOS33_8mA	LVCMOS 3.3 8 mA drive, slow slew rate	0.98	1.10	1.22	ns



Buffer Type	Description	-8	-7	-6	Units
LVCMOS33_12mA	LVCMOS 3.3 12 mA drive, slow slew rate	0.67	0.77	0.86	ns
LVCMOS33_16mA	LVCMOS 3.3 16 mA drive, slow slew rate	0.97	1.09	1.21	ns
LVCMOS33_20mA	LVCMOS 3.3 20 mA drive, slow slew rate	0.67	0.76	0.85	ns
LVCMOS25_4mA	LVCMOS 2.5 4 mA drive, slow slew rate	1.48	1.63	1.78	ns
LVCMOS25_8mA	LVCMOS 2.5 8 mA drive, slow slew rate	1.02	1.14	1.27	ns
LVCMOS25_12mA	LVCMOS 2.5 12 mA drive, slow slew rate	0.74	0.84	0.94	ns
LVCMOS25_16mA	LVCMOS 2.5 16 mA drive, slow slew rate	1.02	1.14	1.26	ns
LVCMOS25_20mA	LVCMOS 2.5 20 mA drive, slow slew rate	0.74	0.83	0.93	ns
LVCMOS18_4mA	LVCMOS 1.8 4 mA drive, slow slew rate	1.60	1.77	1.93	ns
LVCMOS18_8mA	LVCMOS 1.8 8 mA drive, slow slew rate	1.11	1.25	1.38	ns
LVCMOS18_12mA	LVCMOS 1.8 12 mA drive, slow slew rate	0.87	0.98	1.09	ns
LVCMOS18_16mA	LVCMOS 1.8 16 mA drive, slow slew rate	0.86	0.97	1.07	ns
LVCMOS15_4mA	LVCMOS 1.5 4 mA drive, slow slew rate	1.71	1.89	2.08	ns
LVCMOS15_8mA	LVCMOS 1.5 8 mA drive, slow slew rate	1.20	1.34	1.48	ns
LVCMOS12_2mA	LVCMOS 1.2 2 mA drive, slow slew rate	1.37	1.56	1.74	ns
LVCMOS12_6mA	LVCMOS 1.2 6 mA drive, slow slew rate	1.11	1.27	1.43	ns
PCI33	PCI, VCCIO = 3.3 V	-0.12	-0.13	-0.14	ns

#### Notes:

- 1. Timing adders are characterized but not tested on every device.
- 2. LVCMOS timing measured with the load specified in Switching Test Condition table.
- 3. All other standards tested according to the appropriate specifications.
- 4. Not all I/O standards and drive strengths are supported for all banks. See the Architecture section of this data sheet for details.
- Commercial timing numbers are shown. Industrial numbers are typically slower and can be extracted from the Diamond or ispLEVER software.
- 6. This data does not apply to the LatticeECP3-17EA device.
- 7. For details on –9 speed grade devices, contact your Lattice Sales Representative.



## 4.20. LatticeECP3 Maximum I/O Buffer Speed 1, 2, 3, 4, 5, 6

**Over Recommended Operating Conditions** 

Buffer	Description	Max.	Units
Maximum Input Frequency			
LVDS25	LVDS, V <sub>CCIO</sub> = 2.5 V	400	MHz
MLVDS25	MLVDS, Emulated, V <sub>CCIO</sub> = 2.5 V	400	MHz
BLVDS25	BLVDS, Emulated, V <sub>CCIO</sub> = 2.5 V	400	MHz
PPLVDS	Point-to-Point LVDS	400	MHz
TRLVDS	Transition-Reduced LVDS	612	MHz
Mini LVDS	Mini LVDS	400	MHz
LVPECL33	LVPECL, Emulated, V <sub>CCIO</sub> = 3.3 V	400	MHz
HSTL18 (all supported classes)	HSTL_18 class I, II, V <sub>CCIO</sub> = 1.8 V	400	MHz
HSTL15	HSTL_15 class I, V <sub>CCIO</sub> = 1.5 V	400	MHz
SSTL33 (all supported classes)	SSTL_3 class I, II, V <sub>CCIO</sub> = 3.3 V	400	MHz
SSTL25 (all supported classes)	SSTL_2 class I, II, V <sub>CCIO</sub> = 2.5 V	400	MHz
SSTL18 (all supported classes)	SSTL_18 class I, II, V <sub>CCIO</sub> = 1.8 V	400	MHz
LVTTL33	LVTTL, V <sub>CCIO</sub> = 3.3 V	166	MHz
LVCMOS33	LVCMOS, V <sub>CCIO</sub> = 3.3 V	166	MHz
LVCMOS25	LVCMOS, V <sub>CCIO</sub> = 2.5 V	166	MHz
LVCMOS18	LVCMOS, V <sub>CCIO</sub> = 1.8 V	166	MHz
LVCMOS15	LVCMOS 1.5, V <sub>CCIO</sub> = 1.5 V	166	MHz
LVCMOS12	LVCMOS 1.2, V <sub>CCIO</sub> = 1.2 V	166	MHz
PCI33	PCI, VCCIO = 3.3 V	66	MHz
Maximum Output Frequency		·	
LVDS25E	LVDS, Emulated, V <sub>CCIO</sub> = 2.5 V	300	MHz
LVDS25	LVDS, V <sub>CCIO</sub> = 2.5 V	612	MHz
MLVDS25	MLVDS, Emulated, V <sub>CCIO</sub> = 2.5 V	300	MHz
RSDS25	RSDS, Emulated, V <sub>CCIO</sub> = 2.5 V	612	MHz
BLVDS25	BLVDS, Emulated, V <sub>CCIO</sub> = 2.5 V	300	MHz
PPLVDS	Point-to-point LVDS	612	MHz
LVPECL33	LVPECL, Emulated, V <sub>CCIO</sub> = 3.3 V	612	MHz
Mini-LVDS	Mini LVDS	612	MHz
HSTL18 (all supported classes)	HSTL_18 class I, II, V <sub>CCIO</sub> = 1.8 V	200	MHz
HSTL15 (all supported classes)	HSTL_15 class I, V <sub>CCIO</sub> = 1.5 V	200	MHz
SSTL33 (all supported classes)	SSTL_3 class I, II, V <sub>CCIO</sub> = 3.3 V	233	MHz
SSTL25 (all supported classes)	SSTL_2 class I, II, V <sub>CCIO</sub> = 2.5 V	233	MHz
SSTL18 (all supported classes)	SSTL_18 class I, II, V <sub>CCIO</sub> = 1.8 V	266	MHz
LVTTL33	LVTTL, V <sub>CCIO</sub> = 3.3 V	166	MHz
LVCMOS33 (For all drives)	LVCMOS, 3.3 V	166	MHz
LVCMOS25 (For all drives)	LVCMOS, 2.5 V	166	MHz
LVCMOS18 (For all drives)	LVCMOS, 1.8 V	166	MHz
LVCMOS15 (For all drives)	LVCMOS, 1.5 V	166	MHz
LVCMOS12 (For all drives except 2 mA)	LVCMOS, V <sub>CCIO</sub> = 1.2 V	166	MHz
LVCMOS12 (2 mA drive)	LVCMOS, V <sub>CCIO</sub> = 1.2 V	100	MHz
PCI33	PCI, V <sub>CCIO</sub> = 3.3 V	66	MHz

#### Notes:

- 1. These maximum speeds are characterized but not tested on every device.
- 2. Maximum I/O speed for differential output standards emulated with resistors depends on the layout.
- 3. LVCMOS timing is measured with the load specified in the Switching Test Conditions table of this document.



- 4. All speeds are measured at fast slew.
- 5. Actual system operation may vary depending on user logic implementation.
- 6. Maximum data rate equals 2 times the clock rate when utilizing DDR.

## 4.21. sysCLOCK PLL Timing

**Over Recommended Operating Conditions** 

Parameter	Descriptions	Conditions	Clock	Min.	Тур.	Max.	Units
$f_{\text{IN}}$	Input clock frequency (CLKI, CLKFB)	_	Edge clock	2	_	500	MHz
			Primary clock <sup>4</sup>	2	_	420	MHz
f <sub>OUT</sub>	Output clock frequency (CLKOP, CLKOS)	_	Edge clock	4	_	500	MHz
			Primary clock <sup>4</sup>	4	_	420	MHz
f <sub>OUT1</sub>	K-Divider output frequency	CLKOK	_	0.03125	_	250	MHz
f <sub>OUT2</sub>	K2-Divider output frequency	CLKOK2	_	0.667	_	166	MHz
f <sub>VCO</sub>	PLL VCO frequency	_	_	500	_	1000	MHz
f <sub>PFD</sub> <sup>3</sup>	Phase detector input frequency	_	Edge clock	2	_	500	MHz
equency			Primary clock <sup>4</sup>	2	_	420	MHz
AC Characte	eristics	<u> </u>					
t <sub>PA</sub>	Programmable delay unit	_	_	65	130	260	ps
Output clock duty cycle (CLKOS, at 50% setting)		_	Edge clock	45	50	55	%
		f <sub>OUT</sub> ≤ 250 MHz	Primary clock	45	50	55	%
		f <sub>OUT</sub> > 250 MHz	Primary clock	30	50	70	%
t <sub>CPA</sub>	Coarse phase shift error (CLKOS, at all settings)	_	_	-5	0	+5	% of period
t <sub>OPW</sub>	Output clock pulse width high or low (CLKOS)	_	_	1.8	_	_	ns
t <sub>OPJIT</sub> 1	Output clock period jitter	f <sub>OUT</sub> ≥ 420 MHz	_	_	_	200	ps
		420 MHz > f <sub>OUT</sub> ≥ 100 MHz	_	_	_	250	ps
		f <sub>OUT</sub> < 100 MHz	_	_	_	0.025	UIPP
t <sub>SK</sub>	Input clock to output clock skew when N/M = integer	_	_	_	_	500	ps
t <sub>LOCK</sub> <sup>2</sup>	Lock time	2 to 25 MHz	_	_	_	200	us
		25 to 500 MHz	_	_	_	50	us
t <sub>UNLOCK</sub>	Reset to PLL unlock time to ensure fast reset	_	_	_	_	50	ns
t <sub>HI</sub>	Input clock high time	90% to 90%	_	0.5	_	_	ns
t <sub>LO</sub>	Input clock low time	10% to 10%	_	0.5	_	_	ns
t <sub>IPJIT</sub>	Input clock period jitter	_	_	_	_	400	ps
t <sub>RST</sub>	Reset signal pulse width high, RSTK	_	_	10	_	_	ns
	Reset signal pulse width high, RST	_	_	500	_	_	ns



#### Notes:

- 1. Jitter sample is taken over 10,000 samples of the primary PLL output with clean reference clock with no additional I/O toggling.
- 2. Output clock is valid after  $t_{LOCK}$  for PLL reset and dynamic delay adjustment.
- 3. Period jitter and cycle-to-cycle jitter numbers are guaranteed for  $f_{PFD} > 4$  MHz. For  $f_{PFD} < 4$  MHz, the jitter numbers may not be met in certain conditions. Contact the factory for  $f_{PFD} < 4$  MHz.
- 4. When using internal feedback, maximum can be up to 500 MHz.

## 4.22. DLL Timing

**Over Recommended Operating Conditions** 

Parameter	Description	Condition	Min.	Тур.	Max.	Units
f <sub>REF</sub>	Input reference clock frequency (on-chip or off- chip)	_	133	_	500	MHz
$f_{FB}$	Feedback clock frequency (on-chip or off-chip)	_	133	_	500	MHz
f <sub>CLKOP</sub> <sup>1</sup>	Output clock frequency, CLKOP	_	133	_	500	MHz
f <sub>CLKOS</sub> <sup>2</sup>	Output clock frequency, CLKOS	_	33.3	_	500	MHz
t <sub>PJIT</sub>	Output clock period jitter (clean input)	_	_	_	200	ps p-p
t <sub>DUTY</sub>	Output clock duty cycle (at 50% levels, 50% duty	Edge Clock	40	_	60	%
	cycle input clock, 50% duty cycle circuit turned off, time reference delay mode)	Primary Clock	30	_	70	%
t <sub>DUTYTRD</sub>	Output clock duty cycle (at 50% levels, arbitrary	Primary Clock < 250 MHz	45	_	55	%
	duty cycle input clock, 50% duty cycle circuit	Primary Clock ≥ 250 MHz	30	_	70	%
	enabled, time reference delay mode)	Edge Clock	45	_	55	%
t <sub>DUTYCIR</sub>	Output clock duty cycle (at 50% levels, arbitrary	Primary Clock < 250 MHz	40	_	60	%
	duty cycle input clock, 50% duty cycle circuit	Primary Clock ≥ 250 MHz	30	_	70	%
	enabled, clock injection removal mode) with DLL cascading	Edge Clock	45	_	55	%
t <sub>SKEW</sub> <sup>3</sup>	Output clock to clock skew between two outputs with the same phase setting	_	_	_	100	ps
t <sub>PHASE</sub>	Phase error measured at device pads between off-chip reference clock and feedback clocks	_	_	_	+/- 400	ps
t <sub>PWH</sub>	Input clock minimum pulse width high (at 80% level)	_	550	_	_	ps
t <sub>PWL</sub>	Input clock minimum pulse width low (at 20% level)	_	550	_	_	ps
t <sub>INSTB</sub>	Input clock period jitter	_	_	_	500	ps
t <sub>LOCK</sub>	DLL lock time	_	8	_	8200	cycles
t <sub>RSWD</sub>	Digital reset minimum pulse width (at 80% level)	_	3	_	_	ns
t <sub>DEL</sub>	Delay step size	_	27	45	70	ps
t <sub>RANGE1</sub>	Max. delay setting for single delay block (64 taps)	_	1.9	3.1	4.4	ns
t <sub>RANGE4</sub>	Max. delay setting for four chained delay blocks	_	7.6	12.4	17.6	ns

#### Notes:

- 1. CLKOP runs at the same frequency as the input clock.
- 2. CLKOS minimum frequency is obtained with divide by 4.
- 3. This is intended to be a *path-matching* design guideline and is not a measurable specification.



## 4.23. SERDES High-Speed Data Transmitter<sup>1</sup>

#### **Table 4.6. Serial Output Timing and Levels**

Symbol	Description	Frequency	Min.	Тур.	Max.	Units
V <sub>TX-DIFF-P-P-1.44</sub>	Differential swing (1.44 V setting) <sup>1, 2</sup>	0.15 to 3.125 Gbps	1150	1440	1730	mV, p-p
V <sub>TX-DIFF-P-P-1.35</sub>	Differential swing (1.35 V setting) <sup>1, 2</sup>	0.15 to 3.125 Gbps	1080	1350	1620	mV, p-p
V <sub>TX-DIFF-P-P-1.26</sub>	Differential swing (1.26 V setting) <sup>1, 2</sup>	0.15 to 3.125 Gbps	1000	1260	1510	mV, p-p
V <sub>TX-DIFF-P-P-1.13</sub>	Differential swing (1.13 V setting) <sup>1, 2</sup>	0.15 to 3.125 Gbps	840	1130	1420	mV, p-p
V <sub>TX-DIFF-P-P-1.04</sub>	Differential swing (1.04 V setting) <sup>1, 2</sup>	0.15 to 3.125 Gbps	780	1040	1300	mV, p-p
V <sub>TX-DIFF-P-P-0.92</sub>	Differential swing (0.92 V setting) <sup>1, 2</sup>	0.15 to 3.125 Gbps	690	920	1150	mV, p-p
V <sub>TX-DIFF-P-P-0.87</sub>	Differential swing (0.87 V setting) <sup>1, 2</sup>	0.15 to 3.125 Gbps	650	870	1090	mV, p-p
V <sub>TX-DIFF-P-P-0.78</sub>	Differential swing (0.78 V setting) <sup>1, 2</sup>	0.15 to 3.125 Gbps	585	780	975	mV, p-p
V <sub>TX-DIFF-P-P-0.64</sub>	Differential swing (0.64 V setting) <sup>1, 2</sup>	0.15 to 3.125 Gbps	480	640	800	mV, p-p
V <sub>OCM</sub>	Output common mode voltage	_	V <sub>CCOB</sub> -0.75	V <sub>CCOB</sub> -0.60	V <sub>CCOB</sub> -0.45	V
T <sub>TX-R</sub>	Rise time (20% to 80%)	_	145	185	265	ps
T <sub>TX-F</sub>	Fall time (80% to 20%)	_	145	185	265	ps
Z <sub>TX-OI-SE</sub>	Output Impedance 50/75/HiZ Ohms (single ended)	_	-20%	50/75/ Hi Z	+20%	Ω
R <sub>LTX-RL</sub>	Return loss (with package)	_	10			dB
T <sub>TX-INTRASKEW</sub>	Lane-to-lane TX skew within a SERDES quad block (intra-quad)	_	_	_	200	ps
T <sub>TX-INTERSKEW</sub> <sup>3</sup>	Lane-to-lane skew between SERDES quad blocks (inter-quad)	_	_	_	1UI +200	ps

#### Notes:

- 1. All measurements are with 50 Ohm impedance.
- 2. See LatticeECP3 SERDES/PCS Usage Guide (FPGA-TN-02190) for actual binary settings and the min-max range.
- 3. Inter-quad skew is between all SERDES channels on the device and requires the use of a low skew internal reference clock.



