## Features

## Includes ST state-of-the-art patented technology

- Ultra-low-power platform
- 1.65 V to 3.6 V power supply
- $-40^{\circ} \mathrm{C}$ to $105^{\circ} \mathrm{C}$ temperature range
- 305 nA Standby mode (3 wakeup pins)
- $1.15 \mu \mathrm{~A}$ Standby mode + RTC
- $0.475 \mu \mathrm{~A}$ Stop mode (16 wakeup lines)
- $1.35 \mu \mathrm{~A}$ Stop mode + RTC
- $11 \mu \mathrm{~A}$ Low-power run mode
- $230 \mu \mathrm{~A} / \mathrm{MHz}$ Run mode
- 10 nA ultra-low I/O leakage
- $8 \mu \mathrm{~s}$ wakeup time
- Core: Arm ${ }^{\circledR}$ Cortex ${ }^{\circledR}$-M3 32-bit CPU
- From 32 kHz up to 32 MHz max
- 33.3 DMIPS peak (Dhrystone 2.1)
- Memory protection unit
- Up to 34 capacitive sensing channels
- CRC calculation unit, 96-bit unique ID
- Reset and supply management
- Low-power, ultrasafe BOR (brownout reset) with 5 selectable thresholds
- Ultra-low-power POR/PDR
- Programmable voltage detector (PVD)
- Clock sources
- 1 to 24 MHz crystal oscillator
- 32 kHz oscillator for RTC with calibration
- High Speed Internal 16 MHz factorytrimmed RC (+/-1\%)
- Internal low-power 37 kHz RC
- Internal multispeed low-power 65 kHz to 4.2 MHz
- PLL for CPU clock and USB (48 MHz)
- Pre-programmed bootloader
- USB and USART supported

- Serial wire debug, JTAG and trace
- Up to 116 fast I/Os (102 I/Os 5V tolerant), all mappable on 16 external interrupt vectors
- Memories
- 384 Kbytes of Flash memory with ECC (with 2 banks of 192 Kbytes enabling Rww capability)
- 48 Kbytes of RAM
- 12 Kbytes of true EEPROM with ECC
- 128-byte backup register
- Memory interface controller supporting SRAM, PSRAM and NOR Flash
- LCD driver (except STM32L151xD devices) up to $8 x 40$ segments, contrast adjustment, blinking mode, step-up converter
- Rich analog peripherals (down to 1.8 V )
- $3 x$ operational amplifiers
- 12-bit ADC 1 Msps up to 40 channels
- 12-bit DAC 2 ch with output buffers
- 2x ultra-low-power-comparators (window mode and wakeup capability)
- DMA controller $12 x$ channels
- $12 x$ peripheral communication interfaces
- 1x USB 2.0 (internal 48 MHz PLL)
- 5x USARTs
- Up to $8 x$ SPIs ( $2 x$ I2S, $3 x 16 \mathrm{Mbit} / \mathrm{s}$ )
- 2x I2Cs (SMBus/PMBus)
- 1x SDIO interface
- 11x timers: $1 \times 32$-bit, $6 \times 16$-bit with up to 4 IC/OC/PWM channels, $2 \times 16$-bit basic timers, $2 x$ watchdog timers (independent and window)

Table 1. Device summary

| Reference | Part number |
| :--- | :--- |
| STM32L151xD | STM32L151QD, STM32L151RD, STM32L151VD, STM32L151ZD |
| STM32L152xD | STM32L152QD, STM32L152RD, STM32L152VD, STM32L152ZD |

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## 1 Introduction

This datasheet provides the ordering information and mechanical device characteristics of the STM32L151xD and STM32L152xD ultra-low-power Arm ${ }^{\circledR}$ Cortex ${ }^{\circledR}$-M3 based microcontroller product line.

The STM32L151xD and STM32L152xD microcontrollers feature 384 Kbytes of Flash memory.

The ultra-low-power STM32L151xD and STM32L152xD family includes devices in 5 different package types: from 64 pins to 144 pins. Depending on the device chosen, different sets of peripherals are included, the description below gives an overview of the complete range of peripherals proposed in this family.

These features make the ultra-low-power STM32L151xD and STM32L152xD microcontroller family suitable for a wide range of applications:

- Medical and handheld equipment
- Application control and user interface
- PC peripherals, gaming, GPS and sport equipment
- Alarm systems, wired and wireless sensors, video intercom
- Utility metering

This STM32L151xD and STM32L152xD datasheet must be read in conjunction with the STM32L1xxxx reference manual (RM0038). The application note "Getting started with STM32L1xxxx hardware development" (AN3216) gives a hardware implementation overview. Both documents are available from the STMicroelectronics website www.st.com.

For information on the Arm ${ }^{\circledR(a)}$ Cortex ${ }^{\circledR}-\mathrm{M} 3$ core, refer to the Arm ${ }^{\circledR}$ Cortex ${ }^{\circledR}$-M3 technical reference manual, available from the www.arm.com website. Figure 1 shows the general block diagram of the device family.

For information on the device errata with respect to the datasheet and reference manual, refer to the STM32L151xD and STM32L152xD errata sheet (ES0240), available on the STMicroelectronics website www.st.com.


[^0]
## 2 Description

The ultra-low-power STM32L151xD and STM32L152xD devices incorporate the connectivity power of the universal serial bus (USB) with the high-performance Arm ${ }^{\circledR}$ Cortex ${ }^{\circledR}$-M3 32-bit RISC core operating at a frequency of 32 MHz (33.3 DMIPS), a memory protection unit (MPU), high-speed embedded memories (Flash memory up to 384 Kbytes and RAM up to 48 Kbytes), a flexible static memory controller (FSMC) interface (for devices with packages of 100 pins and more) and an extensive range of enhanced I/Os and peripherals connected to two APB buses.

The STM32L151xD and STM32L152xD devices offer three operational amplifiers, one 12bit ADC, two DACs, two ultra-low-power comparators, one general-purpose 32-bit timer, six general-purpose 16-bit timers and two basic timers, which can be used as time bases.

Moreover, the STM32L151xD and STM32L152xD devices contain standard and advanced communication interfaces: up to two I2Cs, three SPIs, two I2S, one SDIO, three USARTs, two UARTs, and an USB. The STM32L151xD and STM32L152xD devices offer up to 34 capacitive sensing channels to simply add a touch sensing functionality to any application.

They also include a real-time clock and a set of backup registers that remain powered in Standby mode.
Finally, the integrated LCD controller (except STM32L151xD devices) has a built-in LCD voltage generator that allows to drive up to 8 multiplexed LCDs with the contrast independent of the supply voltage.

The ultra-low-power STM32L151xD and STM32L152xD devices operate from a 1.8 to 3.6 V power supply (down to 1.65 V at power down) with BOR and from a 1.65 to 3.6 V power supply without BOR option. They are available in the -40 to $+85^{\circ} \mathrm{C}$ and -40 to $+105^{\circ} \mathrm{C}$ temperature ranges. A comprehensive set of power-saving modes allows the design of lowpower applications.

### 2.1 Device overview

Table 2. Ultra-low-power STM32L151xD and STM32L152xD device features and peripheral counts

| Peripheral |  | STM32L15xRD | STM32L15xVD | STM32L15xQD | STM32L15xZD |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Flash (Kbytes) |  | 384 |  |  |  |
| Data EEPROM (Kbytes) |  | 12 |  |  |  |
| RAM (Kbytes) |  | 48 |  |  |  |
| FSMC |  | No | multiplexed only | Yes |  |
| Timers | 32 bit | 1 |  |  |  |
|  | Generalpurpose | 6 |  |  |  |
|  | Basic | 2 |  |  |  |
| Communication interfaces | SPI | $8(3)^{(1)}$ |  |  |  |
|  | $1^{2} S$ | 2 |  |  |  |
|  | $\mathrm{I}^{2} \mathrm{C}$ | 2 |  |  |  |
|  | USART | 5 |  |  |  |
|  | USB | 1 |  |  |  |
|  | SDIO | 1 |  |  |  |
| GPIOs |  | 51 | 83 | 109 | 115 |
| Operation amplifiers |  | 3 |  |  |  |
| 12-bit synchronized ADC Number of channels |  | $\begin{gathered} 1 \\ 21 \end{gathered}$ | $\begin{gathered} 1 \\ 25 \end{gathered}$ | $\begin{gathered} 1 \\ 40 \end{gathered}$ | $\begin{gathered} 1 \\ 40 \end{gathered}$ |
| 12-bit DAC <br> Number of channels |  | 2 |  |  |  |
| LCD (STM32L152xx devices only) <br> COM x SEG |  | $\begin{gathered} 1 \\ 4 \times 32 \text { or } 8 \times 28 \end{gathered}$ | $1$ |  |  |
| Comparators |  | 2 |  |  |  |
| Capacitive sensing channels |  | 23 |  | 33 | 34 |
| Max. CPU frequency |  | 32 MHz |  |  |  |
| Operating voltage |  | 1.8 V to 3.6 V (down to 1.65 V at power-down) with BOR option 1.65 V to 3.6 V without BOR option |  |  |  |

Table 2. Ultra-low-power STM32L151xD and STM32L152xD device features and peripheral counts (continued)

| Peripheral | STM32L15xRD | STM32L15xVD | STM32L15xQD | STM32L15xZD |
| :--- | :---: | :---: | :---: | :---: |
| Operating temperatures | Ambient operating temperature: $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C} /-40^{\circ} \mathrm{C}$ to $105^{\circ} \mathrm{C}$ |  |  |  |
| Junction temperature: -40 to $+110^{\circ} \mathrm{C}$ |  |  |  |  |

1. 5 SPIs are USART configured in synchronous mode emulating SPI master.

### 2.2 Ultra-low-power device continuum

The ultra-low-power family offers a large choice of cores and features. From proprietary 8bit to up to Cortex-M3, including the Cortex-M0+, the STM32Lx series are the best choice to answer the user needs, in terms of ultra-low-power features. The STM32 ultra-low-power series are the best fit, for instance, for gas/water meter, keyboard/mouse or fitness and healthcare, wearable applications. Several built-in features like LCD drivers, dual-bank memory, Low-power run mode, op-amp, AES 128-bit, DAC, USB crystal-less and many others clearly allow very cost-optimized applications to be built by reducing BOM.
Note: $\quad$ STMicroelectronics as a reliable and long-term manufacturer ensures as much as possible the pin-to-pin compatibility between any STM8Lxxxxx and STM32Lxxxxx devices and between any of the STM32Lx and STM32Fx series. Thanks to this unprecedented scalability, the old applications can be upgraded to respond to the latest market features and efficiency demand.

### 2.2.1 Performance

All the families incorporate highly energy-efficient cores with both Harvard architecture and pipelined execution: advanced STM8 core for STM8L families and Arm Cortex-M3 core for STM32L family. In addition specific care for the design architecture has been taken to optimize the $\mathrm{mA} / \mathrm{DMIPS}$ and $\mathrm{mA} / \mathrm{MHz}$ ratios.

This allows the ultra-low-power performance to range from 5 up to 33.3 DMIPs.

### 2.2.2 Shared peripherals

STM8L15xxx, STM32L15xxx and STM32L162xx share identical peripherals which ensure a very easy migration from one family to another:

- Analog peripherals: ADC, DAC and comparators
- Digital peripherals: RTC and some communication interfaces


### 2.2.3 Common system strategy.

To offer flexibility and optimize performance, the STM8L15xxx, STM32L15xxx and STM32L162xx family uses a common architecture:

- $\quad$ Same power supply range from 1.65 V to 3.6 V
- Architecture optimized to reach ultra-low consumption both in low-power modes and Run mode
- Fast startup strategy from low-power modes
- Flexible system clock
- Ultrasafe reset: same reset strategy including power-on reset, power-down reset, brownout reset and programmable voltage detector


### 2.2.4 Features

ST ultra-low-power continuum also lies in feature compatibility:

- More than 15 packages with pin count from 20 to 144 pins and size down to $3 \times 3 \mathrm{~mm}$
- Memory density ranging from 2 to 512 Kbytes


## 3 Functional overview

Figure 1. Ultra-low-power STM32L151xD and STM32L152xD block diagram


### 3.1 Low-power modes

The ultra-low-power STM32L151xD and STM32L152xD devices support dynamic voltage scaling to optimize its power consumption in run mode. The voltage from the internal lowdrop regulator that supplies the logic can be adjusted according to the system's maximum operating frequency and the external voltage supply.

There are three power consumption ranges:

- Range $1\left(\mathrm{~V}_{\mathrm{DD}}\right.$ range limited to $1.71 \mathrm{~V}-3.6 \mathrm{~V}$ ), with the CPU running at up to 32 MHz
- Range 2 (full $\mathrm{V}_{\mathrm{DD}}$ range), with a maximum CPU frequency of 16 MHz
- $\quad$ Range 3 (full $\mathrm{V}_{\mathrm{DD}}$ range), with a maximum CPU frequency limited to 4 MHz (generated only with the multispeed internal RC oscillator clock source)

Seven low-power modes are provided to achieve the best compromise between low-power consumption, short startup time and available wakeup sources:

- Sleep mode

In Sleep mode, only the CPU is stopped. All peripherals continue to operate and can wake up the CPU when an interrupt/event occurs. Sleep mode power consumption at 16 MHz is about 1 mA with all peripherals off.

- Low-power run mode

This mode is achieved with the multispeed internal (MSI) RC oscillator set to the MSI range 0 or MSI range 1 clock range (maximum 131 kHz ), execution from SRAM or Flash memory, and internal regulator in low-power mode to minimize the regulator's operating current. In low-power run mode, the clock frequency and the number of enabled peripherals are both limited.

- Low-power sleep mode

This mode is achieved by entering Sleep mode with the internal voltage regulator in Low-power mode to minimize the regulator's operating current. In Low-power sleep mode, both the clock frequency and the number of enabled peripherals are limited; a typical example would be to have a timer running at 32 kHz .
When wakeup is triggered by an event or an interrupt, the system reverts to the run mode with the regulator on.

- Stop mode with RTC

Stop mode achieves the lowest power consumption while retaining the RAM and register contents and real time clock. All clocks in the $\mathrm{V}_{\text {CORE }}$ domain are stopped, the PLL, MSI RC, HSI RC and HSE crystal oscillators are disabled. The LSE or LSI is still running. The voltage regulator is in the low-power mode.
The device can be woken up from Stop mode by any of the EXTI line, in $8 \mu \mathrm{~s}$. The EXTI line source can be one of the 16 external lines. It can be the PVD output, the Comparator 1 event or Comparator 2 event (if internal reference voltage is on), it can be the RTC alarm(s), the USB wakeup, the RTC tamper events, the RTC timestamp event or the RTC wakeup.

- Stop mode without RTC

Stop mode achieves the lowest power consumption while retaining the RAM and register contents. All clocks are stopped, the PLL, MSI RC, HSI and LSI RC, LSE and HSE crystal oscillators are disabled. The voltage regulator is in the low-power mode. The device can be woken up from Stop mode by any of the EXTI line, in $8 \mu \mathrm{~s}$. The EXTI line source can be one of the 16 external lines. It can be the PVD output, the Comparator 1 event or Comparator 2 event (if internal reference voltage is on). It can also be wakened by the USB wakeup.

- Standby mode with RTC

Standby mode is used to achieve the lowest power consumption and real time clock. The internal voltage regulator is switched off so that the entire $V_{\text {CORE }}$ domain is powered off. The PLL, MSI RC, HSI RC and HSE crystal oscillators are also switched off. The LSE or LSI is still running. After entering Standby mode, the RAM and register contents are lost except for registers in the Standby circuitry (wakeup logic, IWDG, RTC, LSI, LSE Crystal 32K osc, RCC_CSR).
The device exits Standby mode in $60 \mu \mathrm{~s}$ when an external reset (NRST pin), an IWDG reset, a rising edge on one of the three WKUP pins, RTC alarm (Alarm A or Alarm B), RTC tamper event, RTC timestamp event or RTC Wakeup event occurs.

- Standby mode without RTC

Standby mode is used to achieve the lowest power consumption. The internal voltage regulator is switched off so that the entire $\mathrm{V}_{\text {CORE }}$ domain is powered off. The PLL, MSI RC, HSI and LSI RC, HSE and LSE crystal oscillators are also switched off. After entering Standby mode, the RAM and register contents are lost except for registers in the Standby circuitry (wakeup logic, IWDG, RTC, LSI, LSE Crystal 32K osc, RCC_CSR).
The device exits Standby mode in $60 \mu s$ when an external reset (NRST pin) or a rising edge on one of the three WKUP pin occurs.

Note: The RTC, the IWDG, and the corresponding clock sources are not stopped automatically by entering Stop or Standby mode.

Table 3. Functionalities depending on the operating power supply range

| - | Functionalities depending on the operating power supply <br> range ${ }^{(1)}$ |  |  |
| :---: | :---: | :---: | :---: |
| Operating power supply <br> range | DAC and ADC <br> operation | USB | Dynamic voltage scaling <br> range |
| $V_{D D}=V_{D D A}=1.65$ to 1.71 V | Not functional | Not functional | Range 2 or Range 3 |
| $V_{D D}=V_{D D A}=1.71$ to $1.8 \mathrm{~V}^{(2)}$ | Not functional | Not functional | Range 1, Range 2 or <br> Range 3 |
| $V_{D D}=V_{D D A}=1.8$ to $2.0 \mathrm{~V}^{(2)}$ | Conversion time up <br> to 500 Ksps | Not functional | Range 1, Range 2 or <br> Range 3 |

Table 3. Functionalities depending on the operating power supply range (continued)

| - | Functionalities depending on the operating power supply range ${ }^{(1)}$ |  |  |
| :---: | :---: | :---: | :---: |
| Operating power supply range | DAC and ADC operation | USB | Dynamic voltage scaling range |
| $\mathrm{V}_{\mathrm{DD}}=\mathrm{V}_{\mathrm{DDA}}=2.0$ to 2.4 V | Conversion time up to 500 Ksps | Functional ${ }^{(3)}$ | Range 1, Range 2 or Range 3 |
| $\mathrm{V}_{\mathrm{DD}}=\mathrm{V}_{\mathrm{DDA}}=2.4$ to 3.6 V | Conversion time up to 1 Msps | Functional ${ }^{(3)}$ | Range 1, Range 2 or Range 3 |

1. The GPIO speed also depends from VDD voltage and the user has to refer to Table 52: I/O AC characteristics for more information about I/O speed.
2. CPU frequency changes from initial to final must respect " $F_{\text {CPU }}$ initial $<4^{\star} F_{\text {CPU }}$ final" to limit $V_{\text {CORE }}$ drop due to current consumption peak when frequency increases. It must also respect $5 \mu$ s delay between two changes. For example to switch from 4.2 MHz to 32 MHz , the user can switch from 4.2 MHz to 16 MHz , wait $5 \mu \mathrm{~s}$, then switch from 16 MHz to 32 MHz .
3. Must be USB compliant from I/O voltage standpoint, the minimum $\mathrm{V}_{\mathrm{DD}}$ is 3.0 V .