**Table 4.7. Channel Output Jitter** 

Description	Frequency	Min.	Тур.	Max.	Units
Deterministic	3.125 Gbps	_	_	0.17	UI, p-p
Random	3.125 Gbps	_	_	0.25	UI, p-p
Total	3.125 Gbps	_	_	0.35	UI, p-p
Deterministic	2.5 Gbps	_	_	0.17	UI, p-p
Random	2.5 Gbps	_	_	0.20	UI, p-p
Total	2.5 Gbps	_	_	0.35	UI, p-p
Deterministic	1.25 Gbps	_	_	0.10	UI, p-p
Random	1.25 Gbps	_	_	0.22	UI, p-p
Total	1.25 Gbps	_	_	0.24	UI, p-p
Deterministic	622 Mbps	_	_	0.10	UI, p-p
Random	622 Mbps	_	_	0.20	UI, p-p
Total	622 Mbps	_	_	0.24	UI, p-p
Deterministic	250 Mbps	_	_	0.10	UI, p-p
Random	250 Mbps	_	_	0.18	UI, p-p
Total	250 Mbps	_	_	0.24	UI, p-p
Deterministic	150 Mbps	_	_	0.10	UI, p-p
Random	150 Mbps	_	_	0.18	UI, p-p
Total	150 Mbps	_	_	0.24	UI, p-p

**Note**: Values are measured with PRBS 2<sup>7</sup>-1, all channels operating, FPGA logic active, I/O around SERDES pins quiet, reference clock @ 10X mode.



## 4.24. SERDES/PCS Block Latency

Table 4.8 describes the latency of each functional block in the transmitter and receiver. Latency is given in parallel clock cycles. Figure 4.12 shows the location of each block.

Table 4.8. SERDES/PCS Latency Breakdown

Item	Description	Min.	Avg.	Max.	Fixed	Bypass	Units
Transm	it Data Latency¹						
T1	FPGA Bridge - Gearing disabled with different clocks	1	3	5	_	1	word clk
	FPGA Bridge - Gearing disabled with same clocks	_	_	_	3	1	word clk
	FPGA Bridge - Gearing enabled	1	3	5	_	_	word clk
T2	8b10b Encoder	_	_	_	2	1	word clk
T3	SERDES Bridge transmit	_	_	_	2	1	word clk
T4	Serializer: 8-bit mode	_	_	_	15 + Δ1	_	UI + ps
	Serializer: 10-bit mode	_	_	_	18 + Δ1	_	UI + ps
T5	Pre-emphasis ON	_	_	_	1 + Δ2	_	UI + ps
	Pre-emphasis OFF	_	_	_	0 + Δ3	_	UI + ps
Receive	e Data Latency <sup>2</sup>						
R1	Equalization ON	_	_	_	Δ1	_	UI + ps
	Equalization OFF	_	_	_	Δ2	_	UI + ps
R2	Deserializer: 8-bit mode	_	_	_	10 + Δ3	_	UI + ps
	Deserializer: 10-bit mode	_	_	_	12 + Δ3	_	UI + ps
R3	SERDES Bridge receive	_	_	_	2	_	word clk
R4	Word alignment	3.1	_	4	_	_	word clk
R5	8b10b decoder	_	_	_	1	_	word clk
R6	Clock Tolerance Compensation	7	15	23	1	1	word clk
R7	FPGA Bridge - Gearing disabled with different clocks	1	3	5	_	1	word clk
	FPGA Bridge - Gearing disabled with same clocks	_	_	_	3	1	word clk
	FPGA Bridge - Gearing enabled	1	3	5	_	_	word clk

#### Notes:

- 1.  $\Delta 1 = -245 \text{ ps}, \Delta 2 = +88 \text{ ps}, \Delta 3 = +112 \text{ ps}.$
- 2.  $\Delta 1 = +118 \text{ ps}, \Delta 2 = +132 \text{ ps}, \Delta 3 = +700 \text{ ps}.$

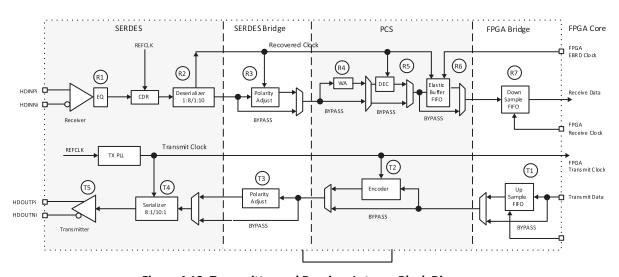


Figure 4.12. Transmitter and Receiver Latency Block Diagram

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## 4.25. SERDES High Speed Data Receiver

**Table 4.9. Serial Input Data Specifications** 

Symbol	Description		Min.	Тур.	Max.	Units
RX-CID <sub>S</sub>	Stream of nontransitions <sup>1</sup>	3.125	_	_	136	Bits
	(CID = Consecutive Identical Digits) @ 10 <sup>-12</sup> BER	G				
		2.5 G	_	_	144	
		1.485	_	_	160	
		G				
		622	_	_	204	
		M				
		270	_	_	228	
		M				
		150	_	_	296	
		М				
V <sub>RX-DIFF-S</sub>	Differential input sensitivity		150	_	1760	mV, p-p
$V_{RX-IN}$	Input levels		0	_	V <sub>CCA</sub> +0.5 <sup>4</sup>	V
$V_{RX-CM-DC}$	Input common mode range (DC coupled)		0.6	_	V <sub>CCA</sub>	V
V <sub>RX-CM-AC</sub>	Input common mode range (AC coupled) <sup>3</sup>		0.1	_	V <sub>CCA</sub> +0.2	V
T <sub>RX-RELOCK</sub>	SCDR re-lock time <sup>2</sup>	•	_	1000	_	Bits
Z <sub>RX-TERM</sub>	Input termination 50/75 Ω/High Z		-20%	50/75/HiZ	+20%	Ω
RL <sub>RX-RL</sub>	Return loss (without package)		10	_	_	dB

#### Notes:

- 1. This is the number of bits allowed without a transition on the incoming data stream when using DC coupling.
- 2. This is the typical number of bit times to re-lock to a new phase or frequency within +/- 300 ppm, assuming 8b10b encoded data.
- 3. AC coupling is used to interface to LVPECL and LVDS. LVDS interfaces are found in laser drivers and Fibre Channel equipment. LVDS interfaces are generally found in 622 Mbps SERDES devices.
- 4. Up to 1.76 V.

#### 4.25.1. Input Data Jitter Tolerance

A receiver's ability to tolerate incoming signal jitter is very dependent on jitter type. High speed serial interface standards have recognized the dependency on jitter type and have specifications to indicate tolerance levels for different jitter types as they relate to specific protocols. Sinusoidal jitter is considered to be a worst case jitter type.

**Table 4.10. Receiver Total Jitter Tolerance Specification** 

Description	Frequency	Condition	Min.	Тур.	Max.	Units
Deterministic	3.125 Gbps	600 mV differential eye	_	_	0.47	UI, p-p
Random		600 mV differential eye	_	_	0.18	UI, p-p
Total		600 mV differential eye	_	_	0.65	UI, p-p
Deterministic	2.5 Gbps	600 mV differential eye	_	_	0.47	UI, p-p
Random		600 mV differential eye	_	_	0.18	UI, p-p
Total		600 mV differential eye	_	_	0.65	UI, p-p
Deterministic	1.25 Gbps	600 mV differential eye	_	_	0.47	UI, p-p
Random		600 mV differential eye	_	_	0.18	UI, p-p
Total		600 mV differential eye	_	_	0.65	UI, p-p
Deterministic	622 Mbps	600 mV differential eye	_	_	0.47	UI, p-p
Random		600 mV differential eye	_	_	0.18	UI, p-p
Total		600 mV differential eye	_	_	0.65	UI, p-p

**Note:** Values are measured with CJPAT, all channels operating, FPGA Logic active, I/O around SERDES pins quiet, voltages are nominal, room temperature.



**Table 4.11. Periodic Receiver Jitter Tolerance Specification** 

Description	Frequency	Condition	Min.	Тур.	Max.	Units
Periodic	2.97 Gbps	600 mV differential eye	_	_	0.24	UI, p-p
Periodic	2.5 Gbps	600 mV differential eye	_	_	0.22	UI, p-p
Periodic	1.485 Gbps	600 mV differential eye	-	_	0.24	UI, p-p
Periodic	622 Mbps	600 mV differential eye	_	_	0.15	UI, p-p
Periodic	150 Mbps	600 mV differential eye	_	_	0.5	UI, p-p

**Note**: Values are measured with PRBS 27–1, all channels operating, FPGA Logic active, I/O around SERDES pins quiet, voltages are nominal, room temperature.

#### 4.25.2. SERDES External Reference Clock

The external reference clock selection and its interface are a critical part of system applications for this product. Table 4.12 specifies reference clock requirements, over the full range of operating conditions.

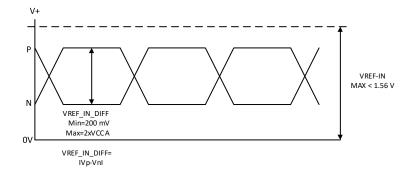
Table 4.12. External Reference Clock Specification (refclkp/refclkn)

Symbol	Description	Min.	Тур.	Max.	Units
F <sub>REF</sub>	Frequency range	15	_	320	MHz
F <sub>REF-PPM</sub>	Frequency tolerance <sup>1</sup>	-1000	_	1000	ppm
V <sub>REF-IN-SE</sub>	Input swing, single-ended clock <sup>2</sup>	200	_	V <sub>CCA</sub>	mV, p-p
V <sub>REF-IN-DIFF</sub>	Input swing, differential clock	200	_	2*V <sub>CCA</sub>	mV, p-p differential
V <sub>REF-IN</sub>	Input levels	0	_	V <sub>CCA</sub> + 0.3	V
D <sub>REF</sub>	Duty cycle <sup>3</sup>	40	_	60	%
T <sub>REF-R</sub>	Rise time (20% to 80%)	200	500	1000	ps
T <sub>REF-F</sub>	Fall time (80% to 20%)	200	500	1000	ps
Z <sub>REF-IN-TERM-DIFF</sub>	Differential input termination	-20%	100/2K	+20%	Ω
C <sub>REF-IN-CAP</sub>	Input capacitance	_	_	7	pF

#### Notes:

- Depending on the application, the PLL\_LOL\_SET and CDR\_LOL\_SET control registers may be adjusted for other tolerance values as described in LatticeECP3 SERDES/PCS Usage Guide (FPGA-TN-02190).
- 2. The signal swing for a single-ended input clock must be as large as the p-p differential swing of a differential input clock to get the same gain at the input receiver. Lower swings for the clock may be possible, but will tend to increase jitter.
- 3. Measured at 50% amplitude.





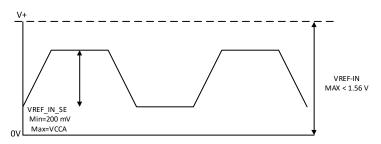


Figure 4.13. SERDES External Reference Clock Waveforms

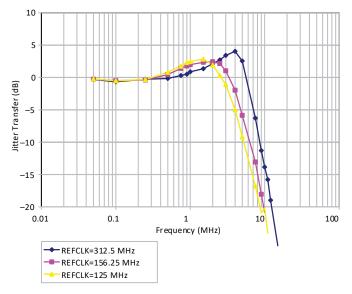


Figure 4.14. Jitter Transfer – 3.125 Gbps



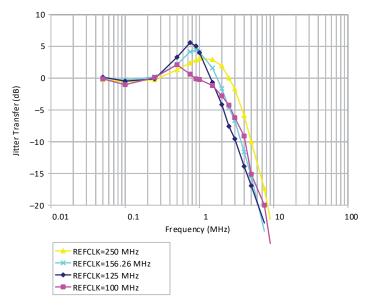


Figure 4.15. Jitter Transfer – 2.5 Gbps

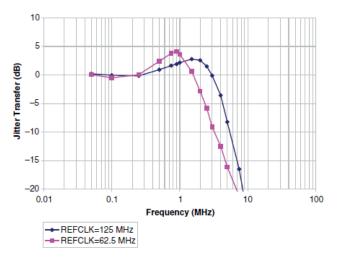


Figure 4.16. Jitter Transfer - 1.25 Gbps

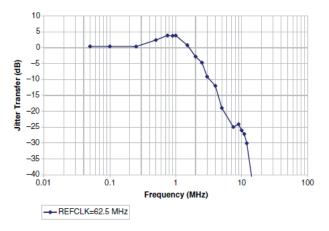


Figure 4.17. Jitter Transfer - 622 Mbps

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FPGA-DS-02074-3.2



## 4.26. PCI Express Electrical and Timing Characteristics

#### 4.26.1. AC and DC Characteristics

**Over Recommended Operating Conditions** 

Symbol	Description	Test Conditions	Min	Тур	Max	Units
Transmit <sup>1</sup>					1	
UI	Unit interval	_	399.88	400	400.12	ps
V <sub>TX-DIFF_P-P</sub>	Differential peak-to-peak output voltage	_	0.8	1.0	1.2	V
V <sub>TX-DE-RATIO</sub>	De-emphasis differential output voltage ratio	_	-3	-3.5	-4	dB
$V_{TX-CM-AC\_P}$	RMS AC peak common-mode output voltage		_	_	20	mV
V <sub>TX-RCV-DETECT</sub>	Amount of voltage change allowed during receiver detection	_	_	_	600	mV
V <sub>TX-DC-CM</sub>	Tx DC common mode voltage	_	0	_	VCCOB + 5%	V
I <sub>TX-SHORT</sub>	Output short circuit current	V <sub>TX-D+</sub> =0.0 V V <sub>TX-D-</sub> =0.0 V	_	_	90	mA
Z <sub>TX-DIFF-DC</sub>	Differential output impedance	_	80	100	120	Ohms
RL <sub>TX-DIFF</sub>	Differential return loss	_	10	_	_	dB
RL <sub>TX-CM</sub>	Common mode return loss	_	6.0	_	_	dB
T <sub>TX-RISE</sub>	Tx output rise time	20 to 80%	0.125	_	_	UI
T <sub>TX-FALL</sub>	Tx output fall time	20 to 80%	0.125	_	_	UI
L <sub>TX-SKEW</sub>	Lane-to-lane static output skew for all lanes in port/link	_	_	_	1.3	ns
T <sub>TX-EYE</sub>	Transmitter eye width	_	0.75	_	_	UI
T <sub>TX-EYE-MEDIAN-TO-MAX-JITTER</sub>	Maximum time between jitter median and maximum deviation from median		_	_	0.125	UI
Receive <sup>1, 2</sup>						
UI	Unit Interval	_	399.88	400	400.12	ps
V <sub>RX-DIFF_P-P</sub>	Differential peak-to-peak input voltage	_	0.34 <sup>3</sup>	_	1.2	V
V <sub>RX-IDLE-DET-DIFF_P-P</sub>	Idle detect threshold voltage	_	65	_	340 <sup>3</sup>	mV
V <sub>RX-CM-AC_P</sub>	Receiver common mode voltage for AC coupling	_	_	_	150	mV
Z <sub>RX-DIFF-DC</sub>	DC differential input impedance	_	80	100	120	Ω
Z <sub>RX-DC</sub>	DC input impedance	_	40	50	60	Ω
Z <sub>RX-HIGH-IMP-DC</sub>	Power-down DC input impedance	_	200K	_	_	Ω
R <sub>LRX-DIFF</sub>	Differential return loss	_	10		_	dB
R <sub>LRX-CM</sub>	Common mode return loss	_	6.0	_	_	dB
T <sub>RX-IDLE-DET-DIFF-ENTERTIME</sub>	Maximum time required for receiver to recognize and signal an unexpected idle on link	_	_	_	_	ms

#### Notes:

- 1. Values are measured at 2.5 Gbps.
- 2. Measured with external AC-coupling on the receiver.
- 3. Not in compliance with PCI Express 1.1 standard.



## 4.27. XAUI/Serial Rapid I/O Type 3/CPRI LV E.30 Electrical and Timing Characteristics

#### 4.27.1. AC and DC Characteristics

#### Table 4.13. Transmit

**Over Recommended Operating Conditions** 

Symbol	Description	Test Conditions	Min.	Тур.	Max.	Units	
T <sub>RF</sub>	Differential rise/fall time	20%-80%	_	80	_	ps	
Z <sub>TX_DIFF_DC</sub>	Differential impedance	_	80	100	120	Ω	
J <sub>TX_DDJ</sub> <sup>2, 3, 4</sup>	Output data deterministic jitter	_	_	_	0.17	UI	
J <sub>TX_TJ</sub> 1, 2, 3, 4	Total output data jitter	_	_	_	0.35	UI	

#### Notes:

- 1. Total jitter includes both deterministic jitter and random jitter.
- 2. Jitter values are measured with each CML output AC coupled into a 50-Ohm impedance (100-Ohm differential impedance).
- 3. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.
- 4. Values are measured at 2.5 Gbps.