Table 4. CPU frequency range depending on dynamic voltage scaling

| CPU frequency range | Dynamic voltage scaling range |
| :---: | :---: |
| 16 MHz to 32 MHz (1ws) 32 kHz to 16 MHz (0ws) | Range 1 |
| 8 MHz to 16 MHz (1ws) 32 kHz to 8 MHz (0ws) | Range 2 |
| 2.1 MHz to 4.2 MHz (1ws) <br> 32 kHz to 2.1 MHz (0ws) | Range 3 |

Table 5. Functionalities depending on the working mode (from Run/active down to standby)

| Ips | Run/Active | Sleep | Lowpower Run | Lowpower Sleep | Stop |  | Standby |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | - | Wakeup capability | - | Wakeup capability |
| CPU | Y | -- | Y | -- | -- | -- | -- | -- |
| Flash | Y | Y | Y | Y | -- | -- | -- | -- |
| RAM | Y | Y | Y | Y | Y | -- | -- | -- |
| Backup Registers | Y | Y | Y | Y | Y | -- | Y | -- |
| EEPROM | Y | Y | Y | Y | Y | -- | -- | -- |
| Brown-out rest (BOR) | Y | Y | Y | Y | Y | Y | Y | -- |
| DMA | Y | Y | Y | Y | -- | -- | -- | -- |
| Programmable Voltage Detector (PVD) | Y | Y | Y | Y | Y | Y | Y | -- |
| Power On Reset (POR) | Y | Y | Y | Y | Y | Y | Y | -- |
| Power Down Rest (PDR) | Y | Y | Y | Y | Y | -- | Y | -- |
| High Speed Internal (HSI) | Y | Y | -- | -- | -- | -- | -- | -- |
| High Speed External (HSE) | Y | Y | -- | -- | -- | -- | -- | -- |
| Low Speed Internal (LSI) | Y | Y | Y | Y | Y | -- | Y | -- |
| Low Speed External (LSE) | Y | Y | Y | Y | Y | -- | Y | -- |
| Multi-Speed Internal (MSI) | Y | Y | Y | Y | -- | -- | -- | -- |
| Inter-Connect Controller | Y | Y | Y | Y | -- | -- | -- | -- |
| RTC | Y | Y | Y | Y | Y | Y | Y | -- |
| RTC Tamper | Y | Y | Y | Y | Y | Y | Y | Y |
| Auto WakeUp (AWU) | Y | Y | Y | Y | Y | Y | Y | Y |
| LCD | Y | Y | Y | Y | Y | -- | -- | -- |
| USB | Y | Y | -- | -- | -- | Y | -- | -- |
| USART | Y | Y | Y | Y | Y | (1) | -- | -- |
| SPI | Y | Y | Y | Y | -- | -- | -- | -- |
| I2C | Y | Y | -- | -- | -- | (1) | -- | -- |

Table 5. Functionalities depending on the working mode (from Run/active down to standby) (continued)

| Ips | Run/Active | Sleep | Lowpower Run | Low- <br> power <br> Sleep | Stop |  | Standby |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  | Wakeup capability | - | Wakeup capability |
| ADC | Y | Y | -- | -- | - | -- | -- | -- |
| DAC | Y | Y | Y | Y | Y | -- | -- | -- |
| Tempsensor | Y | Y | Y | Y | $Y$ | -- | -- | -- |
| OP amp | Y | Y | Y | Y | Y | -- | -- | -- |
| Comparators | Y | Y | Y | Y | Y | Y | -- | -- |
| 16-bit and 32-bit Timers | Y | Y | Y | Y | -- | -- | -- | -- |
| IWDG | Y | Y | Y | Y | Y | Y | Y | Y |
| WWDG | Y | Y | Y | Y | -- | -- | -- | -- |
| Touch sensing | Y | Y | -- | -- | - | -- | -- | -- |
| Systic Timer | Y | Y | Y | Y | -- | -- | -- | -- |
| GPIOs | Y | Y | Y | Y | Y | Y | -- | 3 pins |
| Wakeup time to Run mode | $0 \mu \mathrm{~s}$ | $0.4 \mu \mathrm{~s}$ | $3 \mu \mathrm{~s}$ | $46 \mu \mathrm{~s}$ |  | < $8 \mu \mathrm{~s}$ |  | $58 \mu \mathrm{~s}$ |
| Consumption <br> $\mathrm{V}_{\mathrm{DD}}=1.8$ to 3.6 V <br> (Typ) | Down to 230 $\mu \mathrm{A} / \mathrm{MHz}$ (from Flash) | Down to 43 $\mu \mathrm{A} / \mathrm{MHz}$ (from Flash) | Down to $11 \mu \mathrm{~A}$ | Down to$4.4 \mu \mathrm{~A}$ |  | $0.475 \mu \mathrm{~A}$ (no RTC) $V_{D D}=1.8 \mathrm{~V}$ |  | $\begin{aligned} & 0.305 \mu \mathrm{~A} \\ & \text { (no RTC) } \\ & \mathrm{V}_{\mathrm{DD}}=1.8 \mathrm{~V} \end{aligned}$ |
|  |  |  |  |  |  | $1.1 \mu \mathrm{~A}$ <br> with RTC) <br> $V_{D D}=1.8 \mathrm{~V}$ |  | $0.82 \mu \mathrm{~A}$ (with RTC) $\mathrm{V}_{\mathrm{DD}}=1.8 \mathrm{~V}$ |
|  |  |  |  |  |  | $0.475 \mu \mathrm{~A}$ (no RTC) $V_{D D}=3.0 \mathrm{~V}$ |  | $\begin{aligned} & 0.305 \mu \mathrm{~A} \\ & \text { (no RTC) } \\ & \mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V} \end{aligned}$ |
|  |  |  |  |  |  | $\begin{gathered} 1.35 \mu \mathrm{~A} \\ \text { with } \mathrm{RTC}) \\ \mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V} \end{gathered}$ |  | $1.15 \mu \mathrm{~A}$ with RTC) $V_{D D}=3.0 \mathrm{~V}$ |

1. The startup on communication line wakes the CPU which was made possible by an EXTI, this induces a delay before entering run mode.

### 3.2 Arm ${ }^{\circledR}$ Cortex ${ }^{\circledR}$-M3 core with MPU

The Arm ${ }^{\circledR}$ Cortex ${ }^{\circledR}-\mathrm{M} 3$ processor is the industry leading processor for embedded systems. It has been developed to provide a low-cost platform that meets the needs of MCU implementation, with a reduced pin count and low-power consumption, while delivering outstanding computational performance and an advanced system response to interrupts.
The Arm ${ }^{\circledR}$ Cortex ${ }^{\circledR}$-M3 32-bit RISC processor features exceptional code-efficiency, delivering the high-performance expected from an Arm core in the memory size usually associated with 8 - and 16-bit devices.

The memory protection unit (MPU) improves system reliability by defining the memory attributes (such as read/write access permissions) for different memory regions. It provides up to eight different regions and an optional predefined background region.

Owing to its embedded Arm core, the STM32L151xD and STM32L152xD devices are compatible with all Arm tools and software.

## Nested vectored interrupt controller (NVIC)

The ultra-low-power STM32L151xD and STM32L152xD devices embed a nested vectored interrupt controller able to handle up to 56 maskable interrupt channels (not including the 16 interrupt lines of Arm ${ }^{\circledR}$ Cortex ${ }^{\circledR}$-M3) and 16 priority levels.

- Closely coupled NVIC gives low-latency interrupt processing
- Interrupt entry vector table address passed directly to the core
- Closely coupled NVIC core interface
- Allows early processing of interrupts
- Processing of late arriving, higher-priority interrupts
- Support for tail-chaining
- Processor state automatically saved on interrupt entry, and restored on interrupt exit, with no instruction overhead

This hardware block provides flexible interrupt management features with minimal interrupt latency.

### 3.3 Reset and supply management

### 3.3.1 Power supply schemes

- $\quad \mathrm{V}_{\mathrm{DD}}=1.65$ to 3.6 V : external power supply for I/Os and the internal regulator. Provided externally through $V_{D D}$ pins.
- $\quad V_{S S A}, V_{D D A}=1.65$ to 3.6 V : external analog power supplies for ADC, reset blocks, RCs and PLL (minimum voltage to be applied to $\mathrm{V}_{\text {DDA }}$ is 1.8 V when the $A D C$ is used). $\mathrm{V}_{\text {DDA }}$ and $V_{S S A}$ must be connected to $V_{D D}$ and $V_{S S}$, respectively.


### 3.3.2 Power supply supervisor

The device has an integrated ZEROPOWER power-on reset (POR)/power-down reset (PDR) that can be coupled with a brownout reset (BOR) circuitry.

The device exists in two versions:

- The version with BOR activated at power-on operates between 1.8 V and 3.6 V .
- The other version without BOR operates between 1.65 V and 3.6 V .

After the $\mathrm{V}_{\mathrm{DD}}$ threshold is reached ( 1.65 V or 1.8 V depending on the BOR which is active or not at power-on), the option byte loading process starts, either to confirm or modify default thresholds, or to disable the BOR permanently: in this case, the $\mathrm{V}_{\mathrm{DD}}$ min value becomes 1.65 V (whatever the version, BOR active or not, at power-on).

When BOR is active at power-on, it ensures proper operation starting from 1.8 V whatever the power ramp-up phase before it reaches 1.8 V . When BOR is not active at power-up, the power ramp-up must guarantee that 1.65 V is reached on $\mathrm{V}_{\mathrm{DD}}$ at least 1 ms after it exits the POR area.

Five BOR thresholds are available through option bytes, starting from 1.8 V to 3 V . To reduce the power consumption in Stop mode, it is possible to automatically switch off the internal reference voltage ( $\mathrm{V}_{\text {REFINT }}$ ) in Stop mode. The device remains in reset mode when $V_{D D}$ is below a specified threshold, $V_{P O R / P D R}$ or $V_{B O R}$, without the need for any external reset circuit.

Note: $\quad$ The start-up time at power-on is typically 3.3 ms when BOR is active at power-up, the startup time at power-on can be decreased down to 1 ms typically for devices with BOR inactive at power-up.

The device features an embedded programmable voltage detector (PVD) that monitors the $\mathrm{V}_{\mathrm{DD}} / \mathrm{V}_{\text {DDA }}$ power supply and compares it to the $\mathrm{V}_{\text {PVD }}$ threshold. This PVD offers 7 different levels between 1.85 V and 3.05 V , chosen by software, with a step around 200 mV . An interrupt can be generated when $\mathrm{V}_{\mathrm{DD}} / \mathrm{V}_{\mathrm{DDA}}$ drops below the $\mathrm{V}_{\mathrm{PVD}}$ threshold and/or when $V_{D D} / V_{D D A}$ is higher than the $V_{P V D}$ threshold. The interrupt service routine can then generate a warning message and/or put the MCU into a safe state. The PVD is enabled by software.

### 3.3.3 Voltage regulator

The regulator has three operation modes: main (MR), low-power (LPR) and power down.

- MR is used in Run mode (nominal regulation)
- LPR is used in the Low-power run, Low-power sleep and Stop modes
- Power down is used in Standby mode. The regulator output is high impedance, the kernel circuitry is powered down, inducing zero consumption but the contents of the registers and RAM are lost except for the standby circuitry (wakeup logic, IWDG, RTC, LSI, LSE crystal 32K osc, RCC_CSR).


### 3.3.4 Boot modes

At startup, boot pins are used to select one of three boot options:

- Boot from Flash memory
- Boot from System memory
- Boot from embedded RAM

The boot loader is located in System memory. It is used to reprogram the Flash memory by using USART1, USART2 or USB. See Application note "STM32 microcontroller system memory boot mode" (AN2606) for details.

### 3.4 Clock management

The clock controller distributes the clocks coming from different oscillators to the core and the peripherals. It also manages clock gating for low-power modes and ensures clock robustness. It features:

- Clock prescaler: to get the best trade-off between speed and current consumption, the clock frequency to the CPU and peripherals can be adjusted by a programmable prescaler.
- Safe clock switching: clock sources can be changed safely on the fly in run mode through a configuration register.
- Clock management: to reduce power consumption, the clock controller can stop the clock to the core, individual peripherals or memory.
- System clock source: three different clock sources can be used to drive the master clock SYSCLK:
- 1-24 MHz high-speed external crystal (HSE), that can supply a PLL
- 16 MHz high-speed internal RC oscillator (HSI), trimmable by software, that can supply a PLL
- Multispeed internal RC oscillator (MSI), trimmable by software, able to generate 7 frequencies ( $65 \mathrm{kHz}, 131 \mathrm{kHz}, 262 \mathrm{kHz}, 524 \mathrm{kHz}, 1.05 \mathrm{MHz}, 2.1 \mathrm{MHz}, 4.2 \mathrm{MHz}$ ). When a 32.768 kHz clock source is available in the system (LSE), the MSI frequency can be trimmed by software down to a $\pm 0.5 \%$ accuracy.
- Auxiliary clock source: two ultra-low-power clock sources that can be used to drive the LCD controller and the real-time clock:
- $\quad 32.768 \mathrm{kHz}$ low-speed external crystal (LSE)
- $\quad 37 \mathrm{kHz}$ low-speed internal RC (LSI), also used to drive the independent watchdog. The LSI clock can be measured using the high-speed internal RC oscillator for greater precision.
- RTC and LCD clock sources: the LSI, LSE or HSE sources can be chosen to clock the RTC and the LCD, whatever the system clock.
- USB clock source: the embedded PLL has a dedicated 48 MHz clock output to supply the USB interface.
- Startup clock: after reset, the microcontroller restarts by default with an internal 2 MHz clock (MSI). The prescaler ratio and clock source can be changed by the application program as soon as the code execution starts.
- Clock security system (CSS): this feature can be enabled by software. If a HSE clock failure occurs, the master clock is automatically switched to HSI and a software interrupt is generated if enabled.
- Clock-out capability (MCO: microcontroller clock output): it outputs one of the internal clocks for external use by the application.
Several prescalers allow the configuration of the AHB frequency, each APB (APB1 and APB2) domains. The maximum frequency of the AHB and the APB domains is 32 MHz . See Figure 2 for details on the clock tree.

Figure 2. Clock tree


### 3.5 Low-power real-time clock and backup registers

The real-time clock (RTC) is an independent BCD timer/counter. Dedicated registers contain the sub-second, second, minute, hour (12/24 hour), week day, date, month, year, in BCD (binary-coded decimal) format. Correction for 28, 29 (leap year), 30, and 31 day of the month are made automatically. The RTC provides two programmable alarms and programmable periodic interrupts with wakeup from Stop and Standby modes.

The programmable wakeup time ranges from $120 \mu s$ to 36 hours.
The RTC can be calibrated with an external 512 Hz output, and a digital compensation circuit helps reduce drift due to crystal deviation.

The RTC can also be automatically corrected with a $50 / 60 \mathrm{~Hz}$ stable powerline.
The RTC calendar can be updated on the fly down to sub second precision, which enables network system synchronization.

A time stamp can record an external event occurrence, and generates an interrupt.
There are thirty-two 32-bit backup registers provided to store 128 bytes of user application data. They are cleared in case of tamper detection.

Three pins can be used to detect tamper events. A change on one of these pins can reset backup register and generate an interrupt. To prevent false tamper event, like ESD event, these three tamper inputs can be digitally filtered.

### 3.6 GPIOs (general-purpose inputs/outputs)

Each of the GPIO pins can be configured by software as output (push-pull or open-drain), as input (with or without pull-up or pull-down) or as peripheral alternate function. Most of the GPIO pins are shared with digital or analog alternate functions, and can be individually remapped using dedicated AFIO registers. All GPIOs are high current capable. The alternate function configuration of I/Os can be locked if needed following a specific sequence in order to avoid spurious writing to the I/O registers. The I/O controller is connected to the AHB with a toggling speed of up to 16 MHz .

## External interrupt/event controller (EXTI)

The external interrupt/event controller consists of 24 edge detector lines used to generate interrupt/event requests. Each line can be individually configured to select the trigger event (rising edge, falling edge, both) and can be masked independently. A pending register maintains the status of the interrupt requests. The EXTI can detect an external line with a pulse width shorter than the Internal APB2 clock period. Up to 115 GPIOs can be connected to the 16 external interrupt lines. The 8 other lines are connected to RTC, PVD, USB, comparator events or capacitive sensing acquisition.

### 3.7 Memories

The STM32L151xD and STM32L152xD devices have the following features:

- 48 Kbytes of embedded RAM accessed (read/write) at CPU clock speed with 0 wait states. With the enhanced bus matrix, operating the RAM does not lead to any performance penalty during accesses to the system bus (AHB and APB buses).
- The non-volatile memory is divided into three arrays:
- 384 Kbytes of embedded Flash program memory
- 12 Kbytes of data EEPROM
- Options bytes

Flash program and data EEPROM are divided into two banks, this enables writing in one bank while running code or reading data in the other bank.

The options bytes are used to write-protect or read-out protect the memory (with 4 Kbytes granularity) and/or readout-protect the whole memory with the following options:

- Level 0: no readout protection
- Level 1: memory readout protection, the Flash memory cannot be read from or written to if either debug features are connected or boot in RAM is selected
- Level 2: chip readout protection, debug features (Arm Cortex-M3 JTAG and serial wire) and boot in RAM selection disabled (JTAG fuse)
The whole non-volatile memory embeds the error correction code (ECC) feature.


### 3.8 FSMC (flexible static memory controller)

The FSMC supports the following modes: SRAM, PSRAM, NOR/OneNAND Flash.
Functionality overview:

- Up to 26 bit address bus
- Up to 16-bit data bus
- Write FIFO
- Burst mode
- Code execution from external memory
- Four chip select signals
- Up to 32 MHz external access


### 3.9 DMA (direct memory access)

The flexible 12-channel, general-purpose DMA is able to manage memory-to-memory, peripheral-to-memory and memory-to-peripheral transfers. The DMA controller supports circular buffer management, avoiding the generation of interrupts when the controller reaches the end of the buffer.

Each channel is connected to dedicated hardware DMA requests, with software trigger support for each channel. Configuration is done by software and transfer sizes between source and destination are independent.

The DMA can be used with the main peripherals: SPI, $I^{2}$ C, USART, SDIO, general-purpose timers, DAC and ADC.

### 3.10 LCD (liquid crystal display)

The LCD drives up to 8 common terminals and 44 segment terminals to drive up to 320 pixels.

- Internal step-up converter to guarantee functionality and contrast control irrespective of $\mathrm{V}_{\mathrm{DD}}$. This converter can be deactivated, in which case the $\mathrm{V}_{\mathrm{LCD}}$ pin is used to provide the voltage to the LCD
- $\quad$ Supports static, $1 / 2,1 / 3,1 / 4$ and $1 / 8$ duty
- Supports static, $1 / 2,1 / 3$ and $1 / 4$ bias
- Phase inversion to reduce power consumption and EMI
- Up to 8 pixels can be programmed to blink
- Unneeded segments and common pins can be used as general I/O pins
- LCD RAM can be updated at any time owing to a double-buffer
- The LCD controller can operate in Stop mode


### 3.11 ADC (analog-to-digital converter)

A 12-bit analog-to-digital converters is embedded into STM32L151xD and STM32L152xD devices with up to 40 external channels, performing conversions in single-shot or scan mode. In scan mode, automatic conversion is performed on a selected group of analog inputs with up to 28 external channels in a group.
The ADC can be served by the DMA controller.
An analog watchdog feature allows very precise monitoring of the converted voltage of one, some or all scanned channels. An interrupt is generated when the converted voltage is outside the programmed thresholds.
The events generated by the general-purpose timers (TIMx) can be internally connected to the ADC start triggers, to allow the application to synchronize A/D conversions and timers. An injection mode allows high priority conversions to be done by interrupting a scan mode which runs in as a background task.

The ADC includes a specific low-power mode. The converter is able to operate at maximum speed even if the CPU is operating at a very low frequency and has an auto-shutdown function. The ADC's runtime and analog front-end current consumption are thus minimized whatever the MCU operating mode.

### 3.11.1 Temperature sensor

The temperature sensor (TS) generates a voltage $\vee_{\text {SENSE }}$ that varies linearly with temperature.

The temperature sensor is internally connected to the ADC_IN16 input channel which is used to convert the sensor output voltage into a digital value.
The sensor provides good linearity but it has to be calibrated to obtain good overall accuracy of the temperature measurement. As the offset of the temperature sensor varies
from chip to chip due to process variation, the uncalibrated internal temperature sensor is suitable for applications that detect temperature changes only.

To improve the accuracy of the temperature sensor measurement, each device is individually factory-calibrated by ST. The temperature sensor factory calibration data are stored by ST in the system memory area, accessible in read-only mode. See Table 69: Temperature sensor calibration values.

### 3.11.2 Internal voltage reference ( $\mathrm{V}_{\text {REFINT }}$ )

The internal voltage reference ( $\mathrm{V}_{\text {REFINT }}$ ) provides a stable (bandgap) voltage output for the ADC and Comparators. $\mathrm{V}_{\text {REFINT }}$ is internally connected to the ADC_IN17 input channel. It enables accurate monitoring of the $\mathrm{V}_{\mathrm{DD}}$ value (when no external voltage, VREF+, is available for ADC). The precise voltage of $\mathrm{V}_{\text {REFINT }}$ is individually measured for each part by ST during production test and stored in the system memory area. It is accessible in readonly mode. See Table 15: Embedded internal reference voltage calibration values.

### 3.12 DAC (digital-to-analog converter)

The two 12-bit buffered DAC channels can be used to convert two digital signals into two analog voltage signal outputs. The chosen design structure is composed of integrated resistor strings and an amplifier in non-inverting configuration.

This dual digital Interface supports the following features:

- Two DAC converters: one for each output channel
- 8-bit or 12-bit monotonic output
- Left or right data alignment in 12-bit mode
- Synchronized update capability
- Noise-wave generation
- Triangular-wave generation
- Dual DAC channels, independent or simultaneous conversions
- DMA capability for each channel (including the underrun interrupt)
- External triggers for conversion
- Input reference voltage $\mathrm{V}_{\text {REF+ }}$

Eight DAC trigger inputs are used in the STM32L151xD and STM32L152xD devices. The DAC channels are triggered through the timer update outputs that are also connected to different DMA channels.

### 3.13 Operational amplifier

The STM32L151xD and STM32L152xD devices embed three operational amplifiers with external or internal follower routing capability (or even amplifier and filter capability with external components). When one operational amplifier is selected, one external ADC channel is used to enable output measurement.

The operational amplifiers feature:

- Low input bias current
- Low offset voltage
- Low-power mode
- Rail-to-rail input


### 3.14 Ultra-low-power comparators and reference voltage

The STM32L151xD and STM32L152xD devices embed two comparators sharing the same current bias and reference voltage. The reference voltage can be internal or external (coming from an I/O).

- One comparator with fixed threshold
- One comparator with rail-to-rail inputs, fast or slow mode. The threshold can be one of the following:
- DAC output
- External I/O
- Internal reference voltage $\left(\mathrm{V}_{\text {REFINT }}\right)$ or a sub-multiple (1/4, $1 / 2,3 / 4$ )

Both comparators can wake up from Stop mode, and be combined into a window comparator.

The internal reference voltage is available externally via a low-power / low-current output buffer (driving current capability of $1 \mu \mathrm{~A}$ typical).

### 3.15 System configuration controller and routing interface

The system configuration controller provides the capability to remap some alternate functions on different l/O ports.

The highly flexible routing interface allows the application firmware to control the routing of different I/Os to the TIM2, TIM3 and TIM4 timer input captures. It also controls the routing of internal analog signals to ADC1, COMP1 and COMP2 and the internal reference voltage $V_{\text {REFINT }}$.

### 3.16 Touch sensing

The STM32L151xD and STM32L152xD devices provide a simple solution for adding capacitive sensing functionality to any application. Thesedevices offer up to 34 capacitive sensing channels distributed over 11 analog I/O groups. Both software and timer capacitive sensing acquisition modes are supported.

Capacitive sensing technology is able to detect the presence of a finger near a sensor which is protected from direct touch by a dielectric (glass, plastic...). The capacitive variation introduced by the finger (or any conductive object) is measured using a proven implementation based on a surface charge transfer acquisition principle. It consists of charging the sensor capacitance and then transferring a part of the accumulated charges into a sampling capacitor until the voltage across this capacitor has reached a specific threshold. The capacitive sensing acquisition only requires few external components to
operate. This acquisition is managed directly by the GPIOs, timers and analog I/O groups (see Section 3.15: System configuration controller and routing interface).

Reliable touch sensing functionality can be quickly and easily implemented using the free STM32L1xx STMTouch touch sensing firmware library.

### 3.17 Timers and watchdogs

The ultra-low-power STM32L151xD and STM32L152xD devices include seven generalpurpose timers, two basic timers, and two watchdog timers.
Table 6 compares the features of the general-purpose and basic timers.
Table 6. Timer feature comparison

| Timer | Counter <br> resolution | Counter type | Prescaler factor | DMA <br> request <br> generation | Capture/compare <br> channels | Complementary <br> outputs |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| TIM2, <br> TIM3, <br> TIM4 | 16-bit | Up, down, <br> up/down | Any integer between <br> 1 and 65536 | Yes | 4 | No |
| TIM5 | 32-bit | Up, down, <br> up/down | Any integer between <br> 1 and 65536 | Yes | 4 | No |
| TIM9 | 16-bit | Up, down, <br> up/down | Any integer between <br> 1 and 65536 | No | 2 | No |
| TIM10, <br> TIM11 | 16-bit | Up | Any integer between <br> 1 and 65536 | No | 1 | No |
| TIM6, <br> TIM7 | 16-bit | Up | Any integer between <br> 1 and 65536 | Yes | 0 | No |

### 3.17.1 General-purpose timers (TIM2, TIM3, TIM4, TIM5, TIM9, TIM10 and TIM11)

There are seven synchronizable general-purpose timers embedded in the STM32L151xD and STM32L152xD devices (see Table 6 for differences).

## TIM2, TIM3, TIM4, TIM5

TIM2, TIM3, TIM4 are based on 16-bit auto-reload up/down counter. TIM5 is based on a 32bit auto-reload up/down counter. They include a 16-bit prescaler. They feature four independent channels each for input capture/output compare, PWM or one-pulse mode output. This gives up to 16 input captures/output compares/PWMs on the largest packages.

TIM2, TIM3, TIM4, TIM5 general-purpose timers can work together or with the TIM10, TIM11 and TIM9 general-purpose timers via the Timer Link feature for synchronization or event chaining. Their counter can be frozen in debug mode. Any of the general-purpose timers can be used to generate PWM outputs.
TIM2, TIM3, TIM4, TIM5 all have independent DMA request generation.
These timers are capable of handling quadrature (incremental) encoder signals and the digital outputs from 1 to 3 hall-effect sensors.

## TIM10, TIM11 and TIM9

TIM10 and TIM11 are based on a 16-bit auto-reload upcounter. TIM9 is based on a 16-bit auto-reload up/down counter. They include a 16-bit prescaler. TIM10 and TIM11 feature one independent channel, whereas TIM9 has two independent channels for input capture/output compare, PWM or one-pulse mode output. They can be synchronized with the TIM2, TIM3, TIM4, TIM5 full-featured general-purpose timers.

They can also be used as simple time bases and be clocked by the LSE clock source $(32.768 \mathrm{kHz})$ to provide time bases independent from the main CPU clock.

### 3.17.2 Basic timers (TIM6 and TIM7)

These timers are mainly used for DAC trigger generation. They can also be used as generic 16-bit time bases.

### 3.17.3 SysTick timer

This timer is dedicated to the OS, but could also be used as a standard downcounter. It is based on a 24-bit downcounter with autoreload capability and a programmable clock source. It features a maskable system interrupt generation when the counter reaches 0 .

### 3.17.4 Independent watchdog (IWDG)

The independent watchdog is based on a 12-bit downcounter and 8-bit prescaler. It is clocked from an independent 37 kHz internal RC and, as it operates independently of the main clock, it can operate in Stop and Standby modes. It can be used either as a watchdog to reset the device when a problem occurs, or as a free-running timer for application timeout management. It is hardware- or software-configurable through the option bytes. The counter can be frozen in debug mode.

### 3.17.5 Window watchdog (WWDG)

The window watchdog is based on a 7-bit downcounter that can be set as free-running. It can be used as a watchdog to reset the device when a problem occurs. It is clocked from the main clock. It has an early warning interrupt capability and the counter can be frozen in debug mode.

### 3.18 Communication interfaces

### 3.18.1 $\quad I^{2} \mathrm{C}$ bus

Up to two $I^{2} \mathrm{C}$ bus interfaces can operate in multimaster and slave modes. They can support standard and fast modes.

They support dual slave addressing (7-bit only) and both 7-and 10-bit addressing in master mode. A hardware CRC generation/verification is embedded.

They can be served by DMA and they support SM Bus 2.0/PM Bus.

### 3.18.2 Universal synchronous/asynchronous receiver transmitter (USART)

The three USART and two UART interfaces are able to communicate at speeds of up to 4 Mbit/s. They support IrDA SIR ENDEC and have LIN Master/Slave capability. The three USARTs provide hardware management of the CTS and RTS signals and are ISO 7816 compliant.

All USART/UART interfaces can be served by the DMA controller.

### 3.18.3 Serial peripheral interface (SPI)

Up to three SPIs are able to communicate at up to $16 \mathrm{Mbits} / \mathrm{s}$ in slave and master modes in full-duplex and half-duplex communication modes. The 3-bit prescaler gives 8 master mode frequencies and the frame is configurable to 8 bits or 16 bits. The hardware CRC generation/verification supports basic SD Card/MMC modes.

The SPIs can be served by the DMA controller.

### 3.18.4 Inter-integrated sound ( $\mathbf{I}^{2} \mathrm{~S}$ )

Two standard I2S interfaces (multiplexed with SPI2 and SPI3) are available. They can operate in master or slave mode, and can be configured to operate with a 16-/32-bit resolution as input or output channels. Audio sampling frequencies from 8 kHz up to 192 kHz are supported. When either or both of the I2S interfaces is/are configured in master mode, the master clock can be output to the external DAC/CODEC at 256 times the sampling frequency.

The I2Ss can be served by the DMA controller.

### 3.18.5 SDIO

An SD/SDIO/MMC host interface is available, that supports MultiMediaCard System Specification Version 4.2 in three different databus modes: 1-bit (default), 4-bit and 8-bit.

The interface allows data transfer at up to 24 MHz in 8-bit mode, and is compliant with the SD Memory Card Specification Version 2.0.

The SDIO Card Specification Version 2.0 is also supported with two different databus modes: 1-bit (default) and 4-bit.

The current version supports only one SD/SDIO/MMC4.2 card at any one time and a stack of MMC4.1 or previous.
In addition to SD/SDIO/MMC, this interface is fully compliant with the CE-ATA digital protocol Rev1.1.

### 3.18.6 Universal serial bus (USB)

The STM32L151xD and STM32L152xD devices embed a USB device peripheral compatible with the USB full-speed $12 \mathrm{Mbit} / \mathrm{s}$. The USB interface implements a full-speed (12 Mbit/s) function interface. It has software-configurable endpoint setting and supports suspend/resume. The dedicated 48 MHz clock is generated from the internal main PLL (the clock source must use a HSE crystal oscillator).

### 3.19 CRC (cyclic redundancy check) calculation unit

The CRC (cyclic redundancy check) calculation unit is used to get a CRC code from a 32-bit data word and a fixed generator polynomial.

Among other applications, CRC-based techniques are used to verify data transmission or storage integrity. In the scope of the EN/IEC 60335-1 standard, they offer a means of verifying the Flash memory integrity. The CRC calculation unit helps compute a signature of the software during runtime, to be compared with a reference signature generated at linktime and stored at a given memory location.

### 3.20 Development support

### 3.20.1 Serial wire JTAG debug port (SWJ-DP)

The Arm SWJ-DP interface is embedded, and is a combined JTAG and serial wire debug port that enables either a serial wire debug or a JTAG probe to be connected to the target. The JTAG JTMS and JTCK pins are shared with SWDAT and SWCLK, respectively, and a specific sequence on the JTMS pin is used to switch between JTAG-DP and SW-DP.

The JTAG port can be permanently disabled with a JTAG fuse.

### 3.20.2 Embedded Trace Macroce Il $^{\text {TM }}$

The Arm ${ }^{\circledR}$ Embedded Trace Macrocell provides a greater visibility of the instruction and data flow inside the CPU core by streaming compressed data at a very high rate from the STM32L151xD and STM32L152xD device through a small number of ETM pins to an external hardware trace port analyzer (TPA) device. The TPA is connected to a host computer using USB, Ethernet, or any other high-speed channel. Real-time instruction and data flow activity can be recorded and then formatted for display on the host computer running debugger software. TPA hardware is commercially available from common development tool vendors. It operates with third party debugger software tools.

## 4 Pin descriptions

Figure 3. STM32L15xRD LQFP64 pinout


1. This figure shows the package top view.

Figure 4. STM32L15xRD WLCSP64 ballout
CABC

1. This figure shows the package top view.

Figure 5. STM32L15xVD LQFP100 pinout


1. This figure shows the package top view.

Figure 6. STM32L15xQD UFBGA132 ballout


1. This figure shows the package top view.

Figure 7. STM32L15xZD LQFP144 pinout


1. This figure shows the package top view.

Table 7. Legend/abbreviations used in the pinout table

| Name |  | Abbreviation | Definition |
| :---: | :---: | :---: | :---: |
| Pin name |  | Unless otherwise specified in brackets below the pin name, the pin function during and after reset is the same as the actual pin name |  |
| Pin type |  | S Supply pin |  |
|  |  | 1 | Input only pin |
|  |  | I/O | Input / output pin |
| I/O structure |  | FT | 5 V tolerant I/O |
|  |  | TC | Standard $3.3 \mathrm{VI} / \mathrm{O}$ |
|  |  | B | Dedicated BOOTO pin |
|  |  | RST | Bidirectional reset pin with embedded weak pull-up resistor |
| Notes |  | Unless otherwise specified by a note, all I/Os are set as floating inputs during and after reset |  |
| Pin functions | Alternate functions | Functions selected through GPIOx_AFR registers |  |
|  | Additional functions | Functions directly selected/enabled through peripheral registers |  |

Table 8. STM32L151xD and STM32L152xD pin definitions

| Pins |  |  |  |  | Pin name | $\begin{aligned} & \underset{0}{E} \\ & \stackrel{0}{2} \\ & \underset{i n}{2} \end{aligned}$ |  | Main function ${ }^{(2)}$ (after reset) | Pin functions |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & \underset{\sim}{\underset{1}{4}} \\ & \text { O} \end{aligned}$ | $\begin{aligned} & \mathbb{N} \\ & \stackrel{\text { N}}{\mathbf{N}} \\ & \stackrel{1}{J} \end{aligned}$ | $\begin{aligned} & \text { 을 } \\ & \text { 닌 } \end{aligned}$ |  | $\begin{aligned} & \text { I } \\ & 00 \\ & 0 \\ & \vdots \\ & 3 \end{aligned}$ |  |  |  |  | Alternate functions | Additional functions |
| 1 | B2 | 1 | - | - | PE2 | I/O | FT | PE2 | TIM3_ETR/LCD_SEG38/ TRACECLK/FSMC_A23 | - |
| 2 | A1 | 2 | - | - | PE3 | I/O | FT | PE3 | TIM3_CH1/LCD_SEG39/ TRACEDO/FSMC_A19 | - |
| 3 | B1 | 3 | - | - | PE4 | I/O | FT | PE4 | TIM3_CH2/TRACED1 /FSMC_A20 | - |
| 4 | C2 | 4 | - | - | PE5 | I/O | FT | PE5 | TIM9_CH1/TRACED2 /FSMC_A21 | - |
| 5 | D2 | 5 | - | - | PE6WKUP3 | I/O | FT | PE6 | TIM9_CH2/TRACED3 | WKUP3/ RTC_TAMP3 |
| 6 | E2 | 6 | 1 | C6 | $\mathrm{V}_{\text {LCD }}{ }^{(3)}$ | S | - | $\mathrm{V}_{\text {LCD }}$ | - | - |

Table 8. STM32L151xD and STM32L152xD pin definitions (continued)

| Pins |  |  |  |  | Pin name |  | I/O structure | Main function ${ }^{(2)}$ (after reset) | Pin functions |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\begin{aligned} & \mathbb{N} \\ & \stackrel{N}{\overleftarrow{4}} \\ & \stackrel{M}{1} \end{aligned}$ |  | $\begin{aligned} & \text { U } \\ & \text { O! } \\ & \text { O } \end{aligned}$ |  |  |  |  |  | Alternate functions | Additional functions |
| 7 | C1 | 7 | 2 | C8 | PC13WKUP2 | I/O | FT | PC13 | - | WKUP2/ <br> RTC_TAMP1/ RTC_TS/ RTC_OUT |
| 8 | D1 | 8 | 3 | B8 | $\begin{gathered} \text { PC14- } \\ \text { OSC32_IN(4) } \end{gathered}$ | I/O | TC | PC14 | - | OSC32_IN |
| 9 | E1 | 9 | 4 | B7 | $\begin{gathered} \text { PC15- } \\ \text { OSC32_OUT } \end{gathered}$ | I/O | TC | PC15 | - | OSC32_OUT |
| 10 | D6 | - | - | - | PF0 | I/O | FT | PF0 | FSMC_A0 | - |
| 11 | D5 | - | - | - | PF1 | I/O | FT | PF1 | FSMC_A1 | - |
| 12 | D4 | - | - | - | PF2 | I/O | FT | PF2 | FSMC_A2 | - |
| 13 | E4 | - | - | - | PF3 | I/O | FT | PF3 | FSMC_A3 | - |
| 14 | F3 | - | - | - | PF4 | I/O | FT | PF4 | FSMC_A4 | - |
| 15 | F4 | - | - | - | PF5 | I/O | FT | PF5 | FSMC_A5 | - |
| 16 | F2 | 10 | - | - | $\mathrm{V}_{\text {SS_5 }}$ | S | - | $\mathrm{V}_{\text {SS_5 }}$ | - | - |
| 17 | G2 | 11 | - | - | $\mathrm{V}_{\text {DD_5 }}$ | S | - | $\mathrm{V}_{\text {DD_5 }}$ | - | - |
| 18 | G3 | - | - | - | PF6 | I/O | FT | PF6 | TIM5_CH1/TIM5_ETR | ADC_IN27 |
| 19 | G4 | - | - | - | PF7 | I/O | FT | PF7 | TIM5_CH2 | $\begin{aligned} & \text { ADC_IN28/ } \\ & \text { COMP1_INP } \end{aligned}$ |
| 20 | H4 | - | - | - | PF8 | I/O | FT | PF8 | TIM5_CH3 | $\begin{aligned} & \text { ADC_IN29/ } \\ & \text { COMP1_INP } \end{aligned}$ |
| 21 | J6 | - | - | - | PF9 | I/O | FT | PF9 | TIM5_CH4 | $\begin{aligned} & \text { ADC_IN30/ } \\ & \text { COMP1_INP } \end{aligned}$ |
| 22 | - | - | - | - | PF10 | I/O | FT | PF10 | - | $\begin{aligned} & \text { ADC_IN31/ } \\ & \text { COMP1_INP } \end{aligned}$ |
| 23 | F1 | 12 | 5 | D8 | $\begin{gathered} \text { PHO- } \\ \text { OSC_IN(5) } \end{gathered}$ | I/O | TC | PHO | - | OSC_IN |
| 24 | G1 | 13 | 6 | D7 | $\begin{gathered} \text { PH1- } \\ \text { OSC_OUT }{ }^{(5)} \end{gathered}$ | I/O | TC | PH1 | - | OSC_OUT |
| 25 | H2 | 14 | 7 | C7 | NRST | I/O | RST | NRST | - | - |
| 26 | H1 | 15 | 8 | E8 | PC0 | I/O | FT | PC0 | LCD_SEG18 | $\begin{aligned} & \text { ADC_IN10/ } \\ & \text { COMP1_INP } \end{aligned}$ |

Table 8. STM32L151xD and STM32L152xD pin definitions (continued)

| Pins |  |  |  |  | Pin name |  | 00.000000 | $\begin{array}{\|l} \text { Main } \\ \text { function } \\ \text { (2) } \\ \text { (after reset) } \end{array}$ | Pin functions |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | $\begin{aligned} & \text { oi } \\ & \text { a } \\ & \text { O } \end{aligned}$ | $\begin{aligned} & \text { U } \\ & \text { OM} \\ & \text { OU } \end{aligned}$ | $\begin{aligned} & \mathbf{~} \\ & 0 \\ & 0 \\ & 0 \\ & 3 \\ & 3 \end{aligned}$ |  |  |  |  | Alternate functions | Additional functions |
| 27 | J2 | 16 | 9 | F8 | PC1 | I/O | FT | PC1 | LCD_SEG19 | $\begin{gathered} \text { ADC_IN11/ } \\ \text { COMP1_INP } \\ \text { OPAMP3_VINP } \end{gathered}$ |
| 28 | - | 17 | 10 | D6 | PC2 | I/O | FT | PC2 | LCD_SEG20 | ADC_IN12/ COMP1_INP OPAMP3_VINM |
| - | J3 | - | - | - | PC2 | I/O | FT | PC2 | LCD_SEG20 | ADC_IN12/ COMP1_INP |
| - | K1 | - | - | - | $\begin{gathered} \hline \text { OPAMP3_VI } \\ \text { NM } \end{gathered}$ | 1 | - | OPAMP3 _VINM | - | - |
| 29 | K2 | 18 | 11 | F7 | PC3 | I/O | TC | PC3 | LCD_SEG21 | $\begin{gathered} \text { ADC_IN13/ } \\ \text { COMP1_INP/ } \\ \text { OPAMP3_VOUT } \end{gathered}$ |
| 30 | J1 | 19 | 12 | E7 | $\mathrm{V}_{\text {SSA }}$ | S | - | $\mathrm{V}_{\text {SSA }}$ | - | - |
| 31 | - | 20 | - | - | $\mathrm{V}_{\text {REF- }}$ | S | - | $V_{\text {REF- }}$ | - | - |
| 32 | L1 | 21 | - | - | $\mathrm{V}_{\text {REF+ }}$ | S | - | $V_{\text {REF+ }}$ | - | - |
| 33 | M1 | 22 | 13 | G8 | $\mathrm{V}_{\text {DDA }}$ | S | - | $\mathrm{V}_{\text {DDA }}$ | - | - |
| 34 | L2 | 23 | 14 | F6 | PA0-WKUP1 | I/O | FT | PAO | TIM2 CH1 ETR/ <br> TIM5_CH1/ USART2_CTS | WKUP1/ RTC_TAMP2/ ADC_INO/ COMP1_INP |
| 35 | M2 | 24 | 15 | E6 | PA1 | I/O | FT | PA1 | TIM2_CH2/TIM5_CH2/ USART2_RTS/ LCD_SEG0 | $\begin{gathered} \text { ADC_IN1/ } \\ \text { COMP1_INP/ } \\ \text { OPAMP1_VINP } \end{gathered}$ |
| 36 | - | 25 | 16 | H8 | PA2 | I/O | FT | PA2 | $\begin{gathered} \hline \text { TIM2_CH3/TIM5_CH3/ } \\ \text { TIM9_CH1// } \\ \text { USART2_TX/LCD_SEG1 } \end{gathered}$ | $\begin{gathered} \text { ADC_IN2/ } \\ \text { COMP1_INP/ } \\ \text { OPAMP1_VINM } \end{gathered}$ |
| - | K3 | - | - | - | PA2 | I/O | FT | PA2 | TIM2_CH3/TIM5_CH3/ TIM9_CH1/ USART2_TX/LCD_SEG1 | $\begin{gathered} \text { ADC_IN2/ } \\ \text { COMP1_INP } \end{gathered}$ |
| - | M3 | - | - | - | $\begin{gathered} \hline \text { OPAMP1_VI } \\ \text { NM } \end{gathered}$ | 1 | TC | $\begin{gathered} \text { OPAMP1_ } \\ \text { VINM } \end{gathered}$ | - | - |
| 37 | L3 | 26 | 17 | G7 | PA3 | I/O | TC | PA3 | $\begin{gathered} \text { TIM2_CH4/TIM5_CH4/ } \\ \text { TIM9_CH2/ } \\ \text { USART2_RX/LCD_SEG2 } \end{gathered}$ | $\begin{gathered} \text { ADC_IN3/ } \\ \text { COMP1_INP/ } \\ \text { OPAMP1_VOUT } \end{gathered}$ |