#### **Table 4.14. Receive and Jitter Tolerance**

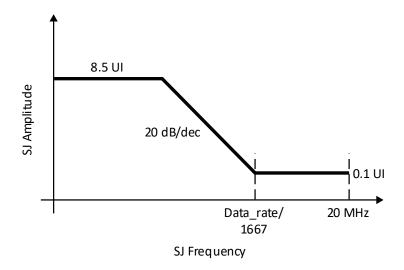
**Over Recommended Operating Conditions** 

Symbol	Description	Test Conditions	Min.	Тур.	Max.	Units
RL <sub>RX_DIFF</sub>	Differential return loss	From 100 MHz	10	_	_	dB
		to 3.125 GHz				
RL <sub>RX_CM</sub>	Common mode return loss	From 100 MHz	6	_	_	dB
		to 3.125 GHz				
Z <sub>RX_DIFF</sub>	Differential termination resistance	_	80	100	120	Ω
J <sub>RX_DJ</sub> 1, 2, 3	Deterministic jitter tolerance (peak-to-peak)	_	_	_	0.37	UI
J <sub>RX_RJ</sub> 1, 2, 3	Random jitter tolerance (peak-to-peak)	_	_	_	0.18	UI
J <sub>RX_SJ</sub> 1, 2, 3	Sinusoidal jitter tolerance (peak-to-peak)	_	_	_	0.10	UI
J <sub>RX_TJ</sub> 1, 2, 3	Total jitter tolerance (peak-to-peak)	_	_	_	0.65	UI
T <sub>RX_EYE</sub>	Receiver eye opening	_	0.35	_	_	UI

#### Notes:

- 1. Total jitter includes deterministic jitter, random jitter and sinusoidal jitter. The sinusoidal jitter tolerance mask is shown in Figure 4.18.
- 2. Jitter values are measured with each high-speed input AC coupled into a  $50-\Omega$  impedance.
- 3. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.
- 4. Jitter tolerance parameters are characterized when Full Rx Equalization is enabled.
- 5. Values are measured at 2.5 Gbps.





Note: The sinusoidal jitter tolerance is measured with at least 0.37 Ulpp of Deterministic jitter (Dj) and the sum of Dj and Rj (random jitter) is at least 0.55 Ulpp. Therefore, the sum of Dj, Rj and Sj (sinusoidal jitter) is at least 0.65 Ulpp (Dj = 0.37, Rj = 0.18, Sj = 0.1).

Figure 4.18. XAUI Sinusoidal Jitter Tolerance Mask

### 4.28. Serial Rapid I/O Type 2/CPRI LV E.24 Electrical and Timing Characteristics

#### 4.28.1. AC and DC Characteristics

Table 4.15. Transmit

Symbol	Description	<b>Test Conditions</b>	Min.	Тур.	Max.	Units
T <sub>RF</sub> <sup>1</sup>	Differential rise/fall time	20%-80%	ı	80	-	ps
Z <sub>TX_DIFF_DC</sub>	Differential impedance	-	80	100	120	Ω
J <sub>TX_DDJ</sub> <sup>3, 4, 5</sup>	Output data deterministic jitter	_	_	_	0.17	UI
J <sub>TX_TJ</sub> <sup>2, 3, 4, 5</sup>	Total output data jitter		1	1	0.35	UI

#### Notes:

- 1. Rise and Fall times measured with board trace, connector and approximately 2.5pf load.
- 2. Total jitter includes both deterministic jitter and random jitter. The random jitter is the total jitter minus the actual deterministic jitter.
- 3. Jitter values are measured with each CML output AC coupled into a 50-0hm impedance (100-0hm differential impedance).
- 4. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.
- 5. Values are measured at 2.5 Gbps.



#### Table 4.16. Receive and Jitter Tolerance

Symbol	Description	Test Conditions	Min.	Тур.	Max.	Units
RL <sub>RX_DIFF</sub>	Differential return loss	From 100 MHz to 2.5 GHz	10	_	_	dB
RL <sub>RX_CM</sub>	Common mode return loss	From 100 MHz to 2.5 GHz	6	_	_	dB
Z <sub>RX_DIFF</sub>	Differential termination resistance	_	80	100	120	Ω
J <sub>RX_DJ</sub> <sup>2, 3, 4, 5</sup>	Deterministic jitter tolerance (peak-to-peak)	_	_	_	0.37	UI
<b>J</b> <sub>RX_RJ</sub> <sup>2, 3, 4, 5</sup>	Random jitter tolerance (peak-to-peak)	_	_	_	0.18	UI
<b>J</b> <sub>RX_SJ</sub> <sup>2, 3, 4, 5</sup>	Sinusoidal jitter tolerance (peak-to-peak)	_	_	_	0.10	UI
J <sub>RX_TJ</sub> 1, 2, 3, 4, 5	Total jitter tolerance (peak-to-peak)	_	_	_	0.65	UI
T <sub>RX_EYE</sub>	Receiver eye opening	_	0.35	_	_	UI

#### Notes:

- Total jitter includes deterministic jitter, random jitter and sinusoidal jitter. The sinusoidal jitter tolerance mask is shown in Figure 4.18.
- 2. Jitter values are measured with each high-speed input AC coupled into a 50-Ohm impedance.
- 3. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.
- 4. Jitter tolerance, Differential Input Sensitivity and Receiver Eye Opening parameters are characterized when Full Rx Equalization is enabled.
- 5. Values are measured at 2.5 Gbps.

## 4.29. Gigabit Ethernet/Serial Rapid I/O Type 1/SGMII/CPRI LV E.12 Electrical and Timing Characteristics

#### 4.29.1. AC and DC Characteristics

Table 4.17. Transmit

Symbol	Description	<b>Test Conditions</b>	Min.	Тур.	Max.	Units
$T_{RF}$	Differential rise/fall time	20%-80%	ı	80	ı	ps
Z <sub>TX_DIFF_DC</sub>	Differential impedance	_	80	100	120	Ω
J <sub>TX_DDJ</sub> <sup>3, 4, 5</sup>	Output data deterministic jitter	_	1	-	0.10	UI
J <sub>TX_TJ</sub> 2, 3, 4, 5	Total output data jitter	_	_	_	0.24	UI

#### Notes:

- 1. Rise and fall times measured with board trace, connector and approximately 2.5 pf load.
- 2. Total jitter includes both deterministic jitter and random jitter. The random jitter is the total jitter minus the actual deterministic jitter.
- 3. Jitter values are measured with each CML output AC coupled into a 50-Ohm impedance (100-Ohm differential impedance).
- 4. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.
- 5. Values are measured at 1.25 Gbps.



#### Table 4.18. Receive and Jitter Tolerance

Symbol	Description	Test Conditions	Min.	Тур.	Max.	Units
RL <sub>RX_DIFF</sub>	Differential return loss	From 100 MHz to 1.25 GHz	10	_	_	dB
RL <sub>RX_CM</sub>	Common mode return loss	From 100 MHz to 1.25 GHz	6	_	_	dB
Z <sub>RX_DIFF</sub>	Differential termination resistance	_	80	100	120	Ω
J <sub>RX_DJ</sub> 1, 2, 3, 4, 5	Deterministic jitter tolerance (peak-to-peak)	_	_	_	0.34	UI
J <sub>RX_RJ</sub> 1, 2, 3, 4, 5	Random jitter tolerance (peak-to-peak)	_	_	_	0.26	UI
J <sub>RX_SJ</sub> <sup>1, 2, 3, 4, 5</sup>	Sinusoidal jitter tolerance (peak-to-peak)	_	_	_	0.11	UI
J <sub>RX_TJ</sub> 1, 2, 3, 4, 5	Total jitter tolerance (peak-to-peak)	_	_	_	0.71	UI
T <sub>RX_EYE</sub>	Receiver eye opening	_	0.29	_	_	UI

#### Notes:

- 1. Total jitter includes deterministic jitter, random jitter and sinusoidal jitter. The sinusoidal jitter tolerance mask is shown in Figure 4.18.
- 2. Jitter values are measured with each high-speed input AC coupled into a 50-Ohm impedance.
- 3. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.
- 4. Jitter tolerance, Differential Input Sensitivity and Receiver Eye Opening parameters are characterized when Full Rx Equalization is enabled.
- 5. Values are measured at 1.25 Gbps.

# 4.30. SMPTE SD/HD-SDI/3G-SDI (Serial Digital Interface) Electrical and Timing Characteristics

#### 4.30.1. AC and DC Characteristics

#### Table 4.19. Transmit

Symbol	Description	Test Conditions	Min.	Тур.	Max.	Units
BR <sub>SDO</sub>	Serial data rate	_	270	_	2975	Mbps
T <sub>JALIGNMENT</sub> <sup>2</sup>	Serial output jitter, alignment	270 Mbps	_	_	0.20	UI
T <sub>JALIGNMENT</sub> <sup>2</sup>	Serial output jitter, alignment	1485 Mbps	_	_	0.20	UI
T <sub>JALIGNMENT</sub> 1, 2	Serial output jitter, alignment	2970Mbps	_	_	0.30	UI
T <sub>JTIMING</sub>	Serial output jitter, timing	270 Mbps	_	_	0.20	UI
T <sub>JTIMING</sub>	Serial output jitter, timing	1485 Mbps	_	_	1.0	UI
T <sub>JTIMING</sub>	Serial output jitter, timing	2970 Mbps	_	_	2.0	UI

#### Notes:

- Timing jitter is measured in accordance with SMPTE RP 184-1996, SMPTE RP 192-1996 and the applicable serial data transmission standard, SMPTE 259M-1997 or SMPTE 292M (proposed). A color bar test pattern is used. The value of fSCLK is 270 MHz or 360 MHz for SMPTE 259M, 540 MHz for SMPTE 344M or 1485 MHz for SMPTE 292M serial data rates. See the Timing Jitter Bandpass section.
- 2. Jitter is defined in accordance with SMPTE RP1 184-1996 as: jitter at an equipment output in the absence of input jitter.
- All Tx jitter is measured at the output of an industry standard cable driver; connection to the cable driver is via a 50 Ω impedance differential signal from the Lattice SERDES device.
- 4. The cable driver drives: RL=75  $\Omega$ , AC-coupled at 270, 1485, or 2970 Mbps, RREFLVL=RREFPRE=4.75 k $\Omega$  1%.



#### Table 4.20. Receive

Symbol	Description	Test Conditions	Min.	Тур.	Max.	Units
BR <sub>SDI</sub>	Serial input data rate		270	ı	2970	Mbps
CID	Stream of non- transitions (=Consecutive Identical Digits)	_	7(3G)/26(SMP TE Triple rates) @ 10-12 BER	1	1	Bits

#### **Table 4.21. Reference Clock**

Symbol	Description	Test Conditions	Min.	Тур.	Max.	Units
F <sub>VCLK</sub>	Video output clock frequency	_	27	1	74.25	MHz
$DC_V$	Duty cycle, video clock	_	45	50	55	%

# 4.31. HDMI (High-Definition Multimedia Interface) Electrical and Timing Characteristics

#### 4.31.1. AC and DC Characteristics

#### Table 4.22. Transmit and Receive<sup>1, 2</sup>

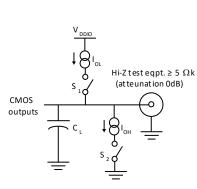
Symbol	Description	Spec. Compliance		Units
		Min. Spec.	Max. Spec.	
Transmit				
Intra-pair Skew	_	_	75	ps
Inter-pair Skew	_	_	800	ps
TMDS Differential Clock Jitter	_	_	0.25	UI
Receive				
R <sub>T</sub>	Termination Resistance	40	60	Ω
V <sub>ICM</sub>	Input AC Common Mode Voltage (50-Ω Setting)	_	50	mV
TMDS Clock Jitter	Clock Jitter Tolerance	_	0.25	UI

#### Notes:

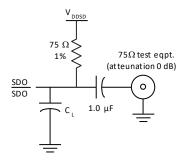
- 1. Output buffers must drive a translation device. Max. speed is 2 Gbps. If translation device does not modify rise/fall time, the maximum speed is 1.5 Gbps.
- 2. Input buffers must be AC coupled in order to support the 3.3 V common mode. Generally, HDMI inputs are terminated by an external cable equalizer before data/clock is forwarded to the LatticeECP3 device.

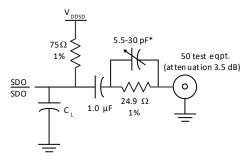


#### Test Loads



C  $_{\rm L}$  including probe and jig capacitance, 3 pF max. S  $_{\rm 1}$  - open, S  $_{\rm 2}$  - closed for V  $_{\rm OH}$  measurement S  $_{\rm 1}$  - closed, S  $_{\rm 2}$  - open for V  $_{\rm OL}$  measurement





\*Risetime compensation.

#### Timing Jitter Bandpass

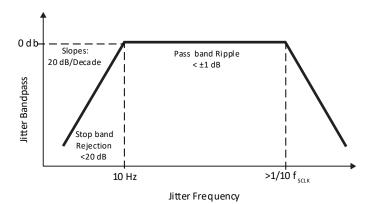


Figure 4.19. Test Loads



## 4.32. LatticeECP3 sysCONFIG Port Timing Specifications

**Over Recommended Operating Conditions** 

Parameter	Description		Min.	Max.	Units
POR, Configu	uration Initialization, and Wakeup				
t <sub>ICFG</sub>	Time from the Application of VCC, VCCAUX or VCCIO8* (Whichever is the Last to Cross the POR Trip Point) to the	Master mode Slave mode		23 6	ms ms
+ .	Rising Edge of INITN  Time from tICFG to the Valid Master MCLK			5	116
t <sub>VMC</sub>			25	5	μs
t <sub>PRGM</sub>	PROGRAMN Low Time to Start Configuration		25	10	ns
t <sub>PRGMRJ</sub>		PROGRAMN Pin Pulse Rejection		10 37	ns
t <sub>DPPINIT</sub>	,	Delay Time from PROGRAMN Low to INITN Low		1	ns
t <sub>DPPDONE</sub>	Delay Time from PROGRAMN Low to DONE Low			37	ns
t <sub>DINIT</sub> <sup>1</sup>	PROGRAMN High to INITN High Delay	-h	100	500	ms
t <sub>MWC</sub>	Additional Wake Master Clock Signals After DONE Pin is Hig	gn	100	500	cycles
t <sub>CZ</sub>	MCLK From Active To Low To High-Z		+-	300	ns
t <sub>IODISS</sub>	User I/O Disable from PROGRAMN Low			100	ns
t <sub>IOENSS</sub>	User I/O Enabled Time from CCLK Edge During Wake-up Sec	quence		100	ns
All Configura					
t <sub>SUCDI</sub>	Data Setup Time to CCLK/MCLK		5		ns
t <sub>HCDI</sub>	Data Hold Time to CCLK/MCLK  CCLK/MCLK to DOUT in Flourthrough Mode		1		ns
t <sub>CODO</sub>	CCLK/MCLK to DOUT in Flowthrough Mode		-0.2	12	ns
Slave Serial	T		I	1	
t <sub>SSCH</sub>	CCLK Minimum High Pulse		5		ns
t <sub>SSCL</sub>	CCLK Minimum Low Pulse		5		ns
f <sub>CCLK</sub>	CCLK Frequency	Without encryption With encryption		33 20	MHz MHz
Master and S	Slave Parallel			•	
t <sub>SUCS</sub>	CSN[1:0] Setup Time to CCLK/MCLK		7	_	ns
t <sub>HCS</sub>	CSN[1:0] Hold Time to CCLK/MCLK		1	_	ns
t <sub>SUWD</sub>	WRITEN Setup Time to CCLK/MCLK		7	_	ns
t <sub>HWD</sub>	WRITEN Hold Time to CCLK/MCLK		1	_	ns
t <sub>DCB</sub>	CCLK/MCLK to BUSY Delay Time		_	12	ns
t <sub>CORD</sub>	CCLK to Out for Read Data		_	12	ns
t <sub>BSCH</sub>	CCLK Minimum High Pulse		6	_	ns
t <sub>BSCL</sub>	CCLK Minimum Low Pulse		6	_	ns
t <sub>BSCYC</sub>	Byte Slave Cycle Time		30	_	ns
f <sub>CCLK</sub>	CCLK/MCLK Frequency	Without encryption	_	33	MHz
		With encryption	_	20	MHz
Master and S	Slave SPI				
t <sub>CFGX</sub>	INITN High to MCLK Low		_	80	ns
t <sub>CSSPI</sub>	INITN High to CSSPIN Low		0.2	2	μs
t <sub>SOCDO</sub>	MCLK Low to Output Valid		_	15	ns
t <sub>CSPID</sub>	CSSPIN[0:1] Low to First MCLK Edge Setup Time		0.3	_	μs
f <sub>CCLK</sub>	CCLK Frequency	Without encryption	_	33	MHz
		With encryption	_	20	MHz
t <sub>SSCH</sub>	CCLK Minimum High Pulse		5	_	ns
t <sub>SSCL</sub>	CCLK Minimum Low Pulse		5	_	ns
t <sub>HLCH</sub>	HOLDN Low Setup Time (Relative to CCLK)		5	_	ns
t <sub>CHHH</sub>	HOLDN Low Hold Time (Relative to CCLK)		5	<u> </u>	ns



Parameter	Description	Min.	Max.	Units			
Master and SI	Master and Slave SPI						
t <sub>CHHL</sub>	HOLDN High Hold Time (Relative to CCLK)	5	_	ns			
t <sub>HHCH</sub>	HOLDN High Setup Time (Relative to CCLK)	5	_	ns			
t <sub>HLQZ</sub>	HOLDN to Output High-Z	_	9	ns			
t <sub>HHQX</sub>	HOLDN to Output Low-Z	_	9	ns			

Note: Re-toggling the PROGRAMN pin is not permitted until the INITN pin is high. Avoid consecutive toggling of the PROGRAMN.

Parameter	Min.	Max.	Units
Master Clock Frequency	Selected value - 15%	Selected value + 15%	MHz
Duty Cycle	40	60	%

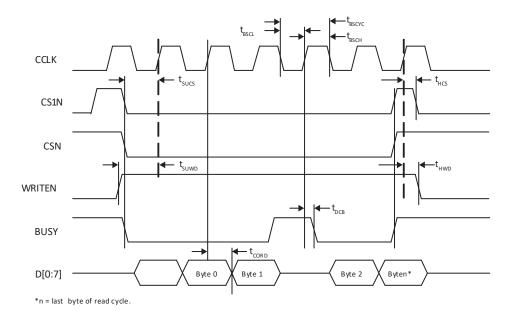
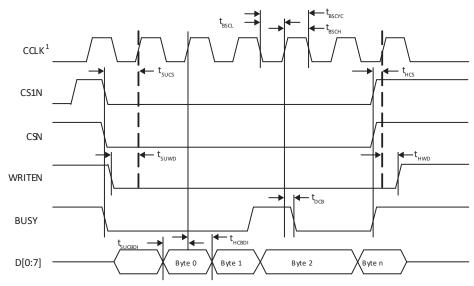


Figure 4.20. sysCONFIG Parallel Port Read Cycle



1. In Master Parallel Mode the FPGA provides CCLK (MCLK). In Slave Parallel Mode the external device provides CCLK.

Figure 4.21. sysCONFIG Parallel Port Write Cycle



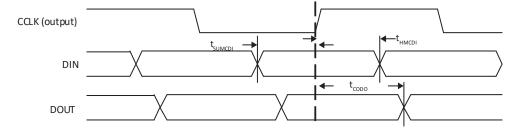


Figure 4.22. sysCONFIG Master Serial Port Timing

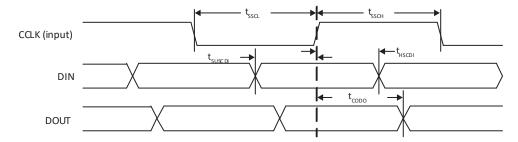
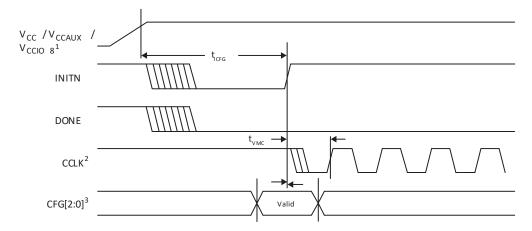


Figure 4.23. sysCONFIG Slave Serial Port Timing



- 1. Time taken from VCC, VCCAUX or VCCIO8, whichever is the last to cross the POR trip point.
- 2. Device is in a Master Mode (SPI, SPIm).
- 3. The CFG pins are normally static (hard wired).

Figure 4.24. Power-On-Reset (POR) Timing



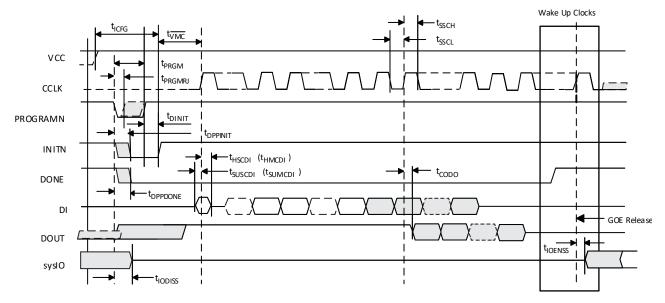
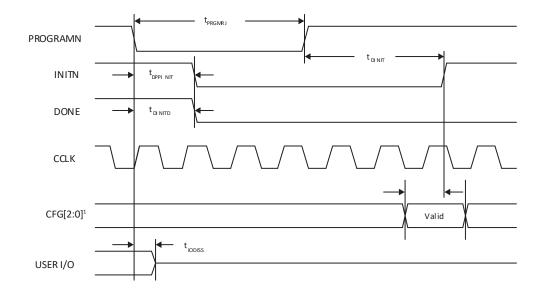


Figure 4.25. sysCONFIG Port Timing



1. The CFG pins are normally static (hardwired)

Figure 4.26. Configuration from PROGRAMN Timing



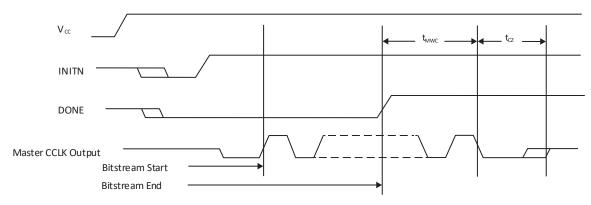


Figure 4.27. Wake-Up Timing

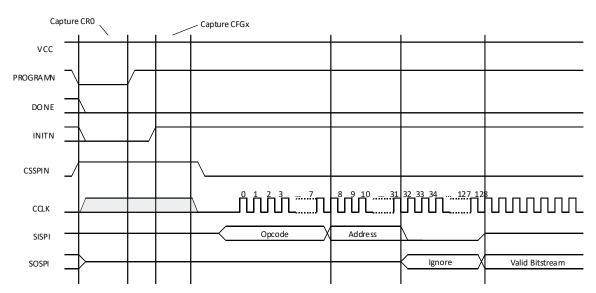


Figure 4.28. Master SPI Configuration Waveforms

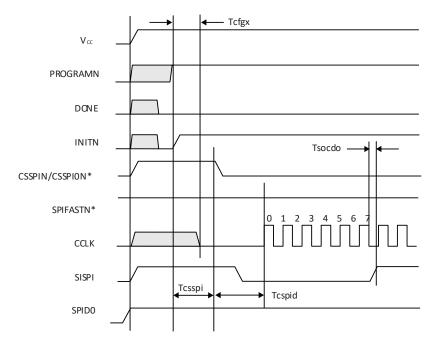


Figure 4.29. Master SPI POR Waveforms

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Figure 4.30. SPI Configuration Waveforms

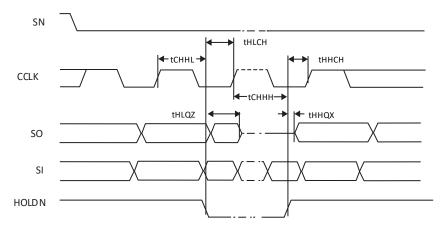


Figure 4.31. Slave SPI HOLDN Waveforms



## 4.33. JTAG Port Timing Specifications

**Over Recommended Operating Conditions** 

Symbol	Parameter	Min	Max	Units
f <sub>MAX</sub>	TCK clock frequency	_	25	MHz
t <sub>BTCP</sub>	TCK [BSCAN] clock pulse width	40	_	ns
t <sub>BTCPH</sub>	TCK [BSCAN] clock pulse width high	20	_	ns
t <sub>BTCPL</sub>	TCK [BSCAN] clock pulse width low	20	_	ns
t <sub>BTS</sub>	TCK [BSCAN] setup time	10	_	ns
t <sub>BTH</sub>	TCK [BSCAN] hold time	8	_	ns
t <sub>BTRF</sub>	TCK [BSCAN] rise/fall time	50	_	mV/ns
t <sub>BTCO</sub>	TAP controller falling edge of clock to valid output	_	10	ns
t <sub>BTCODIS</sub>	TAP controller falling edge of clock to valid disable	_	10	ns
t <sub>BTCOEN</sub>	TAP controller falling edge of clock to valid enable	_	10	ns
t <sub>BTCRS</sub>	BSCAN test capture register setup time	8	_	ns
t <sub>BTCRH</sub>	BSCAN test capture register hold time	25	_	ns
t <sub>BUTCO</sub>	BSCAN test update register, falling edge of clock to valid output	_	25	ns
t <sub>BTUODIS</sub>	BSCAN test update register, falling edge of clock to valid disable		25	ns
t <sub>BTUPOEN</sub>	BSCAN test update register, falling edge of clock to valid enable	_	25	ns

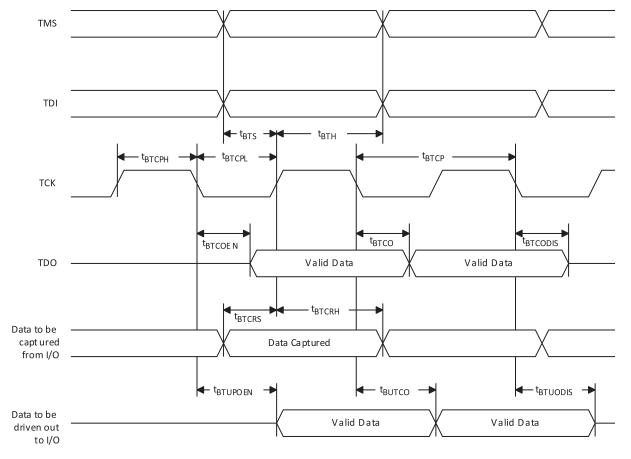


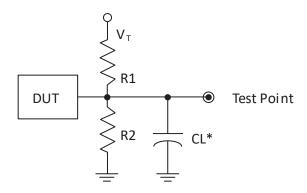
Figure 4.32. JTAG Port Timing Waveforms

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## 4.34. Switching Test Conditions

Figure 4.33 shows the output test load that is used for AC testing. The specific values for resistance, capacitance, voltage, and other test conditions are shown in Figure 4.23.



<sup>\*</sup>CLIncludes Test Fixture and Probe Capacitance

Figure 4.33. Output Test Load, LVTTL and LVCMOS Standards

Table 4.23. Test Fixture Required Components, Non-Terminated Interfaces

Total Constitution			_	Timing Def	
Test Condition	R <sub>1</sub>	R <sub>2</sub>	C <sub>L</sub>	Timing Ref.	V <sub>T</sub>
LVTTL and other LVCMOS settings (L -> H, H -> L)	∞	∞	0 pF	LVCMOS 3.3 = 1.5 V	_
				LVCMOS $2.5 = V_{CCIO}/2$	_
				LVCMOS 1.8 = V <sub>CCIO</sub> /2	_
				LVCMOS 1.5 = $V_{CCIO}/2$	_
				LVCMOS 1.2 = V <sub>CCIO</sub> /2	_
LVCMOS 2.5 I/O (Z -> H)	∞	1 M∞	0 pF	V <sub>CCIO</sub> /2	_
LVCMOS 2.5 I/O (Z -> L)	1 ΜΩ	∞	0 pF	V <sub>CCIO</sub> /2	V <sub>ccio</sub>
LVCMOS 2.5 I/O (H -> Z)	∞	100	0 pF	V <sub>OH</sub> - 0.10	_
LVCMOS 2.5 I/O (L -> Z)	100	8	0 pF	V <sub>OL</sub> + 0.10	V <sub>CCIO</sub>

Note: Output test conditions for all other interfaces are determined by the respective standards.

## 4.35. sysI/O Differential Electrical Characteristics

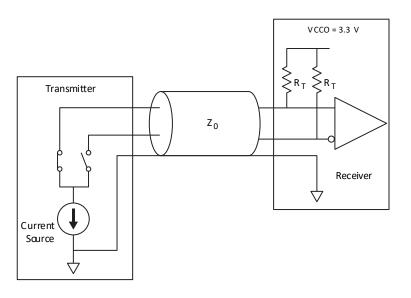
#### 4.35.1. Transition Reduced LVDS (TRLVDS DC Specification)

**Over Recommended Operating Conditions** 

Symbol	Description	Min.	Nom.	Max.	Units
V <sub>CCO</sub>	Driver supply voltage (+/– 5%)	3.14	3.3	3.47	V
V <sub>ID</sub>	Input differential voltage	150	1	1200	mV
V <sub>ICM</sub>	Input common mode voltage	3	_	3.265	V
V <sub>cco</sub>	Termination supply voltage	3.14	3.3	3.47	V
R <sub>T</sub>	Termination resistance (off-chip)	45	50	55	Ω

Note: LatticeECP3 supports only the TRLVDS receiver.





#### 4.35.2. Mini LVDS

**Over Recommended Operating Conditions** 

Symbol	Description	Min.	Тур.	Max.	Units
Z <sub>0</sub>	Single-ended PCB trace impedance	30	50	75	Ohms
R <sub>T</sub>	Differential termination resistance	50	100	150	Ohms
V <sub>OD</sub>	Output voltage, differential,  V <sub>OP</sub> – V <sub>OM</sub>	300	_	600	mV
Vos	Output voltage, common mode,  V <sub>OP</sub> + V <sub>OM</sub>  /2	1	1.2	1.4	V
$\Delta V_{\text{OD}}$	Change in V <sub>OD</sub> , between H and L	_	_	50	mV
$\Delta V_{ID}$	Change in V <sub>OS</sub> , between H and L	_	_	50	mV
V <sub>THD</sub>	Input voltage, differential,  V <sub>INP</sub> – V <sub>INM</sub>	200	_	600	mV
V <sub>CM</sub>	Input voltage, common mode,  V <sub>INP</sub> + V <sub>INM</sub>  /2	0.3+(V <sub>THD</sub> /2)	_	2.1-(V <sub>THD</sub> /2)	
T <sub>R</sub> , T <sub>F</sub>	Output rise and fall times, 20% to 80%	_	_	550	ps
T <sub>ODUTY</sub>	Output clock duty cycle	40	_	60	%

Note: Data is for 6 mA differential current drive. Other differential driver current options are available.

## 4.35.3. Point-to-Point LVDS (PPLVDS)

**Over Recommended Operating Conditions** 

Description	Min.	Тур.	Max.	Units
Output driver supply (+/- 5%)	3.14	3.3	3.47	V
	2.25	2.5	2.75	V
Input differential voltage	100	_	400	mV
Input common mode voltage	0.2	_	2.3	V
Output differential voltage	130	_	400	mV
Output common mode voltage	0.5	0.8	1.4	V

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#### 4.35.4. RSDS

**Over Recommended Operating Conditions** 

Symbol	Description	Min.	Тур.	Max.	Units
V <sub>OD</sub>	Output voltage, differential, $R_T = 100 \Omega$	100	200	600	mV
Vos	Output voltage, common mode	0.5	1.2	1.5	V
I <sub>RSDS</sub>	Differential driver output current	1	2	6	mA
V <sub>THD</sub>	Input voltage differential	100	_	_	mV
V <sub>CM</sub>	Input common mode voltage	0.3	_	1.5	V
T <sub>R</sub> , T <sub>F</sub>	Output rise and fall times, 20% to 80%	_	500	_	ps
T <sub>ODUTY</sub>	Output clock duty cycle	35	50	65	%

Note: Data is for 2 mA drive. Other differential driver current options are available.



## 5. Pinout Information

## 5.1. Signal Descriptions

5.1. Signal Descriptions		
Signal Name	I/O	Description
General Purpose		
P[Edge] [Row/Column Number]_[A/B]	I/O	[Edge] indicates the edge of the device on which the pad is located. Valid edge designations are L (Left), B (Bottom), R (Right), T (Top).  [Row/Column Number] indicates the PFU row or the column of the device on which the PIC exists. When Edge is T (Top) or B (Bottom), only need to spec- ify Column Number. When Edge is L (Left) or R (Right), only need to specify Row Number.  [A/B] indicates the PIO within the PIC to which the pad is connected. Some of these user-programmable pins are shared with special function pins. These pins, when not used as special purpose pins, can be programmed as I/O for user logic. During configuration the user-programmable I/O are tri-stated with an internal pull-up resistor enabled. If any pin is not used (or not bonded to a package pin), it is also tri-stated with an internal pull-up resistor enabled after configuration.
P[Edge][Row Number]E_[A/B/C/D]	l	These general purpose signals are input-only pins and are located near the PLLs.
GSRN	I	Global RESET signal (active low). Any I/O pin can be GSRN.
NC	_	No connect.
RESERVED	_	This pin is reserved and should not be connected to anything on the board.
GND	_	Ground. Dedicated pins.
V <sub>CC</sub>	_	Power supply pins for core logic. Dedicated pins.
V <sub>CCAUX</sub>	-	Auxiliary power supply pin. This dedicated pin powers all the differential and referenced input buffers.
V <sub>CCIOx</sub>	_	Dedicated power supply pins for I/O bank x.
V <sub>CCA</sub>	I	SERDES, transmit, receive, PLL and reference clock buffer power supply. All VCCA supply pins must always be powered to the recommended operating voltage range. If no SERDES channels are used, connect VCCA to VCC.
V <sub>CCPLL_[LOC]</sub>	١	General purpose PLL supply pins where LOC=L (left) or R (right).
V <sub>REF1_x</sub> , V <sub>REF2_x</sub>	ı	Reference supply pins for I/O bank x. Pre-determined pins in each bank are assigned as VREF inputs. When not used, they may be used as I/O pins.
VTTx	1	Power supply for on-chip termination of I/O.
XRES <sup>1</sup>	_	$10~k\Omega$ +/-1% resistor must be connected between this pad and ground.
PLL, DLL, and Clock Functions		
[LOC][num]_GPLL[T, C]_IN_[index]	I	General Purpose PLL (GPLL) input pads: LUM, LLM, RUM, RLM, num = row from center, T = true and C = complement, index A,B,Cat each side.
[LOC][num]_GPLL[T, C]_FB_[index]	I	Optional feedback GPLL input pads: LUM, LLM, RUM, RLM, num = row from center, T = true and C = complement, index A,B,Cat each side.
[LOC]0_GDLLT_IN_[index]2	I/O	General Purpose DLL (GDLL) input pads where LOC=RUM or LUM, T is True Complement, index is A or B.
[LOC]0_GDLLT_FB_[index]2	I/O	Optional feedback GDLL input pads where LOC=RUM or LUM, T is True Complement, index is A or B.
PCLK[T, C][n:0]_[3:0]2	I/O	Primary Clock pads, T = true and C = complement, n per side, indexed by bank and 0, 1, 2, 3 within bank.
[LOC]DQS[num]	I/O	DQ input/output pads: T (top), R (right), B (bottom), L (left), DQS, num = ball function number.
[LOC]DQ[num]	I/O	DQ input/output pads: T (top), R (right), B (bottom), L (left), DQ, associated DQS number.