Table 8. STM32L151xD and STM32L152xD pin definitions (continued)

| Pins |  |  |  |  | Pin name |  | 003000000 | $\begin{aligned} & \text { Main } \\ & \text { function }{ }^{(2)} \\ & \text { (after reset) } \end{aligned}$ | Pin functions |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & J \\ & \underset{\sim}{i} \\ & \text { O} \end{aligned}$ |  | $\begin{aligned} & \text { 응 } \\ & \stackrel{1}{4} \\ & \text { O] } \end{aligned}$ | $\begin{aligned} & \text { U } \\ & \text { OU1 } \\ & \text { OU } \end{aligned}$ | $\begin{aligned} & \mathbf{~} \\ & 0 \\ & 0 \\ & 0 \\ & 3 \end{aligned}$ |  |  |  |  | Alternate functions | Additional functions |
| 38 | - | 27 | 18 | F5 | $\mathrm{V}_{\text {SS_4 }}$ | S | - | $\mathrm{V}_{\text {SS_4 }}$ | - | - |
| 39 | - | 28 | 19 | G6 | $\mathrm{V}_{\mathrm{DD} \text { _4 }}$ | S | - | $\mathrm{V}_{\mathrm{DD} \text { _4 }}$ | - | - |
| 40 | J4 | 29 | 20 | H7 | PA4 | I/O | TC | PA4 | SPI1 NSS/SPI3 NSS/ I2S3_WS/USART2_CK | ADC_IN4/ DAC_OUT1/ COMP1_INP |
| 41 | K4 | 30 | 21 | E5 | PA5 | I/O | TC | PA5 | TIM2_CH1_ETR/ SPI1 SCK SPI_SCK | ADC IN5/ DAC_OUT2/ COMP1_INP |
| 42 | L4 | 31 | 22 | G5 | PA6 | I/O | FT | PA6 | TIM3_CH1/TIM10_CH1/ SPI1_MISO/ LCD_SEG3 | ADC IN6/ COMP1_INP/ OPAMP2_VINP |
| 43 | - | 32 | 23 | G4 | PA7 | I/O | FT | PA7 | TIM3_CH2/TIM11_CH1/ <br> SPI1_MOSI/ <br> LCD_SEG4 | $\begin{gathered} \text { ADC_IN7/ } \\ \text { COMP1_INP/ } \\ \text { OPAMP2_VINM } \end{gathered}$ |
| - | J5 | - | - | - | PA7 | I/O | FT | PA7 | TIM3_CH2/TIM11_CH1/ <br> SPI1_MOSI/ <br> LCD_SEG4 | $\begin{gathered} \text { ADC_IN7/ } \\ \text { COMP1_INP } \end{gathered}$ |
| - | M4 | - | - | - | $\begin{gathered} \text { OPAMP2_VI } \\ \text { NM } \end{gathered}$ | 1 | TC | $\begin{gathered} \text { OPAMP2_V } \\ \text { INM } \end{gathered}$ | - | - |
| 44 | K5 | 33 | 24 | H6 | PC4 | I/O | FT | PC4 | LCD_SEG22 | ADC_IN14/ COMP1_INP |
| 45 | L5 | 34 | 25 | H5 | PC5 | I/O | FT | PC5 | LCD_SEG23 | ADC_IN15/ COMP1_INP |
| 46 | M5 | 35 | 26 | H4 | PB0 | I/O | TC | PB0 | TIM3_CH3/LCD_SEG5 | ```ADC_IN8/ COMP1_INP/ OPAMP2_VOUT/ VREF OUT``` |
| 47 | M6 | 36 | 27 | F4 | PB1 | I/O | FT | PB1 | TIM3_CH4/LCD_SEG6 | $\begin{gathered} \text { ADC_IN9/ } \\ \text { COMP1_INP/ } \\ \text { VREF_OUT } \end{gathered}$ |
| 48 | L6 | 37 | 28 | H3 | PB2 | I/O | FT | $\begin{gathered} \text { PB2/ } \\ \text { BOOT1 } \end{gathered}$ | B00T1 | ADC_INOb |
| 49 | K6 | - | - | - | PF11 | I/O | FT | PF11 | - | ADC_IN1b |
| 50 | J7 | - | - | - | PF12 | I/O | FT | PF12 | FSMC_A6 | ADC_IN2b |
| 51 | E3 | - | - | - | $\mathrm{V}_{\text {SS_6 }}$ | S | - | $\mathrm{V}_{\text {SS_6 }}$ | - | - |

Table 8. STM32L151xD and STM32L152xD pin definitions (continued)

| Pins |  |  |  |  | Pin name | $\begin{aligned} & \underset{O}{\mathcal{O}} \\ & \stackrel{0}{\lambda} \\ & \underset{\mathrm{I}}{\mathrm{I}} \end{aligned}$ | 00005000 | $\begin{aligned} & \text { Main } \\ & \text { function }{ }^{(2)} \\ & \text { (after reset) } \end{aligned}$ | Pin functions |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\begin{aligned} & \mathbb{N} \\ & \stackrel{N}{4} \\ & \text { M } \\ & \stackrel{1}{3} \end{aligned}$ | $\begin{aligned} & \text { 응 } \\ & \text { N } \\ & \text { OU } \end{aligned}$ | $\begin{aligned} & \text { U } \\ & \text { D } \\ & \text { O } \end{aligned}$ | $\begin{aligned} & \text { I } \\ & 0 \\ & 0 \\ & 0 \\ & 3 \\ & 3 \end{aligned}$ |  |  |  |  | Alternate functions | Additional functions |
| 52 | H3 | - | - | - | $\mathrm{V}_{\text {DD_6 }}$ | S | - | $\mathrm{V}_{\text {DD_6 }}$ | - | - |
| 53 | K7 | - | - | - | PF13 | I/O | FT | PF13 | FSMC_A7 | ADC_IN3b |
| 54 | J8 | - | - | - | PF14 | I/O | FT | PF14 | FSMC_A8 | ADC_IN6b |
| 55 | J9 | - | - | - | PF15 | I/O | FT | PF15 | FSMC_A9 | ADC_IN7b |
| 56 | H9 | - | - | - | PG0 | I/O | FT | PG0 | FSMC_A10 | ADC_IN8b |
| 57 | G9 | - | - | - | PG1 | I/O | FT | PG1 | FSMC_A11 | ADC_IN9b |
| 58 | M7 | 38 | - | - | PE7 | I/O | TC | PE7 | FSMC_D4 | $\begin{aligned} & \text { ADC_IN22/ } \\ & \text { COMP1_INP } \end{aligned}$ |
| 59 | L7 | 39 | - | - | PE8 | 1/O | TC | PE8 | FSMC_D5 | $\begin{aligned} & \text { ADC_IN23/ } \\ & \text { COMP1_INP } \end{aligned}$ |
| 60 | M8 | 40 | - | - | PE9 | I/O | TC | PE9 | TIM2_CH1_ETR /FSMC_D6 | $\begin{gathered} \text { ADC_IN24/ } \\ \text { COMP1_INP } \end{gathered}$ |
| 61 | - | - | - | - | $\mathrm{V}_{\text {SS_7 }}$ | S | - | $\mathrm{V}_{\text {SS_7 }}$ | - | - |
| 62 | - | - | - | - | $\mathrm{V}_{\text {DD_7 }}$ | S | - | $\mathrm{V}_{\mathrm{DD} \text {-7 }}$ | - | - |
| 63 | L8 | 41 | - | - | PE10 | I/O | TC | PE10 | TIM2_CH2/FSMC_D7 | $\begin{aligned} & \text { ADC_IN25/ } \\ & \text { COMP1_INP } \end{aligned}$ |
| 64 | M9 | 42 | - | - | PE11 | I/O | FT | PE11 | TIM2_CH3/FSMC_D8 | - |
| 65 | L9 | 43 | - | - | PE12 | I/O | FT | PE12 | $\begin{gathered} \text { TIM2_CH4/SPI1_NSS } \\ \text { /FSMC_D9 } \end{gathered}$ | - |
| 66 | M10 | 44 | - | - | PE13 | I/O | FT | PE13 | SPI1_SCK/FSMC_D10 | - |
| 67 | M11 | 45 | - | - | PE14 | I/O | FT | PE14 | SPI1_MISO/FSMC_D11 | - |
| 68 | M12 | 46 | - | - | PE15 | I/O | FT | PE15 | SPI1_MOSI/FSMC_D12 | - |
| 69 | L10 | 47 | 29 | G3 | PB10 | I/O | FT | PB10 | TIM2_CH3/I2C2_SCL/ USART3_TXI LCD_SEG10 | - |
| 70 | L11 | 48 | 30 | F3 | PB11 | I/O | FT | PB11 | TIM2_CH4/ I2C2_SDA/ USART3_RX/ LCD_SEG11 | - |
| 71 | F12 | 49 | 31 | H2 | $\mathrm{V}_{\text {SS_1 }}$ | S | - | $\mathrm{V}_{\text {SS_1 }}$ | - | - |
| 72 | G12 | 50 | 32 | H1 | $\mathrm{V}_{\mathrm{DD} \text { _1 }}$ | S | - | $\mathrm{V}_{\mathrm{DD} \text { _1 }}$ | - | - |
| 73 | L12 | 51 | 33 | G2 | PB12 | I/O | FT | PB12 | TIM10_CH1/I2C2_SMBA/ SPI2_NSS/ I2S2_WS/ USART3_CK/ LCD_SEG12 | ADC_IN18/ COMP1_INP |

Table 8. STM32L151xD and STM32L152xD pin definitions (continued)

| Pins |  |  |  |  | Pin name |  |  | $\begin{array}{\|c} \text { Main }^{(2)} \\ \text { function } \\ \text { (after reset) } \end{array}$ | Pin functions |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\begin{aligned} & \text { N } \\ & \stackrel{\text { N}}{4} \\ & \text { M } \\ & \stackrel{1}{J} \end{aligned}$ | $\begin{aligned} & \text { oi } \\ & \text { a } \\ & \text { O } \end{aligned}$ | $\begin{aligned} & \text { U } \\ & 0 \\ & \text { 010 } \\ & \text { O} \end{aligned}$ | $\begin{aligned} & \mathbf{0} \\ & 0 \\ & 0 \\ & \vdots \\ & 3 \end{aligned}$ |  |  |  |  | Alternate functions | Additional functions |
| 74 | K12 | 52 | 34 | G1 | PB13 | I/O | FT | PB13 | $\begin{gathered} \hline \text { TIM9_CH1/SPI2_SCK/ } \\ \text { I2S2_CK/ } \\ \text { USART3_CTS/ } \\ \text { LCD_SEG13 } \end{gathered}$ | ADC IN19/ COMP1_INP |
| 75 | K11 | 53 | 35 | F2 | PB14 | I/O | FT | PB14 | TIM9_CH2/SPI2_MISO/ USART3_RTS/ LCD_SEG14 | $\begin{aligned} & \text { ADC_IN20/ } \\ & \text { COMP1_INP } \end{aligned}$ |
| 76 | K10 | 54 | 36 | F1 | PB15 | I/O | FT | PB15 | $\begin{gathered} \text { TIM11_CH1/SPI2_MOSI/ } \\ \text { I2S2_SD/ } \\ \text { LCD_SEG15 } \end{gathered}$ | $\begin{aligned} & \text { ADC_IN21/ } \\ & \text { COMP1_INP/ } \\ & \text { RTC_REFIN } \end{aligned}$ |
| 77 | K9 | 55 | - | - | PD8 | I/O | FT | PD8 | USART3_TX/LCD_SEG28/ FSMC_D13 | - |
| 78 | K8 | 56 | - | - | PD9 | I/O | FT | PD9 | USART3_RX/LCD_SEG29/ FSMC_D14 | - |
| 79 | J12 | 57 | - | - | PD10 | I/O | FT | PD10 | USART3_CK/LCD_SEG30/ FSMC_D15 | - |
| 80 | J11 | 58 | - | - | PD11 | I/O | FT | PD11 | USART3_CTS/LCD_SEG31 /FSMC_A16 | - |
| 81 | J10 | 59 | - | - | PD12 | I/O | FT | PD12 | TIM4_CH1/USART3_RTS/ <br> LCD_SEG32/ <br> FSMC_A17 | - |
| 82 | H12 | 60 | - | - | PD13 | I/O | FT | PD13 | TIM4_CH2/LCD_SEG33/ <br> FSMC_A18 | - |
| 83 | - | - | - | - | $\mathrm{V}_{\text {SS_8 }}$ | S | - | $\mathrm{V}_{\text {SS } ~ 8 ~}$ | - | - |
| 84 | - | - | - | - | $\mathrm{V}_{\mathrm{DD} \text { _8 }}$ | S | - | $\mathrm{V}_{\mathrm{DD} \text { _8 }}$ | - | - |
| 85 | H11 | 61 | - | - | PD14 | I/O | FT | PD14 | TIM4_CH3/LCD_SEG34/ FSMC_D0 | - |
| 86 | H10 | 62 | - | - | PD15 | I/O | FT | PD15 | TIM4_CH4/LCD_SEG35 /FSMC_D1 | - |
| 87 | G10 | - | - | - | PG2 | I/O | FT | PG2 | FSMC_A12 | ADC_IN10b |
| 88 | F9 | - | - | - | PG3 | I/O | FT | PG3 | FSMC_A13 | ADC_IN11b |
| 89 | F10 | - | - | - | PG4 | I/O | FT | PG4 | FSMC_A14 | ADC_IN12b |
| 90 | E9 | - | - | - | PG5 | I/O | FT | PG5 | FSMC_A15 | - |
| 91 | - | - | - | - | PG6 | I/O | FT | PG6 | - | - |

Table 8. STM32L151xD and STM32L152xD pin definitions (continued)

| Pins |  |  |  |  | Pin name |  | 000000000 | Main function ${ }^{(2)}$ (after reset) | Pin functions |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\pm$ $\vdots$ $\vdots$ $\vdots$ | $\begin{aligned} & \mathbb{N} \\ & \stackrel{N}{\overleftarrow{4}} \\ & \stackrel{M}{1} \end{aligned}$ | 은 <br> 믄 <br> 1 | $\begin{aligned} & \text { U } \\ & \text { O! } \\ & \text { O } \end{aligned}$ |  |  |  |  |  | Alternate functions | Additiona functions |
| 92 | - | - | - | - | PG7 | I/O | FT | PG7 | - | - |
| 93 | - | - | - | - | PG8 | I/O | FT | PG8 | - | - |
| 94 | F6 | - | - | - | $\mathrm{V}_{\text {SS_9 }}$ | S | - | $\mathrm{V}_{\text {SS_9 }}$ | - | - |
| 95 | G6 | - | - | - | $\mathrm{V}_{\text {DD_9 }}$ | S | - | $\mathrm{V}_{\mathrm{DD} \text { _9 }}$ | - | - |
| 96 | E12 | 63 | 37 | E1 | PC6 | I/O | FT | PC6 | TIM3_CH1/I2S2_MCK/ LCD_SEG24/SDIO_D6 | - |
| 97 | E11 | 64 | 38 | E2 | PC7 | I/O | FT | PC7 | TIM3_CH2/I2S3_MCK/ LCD_SEG25/SDIO_D7 | - |
| 98 | E10 | 65 | 39 | E3 | PC8 | I/O | FT | PC8 | $\begin{gathered} \text { TIM3_CH3/LCD_SEG26/ } \\ \text { SDIO_D0 } \end{gathered}$ | - |
| 99 | D12 | 66 | 40 | D1 | PC9 | I/O | FT | PC9 | $\begin{gathered} \hline \text { TIM3_CH4/LCD_SEG27/ } \\ \text { SDIO_D1 } \end{gathered}$ | - |
| 100 | D11 | 67 | 41 | E4 | PA8 | I/O | FT | PA8 | USART1_CK/MCO/ LCD_COMO | - |
| 101 | D10 | 68 | 42 | D2 | PA9 | I/O | FT | PA9 | USART1_TX / LCD_COM1 | - |
| 102 | C12 | 69 | 43 | D3 | PA10 | I/O | FT | PA10 | USART1_RX / LCD_COM2 | - |
| 103 | B12 | 70 | 44 | C1 | PA11 | I/O | FT | PA11 | USART1_CTS/ SPI1_MISO | USB_DM |
| 104 | A12 | 71 | 45 | C2 | PA12 | I/O | FT | PA12 | USART1_RTS/ SPI1_MOSI | USB_DP |
| 105 | A11 | 72 | 46 | D4 | PA13 | I/O | FT | JTMSSWDIO | JTMS-SWDIO | - |
| 106 | C11 | 73 | - | - | PH2 | I/O | FT | PH2 | FSMC_A22 | - |
| 107 | F11 | 74 | 47 | B1 | $\mathrm{V}_{\text {SS_2 }}$ | S | - | $\mathrm{V}_{\text {SS_2 }}$ | - | - |
| 108 | G11 | 75 | 48 | A1 | $\mathrm{V}_{\mathrm{DD} \text { _2 }}$ | S | - | $\mathrm{V}_{\mathrm{DD} \text { _2 }}$ | - | - |
| 109 | A10 | 76 | 49 | B2 | PA14 | I/O | FT | JTCKSWCLK | JTCK-SWCLK | - |
| 110 | A9 | 77 | 50 | C3 | PA15 | I/O | FT | JTDI | TIM2_CH1_ETR/ <br> SPI1_NSS/SPI3_NSS/ I2S3_WS/LCD_SEG17/ JTDI | - |

Table 8. STM32L151xD and STM32L152xD pin definitions (continued)

| Pins |  |  |  |  | Pin name |  |  | $\begin{array}{\|c} \text { Main }^{(2)} \\ \text { function } \\ \text { (after reset) } \end{array}$ | Pin functions |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | $\begin{aligned} & \text { oi } \\ & \text { a } \\ & \text { O } \end{aligned}$ | $\begin{aligned} & \text { U } \\ & 0 \\ & \text { 010 } \\ & \text { O} \end{aligned}$ | $\begin{aligned} & \text { O } \\ & 0 \\ & \text { O} \\ & \vdots \\ & 3 \end{aligned}$ |  |  |  |  | Alternate functions | Additional functions |
| 111 | B11 | 78 | 51 | A2 | PC10 | 1/O | FT | PC10 | SPI3_SCK/I2S3_CK/ USART3_TX/UART4_TX/ LCD_SEG28/LCD_SEG40/ LCD_COM4/ SDIO D2 | - |
| 112 | C10 | 79 | 52 | B3 | PC11 | 1/O | FT | PC11 | $\begin{gathered} \text { SPI3_MISO/USART3_RXI } \\ \text { UART4_RX/ } \\ \text { LCD_SEG29/LCD_SEG41/ } \\ \text { LCD_COM5// } \\ \text { SDIO_D3 } \end{gathered}$ | - |
| 113 | B10 | 80 | 53 | C4 | PC12 | 1/O | FT | PC12 | SPI3_MOSI/I2S3_SD/ USART3_CK/ <br> UART5_TX/LCD_SEG30/ LCD_SEG42/ <br> LCD_COM6/SDIO_CK | - |
| 114 | C9 | 81 | - | - | PD0 | I/O | FT | PDO | TIM9_CH1/SPI2_NSS/ I2S2 WS/ FSMC D2 | - |
| 115 | B9 | 82 | - | - | PD1 | I/O | FT | PD1 | $\begin{gathered} \text { SPI2_SCK/I2S2_CK } \\ \text { /FSMC_D3 } \end{gathered}$ | - |
| 116 | C8 | 83 | 54 | A3 | PD2 | 1/O | FT | PD2 | TIM3_ETR/ UART5_RX/LCD_SEG31/ LCD_SEG43/LCD_COM7/ SDIO_CMD | - |
| 117 | B8 | 84 | - | - | PD3 | I/O | FT | PD3 | SPI2_MISO/USART2_CTS/ FSMC_CLK | - |
| 118 | B7 | 85 | - | - | PD4 | 1/O | FT | PD4 | SPI2_MOSI/I2S2_SD/ USART2_RTS/ FSMC_NOE | - |
| 119 | A6 | 86 | - | - | PD5 | 1/0 | FT | PD5 | USART2_TX/FSMC_NWE | - |
| 120 | F7 | - | - | - | $\mathrm{V}_{\text {SS_10 }}$ | S | - | $V_{\text {SS_10 }}$ | - | - |
| 121 | G7 | - | - | - | $\mathrm{V}_{\mathrm{DD} \text { _10 }}$ | S | - | $\mathrm{V}_{\mathrm{DD} \text { _10 }}$ | - | - |
| 122 | B6 | 87 | - | - | PD6 | I/O | FT | PD6 | USART2 RX /FSMC_NWAIT | - |
| 123 | A5 | 88 | - | - | PD7 | I/O | FT | PD7 | TIM9_CH2/USART2_CK /FSMC_NE1 | - |
| 124 | D9 | - | - | - | PG9 | 1/O | FT | PG9 | FSMC_NE2 | - |

Table 8. STM32L151xD and STM32L152xD pin definitions (continued)

| Pins |  |  |  |  | Pin name |  |  | $\begin{gathered} \text { Main } \\ \text { function }{ }^{(2)} \\ \text { (after reset) } \end{gathered}$ | Pin functions |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & \underset{寸}{J} \\ & \stackrel{1}{U} \\ & \underset{O}{1} \end{aligned}$ |  | 은 <br> 믄 <br> 1 | $\begin{aligned} & \text { U } \\ & \text { OL} \\ & \text { OU } \end{aligned}$ | $\begin{aligned} & \text { O } \\ & 0 \\ & 0 \\ & 0 \\ & 3 \end{aligned}$ |  |  |  |  | Alternate functions | Additional functions |
| 125 | D8 | - | - | - | PG10 | I/O | FT | PG10 | FSMC_NE3 | - |
| 126 | - | - | - | - | PG11 | I/O | FT | PG11 | - | - |
| 127 | D7 | - | - | - | PG12 | I/O | FT | PG12 | FSMC_NE4 | - |
| 128 | C7 | - | - | - | PG13 | I/O | FT | PG13 | FSMC_A24 | - |
| 129 | C6 | - | - | - | PG14 | I/O | FT | PG14 | FSMC_A25 | - |
| 130 | - | - | - | - | $V_{\text {SS_11 }}$ | S | - | $V_{\text {SS_11 }}$ | - | - |
| 131 | - | - | - | - | $\mathrm{V}_{\text {DD_11 }}$ | S | - | $\mathrm{V}_{\text {DD_11 }}$ | - | - |
| 132 | - | - | - | - | PG15 | I/O | FT | PG15 | - | - |
| 133 | A8 | 89 | 55 | A4 | PB3 | I/O | FT | JTDO | TIM2_CH2/SPI1_SCK/ SPI3_SCK/ I2S3_CK/ LCD_SEG7/JTDO | COMP2_INM |
| 134 | A7 | 90 | 56 | B4 | PB4 | I/O | FT | NJTRST | TIM3_CH1/SPI1_MISO/ SPI3_MISO/ LCD_SEG8/NJTRST | COMP2_INP |
| 135 | C5 | 91 | 57 | A5 | PB5 | I/O | FT | PB5 | ```TIM3_CH2/I2C1_SMBA/``` | COMP2_INP |
| 136 | B5 | 92 | 58 | B5 | PB6 | I/O | FT | PB6 | TIM4_CH1/I2C1_SCL/ USART1_TX/ | COMP2_INP |
| 137 | B4 | 93 | 59 | C5 | PB7 | I/O | FT | PB7 | TIM4 CH2/I2C1 SDA/ USART1_RX/FSMC_NADV | $\begin{gathered} \hline \text { COMP2_INP/ } \\ \text { PVD_IN } \end{gathered}$ |
| 138 | A4 | 94 | 60 | A6 | BOOT0 | 1 | B | BOOT0 | - | - |
| 139 | A3 | 95 | 61 | D5 | PB8 | I/O | FT | PB8 | $\begin{gathered} \text { TIM4_CH3/TIM10_CH1/ } \\ \text { I2C1_SCL/ } \\ \text { LCD_SEG16/SDIO_D4 } \end{gathered}$ | - |
| 140 | B3 | 96 | 62 | B6 | PB9 | I/O | FT | PB9 |  | - |
| 141 | C3 | 97 | - | - | PE0 | I/O | FT | PE0 | TIM4_ETR/TIM10_CH1/ LCD_SEG36/FSMC_NBLO | - |
| 142 | A2 | 98 | - | - | PE1 | I/O | FT | PE1 | TIM11_CH1/LCD_SEG37 /FSMC_NBL1 | - |

Table 8. STM32L151xD and STM32L152xD pin definitions (continued)

| Pins |  |  |  |  | Pin name |  |  | Main function ${ }^{(2)}$ (after reset) | Pin functions |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | $\begin{aligned} & \text { 응 } \\ & \frac{1}{4} \\ & 0 \end{aligned}$ | $\begin{aligned} & \text { む } \\ & \text { OR } \\ & \text { O } \\ & \hline \end{aligned}$ | $\begin{aligned} & \mathbf{~} \\ & 0 \\ & 0 \\ & 0 \\ & 3 \\ & 3 \end{aligned}$ |  |  |  |  | Alternate functions | Additional functions |
| 143 | D3 | 99 | 63 | A7 | $\mathrm{V}_{\text {SS_3 }}$ | S | - | $\mathrm{V}_{\text {SS_3 }}$ | - | - |
| 144 | C4 | 100 | 64 | A8 | $\mathrm{V}_{\mathrm{DD}} 3$ | S | - | $\mathrm{V}_{\mathrm{DD}}$ 3 | - | - |

1. $I=$ input, $O=$ output, $S=$ supply.
2. Function availability depends on the chosen device.
3. Applicable to STM32L152xD devices only. In STM32L151xD devices, this pin should be connected to $V_{D D}$.
4. The PC14 and PC15 I/Os are only configured as OSC32_IN/OSC32_OUT when the LSE oscillator is ON (by setting the LSEON bit in the RCC_CSR register). The LSE oscillator pins OSC32_IN/OSC32_OUT can be used as general-purpose PH0/PH1 I/Os, respectively, when the LSE oscillator is off (after reset, the LSE oscillator is off). The LSE has priority over the GPIO function. For more details, refer to Using the OSC32_IN/OSC32_OUT pins as GPIO PC14/PC15 port pins section in the STM32L151xx, STM32L152xx and STM32L162xx reference manual (RM0038).
5. The PH 0 and PH 1 I/Os are only configured as OSC_IN/OSC_OUT when the HSE oscillator is ON (by setting the HSEON bit in the RCC_CR register). The HSE oscillator pins OSC_IN/OSC_OUT can be used as general-purpose PH0/PH1 I/Os, respectively, when the HSE oscillator is off ( after reset, the HSE oscillator is off ). The HSE has priority over the GPIO function.
Alternate functions
Table 9. Alternate function input/output

Table 9. Alternate function input/output (continued)

| Port name | Digital alternate function number |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | AFIOO | AFIO1 | AFIO2 | AFIO3 | AFIO4 | AFIO5 | AFIO6 | AFIO7 | AFIO8 | .. | AFIO11 | AFIO12 | .. AFIO14 | AFIO15 |
|  | Alternate function |  |  |  |  |  |  |  |  |  |  |  |  |  |
|  | SYSTEM | TIM2 | TIM3/4/5 | $\begin{aligned} & \text { TIM9/ } \\ & \text { 10/11 } \end{aligned}$ | I2C1/2 | SPI1/2 | SPI3 | USART1/2/3 | $\begin{array}{\|c\|} \hline \text { UART4I } \\ 5 \end{array}$ |  | LCD | $\begin{aligned} & \text { FSMC/ } \\ & \text { SDIO } \end{aligned}$ | CPRI | SYSTEM |
| PA12 | - | - | - | - | - | SPI1_MOSI | - | USART1_RTS | - | - | - | - | - TIMx_IC1 | $\begin{array}{\|l\|} \text { EVENT } \\ \text { OUT } \end{array}$ |
| PA13 | JTMS-SWDIO | - | - | - | - | - | - | - | - | - | - | - | - TIMx_IC2 | EVENT OUT |
| PA14 | JTCK-SWCLK | - | - | - | - | - | - | - | - | - | - | - | - TIMx_IC3 | $\left\lvert\, \begin{aligned} & \text { EVEN } \\ & \text { TOUT } \end{aligned}\right.$ |
| PA15 | JTDI | TIM2_CH1_ETR | - | - | - | SPI1_NSS | $\begin{array}{\|l} \text { SPI3_NSS } \\ \text { I2S3_WS } \end{array}$ | - | - | -S | SEG17 | - | - TIMx_IC4 | EVEN TOUT |
| PBO | - | - | TIM3_CH3 | - | - | - | - | - | - | - S | SEG5 | - | - - | EVEN TOUT |
| PB1 | - | - | TIM3_CH4 | - | - | - | - | - | - | -S | SEG6 | - | - - | EVENT OUT |
| PB2 | BOOT1 | - | - | - | - | - | - | - | - | - | - | - | - - | $\begin{array}{\|l} \hline \text { EVENT } \\ \text { OUT } \end{array}$ |
| PB3 | JTDO | TIM2_CH2 | - | - | - | SPI1_SCK | $\begin{aligned} & \text { SPI3_SCK } \\ & \text { I2S3_CK } \end{aligned}$ | - | - | -S | SEG7 | - | - - | $\begin{aligned} & \text { EVENT } \\ & \text { OUT } \end{aligned}$ |
| PB4 | NJTRST | - | TIM3_CH1 | - | - | SPI1_MISO | SPI3_MISO | - | - | -S | SEG8 | - | - - | $\begin{array}{\|l\|l\|} \hline \text { EVENT } \\ \text { OUT } \end{array}$ |
| PB5 | - | - | TIM3_CH2 | - | $\begin{aligned} & 12 \mathrm{IC1} \\ & \text { SMBA } \end{aligned}$ | SPI1_MOSI | $\begin{array}{\|l\|} \hline \text { SPI3_MOSI } \\ \text { I2S3_SD } \end{array}$ | - | - | -S | SEG9 | - | - - | EVENT OUT |
| PB6 | - | - | TIM4_CH1 | - | 12C1_SCL | - | - | USART1_TX | - | - | - | - | - - | $\begin{array}{\|l\|} \text { EVENT } \\ \hline \text { OUT } \\ \hline \end{array}$ |
| PB7 | - | - | TIM4_CH2 | - | 12C1_SDA | - | - | USART1_RX | - | - | - | NADV | - - | $\begin{array}{\|l\|} \text { EVENT } \\ \text { OUT } \end{array}$ |
| PB8 | - | - | TIM4_CH3 | TIM10_CH1 | 12C1_SCL | - | - | - | - | -S | SEG16 | SDIO_D4 | - - | $\begin{aligned} & \text { EVENT } \\ & \text { OUT } \\ & \hline \end{aligned}$ |
| PB9 | - | - | TIM4_CH4 | TIM11_CH1 | 12C1_SDA | - | - | - | - |  | СОМ3 | SDIO_D5 | - - | $\begin{aligned} & \text { EVENT } \\ & \text { OUT } \end{aligned}$ |
| PB10 | - | TIM2_CH3 | - | - | 12C2_SCL | - | - | USART3_TX | - |  | SEG10 | - | - - | $\begin{array}{\|l\|} \text { EVENT } \\ \text { OUT } \end{array}$ |

Table 9. Alternate function input/output (continued)

| Port name | Digital alternate function number |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | AFIOO | AFIO1 | AFIO2 | AFIO3 | AFIO4 | AFIO5 | AFIO6 | AFIO7 | AFIO8 |  | AFIO11 | AFIO12 | .. | AFIO14 | AFIO15 |
|  | Alternate function |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
|  | SYSTEM | TIM2 | TIM3/4/5 | $\begin{aligned} & \text { TIM9/ } \\ & \text { 10/11 } \end{aligned}$ | 12C1/2 | SPI1/2 | SPI3 | USART1/2/3 | $\begin{array}{\|c} \text { UART4I } \\ 5 \end{array}$ |  | LCD | $\begin{aligned} & \text { FSMC/ } \\ & \text { SDIO } \end{aligned}$ |  | CPRI | SYSTEM |
| PB11 | - | TIM2_CH4 | - | - | 12C2_SDA | - | - | USART3_RX | - |  | SEG11 | - | - | - | EVENT OUT |
| PB12 | - | - | - | TIM10_CH1 | 12C2_SMBA | SPI2_NSS I2S2_WS | - | USART3_CK | - |  | SEG12 | - | - | - | EVENT OUT |
| PB13 | - | - | - | TIM9_CH1 | - | SPI2_SCK I2S2_CK | - | USART3_CTS | - |  | SEG13 | - | - | - | EVENT OUT |
| PB14 | - | - | - | TIM9_CH2 | - | SPI2_MISO | - | USART3_RTS | - |  | SEG14 | - | - | - | EVENT OUT |
| PB15 | - | - | - | TIM11_CH1 | - | $\begin{aligned} & \text { SPI2_MOSI } \\ & \text { I2S2_SD } \end{aligned}$ | - | - | - |  | SEG15 | - | - | - | EVENT OUT |
| PC0 | - | - | - | - | - | - | - | - | - |  | SEG18 | - |  | TIMx_IC1 | EVENT OUT |
| PC1 | - | - | - | - | - | - | - | - | - |  | SEG19 | - |  | TIMx_IC2 | EVENT OUT |
| PC2 | - | - | - | - | - | - | - | - | - |  | SEG20 | - |  | TIMx_IC3 | EVENT OUT |
| PC3 | - | - | - | - | - | - | - | - | - |  | SEG21 | - |  | IMx_IC4 | $\begin{aligned} & \text { EVENT } \\ & \text { OUT } \end{aligned}$ |
| PC4 | - | - | - | - | - | - | - | - | - |  | SEG22 | - |  | TIMx_IC1 | $\begin{aligned} & \text { EVENT } \\ & \text { OUT } \end{aligned}$ |
| PC5 | - | - | - | - | - | - | - | - | - |  | SEG23 | - |  | TIMx_IC2 | EVENT OUT |
| PC6 | - | - | TIM3_CH1 | - | - | I2S2_MCK | - | - | - |  | SEG24 | SDIO_D6 |  | TIMx_IC3 | EVENT OUT |
| PC7 | - | - | TIM3_CH2 | - | - | - | I2S3_MCK | - | - |  | SEG25 | SDIO_D7 |  | TIMx_IC4 | $\begin{aligned} & \text { EVENT } \\ & \text { OUT } \end{aligned}$ |
| PC8 | - | - | TIM3_CH3 | - | - | - | - | - | - |  | SEG26 | SDIO_D0 |  | TIMx_IC1 | EVENT OUT |
| PC9 | - | - | TIM3_CH4 | - | - | - | - | - | - |  | SEG27 | SDIO_D1 |  | TIMx_IC2 | EVENT OUT |

Table 9. Alternate function input/output (continued)

| Port name | Digital alternate function number |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | AFIOO | AFIO1 | AFIO2 | AFIO3 | AFIO4 | AFIO5 | AFIO6 | AFIO7 | AFIO8 | .. AFIO11 | AFIO12 | .. AFIO14 | AFIO15 |
|  | Alternate function |  |  |  |  |  |  |  |  |  |  |  |  |
|  | SYSTEM | TIM2 | TIM3/4/5 | $\begin{aligned} & \text { TIM9/ } \\ & \text { 10/11 } \end{aligned}$ | 12C1/2 | SPI1/2 | SPI3 | USART1/2/3 | $\begin{array}{\|c\|} \hline \text { UART4I } \\ 5 \end{array}$ | LCD | $\begin{aligned} & \text { FSMC/ } \\ & \text { SDIO } \end{aligned}$ | CPRI | SYSTEM |
| PC10 | - | - | - | - | - | - | $\begin{aligned} & \text { SPI3_SCK } \\ & \text { 12S3_CK } \end{aligned}$ | USART3_TX | UART4_TX | $\begin{aligned} & \hline \text { COM4/ } \\ & -\mathrm{SEG28/} \\ & \text { SEG40 } \end{aligned}$ | SDIO_D2 | - TIMx_IC3 | EVENT <br> OUT |
| PC11 | - | - | - | - | - | - | SPI3_MISO | USART3_RX | UART4_RX | $\begin{array}{\|l} \text { COM5/ } \\ -\mathrm{SEG29} \\ \hline \text { SEG41 } \end{array}$ | SDIO_D3 | - TIMx_IC4 | $\begin{aligned} & \text { EVENT } \\ & \text { OUT } \end{aligned}$ |
| PC12 | - | - | - | - | - | - | $\left\lvert\, \begin{aligned} & \text { SPI3_MOSI } \\ & \text { I2S3_SD } \end{aligned}\right.$ | USART3_CK | UART5_TX | $\text { - } \begin{aligned} & \text { COM6/ } \\ & \text { - SEG30/ } \\ & \text { SEG42 } \end{aligned}$ | SDIO_CK | - TIMx_IC1 | $\begin{aligned} & \text { EVENT } \\ & \text { OUT } \end{aligned}$ |
| PC13-WKUP2 | - | - | - | - | - | - | - | - | - | - - | - | - TIMx_IC2 | EVENT OUT |
| PC14 OSC32_IN | - | - | - | - | - | - | - | - | - | - - | - | - TIMx_IC3 | EVENT OUT |
| PC15 OSC32_OUT | - | - | - | - | - | - | - | - | - | - - | - | - TIMx_IC4 | EVENT OUT |
| PDO | - | - | - | TIM9_CH1 | - | $\begin{aligned} & \hline \text { SPI2_NSS } \\ & \text { I2S2_WS } \\ & \hline \end{aligned}$ | - | - | - | - - | D2 /DA2 | - TIMx_IC1 | EVENT OUT |
| PD1 | - | - | - | - | - | $\begin{aligned} & \text { SPI2 SCK } \\ & \text { 12S2_CK } \end{aligned}$ | - | - | - | - - | D3/DA3 | - TIMx_IC2 | EVENT OUT |
| PD2 | - | - | TIM3_ETR | - | - | - | - | - | UART5_RX | $\begin{aligned} & \text { COM71 } \\ & -\begin{array}{l} \text { SEG31/ } \\ \text { SEG43 } \end{array} \end{aligned}$ | $\begin{aligned} & \text { SDIO- } \\ & \text { CMD } \end{aligned}$ | - TIMx_IC3 | $\begin{aligned} & \text { EVENT } \\ & \text { OUT } \end{aligned}$ |
| PD3 | - | - | - | - | - | SPI2_MISO |  | USART2_CTS | - | - - | CLK | - TIMx_IC4 | OU <br> EVENT OUT |
| PD4 | - | - | - | - | - | $\begin{array}{\|l\|} \hline \text { SPI2_MOSI } \\ \text { I2S2_SD } \\ \hline \end{array}$ |  | USART2_RTS | - | - - | NOE | - TIMx_IC1 | EVENT OUT |
| PD5 | - | - | - | - | - | - | - | USART2_TX | - | - - | NWE | - TIMx_IC2 | EVENT OUT |
| PD6 | - | - | - | - | - | - | - | USART2_RX | - | - - | NWAIT | - TIMx_IC3 | EVENT OUT |
| PD7 | - | - | - | TIM9_CH2 | - | - | - | USART2_CK | - | - - | NE1 | - TIMx_IC4 | EVENT OUT |