Signal Name	I/O	Description
Test and Programming (Dedicated F	Pins)	
TMS	_	Test Mode Select input, used to control the 1149.1 state machine. Pull-up is enabled during configuration.
TCK	I	Test Clock input pin, used to clock the 1149.1 state machine. No pull-up enabled.
TDI	1	Test Data in pin. Used to load data into device using 1149.1 state machine.  After power-up, this TAP port can be activated for configuration by sending appropriate command. (Note: once a configuration port is selected it is locked. Another configuration port cannot be selected until the power-up sequence). Pull-up is enabled during configuration.
TDO	0	Output pin. Test Data Out pin used to shift data out of a device using 1149.1.
VCCJ	_	Power supply pin for JTAG Test Access Port.
Configuration Pads (Used During sy	sCONFIG)	
CFG[2:0]	I	Mode pins used to specify configuration mode values latched on rising edge of INITN. During configuration, a pull-up is enabled. These are dedicated pins.
INITN	1/0	Open Drain pin. Indicates the FPGA is ready to be configured. During configuration, a pull-up is enabled. It is a dedicated pin.
PROGRAMN	I	Initiates configuration sequence when asserted low. This pin always has an active pull-up. It is a dedicated pin.
DONE	1/0	Open Drain pin. Indicates that the configuration sequence is complete, and the startup sequence is in progress. It is a dedicated pin.
CCLK	I	Input Configuration Clock for configuring an FPGA in Slave SPI, Serial, and CPU modes. It is a dedicated pin.
MCLK	I/O	Output Configuration Clock for configuring an FPGA in SPI, SPIm, and Master configuration modes.
BUSY/SISPI	0	Parallel configuration mode busy indicator. SPI/SPIm mode data output.
CSN/SN/OEN	I/O	Parallel configuration mode active-low chip select. Slave SPI chip select. Parallel burst Flash output enable.
CS1N/HOLDN/RDY	I	Parallel configuration mode active-low chip select. Slave SPI hold input.
WRITEN	1	Write enable for parallel configuration modes.
DOUT/CSON/CSSPI1N	0	Serial data output. Chip select output. SPI/SPIm mode chip select.
D[0]/SPIFASTN	1/0	sysCONFIG Port Data I/O for Parallel mode. Open drain during configuration.  sysCONFIG Port Data I/O for SPI or SPIm. When using the SPI or SPIm mode, this pin should either be tied high or low, must not be left floating. Open drain during configuration.
D1	I/O	Parallel configuration I/O. Open drain during configuration.
D2	1/0	Parallel configuration I/O. Open drain during configuration.
D3/SI	1/0	Parallel configuration I/O. Slave SPI data input. Open drain during configuration.
D4/SO	I/O	Parallel configuration I/O. Slave SPI data output. Open drain during configuration.
D5	I/O	Parallel configuration I/O. Open drain during configuration.
D6/SPID1	I/O	Parallel configuration I/O. SPI/SPIm data input. Open drain during configuration.
D7/SPID0	I/O	Parallel configuration I/O. SPI/SPIm data input. Open drain during configuration.
DI/CSSPION/CEN	I/O	Serial data input for slave serial mode. SPI/SPIm mode chip select.



Signal Name	I/O	Description
Dedicated SERDES Signals <sup>3</sup>		
PCS[Index]_HDINNm	_	High-speed input, negative channel m
PCS[Index]_HDOUTNm	0	High-speed output, negative channel m
PCS[Index]_REFCLKN	I	Negative Reference Clock Input
PCS[Index]_HDINPm	I	High-speed input, positive channel m
PCS[Index]_HDOUTPm	0	High-speed output, positive channel m
PCS[Index]_REFCLKP	I	Positive Reference Clock Input
PCS[Index]_VCCOBm	_	Output buffer power supply, channel m (1.2V/1.5)
PCS[Index]_VCCIBm	_	Input buffer power supply, channel m (1.2V/1.5V)

#### Notes:

- 1. When placing switching I/O around these critical pins that are designed to supply the device with the proper reference or supply voltage, care must be given.
- 2. These pins are dedicated inputs or can be used as general purpose I/O.
- 3. m defines the associated channel in the quad.

## 5.2. PICs and DDR Data (DQ) Pins Associated with the DDR Strobe (DQS) Pin

PICs Associated with DQS Strobe	PIO Within PIC	DDR Strobe (DQS) and Data (DQ) Pins
For Left and Right Edges of the Device		
P[Edge] [n-3]	A	DQ
	В	DQ
P[Edge] [n-2]	Α	DQ
	В	DQ
P[Edge] [n-1]	A	DQ
	В	DQ
P[Edge] [n]	A	[Edge]DQSn
	В	DQ
P[Edge] [n+1]	А	DQ
	В	DQ
P[Edge] [n+2]	Α	DQ
	В	DQ
For Top Edge of the Device		
P[Edge] [n-3]	Α	DQ
	В	DQ
P[Edge] [n-2]	Α	DQ
	В	DQ
P[Edge] [n-1]	Α	DQ
	В	DQ
P[Edge] [n]	A	[Edge]DQSn
	В	DQ
P[Edge] [n+1]	A	DQ
	В	DQ
P[Edge] [n+2]	A	DQ
	В	DQ

**Note:** *n* is a row PIC number.

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## 5.3. Pin Information Summary

Pin Information Summ	ary	E	CP3-17EA		ECP3-35EA			ECP3-70EA		
Pin Type		256 ftBGA	328 csBGA	484 fpBGA	256 ftBGA	484 fpBGA	672 fpBGA	484 fpBGA	672 fpBGA	1156 fpBGA
General Purpose	Bank 0	26	20	36	26	42	48	42	60	86
Inputs/Outputs per	Bank 1	14	10	24	14	36	36	36	48	78
Bank	Bank 2	6	7	12	6	24	24	24	34	36
	Bank 3	18	12	44	16	54	59	54	59	86
	Bank 6	20	11	44	18	63	61	63	67	86
	Bank 7	19	26	32	19	36	42	36	48	54
	Bank 8	24	24	24	24	24	24	24	24	24
General Purpose	Bank 0	0	0	0	0	0	0	0	0	0
Inputs per Bank	Bank 1	0	0	0	0	0	0	0	0	0
	Bank 2	2	2	2	2	4	4	4	8	8
	Bank 3	0	0	0	2	4	4	4	12	12
	Bank 6	0	0	0	2	4	4	4	12	12
	Bank 7	4	4	4	4	4	4	4	8	8
	Bank 8	0	0	0	0	0	0	0	0	0
General Purpose Out-	Bank 0	0	0	0	0	0	0	0	0	0
puts per Bank	Bank 1	0	0	0	0	0	0	0	0	0
	Bank 2	0	0	0	0	0	0	0	0	0
	Bank 3	0	0	0	0	0	0	0	0	0
	Bank 6	0	0	0	0	0	0	0	0	0
	Bank 7	0	0	0	0	0	0	0	0	0
	Bank 8	0	0	0	0	0	0	0	0	0
Total Single-Ended Use	r I/O	133	116	222	133	295	310	295	380	490
VCC		6	16	16	6	16	32	16	32	32
VCCAUX		4	5	8	4	8	12	8	12	16
VTT		4	7	4	4	4	4	4	4	8
VCCA		4	6	4	4	4	8	4	8	16
VCCPLL		2	2	4	2	4	4	4	4	4
VCCIO	Bank 0	2	3	2	2	2	4	2	4	4
	Bank 1	2	3	2	2	2	4	2	4	4
	Bank 2	2	2	2	2	2	4	2	4	4
	Bank 3	2	3	2	2	2	4	2	4	4
	Bank 6	2	3	2	2	2	4	2	4	4
	Bank 7	2	3	2	2	2	4	2	4	4
	Bank 8	1	2	2	1	2	2	2	2	2
VCCJ		1	1	1	1	1	1	1	1	1
TAP		4	4	4	4	4	4	4	4	4
GND, GNDIO		51	126	98	51	98	139	98	139	233
NC		0	0	73	0	0	96	0	0	238
Reserved1		0	0	2	0	2	2	2	2	2
SERDES		26	18	26	26	26	26	26	52	78
Miscellaneous Pins		8	8	8	8	8	8	8	8	8
Total Bonded Pins		256	328	484	256	484	672	484	672	1156



## **Pin Information Summary (Cont.)**

Pin Information Summary			ECP3-17EA		ECP3-35EA			
Pin Type		256	328	484	256	484	672	
		ftBGA	csBGA	fpBGA	ftBGA	fpBGA	fpBGA	
Emulated Differential I/O	Bank 0	13	10	18	13	21	24	
per Bank	Bank 1	7	5	12	7	18	18	
	Bank 2	2	2	4	1	8	8	
	Bank 3	4	2	13	5	20	19	
	Bank 6	5	1	13	6	22	20	
	Bank 7	6	9	10	6	11	13	
	Bank 8	12	12	12	12	12	12	
Highspeed Differential I/O	Bank 0	0	0	0	0	0	0	
per Bank	Bank 1	0	0	0	0	0	0	
	Bank 2	2	2	3	3	6	6	
	Bank 3	5	4	9	4	9	12	
	Bank 6	5	4	9	4	11	12	
	Bank 7	5	6	8	5	9	10	
	Bank 8	0	0	0	0	0	0	
Total Single Ended/ Total	Bank 0	26/13	20/10	36/18	26/13	42/21	48/24	
Differential I/O per Bank	Bank 1	14/7	10/5	24/12	14/7	36/18	36/18	
	Bank 2	8/4	9/4	14/7	8/4	28/14	28/14	
	Bank 3	18/9	12/6	44/22	18/9	58/29	63/31	
	Bank 6	20/10	11/5	44/22	20/10	67/33	65/32	
	Bank 7	23/11	30/15	36/18	23/11	40/20	46/23	
	Bank 8	24/12	24/12	24/12	24/12	24/12	24/12	
DDR Groups Bonded per	Bank 0	2	1	3	2	3	4	
Bank2	Bank 1	1	0	2	1	3	3	
	Bank 2	0	0	1	0	2	2	
	Bank 3	1	0	3	1	3	4	
	Bank 6	1	0	3	1	4	4	
	Bank 7	1	2	2	1	3	3	
	Configuration Bank 8	0	0	0	0	0	0	
SERDES Quads	•	1	1	1	1	1	1	

#### Notes:

- 1. These pins must remain floating on the board.
- 2. Some DQS groups may not support DQS-12. Refer to the device pinout (.csv) file.



Pin Information Summary (Cont.)

Pin Information Summary		ECP3-70EA					
Pin Type		484 fpBGA	672 fpBGA	1156 fpBGA			
Emulated Differential	Bank 0	21	30	43			
I/O per Bank	Bank 1	18	24	39			
	Bank 2	8	12	13			
	Bank 3	20	23	33			
	Bank 6	22	25	33			
	Bank 7	11	16	18			
	Bank 8	12	12	12			
High-Speed Differential	Bank 0	0	0	0			
I/ O per Bank	Bank 1	0	0	0			
	Bank 2	6	9	9			
	Bank 3	9	12	16			
	Bank 6	11	14	16			
	Bank 7	9	12	13			
	Bank 8	0	0	0			
Total Single-Ended/ Total	Bank 0	42/21	60/30	86/43			
Differential I/O per	Bank 1	36/18	48/24	78/39			
Bank	Bank 2	28/14	42/21	44/22			
	Bank 3	58/29	71/35	98/49			
	Bank 6	67/33	78/39	98/49			
	Bank 7	40/20	56/28	62/31			
	Bank 8	24/12	24/12	24/12			
DDR Groups Bonded per	Bank 0	3	5	7			
Bank*	Bank 1	3	4	7			
	Bank 2	2	3	3			
	Bank 3	3	4	5			
	Bank 6	4	4	5			
	Bank 7	3	4	4			
	Configuration Bank 8	0	0	0			
SERDES Quads		1	2	3			

<sup>\*</sup>Note: Some DQS groups may not support DQS-12. Refer to the device pinout (.csv) file.



**Pin Information Summary (Cont.)** 

Pin Information Summary			ECP3-95E/	4	ECP3-150EA		
Pin Type			672 fpBGA	1156 fpBGA	672 fpBGA	1156 fpBGA	
General Purpose	Bank 0	42	60	86	60	94	
Inputs/Outputs per bank	Bank 1	36	48	78	48	86	
	Bank 2	24	34	36	34	58	
	Bank 3	54	59	86	59	104	
	Bank 6	63	67	86	67	104	
	Bank 7	36	48	54	48	76	
	Bank 8	24	24	24	24	24	
General Purpose Inputs per	Bank 0	0	0	0	0	0	
Bank	Bank 1	0	0	0	0	0	
	Bank 2	4	8	8	8	8	
	Bank 3	4	12	12	12	12	
	Bank 6	4	12	12	12	12	
	Bank 7	4	8	8	8	8	
	Bank 8	0	0	0	0	0	
General Purpose Outputs per	Bank 0	0	0	0	0	0	
Bank	Bank 1	0	0	0	0	0	
	Bank 2	0	0	0	0	0	
	Bank 3	0	0	0	0	0	
	Bank 6	0	0	0	0	0	
	Bank 7	0	0	0	0	0	
	Bank 8	0	0	0	0	0	
Total Single-Ended User I/O		295	380	490	380	586	
VCC		16	32	32	32	32	
VCCAUX		8	12	16	12	16	
VTT		4	4	8	4	8	
VCCA		4	8	16	8	16	
VCCPLL		4	4	4	4	4	
VCCIO	Bank 0	2	4	4	4	4	
	Bank 1	2	4	4	4	4	
	Bank 2	2	4	4	4	4	
	Bank 3	2	4	4	4	4	
	Bank 6	2	4	4	4	4	
	Bank 7	2	4	4	4	4	
	Bank 8	2	2	2	2	2	
VCCJ		1	1	1	1	1	
TAP		4	4	4	4	4	
GND, GNDIO		98	139	233	139	233	
NC		0	0	238	0	116	
Reserved*		2	2	2	2	2	
SERDES		26	52	78	52	104	
Miscellaneous Pins		8	8	8	8	8	
Total Bonded Pins		484	672	1156	672	1156	



Pin Information Summary			ECP3-95E/	4	ECP3-150EA		
Pin Type		484 fpBGA	672 fpBGA	1156 fpBGA	672 fpBGA	1156 fpBGA	
Emulated Differential I/O per	Bank 0	21	30	43	30	47	
Bank	Bank 1	18	24	39	24	43	
	Bank 2	8	12	13	12	18	
	Bank 3	20	23	33	23	37	
	Bank 6	22	25	33	25	37	
	Bank 7	11	16	18	16	24	
	Bank 8	12	12	12	12	12	
Highspeed Differential I/O per	Bank 0	0	0	0	0	0	
Bank	Bank 1	0	0	0	0	0	
	Bank 2	6	9	9	9	15	
	Bank 3	9	12	16	12	21	
	Bank 6	11	14	16	14	21	
	Bank 7	9	12	13	12	18	
	Bank 8	0	0	0	0	0	
Total Single Ended/ Total	Bank 0	42/21	60/30	86/43	60/30	94/47	
Differential I/O per Bank	Bank 1	36/18	48/24	78/39	48/24	86/43	
	Bank 2	28/14	42/21	44/22	42/21	66/33	
	Bank 3	58/29	71/35	98/49	71/35	116/58	
	Bank 6	67/33	78/39	98/49	78/39	116/58	
	Bank 7	40/20	56/28	62/31	56/28	84/42	
	Bank 8	24/12	24/12	24/12	24/12	24/12	
DDR Groups Bonded per Bank	Bank 0	3	5	7	5	7	
	Bank 1	3	4	7	4	7	
	Bank 2	2	3	3	3	4	
	Bank 3	3	4	5	4	7	
	Bank 6	4	4	5	4	7	
	Bank 7	3	4	4	4	6	
	Configuration Bank 8	0	0	0	0	0	
SERDES Quads			1	2	2	4	

<sup>\*</sup>Note: These pins must remain floating on the board.

#### 5.3.1. Package Pinout Information

Package pinout information can be found under "Data Sheets" on the LatticeECP3 product pages on the Lattice website at <a href="http://www.latticesemi.com/Products/FPGAandCPLD/LatticeECP3">http://www.latticesemi.com/Products/FPGAandCPLD/LatticeECP3</a> and in the Diamond or ispLEVER software tools. To create pinout information from within ispLEVER Design Planner, select Tools > Spreadsheet View. Then select Select File > Export and choose a type of output file. To create a pin information file from within Diamond select Tools > Spreadsheet View or Tools > Package View; then, select File > Export and choose a type of output file. See Diamond or ispLEVER Help for more information.