Table 9. Alternate function input/output (continued)

| Port name | Digital alternate function number |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | AFIOO | AFIO1 | AFIO2 | AFIO3 | AFIO4 | AFIO5 | AFIO6 | AFIO7 | AFIO8 |  | AFIO11 | AFIO12 | .. AFIO14 | AFIO15 |
|  | Alternate function |  |  |  |  |  |  |  |  |  |  |  |  |  |
|  | SYSTEM | TIM2 | TIM3/4/5 | $\begin{aligned} & \text { TIM9/ } \\ & \text { 10/11 } \end{aligned}$ | 12C1/2 | SPI1/2 | SPI3 | USART1/2/3 | $\begin{array}{\|c\|} \hline \text { UART4I } \\ 5 \end{array}$ |  | LCD | $\begin{aligned} & \text { FSMC/ } \\ & \text { SDIO } \end{aligned}$ | CPRI | SYSTEM |
| PD8 | - | - | - | - | - | - | - | USART3_TX | - |  | SEG28 | D13/DA13 | - TIMx_IC1 | EVENT OUT |
| PD9 | - | - | - | - | - | - | - | USART3_RX | - |  | SEG29 | D14/DA14 | - TIMx_IC2 | EVENT OUT |
| PD10 | - | - | - | - | - | - | - | USART3_CK | - |  | SEG30 | D15/DA15 | - TIMx_IC3 | EVENT OUT |
| PD11 | - | - | - | - | - | - | - | USART3_CTS | - |  | SEG31 | A16 | - TIMx_IC4 | EVENT OUT |
| PD12 | - | - | TIM4_CH1 | - | - | - | - | USART3_RTS | - |  | SEG32 | A17 | - TIMx_IC1 | EVENT OUT |
| PD13 | - | - | TIM4_CH2 | - | - | - | - | - | - |  | SEG33 | A18 | - TIMx_IC2 | EVENT OUT |
| PD14 | - | - | TIM4_CH3 | - | - | - | - | - | - | - S | SEG34 | DO/DAO | - TIMx_IC3 | EVENT OUT |
| PD15 | - | - | TIM4_CH4 | - | - | - | - | - | - |  | SEG35 | D1/DA1 | - TIMx_IC4 | EVENT OUT |
| PE0 | - | - | TIM4_ETR | TIM 10_CH1 | - | - | - | - | - |  | SEG36 | NBLO | - TIMx_IC1 | EVENT OUT |
| PE1 | - | - | - | TIM11_CH1 | - | - | - | - | - |  | SEG37 | NBL1 | - TIMx_IC2 | $\begin{aligned} & \text { EVENT } \\ & \text { OUT } \end{aligned}$ |
| PE2 | TRACECK | - | TIM3_ETR | - | - | - | - | - | - |  | SEG 38 | A23 | - TIMx_IC3 | EVENT OUT |
| PE3 | TRACEDO | - | TIM3_CH1 | - | - | - | - | - | - |  | SEG 39 | A19 | - TIMx_IC4 | EVENT OUT |
| PE4 | TRACED1 | - | TIM3_CH2 | - | - | - | - | - | - | - | - | A20 | - TIMx_IC1 | $\begin{aligned} & \text { EVENT } \\ & \text { OUT } \end{aligned}$ |
| PE5 | TRACED2 | - | - | TIM9_CH1 | - | - | - | - | - | - | - | A21 | - TIMx_IC2 | EVENT OUT |
| PE6-WKUP3 | TRACED3 | - | - | TIM9_CH2 | - | - | - | - | - | - | - | - | - TIMx_IC3 | EVENT OUT |

Table 9. Alternate function input/output (continued)

| Port name | Digital alternate function number |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | AFIOO | AFIO1 | AFIO2 | AFIO3 | AFIO4 | AFIO5 | AFIO6 | AFIO7 | AFIO8 | .. | AFIO11 | AFIO12 |  | AFIO14 | AFIO15 |
|  | Alternate function |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
|  | SYSTEM | TIM2 | TIM3/4/5 | $\begin{aligned} & \text { TIM9/ } \\ & \text { 10/11 } \end{aligned}$ | 12C1/2 | SPI1/2 | SPI3 | USART1/2/3 | $\begin{array}{\|c\|} \hline \text { UART4/ } \\ 5 \end{array}$ | - | LCD | $\begin{aligned} & \text { FSMC/ } \\ & \text { SDIO } \end{aligned}$ | - | CPRI | SYSTEM |
| PE7 | - | - | - | - | - | - | - | - | - | - | - | D4/DA4 |  | TIMx_IC4 | EVENT OUT |
| PE8 | - | - | - | - | - | - | - | - | - | - | - | D5/DA5 |  | TIMx_IC1 | EVENT OUT |
| PE9 | - | TIM2_CH1_ETR | - | - | - | - | - | - | - | - | - | D6/DA6 |  | TIMx_IC2 | EVENT OUT |
| PE10 | - | TIM2_CH2 | - | - | - | - | - | - | - | - | - | D7/DA7 |  | TIMx_IC3 | EVENT OUT |
| PE11 | - | TIM2_CH3 | - | - | - | - | - | - | - | - | - | D8/DA8 |  | TIMx_IC4 | EVENT OUT |
| PE12 | - | TIM2_CH4 | - | - | - | SPI1_NSS | - | - | - | - | - | D9/DA9 |  | TIMx_IC1 | EVENT OUT |
| PE13 | - | - | - | - | - | SPI1_SCK | - | - | - | - | - | D10/DA10 |  | TIMx_IC2 | EVENT OUT |
| PE14 | - | - | - | - | - | SPI1_MISO | - | - | - | - | - | D11/DA11 |  | TIMx_IC3 | EVENT OUT |
| PE15 | - | - | - | - | - | SPI1_MOSI | - | - | - | - | - | D12/DA12 |  | TIMx_IC4 | EVENT OUT |
| PF0 | - | - | - | - | - | - | - | - | - | - | - | A0 | - | - | $\begin{aligned} & \text { EVENT } \\ & \text { OUT } \end{aligned}$ |
| PF1 | - | - | - | - | - | - | - | - | - | - | - | A1 | - | - | EVENT OUT |
| PF2 | - | - | - | - | - | - | - | - | - | - | - | A2 | - | - | EVENT OUT |
| PF3 | - | - | - | - | - | - | - | - | - | - | - | A3 | - | - | $\begin{aligned} & \text { EVENT } \\ & \text { OUT } \end{aligned}$ |
| PF4 | - | - | - | - | - | - | - | - | - | - | - | A4 | - | - | EVENT OUT <br> OU |
| PF5 | - | - | - | - | - | - | - | - | - | - | - | A5 | - | - | EVENT OUT |

Table 9. Alternate function input/output (continued)

Table 9. Alternate function input/output (continued)

| Port name | Digital alternate function number |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | AFIOO | AFIO1 | AFIO2 | AFIO3 | AFIO4 | AFIO5 | AFIO6 | AFIO7 | AFIO8 | .. | AFIO11 | AFIO12 | .. | AFIO14 | AFIO15 |
|  | Alternate function |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
|  | SYSTEM | TIM2 | TIM3/4/5 | $\begin{aligned} & \text { TIM9/ } \\ & \text { 10/11 } \end{aligned}$ | 12C1/2 | SPI1/2 | SPI3 | USART1/2/3 | \| UART4/ $5$ | - | LCD | $\begin{aligned} & \text { FSMC/ } \\ & \text { SDIO } \end{aligned}$ | - | CPRI | SYSTEM |
| PG5 | - | - | - | - | - | - | - | - | - | - | - | A15 | - | - | $\begin{aligned} & \text { EVENT } \\ & \text { OUT } \end{aligned}$ |
| PG6 | - | - | - | - | - | - | - | - | - | - | - | - | - | - | $\begin{aligned} & \text { EVENT } \\ & \text { OUT } \end{aligned}$ |
| PG7 | - | - | - | - | - | - | - | - | - | - | - | - | - | - | $\begin{aligned} & \text { EVENT } \\ & \text { OUT } \end{aligned}$ |
| PG8 | - | - | - | - | - | - | - | - | - | - | - | - | - | - | $\begin{aligned} & \text { EVENT } \\ & \text { OUT } \end{aligned}$ |
| PG9 | - | - | - | - | - | - | - | - | - | - | - | NE2 | - | - | EVENT OUT |
| PG10 | - | - | - | - | - | - | - | - | - | - | - | NE3 | - | - | $\begin{aligned} & \text { EVENT } \\ & \text { OUT } \end{aligned}$ |
| PG11 | - | - | - | - | - | - | - | - | - | - | - | - | - | - | EVENT OUT |
| PG12 | - | - | - | - | - | - | - | - | - | - | - | NE4 | - | - | EVENT OUT |
| PG13 | - | - | - | - | - | - | - | - | - | - | - | A24 | - | - | EVENT OUT |
| PG14 | - | - | - | - | - | - | - | - | - | - | - | A25 | - | - | $\begin{array}{\|l} \text { EVENT } \\ \text { OUT } \\ \hline \end{array}$ |
| PG15 | - | - | - | - | - | - | - | - | - | - | - | - | - | - | EVENT OUT |
| PH0OSC_IN | - | - | - | - | - | - | - | - | - | - | - | - | - | - | - |
| PH1OSC_OUT | - | - | - | - | - | - | - | - | - | - | - | - | - | - | - |
| PH2 | - | - | - | - | - | - | - | - | - | - | - | A22 | - | - | - |

## 5 Memory mapping

Figure 8. Memory map


## 6 Electrical characteristics

### 6.1 Parameter conditions

Unless otherwise specified, all voltages are referenced to $\mathrm{V}_{\mathrm{SS}}$.

### 6.1.1 Minimum and maximum values

Unless otherwise specified the minimum and maximum values are guaranteed in the worst conditions of ambient temperature, supply voltage and frequencies by tests in production on $100 \%$ of the devices with an ambient temperature at $T_{A}=25^{\circ} \mathrm{C}$ and $T_{A}=T_{A}$ max (given by the selected temperature range).

Data based on characterization results, design simulation and/or technology characteristics are indicated in the table footnotes. Based on characterization, the minimum and maximum values refer to sample tests and represent the mean value plus or minus three times the standard deviation (mean $\pm 3 \sigma$ ).

### 6.1.2 Typical values

Unless otherwise specified, typical data are based on $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{DD}}=3.6 \mathrm{~V}$ (for the $1.65 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DD}} \leq 3.6 \mathrm{~V}$ voltage range). They are given only as design guidelines and are not tested.

Typical ADC accuracy values are determined by characterization of a batch of samples from a standard diffusion lot over the full temperature range, where $95 \%$ of the devices have an error less than or equal to the value indicated (mean $\pm 2 \sigma$ ).

### 6.1.3 Typical curves

Unless otherwise specified, all typical curves are given only as design guidelines and are not tested.

### 6.1.4 Loading capacitor

The loading conditions used for pin parameter measurement are shown in Figure 9.

### 6.1.5 Pin input voltage

The input voltage measurement on a pin of the device is described in Figure 10.


### 6.1.6 Power supply scheme

Figure 11. Power supply scheme


### 6.1.7 Optional LCD power supply scheme

Figure 12. Optional LCD power supply scheme


1. Option 1: LCD power supply is provided by a dedicated VLCD supply source, VSEL switch is open.
2. Option 2: LCD power supply is provided by the internal step-up converter, VSEL switch is closed, an external capacitance is needed for correct behavior of this converter.

### 6.1.8 Current consumption measurement

Figure 13. Current consumption measurement scheme


### 6.2 Absolute maximum ratings

Stresses above the absolute maximum ratings listed in Table 10: Voltage characteristics, Table 11: Current characteristics, and Table 12: Thermal characteristics may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these conditions is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

Table 10. Voltage characteristics

| Symbol | Ratings | Min | Max | Unit |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{DD}}-\mathrm{V}_{S S}$ | External main supply voltage (including $\mathrm{V}_{\mathrm{DDA}}$ and $\left.\mathrm{V}_{\mathrm{DD}}\right)^{(1)}$ | -0.3 | 4.0 | V |
| $\mathrm{V}_{\text {IN }}{ }^{(2)}$ | Input voltage on five-volt tolerant pin | $\mathrm{V}_{\text {SS }}-0.3$ | $\mathrm{V}_{\mathrm{DD}}+4.0$ |  |
|  | Input voltage on any other pin | $\mathrm{V}_{\text {SS }}-0.3$ | 4.0 |  |
| $\left\|\Delta V_{\text {DDx }}\right\|$ | Variations between different $\mathrm{V}_{\mathrm{DD}}$ power pins | - | 50 | mV |
| $\left\|\mathrm{V}_{\text {SSX }}-\mathrm{V}_{\text {SS }}\right\|$ | Variations between all different ground pins ${ }^{(3)}$ | - | 50 |  |
| $\mathrm{V}_{\text {REF }+}-\mathrm{V}_{\text {DDA }}$ | Allowed voltage difference for $\mathrm{V}_{\text {REF }+}>\mathrm{V}_{\text {DDA }}$ | - | 0.4 | V |
| $\mathrm{V}_{\text {ESD (HBM) }}$ | Electrostatic discharge voltage (human body model) | see Section 6.3.12 |  | - |

1. All main power $\left(V_{D D}, V_{D D A}\right)$ and ground $\left(V_{S S}, V_{S S A}\right)$ pins must always be connected to the external power supply, in the permitted range.
2. $\mathrm{V}_{\mathrm{IN}}$ maximum must always be respected. Refer to Table 11 for maximum allowed injected current values.
3. Include $\mathrm{V}_{\text {REF- }}$ pin.

Table 11. Current characteristics

| Symbol | Ratings | Max. | Unit |
| :---: | :---: | :---: | :---: |
| $I_{\operatorname{VDD}(\Sigma)}$ | Total current into sum of all $\mathrm{V}_{\text {DD_x }}$ power lines (source) ${ }^{(1)}$ | 100 | mA |
| $\mathrm{I}_{\mathrm{VSS}(\Sigma)}{ }^{(2)}$ | Total current out of sum of all $\mathrm{V}_{\text {SS_x }}$ ground lines (sink) ${ }^{(1)}$ | 100 |  |
| $\mathrm{I}_{\mathrm{VDD}(\mathrm{PIN})}$ | Maximum current into each $\mathrm{V}_{\text {DD_x }}$ p power pin (source) ${ }^{(1)}$ | 70 |  |
| $\mathrm{I}_{\mathrm{VSS}(\text { PIN })}$ | Maximum current out of each VSS_x ground pin (sink) ${ }^{(1)}$ | -70 |  |
| 10 | Output current sunk by any I/O and control pin | 25 |  |
|  | Output current sourced by any I/O and control pin | -25 |  |
| $\Sigma I_{\text {IO(PIN })}$ | Total output current sunk by sum of all IOs and control pins ${ }^{(2)}$ | 60 |  |
|  | Total output current sourced by sum of all IOs and control pins ${ }^{(2)}$ | -60 |  |
| $\mathrm{I}_{\mathrm{INJ}(\mathrm{PIN})}{ }^{(3)}$ | Injected current on five-volt tolerant I/O ${ }^{(4)}$, RST and B pins | -5/+0 |  |
|  | Injected current on any other pin ${ }^{(5)}$ | $\pm 5$ |  |
| $\Sigma l_{1 \times 3(\text { PIN })}$ | Total injected current (sum of all I/O and control pins) ${ }^{(6)}$ | $\pm 25$ |  |

1. All main power $\left(V_{D D}, V_{D D A}\right)$ and ground $\left(V_{S S}, V_{S S A}\right)$ pins must always be connected to the external power supply, in the permitted range.
2. This current consumption must be correctly distributed over all I/Os and control pins. The total output current must not be sunk/sourced between two consecutive power supply pins referring to high pin count LQFP packages.
3. Negative injection disturbs the analog performance of the device. See note in Section 6.3.19.
4. Positive current injection is not possible on these $\mathrm{I} / \mathrm{Os}$. A negative injection is induced by $\mathrm{V}_{\mathbb{I N}}<\mathrm{V}_{S S}$. $\mathrm{I}_{\mathbb{I N J}(\mathrm{PIN})}$ must never be exceeded. Refer to Table 10 for maximum allowed input voltage values.
5. A positive injection is induced by $V_{I N}>V_{D D}$ while a negative injection is induced by $V_{I N}<V_{S S}$. $l_{I N J(P I N)}$ must never be exceeded. Refer to Table 10: Voltage characteristics for the maximum allowed input voltage values.
6. When several inputs are submitted to a current injection, the maximum $\Sigma l_{\operatorname{INJ}(\mathrm{PIN})}$ is the absolute sum of the positive and negative injected currents (instantaneous values).

Table 12. Thermal characteristics

| Symbol | Ratings | Value | Unit |
| :---: | :--- | :---: | :---: |
| $\mathrm{T}_{\text {STG }}$ | Storage temperature range | -65 to +150 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{J}$ | Maximum junction temperature | 150 | ${ }^{\circ} \mathrm{C}$ |

### 6.3 Operating conditions

### 6.3.1 General operating conditions

Table 13. General operating conditions

| Symbol | Parameter | Conditions | Min | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{f}_{\text {HCLK }}$ | Internal AHB clock frequency | - | 0 | 32 | MHz |
| $\mathrm{f}_{\text {PCLK1 }}$ | Internal APB1 clock frequency | - | 0 | 32 |  |
| $\mathrm{f}_{\text {PCLK2 }}$ | Internal APB2 clock frequency | - | 0 | 32 |  |
| $V_{D D}$ | Standard operating voltage | BOR detector disabled | 1.65 | 3.6 | V |
|  |  | BOR detector enabled, at power on | 1.8 | 3.6 |  |
|  |  | BOR detector disabled, after power on | 1.65 | 3.6 |  |
| $V_{\text {DDA }}{ }^{(1)}$ | Analog operating voltage (ADC and DAC not used) | Must be the same voltage as $V_{D D}{ }^{(2)}$ | 1.65 | 3.6 | V |
|  | Analog operating voltage (ADC or DAC used) |  | 1.8 | 3.6 |  |
| $\mathrm{V}_{\mathrm{IN}}$ | I/O input voltage | FT pins; $2.0 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DD}}$ | -0.3 | $5.5{ }^{(3)}$ | V |
|  |  | FT pins; $\mathrm{V}_{\mathrm{DD}}<2.0 \mathrm{~V}$ | -0.3 | $5.25{ }^{(3)}$ |  |
|  |  | BOOT0 pin | 0 | 5.5 |  |
|  |  | Any other pin | -0.3 | $\mathrm{V}_{\mathrm{DD}}+0.3$ |  |
| $\mathrm{P}_{\mathrm{D}}$ | Power dissipation at TA $=85^{\circ} \mathrm{C}$ for suffix 6 or TA $=105^{\circ} \mathrm{C}$ for suffix $7^{(4)}$ | LQFP144 package | - | 500 | mW |
|  |  | LQFP100 package | - | 465 |  |
|  |  | LQFP64 package | - | 435 |  |
|  |  | UFBGA132 | - | 333 |  |
|  |  | WLCSP64 package | - | 435 |  |
| TA | Ambient temperature for 6 suffix version | Maximum power dissipation ${ }^{(5)}$ | -40 | 85 | ${ }^{\circ} \mathrm{C}$ |
|  | Ambient temperature for 7 suffix version | Maximum power dissipation | -40 | 105 |  |

Table 13. General operating conditions (continued)

| Symbol | Parameter | Conditions | Min | Max | Unit |
| :---: | :---: | :--- | :---: | :---: | :---: |
| $\mathrm{T} J$ | Junction temperature range | 6 suffix version | -40 | 105 | ${ }^{\circ} \mathrm{C}$ |
|  |  | 7 suffix version | -40 | 110 |  |

1. When the ADC is used, refer to Table 64: ADC characteristics.
2. It is recommended to power $\mathrm{V}_{D D}$ and $\mathrm{V}_{\mathrm{DDA}}$ from the same source. A maximum difference of 300 mV between $\mathrm{V}_{\mathrm{DD}}$ and $V_{\text {DDA }}$ can be tolerated during power-up.
3. To sustain a voltage higher than VDD+0.3V, the internal pull-up/pull-down resistors must be disabled.
4. If $T_{A}$ is lower, higher $P_{D}$ values are allowed as long as $T_{J}$ does not exceed $T_{J}$ max (see Table 80: Thermal characteristics on page 146).
5. In low-power dissipation state, $T_{A}$ can be extended to $-40^{\circ} \mathrm{C}$ to $105^{\circ} \mathrm{C}$ temperature range as long as $T_{J}$ does not exceed $T_{J}$ max (see Table 80: Thermal characteristics on page 146).

### 6.3.2 Embedded reset and power control block characteristics

The parameters given in the following table are derived from the tests performed under the conditions summarized in Table 13.

Table 14. Embedded reset and power control block characteristics

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\mathrm{VDD}}{ }^{(1)}$ | $V_{D D}$ rise time rate | BOR detector enabled | 0 | - | $\infty$ | $\mu \mathrm{s} / \mathrm{V}$ |
|  |  | BOR detector disabled | 0 | - | 1000 |  |
|  | $V_{\text {DD }}$ fall time rate | BOR detector enabled | 20 | - | $\infty$ |  |
|  |  | BOR detector disabled | 0 | - | 1000 |  |
| $\mathrm{T}_{\text {RSTTEMPO }}{ }^{(1)}$ | Reset temporization | $V_{D D}$ rising, BOR enabled | - | 2 | 3.3 | ms |
|  |  | $V_{\text {DD }}$ rising, $B$ ( ${ }^{\text {d }}$ disabled ${ }^{(2)}$ | 0.4 | 0.7 | 1.6 |  |
| $\mathrm{V}_{\text {POR/PDR }}$ | Power on/power down reset threshold | Falling edge | 1 | 1.5 | 1.65 | V |
|  |  | Rising edge | 1.3 | 1.5 | 1.65 |  |
| $V_{\text {BORO }}$ | Brown-out reset threshold 0 | Falling edge | 1.67 | 1.7 | 1.74 |  |
|  |  | Rising edge | 1.69 | 1.76 | 1.8 |  |
| $\mathrm{V}_{\text {BOR1 }}$ | Brown-out reset threshold 1 | Falling edge | 1.87 | 1.93 | 1.97 |  |
|  |  | Rising edge | 1.96 | 2.03 | 2.07 |  |
| $V_{\text {BOR2 }}$ | Brown-out reset threshold 2 | Falling edge | 2.22 | 2.30 | 2.35 |  |
|  |  | Rising edge | 2.31 | 2.41 | 2.44 |  |

Table 14. Embedded reset and power control block characteristics (continued)

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {BOR3 }}$ | Brown-out reset threshold 3 | Falling edge | 2.45 | 2.55 | 2.6 | V |
|  |  | Rising edge | 2.54 | 2.66 | 2.7 |  |
| $\mathrm{V}_{\text {BOR4 }}$ | Brown-out reset threshold 4 | Falling edge | 2.68 | 2.8 | 2.85 |  |
|  |  | Rising edge | 2.78 | 2.9 | 2.95 |  |
| $\mathrm{V}_{\text {PVDO }}$ | Programmable voltage detector threshold 0 | Falling edge | 1.8 | 1.85 | 1.88 |  |
|  |  | Rising edge | 1.88 | 1.94 | 1.99 |  |
| $\mathrm{V}_{\text {PVD1 }}$ | PVD threshold 1 | Falling edge | 1.98 | 2.04 | 2.09 |  |
|  |  | Rising edge | 2.08 | 2.14 | 2.18 |  |
| $\mathrm{V}_{\text {PVD2 }}$ | PVD threshold 2 | Falling edge | 2.20 | 2.24 | 2.28 |  |
|  |  | Rising edge | 2.28 | 2.34 | 2.38 |  |
| $\mathrm{V}_{\text {PVD3 }}$ | PVD threshold 3 | Falling edge | 2.39 | 2.44 | 2.48 |  |
|  |  | Rising edge | 2.47 | 2.54 | 2.58 |  |
| $\mathrm{V}_{\text {PVD4 }}$ | PVD threshold 4 | Falling edge | 2.57 | 2.64 | 2.69 |  |
|  |  | Rising edge | 2.68 | 2.74 | 2.79 |  |
| $\mathrm{V}_{\text {PVD5 }}$ | PVD threshold 5 | Falling edge | 2.77 | 2.83 | 2.88 |  |
|  |  | Rising edge | 2.87 | 2.94 | 2.99 |  |
| $\mathrm{V}_{\text {PVD6 }}$ | PVD threshold 6 | Falling edge | 2.97 | 3.05 | 3.09 |  |
|  |  | Rising edge | 3.08 | 3.15 | 3.20 |  |
| $V_{\text {hyst }}$ | Hysteresis voltage | BOR0 threshold | - | 40 | - | mV |
|  |  | All BOR and PVD thresholds excepting BOR0 | - | 100 | - |  |

1. Guaranteed by characterization results.
2. Valid for device version without BOR at power up. See option "D" in Ordering information scheme for more details.

### 6.3.3 Embedded internal reference voltage

The parameters given in Table 16 are based on characterization results, unless otherwise specified.

Table 15. Embedded internal reference voltage calibration values

| Calibration value name | Description | Memory address |
| :---: | :--- | :---: |
| VREFINT_CAL | Raw data acquired at <br> temperature of $30^{\circ} \mathrm{C} \pm 5^{\circ} \mathrm{C}$ <br> $V_{\text {DDA }}=3 \mathrm{~V} \pm 10 \mathrm{mV}$ | $0 \times 1 \mathrm{FF} 800 \mathrm{~F} 8-0 \times 1 \mathrm{FF} 800 \mathrm{~F} 9$ |

Table 16. Embedded internal reference voltage

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {REFINT out }}{ }^{(1)}$ | Internal reference voltage | $-40^{\circ} \mathrm{C}<\mathrm{T}_{\mathrm{J}}<+110{ }^{\circ} \mathrm{C}$ | 1.202 | 1.224 | 1.242 | V |
| $\mathrm{I}_{\text {REFINT }}$ | Internal reference current consumption | - | - | 1.4 | 2.3 | $\mu \mathrm{A}$ |
| TVREFINT | Internal reference startup time | - | - | 2 | 3 | ms |
| $\mathrm{V}_{\text {VREF_MEAS }}$ | $V_{\text {DDA }}$ and $V_{\text {REF }}$ voltage during $V_{\text {REFINT }}$ factory measure | - | 2.99 | 3 | 3.01 | V |
| Avref_meas | Accuracy of factory-measured $\mathrm{V}_{\text {REF }}$ value ${ }^{(2)}$ | Including uncertainties due to ADC and $V_{\text {DDA }} / V_{\text {REF }}$ values | - | - | $\pm 5$ | mV |
| $\mathrm{T}_{\text {Coeff }}{ }^{(3)}$ | Temperature coefficient | $-40^{\circ} \mathrm{C}<\mathrm{T}_{\mathrm{J}}<+110^{\circ} \mathrm{C}$ | - | 25 | 100 | $\mathrm{ppm} /{ }^{\circ} \mathrm{C}$ |
| $\mathrm{A}_{\text {Coeff }}{ }^{(3)}$ | Long-term stability | 1000 hours, $\mathrm{T}=25^{\circ} \mathrm{C}$ | - | - | 1000 | ppm |
| $V_{\text {DDCoeff }}{ }^{(3)}$ | Voltage coefficient | $3.0 \mathrm{~V}<\mathrm{V}_{\text {DDA }}<3.6 \mathrm{~V}$ | - | - | 2000 | ppm/V |
| $\mathrm{T}_{\text {__vefint }}{ }^{(3)}$ | ADC sampling time when reading the internal reference voltage | - | 4 | - | - | $\mu \mathrm{s}$ |
| $\mathrm{T}_{\text {ADC_BUF }}{ }^{(3)(4)}$ | Startup time of reference voltage buffer for ADC | - | - | - | 10 | $\mu \mathrm{s}$ |
| $\mathrm{I}_{\text {BUF_ADC }}{ }^{(3)}$ | Consumption of reference voltage buffer for ADC | - | - | 13.5 | 25 | $\mu \mathrm{A}$ |
| $\mathrm{I}_{\text {VREF_OUT }}{ }^{(3)}$ | VREF_OUT output current ${ }^{(5)}$ | - | - | - | 1 | $\mu \mathrm{A}$ |
| $\mathrm{C}_{\text {VREF_OUT }}{ }^{(3)}$ | VREF_OUT output load | - | - | - | 50 | pF |
| $\mathrm{I}_{\text {LPBUF }}{ }^{(3)}$ | Consumption of reference voltage buffer for VREF_OUT and COMP | - | - | 730 | 1200 | nA |
| $\mathrm{V}_{\text {REFINT_DIV1 }}{ }^{(3)}$ | 1/4 reference voltage | - | 24 | 25 | 26 | \% <br> $\mathrm{V}_{\text {REFINT }}$ |
| $\mathrm{V}_{\text {REFINT_DIV2 }}{ }^{(3)}$ | 1/2 reference voltage | - | 49 | 50 | 51 |  |
| $\mathrm{V}_{\text {REFINT_DIV3 }}{ }^{(3)}$ | 3/4 reference voltage | - | 74 | 75 | 76 |  |

1. Guaranteed by test in production.
2. The internal $\mathrm{V}_{\text {REF }}$ value is individually measured in production and stored in dedicated EEPROM bytes.
3. Guaranteed by characterization results.
4. Shortest sampling time can be determined in the application by multiple iterations.
5. To guarantee less than $1 \%$ VREF_OUT deviation.

### 6.3.4 Supply current characteristics

The current consumption is a function of several parameters and factors such as the operating voltage, temperature, I/O pin loading, device software configuration, operating frequencies, I/O pin switching rate, program location in memory and executed binary code. The current consumption is measured as described in Figure 13: Current consumption measurement scheme.

All Run-mode current consumption measurements given in this section are performed with a reduced code that gives a consumption equivalent to the Dhrystone 2.1 code, unless otherwise specified. The current consumption values are derived from tests performed under ambient temperature $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ and $\mathrm{V}_{\mathrm{DD}}$ supply voltage conditions summarized in Table 13: General operating conditions, unless otherwise specified.

The MCU is placed under the following conditions:

- All I/O pins are configured in analog input mode
- All peripherals are disabled except when explicitly mentioned.
- The Flash memory access time, 64-bit access and prefetch is adjusted depending on $\mathrm{f}_{\text {HCLK }}$ frequency and voltage range to provide the best CPU performance.
- When the peripherals are enabled $\mathrm{f}_{\mathrm{APB} 1}=\mathrm{f}_{\mathrm{APB} 2}=\mathrm{f}_{\mathrm{AHB}}$.
- When PLL is ON, the PLL inputs are equal to $\mathrm{HSI}=16 \mathrm{MHz}$ (if internal clock is used) or HSE = 16 MHz (if HSE bypass mode is used).
- The HSE user clock applied to OSCI_IN input follows the characteristic specified in Table 26: High-speed external user clock characteristics.
- For maximum current consumption $\mathrm{V}_{\mathrm{DD}}=\mathrm{V}_{\mathrm{DDA}}=3.6 \mathrm{~V}$ is applied to all supply pins.
- For typical current consumption $\mathrm{V}_{\mathrm{DD}}=\mathrm{V}_{\mathrm{DDA}}=3.0 \mathrm{~V}$ is applied to all supply pins if not specified otherwise.

Table 17. Current consumption in Run mode, code with data processing running from Flash

| Symbol | Parameter | Conditions |  | $\mathrm{f}_{\mathrm{HCLK}}$ [MHz] | Typ | $\underset{(1)}{\operatorname{Max}}$ | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| IDD (Run <br> from <br> Flash) | Supply current in Run mode code executed from Flash | $\mathrm{f}_{\text {HSE }}=\mathrm{f}_{\text {HCLK }}$ up to 16 MHz , included $f_{\text {HSE }}=f_{\text {HCLK }} / 2$ above $16 \mathrm{MHz}(\mathrm{PLL} \mathrm{ON}){ }^{(2)}$ | Range3, <br> $\mathrm{V}_{\text {CORE }}=1.2 \mathrm{~V}$ <br> $\operatorname{VOS}[1: 0]=11$ | 1 | 290 | 500 | $\mu \mathrm{A}$ |
|  |  |  |  | 2 | 505 | 750 |  |
|  |  |  |  | 4 | 955 | 1200 |  |
|  |  |  | Range2, <br> $\mathrm{V}_{\text {CORE }}=1.5 \mathrm{~V}$ <br> $\operatorname{VOS}[1: 0]=10$ | 4 | 1.15 | 1.6 | mA |
|  |  |  |  | 8 | 2.3 | 2.9 |  |
|  |  |  |  | 16 | 4.25 | 5.2 |  |
|  |  |  | Range1, <br> $\mathrm{V}_{\mathrm{CORE}}=1.8 \mathrm{~V}$ <br> $\operatorname{VOS}[1: 0]=01$ | 8 | 2.65 | 3.5 |  |
|  |  |  |  | 16 | 5.35 | 6.5 |  |
|  |  |  |  | 32 | 10.5 | 12 |  |
|  |  | HSI clock source ( 16 MHz ) | Range2, <br> $\mathrm{V}_{\text {CORE }}=1.5 \mathrm{~V}$ <br> $\operatorname{VOS}[1: 0]=10$ | 16 | 4.35 | 5.2 |  |
|  |  |  | Range1, <br> $\mathrm{V}_{\text {CORE }}=1.8 \mathrm{~V}$ <br> $\operatorname{VOS}[1: 0]=01$ | 32 | 10.5 | 12.3 |  |
|  |  | MSI clock, 65 kHZ | Range3,$\begin{aligned} & \mathrm{V}_{\mathrm{CORE}}=1.2 \mathrm{~V} \\ & \mathrm{VOS}[1: 0]=11 \end{aligned}$ | 0.065 | 46 | 130 | $\mu \mathrm{A}$ |
|  |  | MSI clock, 524 kHZ |  | 0.524 | 160 | 250 |  |
|  |  | MSI clock, 4.2 MHZ |  | 4.2 | 965 | 1200 |  |

1. Guaranteed by characterization results, unless otherwise specified.
2. Oscillator bypassed (HSEBYP = 1 in RCC_CR register).

Table 18. Current consumption in Run mode, code with data processing running from RAM

| Symbol | Parameter | Conditions |  | $\mathrm{f}_{\mathrm{HCLK}}$ | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $I_{D D}$ (Run from RAM) | Supply current in Run mode code executed from RAM | $\mathrm{f}_{\text {HSE }}=\mathrm{f}_{\text {HCLK }}$ up to 16 MHz , included $f_{\text {HSE }}=f_{H C}(1)^{\prime} / 2$ above 16 MHz (PLL ON) ${ }^{(1)}$ | Range3, <br> $\mathrm{V}_{\text {CORE }}=1.2 \mathrm{~V}$ <br> $\operatorname{VOS}[1: 0]=11$ | 1 | 230 | 470 | $\mu \mathrm{A}$ |
|  |  |  |  | 2 | 415 | 780 |  |
|  |  |  |  | 4 | 800 | 1200 |  |
|  |  |  | Range2, <br> $\mathrm{V}_{\text {CORE }}=1.5 \mathrm{~V}$ <br> $\operatorname{VOS}[1: 0]=10$ | 4 | 0.935 | 1.5 | mA |
|  |  |  |  | 8 | 1.9 | 3 |  |
|  |  |  |  | 16 | 3.75 | 5 |  |
|  |  |  | Range1, <br> $\mathrm{V}_{\text {CORE }}=1.8 \mathrm{~V}$ <br> $\operatorname{VOS}[1: 0]=01$ | 8 | 2.25 | 3.5 |  |
|  |  |  |  | 16 | 4.45 | 5.55 |  |
|  |  |  |  | 32 | 9.05 | 10.9 |  |
|  |  | HSI clock source (16 MHz) | Range2, <br> $\mathrm{V}_{\mathrm{CORE}}=1.5 \mathrm{~V}$ <br> $\operatorname{VOS}[1: 0]=10$ | 16 | 3.75 | 4.8 |  |
|  |  |  | Range1, <br> $\mathrm{V}_{\mathrm{CORE}}=1.8 \mathrm{~V}$ <br> $\operatorname{VOS}[1: 0]=01$ | 32 | 8.95 | 11.7 |  |
|  |  | MSI clock, 65 kHZ | Range3, <br> $\mathrm{V}_{\mathrm{CORE}}=1.2 \mathrm{~V}$ <br> $\operatorname{VOS}[1: 0]=11$ | 0.065 | 43.5 | 100 | $\mu \mathrm{A}$ |
|  |  | MSI clock, 524 kHZ |  | 0.524 | 135 | 215 |  |
|  |  | MSI clock, 4.2 MHZ |  | 4.2 | 835 | 1100 |  |

[^1]Table 19. Current consumption in Sleep mode

| Symbol | Parameter | Conditions |  | $\mathrm{f}_{\mathrm{HCLK}}$ | Typ | $\underset{(1)}{\operatorname{Max}}$ | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{I}_{\text {DD(SLEEP) }}$ | Supply current in Sleep mode, code executed from RAM, Flash switched OFF | $f_{\text {HSE }}=f_{\text {HCLK }}$ up to 16 MHz , included $\mathrm{f}_{\mathrm{HSE}}=\mathrm{f}_{\mathrm{HCLK}} / 2$ above $16 \mathrm{MHz}(\text { PLL ON })^{(2)}$ | Range3, <br> Vcore=1.2 V <br> $\operatorname{VOS}[1: 0]=11$ | 1 | 58 | 220 | $\mu \mathrm{A}$ |
|  |  |  |  | 2 | 96 | 300 |  |
|  |  |  |  | 4 | 170 | 380 |  |
|  |  |  | Range2, <br> Vcore=1.5 V <br> $\operatorname{VOS}[1: 0]=10$ | 4 | 210 | 500 |  |
|  |  |  |  | 8 | 400 | 700 |  |
|  |  |  |  | 16 | 810 | 1100 |  |
|  |  |  | Range1, <br> Vcore=1.8 V <br> $\operatorname{VOS}[1: 0]=01$ | 8 | 485 | 800 |  |
|  |  |  |  | 16 | 955 | 1250 |  |
|  |  |  |  | 32 | 2100 | 2700 |  |
|  |  | HSI clock source (16 MHz) | Range2, <br> Vcore=1.5 V <br> $\operatorname{VOS}[1: 0]=10$ | 16 | 835 | 1100 |  |
|  |  |  | Range1, <br> Vcore=1.8 V <br> $\operatorname{VOS}[1: 0]=01$ | 32 | 2100 | 2700 |  |
|  |  | MSI clock, 65 kHZ | Range3, <br> Vcore=1.2 V <br> $\operatorname{VOS}[1: 0]=11$ | 0.065 | 18.5 | 72 |  |
|  |  | MSI clock, 524 kHZ |  | 0.524 | 37 | 92 |  |
|  |  | MSI clock, 4.2 MHZ |  | 4.2 | 180 | 273 |  |
|  | Supply current in Sleep mode, Flash switched ON | $\mathrm{f}_{\text {HSE }}=\mathrm{f}_{\text {HCLK }}$ up to 16 MHz , included $\mathrm{f}_{\mathrm{HSE}}=\mathrm{f}_{\mathrm{HCLK}} / 2$ above $16 \mathrm{MHz}(\text { PLL ON })^{(2)}$ | Range3, <br> Vcore=1.2 V <br> $\operatorname{VOS}[1: 0]=11$ | 1 | 75 | 250 |  |
|  |  |  |  | 2 | 115 | 300 |  |
|  |  |  |  | 4 | 200 | 380 |  |
|  |  |  | Range2, <br> Vcore=1.5 V <br> $\operatorname{VOS}[1: 0]=10$ | 4 | 230 | 500 |  |
|  |  |  |  | 8 | 430 | 700 |  |
|  |  |  |  | 16 | 840 | 1120 |  |
|  |  |  | Range1, <br> Vcore=1.8 V <br> $\operatorname{VOS}[1: 0]=01$ | 8 | 500 | 800 |  |
|  |  |  |  | 16 | 980 | 1300 |  |
|  |  |  |  | 32 | 2100 | 2700 |  |
|  |  | HSI clock source (16 MHz) | Range2, Vcore=1.5 V $\operatorname{VOS}[1: 0]=10$ | 16 | 860 | 1160 |  |
|  |  |  | Range1, <br> Vcore=1.8 V <br> $\operatorname{VOS}[1: 0]=01$ | 32 | 2150 | 2800 |  |
|  |  | MSI clock, 65 kHZ | Range3, <br> Vcore=1.2 V <br> $\operatorname{VOS}[1: 0]=11$ | 0.065 | 33,5 | 90 |  |
|  |  | MSI clock, 524 kHZ |  | 0.524 | 53 | 110 |  |
|  |  | MSI clock, 4.2 MHZ |  | 4.2 | 200 | 290 |  |

[^2]2. Oscillator bypassed (HSEBYP $=1$ in RCC_CR register)

Table 20. Current consumption in Low-power run mode

| Symbol | Parameter | Conditions |  |  | Typ | Max ${ }^{(1)}$ | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{I}_{\mathrm{DD}(\mathrm{LP}}$Run) | Supply current in Low-power run mode | All peripherals OFF, code executed from RAM, Flash switched OFF, $V_{D D}$ from 1.65 V to 3.6 V | MSI clock, 65 kHz $\mathrm{f}_{\text {HCLK }}=32 \mathrm{kHz}$ | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 11 | 14 | $\mu \mathrm{A}$ |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ | 26 | 32 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ | 53 | 72 |  |
|  |  |  | MSI clock, 65 kHz$\mathrm{f}_{\mathrm{HCLK}}=65 \mathrm{kHz}$ | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 18 | 21 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ | 33 | 40 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ | 60 | 78 |  |
|  |  |  | MSI clock, 131 kHz$\mathrm{f}_{\mathrm{HCLK}}=131 \mathrm{kHz}$ | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 36 | 41 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=55^{\circ} \mathrm{C}$ | 39 | 44 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ | 50 | 58 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ | 78 | 95 |  |
|  |  | All peripherals OFF, code executed from Flash, $V_{D D}$ from 1.65 V to 3.6 V | MSI clock, 65 kHz$\mathrm{f}_{\mathrm{HCLK}}=32 \mathrm{kHz}$ | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 36 | 40.5 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ | 53 | 60 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ | 81 | 100 |  |
|  |  |  | MSI clock, 65 kHz$\mathrm{f}_{\mathrm{HCLK}}=65 \mathrm{kHz}$ | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 44 | 49 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ | 61 | 67 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ | 89 | 107 |  |
|  |  |  | $\begin{aligned} & \mathrm{MSI} \text { clock, } 131 \mathrm{kHz} \\ & \mathrm{f}_{\mathrm{HCLK}}=131 \mathrm{kHz} \end{aligned}$ | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 64 | 71 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=55^{\circ} \mathrm{C}$ | 68 | 73 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ | 80 | 88 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ | 101 | 110 |  |
| IDD max (LP Run) | Max allowed current in Low-power run mode | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}} \text { from } \\ & 1.65 \mathrm{~V} \text { to } \\ & 3.6 \mathrm{~V} \end{aligned}$ | - | - | - | 200 |  |