### 5.4. Thermal Management

Thermal management is recommended as part of any sound FPGA design methodology. To assess the thermal characteristics of a system, Lattice specifies a maximum allowable junction temperature in all device data sheets. Designers must complete a thermal analysis of their specific design to ensure that the device and package do not exceed the junction temperature limits. Refer to the Thermal Management document to find the device/package specific thermal values.

#### 5.4.1. For Further Information

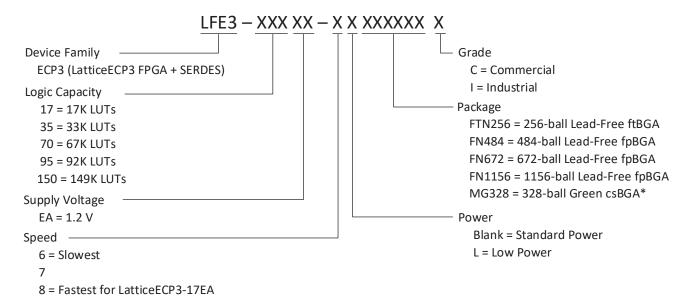
For further information regarding Thermal Management, refer to the following:

- Thermal Management document
- Power Consumption and Management for LatticeECP3 Devices (FPGA-TN-02189)
- Power Calculator tool included with the Diamond and ispLEVER design tools, or as a standalone download from www.latticesemi.com/software



## 6. Ordering Information

## 6.1. LatticeECP3 Part Number Description



<sup>\*</sup> Green = Halogen free and lead free.

## 6.2. Ordering Information

LatticeECP3 devices have top-side markings, for commercial and industrial grades, as shown below:



**Note**: See PCN 05A-12 for information regarding a change to the top-side mark logo.

#### 6.2.1. LatticeECP3 Devices, Green and Lead-Free Packaging

The following devices may have associated errata. Specific devices with associated errata will be notated with a footnote.

#### Commercial

Part Number	Voltage	Grade	Power	Package*	Pins	Temp.	LUTs (K)
LFE3-17EA-6FTN256C	1.2 V	-6	STD	Lead-Free ftBGA	256	СОМ	17
LFE3-17EA-7FTN256C	1.2 V	<b>-</b> 7	STD	Lead-Free ftBGA	256	СОМ	17
LFE3-17EA-8FTN256C	1.2 V	-8	STD	Lead-Free ftBGA	256	СОМ	17
LFE3-17EA-6LFTN256C	1.2 V	-6	LOW	Lead-Free ftBGA	256	СОМ	17
LFE3-17EA-7LFTN256C	1.2 V	-7	LOW	Lead-Free ftBGA	256	СОМ	17
LFE3-17EA-8LFTN256C	1.2 V	-8	LOW	Lead-Free ftBGA	256	СОМ	17
LFE3-17EA-6MG328C	1.2 V	-6	STD	Green csBGA	328	СОМ	17

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LFE3-17EA-7MG328C	1.2 V	-7	STD	Green csBGA	328	СОМ	17
LFE3-17EA-8MG328C	1.2 V	-8	STD	Green csBGA	328	СОМ	17
LFE3-17EA-6LMG328C	1.2 V	-6	LOW	Green csBGA	328	СОМ	17
LFE3-17EA-7LMG328C	1.2 V	<b>-</b> 7	LOW	Green csBGA	328	СОМ	17
LFE3-17EA-8LMG328C	1.2 V	-8	LOW	Green csBGA	328	СОМ	17
LFE3-17EA-6FN484C	1.2 V	-6	STD	Lead-Free fpBGA	484	СОМ	17
LFE3-17EA-7FN484C	1.2 V	<b>-</b> 7	STD	Lead-Free fpBGA	484	СОМ	17
LFE3-17EA-8FN484C	1.2 V	-8	STD	Lead-Free fpBGA	484	СОМ	17
LFE3-17EA-6LFN484C	1.2 V	-6	LOW	Lead-Free fpBGA	484	СОМ	17
LFE3-17EA-7LFN484C	1.2 V	<b>-</b> 7	LOW	Lead-Free fpBGA	484	СОМ	17
LFE3-17EA-8LFN484C	1.2 V	-8	LOW	Lead-Free fpBGA	484	СОМ	17

<sup>\*</sup>Note: Green = Halogen free and lead free.

Part Number	Voltage	Grade*	Power	Package	Pins	Temp.	LUTs (K)
LFE3-35EA-6FTN256C	1.2 V	-6	STD	Lead-Free ftBGA	256	СОМ	33
LFE3-35EA-7FTN256C	1.2 V	<b>-</b> 7	STD	Lead-Free ftBGA	256	СОМ	33
LFE3-35EA-8FTN256C	1.2 V	-8	STD	Lead-Free ftBGA	256	СОМ	33
LFE3-35EA-6LFTN256C	1.2 V	-6	LOW	Lead-Free ftBGA	256	СОМ	33
LFE3-35EA-7LFTN256C	1.2 V	<b>-</b> 7	LOW	Lead-Free ftBGA	256	СОМ	33
LFE3-35EA-8LFTN256C	1.2 V	-8	LOW	Lead-Free ftBGA	256	СОМ	33
LFE3-35EA-6FN484C	1.2 V	-6	STD	Lead-Free fpBGA	484	СОМ	33
LFE3-35EA-7FN484C	1.2 V	<b>-</b> 7	STD	Lead-Free fpBGA	484	СОМ	33
LFE3-35EA-8FN484C	1.2 V	-8	STD	Lead-Free fpBGA	484	СОМ	33
LFE3-35EA-6LFN484C	1.2 V	-6	LOW	Lead-Free fpBGA	484	СОМ	33
LFE3-35EA-7LFN484C	1.2 V	<b>-</b> 7	LOW	Lead-Free fpBGA	484	СОМ	33
LFE3-35EA-8LFN484C	1.2 V	-8	LOW	Lead-Free fpBGA	484	СОМ	33
LFE3-35EA-6FN672C	1.2 V	-6	STD	Lead-Free fpBGA	672	СОМ	33
LFE3-35EA-7FN672C	1.2 V	<b>-</b> 7	STD	Lead-Free fpBGA	672	СОМ	33
LFE3-35EA-8FN672C	1.2 V	-8	STD	Lead-Free fpBGA	672	СОМ	33
LFE3-35EA-6LFN672C	1.2 V	-6	LOW	Lead-Free fpBGA	672	СОМ	33
LFE3-35EA-7LFN672C	1.2 V	<b>-7</b>	LOW	Lead-Free fpBGA	672	СОМ	33
LFE3-35EA-8LFN672C	1.2 V	-8	LOW	Lead-Free fpBGA	672	СОМ	33

<sup>\*</sup>Note: For ordering information on –9 speed grade devices, contact your Lattice Sales Representative.



Part Number	Voltage	Grade*	Power	Package	Pins	Temp.	LUTs (K)
LFE3-70EA-6FN484C	1.2 V	-6	STD	Lead-Free fpBGA	484	СОМ	67
LFE3-70EA-7FN484C	1.2 V	<b>-</b> 7	STD	Lead-Free fpBGA	484	СОМ	67
LFE3-70EA-8FN484C	1.2 V	-8	STD	Lead-Free fpBGA	484	СОМ	67
LFE3-70EA-6LFN484C	1.2 V	-6	LOW	Lead-Free fpBGA	484	СОМ	67
LFE3-70EA-7LFN484C	1.2 V	<b>-</b> 7	LOW	Lead-Free fpBGA	484	СОМ	67
LFE3-70EA-8LFN484C	1.2 V	-8	LOW	Lead-Free fpBGA	484	СОМ	67
LFE3-70EA-6FN672C	1.2 V	-6	STD	Lead-Free fpBGA	672	СОМ	67
LFE3-70EA-7FN672C	1.2 V	<b>-</b> 7	STD	Lead-Free fpBGA	672	СОМ	67
LFE3-70EA-8FN672C	1.2 V	-8	STD	Lead-Free fpBGA	672	СОМ	67
LFE3-70EA-6LFN672C	1.2 V	-6	LOW	Lead-Free fpBGA	672	СОМ	67
LFE3-70EA-7LFN672C	1.2 V	<b>-7</b>	LOW	Lead-Free fpBGA	672	СОМ	67
LFE3-70EA-8LFN672C	1.2 V	-8	LOW	Lead-Free fpBGA	672	СОМ	67
LFE3-70EA-6FN1156C	1.2 V	-6	STD	Lead-Free fpBGA	1156	СОМ	67
LFE3-70EA-7FN1156C	1.2 V	<b>-7</b>	STD	Lead-Free fpBGA	1156	СОМ	67
LFE3-70EA-8FN1156C	1.2 V	-8	STD	Lead-Free fpBGA	1156	СОМ	67
LFE3-70EA-6LFN1156C	1.2 V	-6	LOW	Lead-Free fpBGA	1156	СОМ	67
LFE3-70EA-7LFN1156C	1.2 V	-7	LOW	Lead-Free fpBGA	1156	СОМ	67
LFE3-70EA-8LFN1156C	1.2 V	-8	LOW	Lead-Free fpBGA	1156	СОМ	67

<sup>\*</sup>Note: For ordering information on –9 speed grade devices, contact your Lattice Sales Representative.

Part Number	Voltage	Grade*	Power	Package	Pins	Temp.	LUTs (K)
LFE3-95EA-6FN484C	1.2 V	-6	STD	Lead-Free fpBGA	484	СОМ	92
LFE3-95EA-7FN484C	1.2 V	<b>-</b> 7	STD	Lead-Free fpBGA	484	СОМ	92
LFE3-95EA-8FN484C	1.2 V	-8	STD	Lead-Free fpBGA	484	СОМ	92
LFE3-95EA-6LFN484C	1.2 V	-6	LOW	Lead-Free fpBGA	484	СОМ	92
LFE3-95EA-7LFN484C	1.2 V	<b>-</b> 7	LOW	Lead-Free fpBGA	484	СОМ	92
LFE3-95EA-8LFN484C	1.2 V	-8	LOW	Lead-Free fpBGA	484	СОМ	92
LFE3-95EA-6FN672C	1.2 V	-6	STD	Lead-Free fpBGA	672	СОМ	92
LFE3-95EA-7FN672C	1.2 V	<b>-</b> 7	STD	Lead-Free fpBGA	672	СОМ	92
LFE3-95EA-8FN672C	1.2 V	-8	STD	Lead-Free fpBGA	672	СОМ	92
LFE3-95EA-6LFN672C	1.2 V	-6	LOW	Lead-Free fpBGA	672	СОМ	92
LFE3-95EA-7LFN672C	1.2 V	<b>-</b> 7	LOW	Lead-Free fpBGA	672	СОМ	92
LFE3-95EA-8LFN672C	1.2 V	-8	LOW	Lead-Free fpBGA	672	СОМ	92
LFE3-95EA-6FN1156C	1.2 V	-6	STD	Lead-Free fpBGA	1156	СОМ	92
LFE3-95EA-7FN1156C	1.2 V	<b>-</b> 7	STD	Lead-Free fpBGA	1156	СОМ	92
LFE3-95EA-8FN1156C	1.2 V	-8	STD	Lead-Free fpBGA	1156	СОМ	92
LFE3-95EA-6LFN1156C	1.2 V	-6	LOW	Lead-Free fpBGA	1156	СОМ	92
LFE3-95EA-7LFN1156C	1.2 V	<b>-</b> 7	LOW	Lead-Free fpBGA	1156	СОМ	92
LFE3-95EA-8LFN1156C	1.2 V	-8	LOW	Lead-Free fpBGA	1156	СОМ	92

<sup>\*</sup>Note: For ordering information on –9 speed grade devices, contact your Lattice Sales Representative.



Part Number	Voltage	Grade*	Power	Package	Pins	Temp.	LUTs (K)
LFE3-150EA-6FN672C	1.2 V	-6	STD	Lead-Free fpBGA	672	СОМ	149
LFE3-150EA-7FN672C	1.2 V	<b>-</b> 7	STD	Lead-Free fpBGA	672	СОМ	149
LFE3-150EA-8FN672C	1.2 V	-8	STD	Lead-Free fpBGA	672	СОМ	149
LFE3-150EA-6LFN672C	1.2 V	-6	LOW	Lead-Free fpBGA	672	СОМ	149
LFE3-150EA-7LFN672C	1.2 V	-7	LOW	Lead-Free fpBGA	672	СОМ	149
LFE3-150EA-8LFN672C	1.2 V	-8	LOW	Lead-Free fpBGA	672	СОМ	149
LFE3-150EA-6FN1156C	1.2 V	-6	STD	Lead-Free fpBGA	1156	СОМ	149
LFE3-150EA-7FN1156C	1.2 V	-7	STD	Lead-Free fpBGA	1156	СОМ	149
LFE3-150EA-8FN1156C	1.2 V	-8	STD	Lead-Free fpBGA	1156	СОМ	149
LFE3-150EA-6LFN1156C	1.2 V	-6	LOW	Lead-Free fpBGA	1156	СОМ	149
LFE3-150EA-7LFN1156C	1.2 V	-7	LOW	Lead-Free fpBGA	1156	СОМ	149
LFE3-150EA-8LFN1156C	1.2 V	-8	LOW	Lead-Free fpBGA	1156	СОМ	149

<sup>\*</sup>Note: For ordering information on -9 speed grade devices, contact your Lattice Sales Representative.

Part Number	Voltage	Grade	Power	Package	Pins	Temp.	LUTs (K)
LFE3-150EA-6FN672CTW <sup>1</sup>	1.2 V	-6	STD	Lead-Free fpBGA	672	СОМ	149
LFE3-150EA-7FN672CTW <sup>1</sup>	1.2 V	<b>-</b> 7	STD	Lead-Free fpBGA	672	СОМ	149
LFE3-150EA-8FN672CTW <sup>1</sup>	1.2 V	-8	STD	Lead-Free fpBGA	672	СОМ	149
LFE3-150EA-6FN1156CTW <sup>1</sup>	1.2 V	-6	STD	Lead-Free fpBGA	1156	СОМ	149
LFE3-150EA-7FN1156CTW <sup>1</sup>	1.2 V	<b>-</b> 7	STD	Lead-Free fpBGA	1156	СОМ	149
LFE3-150EA-8FN1156CTW <sup>1</sup>	1.2 V	-8	STD	Lead-Free fpBGA	1156	СОМ	149

#### Note:

Specifications for the LFE3-150EA-spFNpkgCTW and LFE3-150EA-spFNpkgITW devices, (where sp is the speed and pkg is the package), are the same as the LFE3-150EA-spFNpkgC and LFE3-150EA-spFNpkgI devices respectively, except as specified below.

- The CTC (Clock Tolerance Circuit) inside the SERDES hard PCS in the TW device is not functional but it can be bypassed and implemented in soft IP.
- The SERDES XRES pin on the TW device passes CDM testing at 250 V.



#### Industrial

The following devices may have associated errata. Specific devices with associated errata will be notated with a footnote.

Part Number	Voltage	Grade	Power	Package*	Pins	Temp.	LUTs (K)
LFE3-17EA-6FTN256I	1.2 V	-6	STD	Lead-Free ftBGA	256	IND	17
LFE3-17EA-7FTN256I	1.2 V	-7	STD	Lead-Free ftBGA	256	IND	17
LFE3-17EA-8FTN256I	1.2 V	-8	STD	Lead-Free ftBGA	256	IND	17
LFE3-17EA-6LFTN256I	1.2 V	-6	LOW	Lead-Free ftBGA	256	IND	17
LFE3-17EA-7LFTN256I	1.2 V	-7	LOW	Lead-Free ftBGA	256	IND	17
LFE3-17EA-8LFTN256I	1.2 V	-8	LOW	Lead-Free ftBGA	256	IND	17
LFE3-17EA-6MG328I	1.2 V	-6	STD	Lead-Free csBGA	328	IND	17
LFE3-17EA-7MG328I	1.2 V	<b>-</b> 7	STD	Lead-Free csBGA	328	IND	17
LFE3-17EA-8MG328I	1.2 V	-8	STD	Lead-Free csBGA	328	IND	17
LFE3-17EA-6LMG328I	1.2 V	-6	LOW	Green csBGA	328	IND	17
LFE3-17EA-7LMG328I	1.2 V	<b>-</b> 7	LOW	Green csBGA	328	IND	17
LFE3-17EA-8LMG328I	1.2 V	-8	LOW	Green csBGA	328	IND	17
LFE3-17EA-6FN484I	1.2 V	-6	STD	Lead-Free fpBGA	484	IND	17
LFE3-17EA-7FN484I	1.2 V	<b>-</b> 7	STD	Lead-Free fpBGA	484	IND	17
LFE3-17EA-8FN484I	1.2 V	-8	STD	Lead-Free fpBGA	484	IND	17
LFE3-17EA-6LFN484I	1.2 V	-6	LOW	Lead-Free fpBGA	484	IND	17
LFE3-17EA-7LFN484I	1.2 V	-7	LOW	Lead-Free fpBGA	484	IND	17
LFE3-17EA-8LFN484I	1.2 V	-8	LOW	Lead-Free fpBGA	484	IND	17

<sup>\*</sup>Note: Green = Halogen free and lead free.

Part Number	Voltage	Grade1	Power	Package	Pins	Temp.	LUTs (K)
LFE3-35EA-6FTN256I	1.2 V	-6	STD	Lead-Free ftBGA	256	IND	33
LFE3-35EA-7FTN256I	1.2 V	<b>-</b> 7	STD	Lead-Free ftBGA	256	IND	33
LFE3-35EA-8FTN256I	1.2 V	-8	STD	Lead-Free ftBGA	256	IND	33
LFE3-35EA-6LFTN256I	1.2 V	-6	LOW	Lead-Free ftBGA	256	IND	33
LFE3-35EA-7LFTN256I	1.2 V	<b>-</b> 7	LOW	Lead-Free ftBGA	256	IND	33
LFE3-35EA-8LFTN256I	1.2 V	-8	LOW	Lead-Free ftBGA	256	IND	33
LFE3-35EA-6FN484I	1.2 V	-6	STD	Lead-Free fpBGA	484	IND	33
LFE3-35EA-7FN484I	1.2 V	<b>-</b> 7	STD	Lead-Free fpBGA	484	IND	33
LFE3-35EA-8FN484I	1.2 V	-8	STD	Lead-Free fpBGA	484	IND	33
LFE3-35EA-6LFN484I	1.2 V	-6	LOW	Lead-Free fpBGA	484	IND	33
LFE3-35EA-7LFN484I	1.2 V	<b>-</b> 7	LOW	Lead-Free fpBGA	484	IND	33
LFE3-35EA-8LFN484I	1.2 V	-8	LOW	Lead-Free fpBGA	484	IND	33
LFE3-35EA-6FN672I	1.2 V	-6	STD	Lead-Free fpBGA	672	IND	33
LFE3-35EA-7FN672I	1.2 V	<b>-</b> 7	STD	Lead-Free fpBGA	672	IND	33
LFE3-35EA-8FN672I	1.2 V	-8	STD	Lead-Free fpBGA	672	IND	33
LFE3-35EA-6LFN672I	1.2 V	-6	LOW	Lead-Free fpBGA	672	IND	33
LFE3-35EA-7LFN672I	1.2 V	<b>-</b> 7	LOW	Lead-Free fpBGA	672	IND	33
LFE3-35EA-8LFN672I	1.2 V	-8	LOW	Lead-Free fpBGA	672	IND	33

**Note:** For ordering information on -9 speed grade devices, ontact your Lattice Sales Representative.



Part Number	Voltage	Grade*	Power	Package	Pins	Temp.	LUTs (K)
LFE3-70EA-6FN484I	1.2 V	-6	STD	Lead-Free fpBGA	484	IND	67
LFE3-70EA-7FN484I	1.2 V	-7	STD	Lead-Free fpBGA	484	IND	67
LFE3-70EA-8FN484I	1.2 V	-8	STD	Lead-Free fpBGA	484	IND	67
LFE3-70EA-6LFN484I	1.2 V	-6	LOW	Lead-Free fpBGA	484	IND	67
LFE3-70EA-7LFN484I	1.2 V	<b>-7</b>	LOW	Lead-Free fpBGA	484	IND	67
LFE3-70EA-8LFN484I	1.2 V	-8	LOW	Lead-Free fpBGA	484	IND	67
LFE3-70EA-6FN672I	1.2 V	-6	STD	Lead-Free fpBGA	672	IND	67
LFE3-70EA-7FN672I	1.2 V	<b>-7</b>	STD	Lead-Free fpBGA	672	IND	67
LFE3-70EA-8FN672I	1.2 V	-8	STD	Lead-Free fpBGA	672	IND	67
LFE3-70EA-6LFN672I	1.2 V	-6	LOW	Lead-Free fpBGA	672	IND	67
LFE3-70EA-7LFN672I	1.2 V	<b>-7</b>	LOW	Lead-Free fpBGA	672	IND	67
LFE3-70EA-8LFN672I	1.2 V	-8	LOW	Lead-Free fpBGA	672	IND	67
LFE3-70EA-6FN1156I	1.2 V	-6	STD	Lead-Free fpBGA	1156	IND	67
LFE3-70EA-7FN1156I	1.2 V	<b>-7</b>	STD	Lead-Free fpBGA	1156	IND	67
LFE3-70EA-8FN1156I	1.2 V	-8	STD	Lead-Free fpBGA	1156	IND	67
LFE3-70EA-6LFN1156I	1.2 V	-6	LOW	Lead-Free fpBGA	1156	IND	67
LFE3-70EA-7LFN1156I	1.2 V	-7	LOW	Lead-Free fpBGA	1156	IND	67
LFE3-70EA-8LFN1156I	1.2 V	-8	LOW	Lead-Free fpBGA	1156	IND	67

Note: For ordering information on –9 speed grade devices, contact your Lattice Sales Representative.

Part Number	Voltage	Grade*	Power	Package	Pins	Temp.	LUTs (K)
LFE3-95EA-6FN484I	1.2 V	-6	STD	Lead-Free fpBGA	484	IND	92
LFE3-95EA-7FN484I	1.2 V	<b>-</b> 7	STD	Lead-Free fpBGA	484	IND	92
LFE3-95EA-8FN484I	1.2 V	-8	STD	Lead-Free fpBGA	484	IND	92
LFE3-95EA-6LFN484I	1.2 V	-6	LOW	Lead-Free fpBGA	484	IND	92
LFE3-95EA-7LFN484I	1.2 V	<b>-</b> 7	LOW	Lead-Free fpBGA	484	IND	92
LFE3-95EA-8LFN484I	1.2 V	-8	LOW	Lead-Free fpBGA	484	IND	92
LFE3-95EA-6FN672I	1.2 V	-6	STD	Lead-Free fpBGA	672	IND	92
LFE3-95EA-7FN672I	1.2 V	<b>-</b> 7	STD	Lead-Free fpBGA	672	IND	92
LFE3-95EA-8FN672I	1.2 V	-8	STD	Lead-Free fpBGA	672	IND	92
LFE3-95EA-6LFN672I	1.2 V	-6	LOW	Lead-Free fpBGA	672	IND	92
LFE3-95EA-7LFN672I	1.2 V	<b>-</b> 7	LOW	Lead-Free fpBGA	672	IND	92
LFE3-95EA-8LFN672I	1.2 V	-8	LOW	Lead-Free fpBGA	672	IND	92
LFE3-95EA-6FN1156I	1.2 V	-6	STD	Lead-Free fpBGA	1156	IND	92
LFE3-95EA-7FN1156I	1.2 V	<b>-</b> 7	STD	Lead-Free fpBGA	1156	IND	92
LFE3-95EA-8FN1156I	1.2 V	-8	STD	Lead-Free fpBGA	1156	IND	92
LFE3-95EA-6LFN1156I	1.2 V	-6	LOW	Lead-Free fpBGA	1156	IND	92
LFE3-95EA-7LFN1156I	1.2 V	-7	LOW	Lead-Free fpBGA	1156	IND	92
LFE3-95EA-8LFN1156I	1.2 V	-8	LOW	Lead-Free fpBGA	1156	IND	92

<sup>\*</sup>Note: For ordering information on –9 speed grade devices, contact your Lattice Sales Representative.



Part Number	Voltage	Grade*	Power	Package	Pins	Temp.	LUTs (K)
LFE3-150EA-6FN672I	1.2 V	-6	STD	Lead-Free fpBGA	672	IND	149
LFE3-150EA-7FN672I	1.2 V	<b>-</b> 7	STD	Lead-Free fpBGA	672	IND	149
LFE3-150EA-8FN672I	1.2 V	-8	STD	Lead-Free fpBGA	672	IND	149
LFE3-150EA-6LFN672I	1.2 V	-6	LOW	Lead-Free fpBGA	672	IND	149
LFE3-150EA-7LFN672I	1.2 V	<b>-</b> 7	LOW	Lead-Free fpBGA	672	IND	149
LFE3-150EA-8LFN672I	1.2 V	-8	LOW	Lead-Free fpBGA	672	IND	149
LFE3-150EA-6FN1156I	1.2 V	-6	STD	Lead-Free fpBGA	1156	IND	149
LFE3-150EA-7FN1156I	1.2 V	<b>-</b> 7	STD	Lead-Free fpBGA	1156	IND	149
LFE3-150EA-8FN1156I	1.2 V	-8	STD	Lead-Free fpBGA	1156	IND	149
LFE3-150EA-6LFN1156I	1.2 V	-6	LOW	Lead-Free fpBGA	1156	IND	149
LFE3-150EA-7LFN1156I	1.2 V	<b>-</b> 7	LOW	Lead-Free fpBGA	1156	IND	149
LFE3-150EA-8LFN1156I	1.2 V	-8	LOW	Lead-Free fpBGA	1156	IND	149

<sup>\*</sup>Note: For ordering information on –9 speed grade devices, contact your Lattice Sales Representative.

Part Number	Voltage	Grade	Power	Package	Pins	Temp.	LUTs (K)
LFE3-150EA-6FN672ITW*	1.2 V	-6	STD	Lead-Free fpBGA	672	IND	149
LFE3-150EA-7FN672ITW*	1.2 V	<b>-</b> 7	STD	Lead-Free fpBGA	672	IND	149
LFE3-150EA-8FN672ITW*	1.2 V	-8	STD	Lead-Free fpBGA	672	IND	149
LFE3-150EA-6FN1156ITW*	1.2 V	-6	STD	Lead-Free fpBGA	1156	IND	149
LFE3-150EA-7FN1156ITW*	1.2 V	<b>-</b> 7	STD	Lead-Free fpBGA	1156	IND	149
LFE3-150EA-8FN1156ITW*	1.2 V	-8	STD	Lead-Free fpBGA	1156	IND	149

<sup>\*</sup>Note: Specifications for the LFE3-150EA-spFNpkgCTW and LFE3-150EA-spFNpkgITW devices, (where sp is the speed and pkg is the package), are the same as the LFE3-150EA-spFNpkgC and LFE3-150EA-spFNpkgI devices respectively, except as specified below.

- The CTC (Clock Tolerance Circuit) inside the SERDES hard PCS in the TW device is not functional but it can be bypassed and implemented in soft IP.
- The SERDES XRES pin on the TW device passes CDM testing at 250 V.



## 7. Supplemental Information

#### 7.1. For Further Information

A variety of technical notes for the LatticeECP3 family are available on the Lattice website at www.latticesemi.com.

- LatticeECP3 sysCONFIG Usage Guide (FPGA-TN-02192)
- LatticeECP3 SERDES/PCS Usage Guide (FPGA-TN-02190)
- LatticeECP3 sysIO Usage Guide (FPGA-TN-02194)
- LatticeECP3 sysCLOCK PLL/DLL Design and Usage Guide (FPGA-TN-02191)
- LatticeECP3 Memory Usage Guide (FPGA-TN-02188)
- LatticeECP3 High-Speed I/O Interface (FPGA-TN-02184)
- Power Consumption and Management for LatticeECP3 Devices (FPGA-TN-02189)
- LatticeECP3 sysDSP Usage Guide (FPGA-TN-02193)
- LatticeECP3 Soft Error Detection (SED) Usage Guide (FPGA-TN-02207)
- LatticeECP3 Hardware Checklist (FPGA-TN-02183)
- Advanced Security Encryption Key Programming Guide for ECP Device Family (FPGA-TN-02202)
- LatticeECP3, LatticeECP2/M, ECP5 and ECP5-5G Dual Boot and Multiple Boot Feature (FPGA-TN-02203)
- LatticeECP3 Slave SPI Port User's Guide (FPGA-TN-02136)

For further information on interface standards refer to the following websites:

- JEDEC Standards (LVTTL, LVCMOS, SSTL, HSTL): www.jedec.org
- PCI: www.pcisig.com



## **Technical Support Assistance**

Submit a technical support case through www.latticesemi.com/techsupport. For frequently asked questions, refer to the Lattice Answer Database at www.latticesemi.com/Support/AnswerDatabase.



## **Revision History**

#### Revision 3.2, June 2023

Section	Change Summary
All	Changed document title from 'Lattice ECP3 Family Data Sheet' to 'LatticeECP3 Family Data Sheet'.
Disclaimers	Updated the disclaimers.
Architecture	Corrected typo error from ECP2 to ECP3 in the LatticeECP3 sysDSP Slice Architecture Features section.
Technical Support Assistance	Added reference link to the Lattice Answer Database.

#### Revision 3.1, July 2021

Section	Change Summary
DC and Switching Characteristics	Updated footnote for "blank" pattern in section LatticeECP3 Supply Current (Standby).
Supplemental Information	Updated link for Advanced Security Encryption Key Programming Guide for ECP Device Family (FPGA-TN-02202).

#### Revision 3.0, June 2021

Section	Change Summary
sysI/O Single-Ended DC Electrical Characteristics	Added footnote 2 in section 4.10 sysI/O Single-Ended DC Electrical Characteristics.
Supplemental Information	Updated hyperlink for LatticeECP3, LatticeECP2/M, ECP5 and ECP5-5G Dual Boot and Multiple Boot Feature (FPGA-TN-02203) in For Further Information section.

#### Revision 2.9EA. January 2020

Section	Change Summary
All	Changed document number from DS1021 to FPGA-DS-02074.
	Updated document template.
Disclaimers	Added this section.
DC and Switching Characteristics	<ul> <li>Updated Recommended Operating Conditions1 section. Added footnote to V<sub>REF1</sub>.</li> <li>Updated LatticeECP3 External Switching Characteristics <sup>1, 2, 3, 13</sup> section. Corrected header to Generic DDRX2 Output with Clock and Data (&lt;10 Bits Wide) Centered at Pin Using DQSDLL (GDDRX2_TX.DQSDLL. Centered).</li> </ul>
All	<ul> <li>Updated references to:</li> <li>FPGA-TN-02136, LatticeECP3 SPI Slave Port User Guide (previously TN1222)</li> <li>FPGA-TN-02192, LatticeECP3 sysCONFIG Usage Guide (previously TN1169)</li> </ul>

#### Revision 2.8EA, March 2015

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Section	Change Summary	
Pinout Information	Updated Package Pinout Information section. Changed reference to http://www.latticesemi.com/Products/FPGAandCPLD/LatticeECP3.	
All	Minor style/formatting changes.	

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#### Revision 2.7EA, April 2014

Section	Change Summary
DC and Switching Characteristics	Updated LatticeECP3 Supply Current (Standby) table power numbers. Removed speed grade -9 timing numbers in the following sections:
	Typical Building Block Function Performance
	LatticeECP3 External Switching Characteristics
	LatticeECP3 Internal Switching Characteristics
	LatticeECP3 Family Timing Adders
Ordering Information	Removed ordering information for -9 speed grade devices.

### Revision 2.6EA, March 2014

Section	Change Summary
DC and Switching Characteristics	Added information to the sysI/O Single-Ended DC Electrical Characteristics section footnote.

#### Revision 2.5EA, February 2014

Section	Change Summary
DC and Switching Characteristics	Updated Hot Socketing Specifications table. Changed I <sub>Pw</sub> to I <sub>PD</sub> in footnote 3.
	Updated the following figures:
	Figure 4.25, sysCONFIG Port Timing
	Figure 4.27, Wake-Up Timing
Supplemental Information	Added technical note references.

#### Revision 2.4EA, September 2013

Section	Change Summary
DC and Switching Characteristics	Updated the Wake-Up Timing Diagram Added the following figures:  • Master SPI POR Waveforms  • SPI Configuration Waveforms  • Slave SPI HOLDN Waveforms  Added tIODISS and tIOENSS parameters in LatticeECP3 sysCONFIG Port Timing Specifications
	table.



#### Revision 2.3EA, June 2013

Section	Change Summary
Architecture	<ul> <li>sysl/O Buffer Banks text section – Updated description of "Top (Bank 0 and Bank 1) and Bottom syslO Buffer Pairs (Single-Ended Outputs Only)" for hot socketing information.</li> <li>sysl/O Buffer Banks text section – Updated description of "Configuration Bank sysl/O Buffer Pairs (Single-Ended Outputs, Only on Shared Pins When Not Used by Configuration)" for PCI clamp information.</li> <li>On-Chip Oscillator section – clarified the speed of the internal CMOS oscillator (130 MHz +/- 15%).</li> <li>Architecture Overview section – Added information on the state of the register on power up and after configuration.</li> </ul>
DC and Switching Characteristics	<ul> <li>sysI/O Recommended Operating Conditions table – Removed reference to footnote 1 from RSDS standard.</li> <li>sysI/O Single-Ended DC Electrical Characteristics table – Modified footnote 1.</li> <li>Added Oscillator Output Frequency table.</li> <li>LatticeECP3 sysCONFIG Port Timing Specifications table – Updated min. column for t<sub>CODO</sub> parameter.</li> <li>LatticeECP3 Family Timing Adders table – Description column, references to VCCIO = 3.0V changed to 3.3V. For PPLVDS, description changed from emulated to True LVDS and VCCIO = 2.5V changed to VCCIO = 2.5V or 3.3V.</li> <li>LatticeECP3 Maximum I/O Buffer Speed table – Description column, references to VCCIO = 3.0V changed to 3.3 V.</li> <li>Updated SERDES External Reference Clock Waveforms.</li> <li>Transmitter and Receiver Latency Block Diagram – Updated sections of the diagram to match descriptions on the SERDES/PCS Latency Breakdown table.</li> </ul>
Pinout Information	<ul> <li>"Logic Signal Connections" section heading renamed "Package Pinout Information".</li> <li>Software menu selections within this section have been updated.</li> <li>Signal Descriptions table – Updated description for V<sub>CCA</sub> signal.</li> </ul>

#### Revision 2.2EA, April 2012

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Section	Change Summary
Architecture	<ul> <li>Updated first paragraph of Output Register Block section.</li> <li>Updated the information about syslO buffer pairs below Figure 3.7.</li> <li>Updated the information relating to migration between devices in the Density Shifting section.</li> </ul>
DC and Switching Characteristics	Corrected the Definitions in the sysCLOCK PLL Timing table for t <sub>RST</sub> .
Ordering Information	Updated topside marks with new logos in the Ordering Information section.

#### Revision 2.1EA, February 2012

Section	Change Summary
All	Updated document with new corporate logo.

#### Revision 2.0EA, November 2011

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Section	Change Summary
Introduction	Added information for LatticeECP3-17EA, 328-ball csBGA package.
Architecture	Added information for LatticeECP3-17EA, 328-ball csBGA package.
DC and Switching Characteristics	<ul> <li>Updated LatticeECP3 Supply Current table power numbers.</li> <li>Typical Building Block Function Performance table, LatticeECP3 External Switching Characteristics table, LatticeECP3 Internal Switching Characteristics table and LatticeECP3 Family Timing Adders: Added speed grade -9 and updated speed grade -8, -7 and -6 timing numbers.</li> </ul>
Pinout Information	Added information for LatticeECP3-17EA, 328-ball csBGA package.
Ordering Information	<ul> <li>Added information for LatticeECP3-17EA, 328-ball csBGA package.</li> <li>Added ordering information for low power devices and -9 speed grade devices.</li> </ul>

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#### Revision 1.9EA, July 2011

Section	Change Summary
DC and Switching Characteristics	<ul> <li>Removed ESD Performance table and added reference to LatticeECP3 Product Family Qualification Summary document.</li> <li>sysCLOCK PLL TIming table, added footnote 4.</li> <li>External Reference Clock Specification table – removed reference to VREF-CM-AC and removed footnote for VREF-CM-AC.</li> </ul>
Pinout Information	Pin Information Summary table: Corrected VCCIO Bank8 data for LatticeECP3-17EA 256-ball ftBGA package and LatticeECP-35EA 256-ball ftBGA package.

#### Revision 1.8EA, April 2011

Section	Change Summary
Architecture	Updated Secondary Clock/Control Sources text section.
DC and Switching Characteristics	<ul> <li>Added data for 150 Mbps to SERDES Power Supply Requirements table.</li> <li>Updated Frequencies in Table 4.6 Serial Output Timing and Levels</li> <li>Added Data for 150 Mbps to Table 4.7 Channel Output Jitter</li> <li>Corrected External Switching Characteristics table, Description for DDR3 Clock Timing, t<sub>JIT</sub>.</li> <li>Corrected Internal Switching Characteristics table, Description for EBR Timing, t<sub>SUWREN_EBR</sub> and t<sub>HWREN_EBR</sub>.</li> <li>Added footnote 1 to sysConfig Port Timing Specifications table.</li> <li>Updated description for RX-CIDs to 150M in Table 4.9 Serial Input Data Specifications</li> <li>Updated Frequency to 150 Mbps in Table 4.11 Periodic Receiver Jitter Tolerance Specification</li> </ul>



#### Revision 1.7EA, December 2010

Section	Change Summary
Multiple	<ul> <li>Data sheet made final. Removed "preliminary" headings.</li> <li>Removed data for 70E and 95E devices. A separate data sheet is available for these specific devices.</li> <li>Updated for Lattice Diamond design software.</li> </ul>
Introduction	<del>                                     </del>
Introduction Architecture  DC and Switching Characteristics	<ul> <li>Updated for Lattice Diamond design software.</li> <li>Corrected number of user I/O</li> <li>Corrected the package type in Table 3.4 Available SERDES Quad per LatticeECP3 Devices.</li> <li>Updated description of General Purpose PLL</li> <li>Added additional information in the Flexible Quad SERDES Architecture section.</li> <li>Added footnotes and corrected the information in Table 3.6 Selectable master Clock (MCCLK) Frequencies During Configuration (Nominal).</li> <li>Updated Figure 3.6, Per Region Secondary Clock Selection.</li> <li>Updated description for On-Chip Programmable Termination.</li> <li>Added information about number of rows of DSP slices.</li> <li>Updated footnote 2 for Table 3.2, On-Chip Termination Options for Input Modes.</li> <li>Updated information for syslO buffer pairs.</li> <li>Corrected minimum number of General Purpose PLLs (was 4, now 2).</li> <li>Regenerated sysCONFIG Port Timing figure.</li> <li>Added two (clock pulse width) in External Switching Characteristics table.</li> <li>Corrected units, revised and added data, and corrected footnote 1 in External Switching Characteristics table.</li> <li>Added Jitter Transfer figures in SERDES External Reference Clock section.</li> <li>Corrected capacitance information in the DC Electrical Characteristics table.</li> <li>Corrected GDDR Parameter name HOGDDR.</li> <li>Corrected GDDR Parameter name HOGDDR.</li> <li>Corrected GDDR Parameter name HOGDDR.</li> <li>Corrected titles for Figures 4.7 (DDR/DDR2/DDR3 Parameters) and 4.8 (Generic DDR/DDRX2 Parameters).</li> <li>Updated titles for Figures 4.5 (MLVDS25 (Multipoint Low Voltage Differential Signaling)) and 4.6 (Generic DDRX1/DDRX2 (With Clock and Data Edges Aligned)).</li> <li>Updated Supply Current table.</li> <li>Added footnote to syslO Recommended Operating Conditions table.</li> <li>Added footnote to SoslO Recommended Operating Conditions table.</li> <li>Added footnote to SoslO</li></ul>
D: 6	pins".
Pinout Information	Updated description for VTTx.

#### Revision 1.6EA, March 2010

Section	Change Summary
Architecture	Added Read-Before-Write information.
DC and Switching Characteristics	Added footnote #6 to Maximum I/O Buffer Speed table.
	Corrected minimum operating conditions for input and output differential voltages in the Point-to-Point LVDS table.
Pinout Information	Added pin information for the LatticeECP3-70EA and LatticeECP3-95EA devices.
Ordering Information	<ul> <li>Added ordering part numbers for the LatticeECP3-70EA and LatticeECP3-95EA devices.</li> <li>Removed dual mark information.</li> </ul>

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#### Revision 1.5EA, November 2009

Section	Change Summary
Introduction	Updated Embedded SERDES features.
	Added SONET/SDH to Embedded SERDES protocols.
Architecture	Updated Figure 2.4, General Purpose PLL Diagram.
	Updated SONET/SDH to SERDES and PCS protocols.
	Updated Table 3.3, SERDES Standard Support to include SONET/ SDH and updated
	footnote 2.
DC and Switching Characterisitcs	Added footnote to ESD Performance table.
	Updated SERDES Power Supply Requirements table and footnotes.
	Updated Maximum I/O Buffer Speed table.
	Updated Pin-to-Pin Peformance table.
	Updated sysCLOCK PLL Timing table.
	Updated DLL timing table.
	Updated High-Speed Data Transmitter tables.
	Updated High-Speed Data Receiver table.
	Updated footnote for Receiver Total Jitter Tolerance Specification table.
	Updated Periodic Receiver Jitter Tolerance Specification table.
	Updated SERDES External Reference Clock Specification table.
	Updated PCI Express Electrical and Timing AC and DC Characteristics.
	Deleted Reference Clock table for PCI Express Electrical and Timing AC and DC Characteristics.
	Updated SMPTE AC/DC Characteristics Transmit table.
	Updated Mini LVDS table.
	Updated RSDS table.
	Added Supply Current (Standby) table for EA devices.
	Updated Internal Switching Characteristics table.
	<ul> <li>Updated Register-to-Register Performance table.</li> </ul>
	Added HDMI Electrical and Timing Characteristics data.
	Updated Family Timing Adders table.
	Updated sysCONFIG Port Timing Specifications table.
	Updated Recommended Operating Conditions table.
	Updated Hot Socket Specifications table.
	Updated Single-Ended DC table.
	Updated TRLVDS table and figure.
	Updated Serial Data Input Specifications table.
	Updated HDMI Transmit and Receive table.
Ordering Information	Added LFE3-150EA "TW" devices and footnotes to the Commercial and Industrial tables.



#### Revision 1.4EA, September 2009

Section	Change Summary
Architecture	Corrected link in sysMEM Memory Block section.
	Updated information for On-Chip Programmable Termination and modified corresponding figure.
	Added footnote 2 to On-Chip Programmable Termination Options for Input Modes table.
	Corrected Per Quadrant Primary Clock Selection figure.
DC and Switching Characteristics	<ul> <li>Modified -8 Timing data for 1024x18 True-Dual Port RAM (Read-Before-Write, EBR Output Registers)</li> <li>Added ESD Performance table.</li> <li>LatticeECP3 External Switching Characteristics table - updated data for t<sub>DIBGDDR</sub>, t<sub>W_PRI</sub>, t<sub>W_EDGE</sub> and t<sub>SKEW_EDGE_DQS</sub>.</li> <li>LatticeECP3 Internal Switching Characteristics table - updated data for t<sub>COO_PIO</sub> and added footnote #4.</li> </ul>
	<ul> <li>sysCLOCK PLL Timing table - updated data for f<sub>OUT</sub>.</li> <li>External Reference Clock Specification (refclkp/refclkn) table – updated data for V<sub>REF-IN-SE</sub> and V<sub>REF-IN-DIFF</sub>.</li> <li>LatticeECP3 sysCONFIG Port Timing Specifications table – updated data for t<sub>MWC</sub>.</li> <li>Added TRLVDS DC Specification table and diagram.</li> <li>Updated Mini LVDS table.</li> </ul>

#### Revision 1.3EA, August 2009

Section	Change Summary
DC and Switching Characteristics	Corrected truncated numbers for $V_{\text{CCIB}}$ and $V_{\text{CCOB}}$ in Recommended Operating Conditions table.

#### Revision 1.2EA, July 2009

Section	Change Summary
Multiple	Changed references of "multi-boot" to "dual-boot" throughout the data sheet.
Architecture	Updated On-Chip Programmable Termination bullets.
	Updated On-Chip Termination Options for Input Modes table.
	Updated On-Chip Termination figure.
DC and Switching Characteristics	Changed min/max data for FREF_PPM and added footnote 4 in SERDES External Reference Clock Specification table.
	Updated SERDES minimum frequency.
Pinout Information	Corrected MCLK to be I/O and CCLK to be I in Signal Descriptions table



#### Revision 1.1EA, May 2009

added a footnote to indicate 230 Mbps applies to 8b10t  Updated data for ECP3-17 in LatticeECP3 Family Selectic Changed embedded memory from 552 to 700 Kbits in Lic Guide table.  Pupated description for CLKFB in General Purpose PLL D Corrected Primary Clock Sources text section. Corrected Secondary Clock/Control Sources text section Corrected Secondary Clock Regions table. Corrected Inote below Detailed sysDSP Slice Diagram. Corrected Clock, Clock Enable, and Reset Resources text Corrected Clock, Clock Enable, and Reset Resources text Corrected ECP3-17 EBR number in Embedded SRAM in the Added On-Chip Termination Options for Input Modes ta Updated Available SERDES Quads per LatticeECP3 Device Updated Available SERDES Quads per LatticeECP3 Device Updated Simplified Channel Block Diagram for SERDES/f Updated Device Configuration text section. Corrected software default value of MCCLK to be 2.5 Ml DC and Switching Characteristics  Updated VCCOB Min/Max data in Recommended Operating Added added footnote 7 for ts <sub>SEW, PBB</sub> to External Switchia Added 2-to-1 Gearing text section and table. Updated External Reference Clock Specifications - updated Sex Description of the Section of	ange Summary
added a footnote to indicate 230 Mbps applies to 8b10t  Updated data for ECP3-17 in LatticeECP3 Family Selectic Changed embedded memory from 552 to 700 Kbits in Laticustic Council Changed embedded memory from 552 to 700 Kbits in Laticustic Council Changed embedded memory from 552 to 700 Kbits in Laticustic Council Council Changed Council Changed Council Changed Council Changed Council Changed Council Changed Ch	noved references to Parallel burst mode Flash.
Corrected Primary Clock Sources text section. Corrected Secondary Clock Regions table. Corrected Secondary Clock Regions table. Corrected Note below Detailed sysDSP Slice Diagram. Corrected Clock, Clock Enable, and Reset Resources text Corrected ECP3-17 EBR number in Embedded SRAM in the Added On-Chip Termination Options for Input Modes table. Updated Available SERDES Quads per LatticeECP3 Device. Updated Simplified Channel Block Diagram for SERDES/Fulpdated Device Configuration text section. Corrected software default value of MCCLK to be 2.5 Million of	Features - Changed 250 Mbps to 230 Mbps in Embedded SERDES bulleted section and added a footnote to indicate 230 Mbps applies to 8b10b and 10b12b applications.  Updated data for ECP3-17 in LatticeECP3 Family Selection Guide table.  Changed embedded memory from 552 to 700 Kbits in LatticeECP3 Family Selection Guide table.
DC and Switching Characteristics  Updated VCCOB Min/Max data in Recommended Operating Corrected footnote 2 in sysIO Recommended Operating Added added footnote 7 for tskew_PRIB to External Switchie Added 2-to-1 Gearing text section and table. Updated External Reference Clock Specification (refclkp, LatticeECP3 sysCONFIG Port Timing Specifications - updated SysCONFIG Port Timing waveform. Serial Input Data Specifications table, delete Typ data for Added footnote 4 to sysCLOCK PLL Timing table for tpp. Added SERDES/PCS Block Latency Breakdown table. External Reference Clock Specifications table, added footnote 4 to sysCLOCK PLL Timing table for tpp. Added SERDES External Reference Clock Waveforms. Updated Serial Output Timing and Levels table. Pin-to-pin performance table, changed "typically 3% slow Updated timing information Updated SERDES minimum frequency Added data to the following tables: External Switching Characteristics, Family Timing Adders, Maximum Timing, High Speed Data Transmitter, Channel Output Jif Function Performance, Register-to-Register Performance, Requirements.  Updated Serial Input Data Specifications table. Updated Transmit table, Serial Rapid I/O Type 2 Electrica section.	Corrected Secondary Clock/Control Sources text section.  Corrected Secondary Clock Regions table.  Corrected note below Detailed sysDSP Slice Diagram.  Corrected Clock, Clock Enable, and Reset Resources text section.  Corrected ECP3-17 EBR number in Embedded SRAM in the LatticeECP3 Family table.  Added On-Chip Termination Options for Input Modes table.  Updated Available SERDES Quads per LatticeECP3 Devices table.  Updated Simplified Channel Block Diagram for SERDES/PCS Block diagram.
<ul> <li>Corrected footnote 2 in syslO Recommended Operating</li> <li>Added added footnote 7 for t<sub>SKEW_PRIB</sub> to External Switching</li> <li>Added 2-to-1 Gearing text section and table.</li> <li>Updated External Reference Clock Specification (refclkp,</li> <li>LatticeECP3 sysCONFIG Port Timing Specifications - updated sysCONFIG Port Timing waveform.</li> <li>Serial Input Data Specifications table, delete Typ data for Added footnote 4 to sysCLOCK PLL Timing table for t<sub>PFD</sub>.</li> <li>Added SERDES/PCS Block Latency Breakdown table.</li> <li>External Reference Clock Specifications table, added foot vREF-IN-DIFF.</li> <li>Added SERDES External Reference Clock Waveforms.</li> <li>Updated Serial Output Timing and Levels table.</li> <li>Pin-to-pin performance table, changed "typically 3% slow Updated timing information</li> <li>Updated SERDES minimum frequency</li> <li>Added data to the following tables: External Switching C Switching Characteristics, Family Timing Adders, Maximm Timing, High Speed Data Transmitter, Channel Output Jir Function Performance, Register-to-Register Performance Requirements.</li> <li>Updated Serial Input Data Specifications table.</li> <li>Updated Transmit table, Serial Rapid I/O Type 2 Electrical Section.</li> </ul>	Corrected software default value of MCCLK to be 2.5 MHz.
	Updated External Reference Clock Specification (refclkp/refclkn) table.  LatticeECP3 sysCONFIG Port Timing Specifications - updated tDINIT information.  Added sysCONFIG Port Timing waveform.  Serial Input Data Specifications table, delete Typ data for V <sub>RX-DIFF-S</sub> .  Added footnote 4 to sysCLOCK PLL Timing table for t <sub>PFD</sub> .  Added SERDES/PCS Block Latency Breakdown table.  External Reference Clock Specifications table, added footnote 4, add symbol name vREF-IN-DIFF.  Added SERDES External Reference Clock Waveforms.  Updated Serial Output Timing and Levels table.  Pin-to-pin performance table, changed "typically 3% slower" to "typically slower".  Updated timing information  Updated SERDES minimum frequency  Added data to the following tables: External Switching Characteristics, Internal Switching Characteristics, Family Timing Adders, Maximum I/O Buffer Speed, DLL Timing, High Speed Data Transmitter, Channel Output Jitter, Typical Building Block Function Performance, Register-to-Register Performance, and Power Supply Requirements.  Updated Serial Input Data Specifications table.  Updated Transmit table, Serial Rapid I/O Type 2 Electrical and Timing Characteristics
	Updated Signal Description tables. Updated Pin Information Summary tables and added footnote 1.

#### Revision 1.0EA, February 2009

Section	Change Summary
All	Initial release.

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