1. Guaranteed by characterization results, unless otherwise specified.

Table 21. Current consumption in Low-power sleep mode

| Symbol | Parameter | Conditions |  |  | Typ | Max ${ }^{(1)}$ | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & \mathrm{IDD} \\ & (\text { LP Sleep }) \end{aligned}$ | Supply current in Low-power sleep mode | All peripherals OFF, $V_{D D}$ from 1.65 V to 3.6 V | $\begin{aligned} & \text { MSI clock, } 65 \mathrm{kHz} \\ & \mathrm{f}_{\mathrm{HCLK}}=32 \mathrm{kHz} \\ & \text { Flash OFF } \end{aligned}$ | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 4.4 | - | $\mu \mathrm{A}$ |
|  |  |  | MSI clock, 65 kHz <br> $\mathrm{f}_{\mathrm{HCLK}}=32 \mathrm{kHz}$ <br> Flash ON | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 18 | 21 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ | 24 | 27 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ | 35 | 43 |  |
|  |  |  | MSI clock, 65 kHz $\mathrm{f}_{\mathrm{HCLK}}=65 \mathrm{kHz}$, <br> Flash ON | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 18.6 | 21 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ | 24.5 | 28 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ | 35 | 42 |  |
|  |  |  | $\begin{aligned} & \text { MSI clock, } 131 \mathrm{kHz} \\ & \mathrm{f}_{\mathrm{HCLK}}=131 \mathrm{kHz} \text {, } \\ & \text { Flash ON } \end{aligned}$ | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 22 | 25 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=55^{\circ} \mathrm{C}$ | 23.5 | 26 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ | 28.5 | 31 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ | 39 | 45 |  |
|  |  | TIM9 and USART1 enabled, Flash $\mathrm{ON}, \mathrm{V}_{\mathrm{DD}}$ from 1.65 V to 3.6 V | MSI clock, 65 kHz <br> $\mathrm{f}_{\mathrm{HCLK}}=32 \mathrm{kHz}$ | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 18 | 20.5 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ | 24 | 27 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ | 35 | 43 |  |
|  |  |  | MSI clock, 65 kHz$\mathrm{f}_{\mathrm{HCLK}}=65 \mathrm{kHz}$ | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 18.6 | 21 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ | 24.5 | 28 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ | 35 | 42 |  |
|  |  |  | MSI clock, 131 kHz <br> $\mathrm{f}_{\mathrm{HCLK}}=131 \mathrm{kHz}$ | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 22 | 25 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=55^{\circ} \mathrm{C}$ | 23.5 | 26 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ | 28.5 | 31 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ | 39 | 45 |  |
| $I_{D D} \max$ (LP Sleep) | Max allowed current in Low-power sleep mode | $V_{D D} \text { from } 1.65 \mathrm{~V}$ $\text { to } 3.6 \mathrm{~V}$ | - | - | - | 200 |  |

1. Guaranteed by characterization results, unless otherwise specified.

Table 22. Typical and maximum current consumptions in Stop mode

| Symbol | Parameter | Conditions |  |  | Typ | Max ${ }^{(1)}$ | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & \mathrm{I}_{\mathrm{DD}} \text { (Stop } \\ & \text { with RTC) } \end{aligned}$ | Supply current in Stop mode with RTC enabled | RTC clocked by LSI or LSE external clock ( 32.768 kHz ), regulator in LP mode, HSI and HSE OFF (no independent watchdog) | $\begin{aligned} & \text { LCD } \\ & \text { OFF } \end{aligned}$ | $\begin{aligned} & \mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C} \text { to } 25^{\circ} \mathrm{C} \\ & \mathrm{~V}_{\mathrm{DD}}=1.8 \mathrm{~V} \end{aligned}$ | 1.1 | - | $\mu \mathrm{A}$ |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 1.35 | 4 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=55^{\circ} \mathrm{C}$ | 1.95 | 6 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ | 4.35 | 10 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ | 11.0 | 23 |  |
|  |  |  | $\begin{gathered} \text { LCD } \\ \text { ON } \\ \text { (static } \\ \text { duty) }{ }^{(2)} \end{gathered}$ | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 1.65 | 6 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=55^{\circ} \mathrm{C}$ | 2.1 | 7 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ | 4.7 | 12 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ | 11.0 | 27 |  |
|  |  |  | LCDON(1/8 duty) ${ }^{(3)}$ | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 2.5 | 10 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=55^{\circ} \mathrm{C}$ | 4.65 | 11 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ | 7.25 | 16 |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ | 14.0 | 44 |  |
|  |  | RTC clocked by LSE external quartz $(32.768 \mathrm{kHz})$, regulator in LP mode, HSI and HSE OFF (no independent watchdog ${ }^{(4)}$ | $\begin{aligned} & \text { LCD } \\ & \text { OFF } \end{aligned}$ | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 1.7 | - |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=55^{\circ} \mathrm{C}$ | 2.15 | - |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ | 4.7 | - |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ | 11.5 | - |  |
|  |  |  | LCD ON (static duty) ${ }^{(2)}$ | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 1.8 | - |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=55^{\circ} \mathrm{C}$ | 2.35 | - |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ | 4.85 | - |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ | 11.5 | - |  |
|  |  |  | LCDON(1/8 duty) ${ }^{(3)}$ | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 2.45 | - |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=55^{\circ} \mathrm{C}$ | 4.9 | - |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ | 7.7 | - |  |
|  |  |  |  | $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ | 14.5 | - |  |
|  |  |  | $\begin{aligned} & \text { LCD } \\ & \text { OFF } \end{aligned}$ | $\begin{aligned} & \mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C} \text { to } 25^{\circ} \mathrm{C} \\ & \mathrm{~V}_{\mathrm{DD}}=1.8 \mathrm{~V} \end{aligned}$ | 1.35 | - |  |
|  |  |  |  | $\begin{aligned} & \mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C} \text { to } 25^{\circ} \mathrm{C} \\ & \mathrm{~V}_{\mathrm{DD}}=3.0 \mathrm{~V} \end{aligned}$ | 1.7 | - |  |
|  |  |  |  | $\begin{aligned} & \mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C} \text { to } 25^{\circ} \mathrm{C} \\ & \mathrm{~V}_{\mathrm{DD}}=3.6 \mathrm{~V} \end{aligned}$ | 2.0 | - |  |

Table 22. Typical and maximum current consumptions in Stop mode (continued)

| Symbol | Parameter | Conditions |  | Typ | Max ${ }^{(1)}$ | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{I}_{\mathrm{DD}}$ (Stop) | Supply current in Stop mode (RTC disabled) | Regulator in LP mode, HSI and HSE OFF, independent watchdog and LSI enabled | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 1.6 | 2.2 | $\mu \mathrm{A}$ |
|  |  | Regulator in LP mode, LSI, HSI and HSE OFF (no independent watchdog) | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 0.475 | 1 |  |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=55^{\circ} \mathrm{C}$ | 0.915 | 3 |  |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ | 3.35 | 9 |  |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ | 10.0 | $22^{(5)}$ |  |
| IDD (WU from Stop) | Supply current during wakeup from Stop mode | $\mathrm{MSI}=4.2 \mathrm{MHz}$ | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 2 | - | mA |
|  |  | $\mathrm{MSI}=1.05 \mathrm{MHz}$ |  | 1.45 | - |  |
|  |  | MSI $=65 \mathrm{kHz}{ }^{(6)}$ |  | 1.45 | - |  |

1. Guaranteed by characterization results, unless otherwise specified.
2. LCD enabled with external VLCD, static duty, division ratio $=256$, all pixels active, no LCD connected.
3. LCD enabled with external VLCD, $1 / 8$ duty, $1 / 3$ bias, division ratio $=64$, all pixels active, no LCD connected.
4. Based on characterization done with a 32.768 kHz crystal (MC306-G-06Q-32.768, manufacturer JFVNY) with two 6.8 pF loading capacitors.
5. Guaranteed by test in production.
6. When $\mathrm{MSI}=64 \mathrm{kHz}$, the RMS current is measured over the first $15 \mu \mathrm{~s}$ following the wakeup event. For the remaining part of the wakeup period, the current corresponds the Run mode current.

Table 23. Typical and maximum current consumptions in Standby mode

| Symbol | Parameter | Conditions |  | Typ | Max ${ }^{(1)}$ | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $I_{D D}$ (Standby with RTC) | Supply current in Standby mode with RTC enabled | RTC clocked by LSI (no independent watchdog) | $\begin{aligned} & \mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C} \text { to } 25^{\circ} \mathrm{C} \\ & \mathrm{~V}_{\mathrm{DD}}=1.8 \mathrm{~V} \end{aligned}$ | 0.82 | - | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 1.15 | 1.9 |  |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=55^{\circ} \mathrm{C}$ | 1.15 | 2.2 |  |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ | 1.65 | 4 |  |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ | 2.75 | $8.3{ }^{(2)}$ |  |
|  |  | RTC clocked by LSE external quartz (no independent watchdog) ${ }^{(3)}$ | $\begin{aligned} & \mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C} \text { to } 25^{\circ} \mathrm{C} \\ & \mathrm{~V}_{\mathrm{DD}}=1.8 \mathrm{~V} \end{aligned}$ | 1.05 | - |  |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 1.35 | - |  |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=55^{\circ} \mathrm{C}$ | 1.55 | - |  |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ | 2.1 | - |  |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ | 3.3 | - |  |
| $l_{D D}$ (Standby) | Supply current in Standby mode (RTC disabled) | Independent watchdog and LSI enabled | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 1 | 1.7 |  |
|  |  | Independent watchdog and LSI OFF | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 0.305 | 0.6 |  |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=55^{\circ} \mathrm{C}$ | 0.365 | 0.9 |  |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ | 0.66 | 2.75 |  |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ | 2 | $7^{(2)}$ |  |
| $l_{D D}$ (WU from Standby) | Supply current during wakeup time from Standby mode | - | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $25^{\circ} \mathrm{C}$ | 1 | - | mA |

1. Guaranteed by characterization results, unless otherwise specified.
2. Guaranteed by test in production.
3. Based on characterization done with a 32.768 kHz crystal (MC306-G-06Q-32.768, manufacturer JFVNY) with two 6.8 pF loading capacitors.

## On-chip peripheral current consumption

The current consumption of the on-chip peripherals is given in the following table. The MCU is placed under the following conditions:

- all I/O pins are in input mode with a static value at $\mathrm{V}_{\mathrm{DD}}$ or $\mathrm{V}_{\mathrm{SS}}$ (no load)
- all peripherals are disabled unless otherwise mentioned
- the given value is calculated by measuring the current consumption
- with all peripherals clocked off
- with only one peripheral clocked on

Table 24. Peripheral current consumption ${ }^{(1)}$

| Peripheral |  | Typical consumption, $\mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ |  |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | $\begin{gathered} \text { Range 1, } \\ \mathrm{V}_{\text {CORE }}= \\ 1.8 \mathrm{~V} \\ \operatorname{voS}[1: 0]= \\ 01 \end{gathered}$ | $\begin{gathered} \text { Range 2, } \\ \mathrm{V}_{\text {CORE }}= \\ 1.5 \mathrm{~V} \\ \operatorname{voS}[1: 0]= \\ 10 \end{gathered}$ | $\begin{gathered} \text { Range 3, } \\ \mathrm{V}_{\text {CORE }}= \\ 1.2 \mathrm{~V} \\ \operatorname{voS}[1: 0]= \\ 11 \end{gathered}$ | Low-power sleep and run |  |
| APB1 | TIM2 | 14.3 | 12.1 | 9.5 | 12.1 | $\mu \mathrm{A} / \mathrm{MHz}$ <br> ( $\mathrm{f}_{\mathrm{HCLK}}$ ) |
|  | TIM3 | 13.8 | 11.7 | 9.2 | 11.7 |  |
|  | TIM4 | 13.2 | 11.1 | 8.7 | 11.1 |  |
|  | TIM5 | 17.7 | 14.9 | 11.8 | 14.9 |  |
|  | TIM6 | 4.8 | 4.0 | 3.0 | 4.0 |  |
|  | TIM7 | 4.7 | 3.9 | 3.0 | 3.9 |  |
|  | LCD | 5.0 | 4.1 | 3.3 | 4.1 |  |
|  | WWDG | 3.5 | 2.9 | 2.3 | 2.9 |  |
|  | SPI2 | 8.9 | 7.4 | 5.8 | 7.4 |  |
|  | SPI3 | 7.3 | 6.0 | 4.8 | 6.0 |  |
|  | USART2 | 9.4 | 7.7 | 6.1 | 7.7 |  |
|  | USART3 | 9.4 | 7.6 | 6.0 | 7.6 |  |
|  | UART4 | 10.1 | 8.4 | 6.7 | 8.4 |  |
|  | UART5 | 9.5 | 7.9 | 6.3 | 7.9 |  |
|  | I2C1 | 8.9 | 7.4 | 5.8 | 7.4 |  |
|  | I2C2 | 7.9 | 6.4 | 5.1 | 6.4 |  |
|  | USB | 21.2 | 18.0 | 14.3 | 18.0 |  |
|  | PWR | 4.0 | 3.2 | 2.5 | 3.2 |  |
|  | DAC | 6.3 | 5.5 | 4.4 | 5.5 |  |
|  | COMP | 4.9 | 3.9 | 3.2 | 3.9 |  |

Table 24. Peripheral current consumption ${ }^{(1)}$ (continued)

| Peripheral |  | Typical consumption, $\mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ |  |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | $\begin{gathered} \text { Range 1, } \\ \mathrm{V}_{\text {CORE }}= \\ 1.8 \mathrm{~V} \\ \operatorname{vos}[1: 0]= \\ 01 \end{gathered}$ | $\begin{gathered} \text { Range 2, } \\ \mathrm{V}_{\text {CORE }}= \\ 1.5 \mathrm{~V} \\ \operatorname{vos}[1: 0]= \\ 10 \end{gathered}$ | $\begin{gathered} \text { Range 3, } \\ \mathrm{V}_{\text {CORE }}= \\ 1.2 \mathrm{~V} \\ \operatorname{vos}[1: 0]= \\ 11 \end{gathered}$ | Low-power sleep and run |  |
| APB2 | $\begin{aligned} & \text { SYSCFG \& } \\ & \text { RI } \end{aligned}$ | 3.5 | 2.9 | 2.4 | 2.9 | $\mu \mathrm{A} / \mathrm{MHz}$ <br> (f $\mathrm{f}_{\mathrm{HCLK}}$ ) |
|  | TIM9 | 9.0 | 7.4 | 5.8 | 7.4 |  |
|  | TIM10 | 7.1 | 5.8 | 4.6 | 5.8 |  |
|  | TIM11 | 6.5 | 5.3 | 4.3 | 5.3 |  |
|  | ADC ${ }^{(2)}$ | 11.0 | 9.1 | 7.2 | 9.1 |  |
|  | SDIO | 28.4 | 24.2 | 19.1 | 24.2 |  |
|  | SPI1 | 5.1 | 4.2 | 3.3 | 4.2 |  |
|  | USART1 | 9.4 | 7.8 | 6.1 | 7.8 |  |
|  | GPIOA | 7.3 | 6.1 | 4.8 | 6.1 |  |
|  | GPIOB | 7.5 | 6.1 | 4.8 | 6.1 |  |
|  | GPIOC | 8.2 | 6.8 | 5.3 | 6.8 |  |
|  | GPIOD | 8.7 | 7.1 | 5.7 | 7.1 |  |
|  | GPIOE | 7.6 | 6.2 | 4.9 | 6.2 |  |
|  | GPIOF | 7.7 | 6.3 | 5.0 | 6.3 |  |
| AHB | GPIOG | 8.4 | 7.0 | 5.4 | 7.0 |  |
|  | GPIOH | 1.8 | 1.3 | 1.1 | 1.3 |  |
|  | CRC | 0.8 | 0.6 | 0.4 | 0.6 |  |
|  | FLASH | 26.3 | 19.3 | 18.3 | -(3) |  |
|  | DMA1 | 19.0 | 16.0 | 12.8 | 16.0 |  |
|  | DMA2 | 17.0 | 14.5 | 11.5 | 14.5 |  |
|  | FSMC | 16.0 | 13.4 | 10.6 | 13.4 |  |
| All enabled |  | 310 | 246 | 217 | 226.7 |  |

Table 24. Peripheral current consumption ${ }^{(1)}$ (continued)

| Peripheral |  | Typical consumption, $\mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ |  |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | $\begin{gathered} \text { Range 1, } \\ \mathrm{V}_{\text {CORE }}= \\ 1.8 \mathrm{~V} \\ \operatorname{vos}[1: 0]= \\ 01 \end{gathered}$ | $\begin{gathered} \text { Range 2, } \\ \mathrm{V}_{\text {CORE }}= \\ 1.5 \mathrm{~V} \\ \operatorname{vos}[1: 0]= \\ 10 \end{gathered}$ | $\begin{gathered} \text { Range 3, } \\ \mathrm{V}_{\text {CORE }}= \\ 1.2 \mathrm{~V} \\ \operatorname{vos}[1: 0]= \\ 11 \end{gathered}$ | Low-power sleep and run |  |
| $\mathrm{I}_{\mathrm{DD}}$ (RTC) |  | 0.4 |  |  |  | $\mu \mathrm{A}$ |
| $\mathrm{I}_{\mathrm{DD}}(\mathrm{LCD})$ |  | 3.1 |  |  |  |  |
| $\mathrm{I}_{\mathrm{DD}(\mathrm{ADC})^{(4)}}$ |  | 1450 |  |  |  |  |
| $\mathrm{IDD} \mathrm{(DAC)}^{(5)}$ |  |  |  |  |  |  |
| IDD (COMP1) |  | 0.16 |  |  |  |  |
| IDD (COMP2) | Slow mode | 2 |  |  |  |  |
|  | Fast mode | 5 |  |  |  |  |
| $\mathrm{I}_{\mathrm{DD}(\mathrm{PVD} / \mathrm{BOR})^{(6)}}$ |  | 2.6 |  |  |  |  |
| IDD (IWDG) |  | 0.25 |  |  |  |  |

1. Data based on differential $I_{D D}$ measurement between all peripherals OFF an one peripheral with clock enabled, in the following conditions: $f_{\text {HCLK }}=32 \mathrm{MHz}$ (range 1), $\mathrm{f}_{\mathrm{HCLK}}=16 \mathrm{MHz}$ (range 2), $\mathrm{f}_{\mathrm{HCLK}}=4 \mathrm{MHz}$ (range 3), $\mathrm{f}_{\text {HCLK }}=64 \mathrm{kHz}$ (Low-power run/sleep), $\mathrm{f}_{\text {APB1 }}=\mathrm{f}_{\text {HCLK }}, \mathrm{f}_{\text {APB2 }}=\mathrm{f}_{\text {HCLK }}$, default prescaler value for each peripheral. The CPU is in Sleep mode in both cases. No I/O pins toggling.
2. HSI oscillator is OFF for this measure.
3. In Low-power sleep and run mode, the Flash memory must always be in power-down mode.
4. Data based on a differential IDD measurement between ADC in reset configuration and continuous ADC conversion (HSI consumption not included).
5. Data based on a differential IDD measurement between DAC in reset configuration and continuous DAC conversion of VDD/2. DAC is in buffered mode, output is left floating.
6. Including supply current of internal reference voltage.

### 6.3.5 Wakeup time from low-power mode

The wakeup times given in the following table are measured with the MSI RC oscillator. The clock source used to wake up the device depends on the current operating mode:

- Sleep mode: the clock source is the clock that was set before entering Sleep mode
- $\quad$ Stop mode: the clock source is the MSI oscillator in the range configured before entering Stop mode
- $\quad$ Standby mode: the clock source is the MSI oscillator running at 2.1 MHz

All timings are derived from tests performed under the conditions summarized in Table 13.

Table 25. Low-power mode wakeup timings

| Symbol | Parameter | Conditions | Typ | Max ${ }^{(1)}$ | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\text {WUSLEEP }}$ | Wakeup from Sleep mode | $\mathrm{f}_{\text {HCLK }}=32 \mathrm{MHz}$ | 0.4 | - | $\mu \mathrm{s}$ |
| $t_{\text {WUSLEEP_LP }}$ | Wakeup from Low-power sleep mode, $\mathrm{f}_{\mathrm{HCLK}}=262 \mathrm{kHz}$ | $\mathrm{f}_{\text {HCLK }}=262 \mathrm{kHz}$ Flash enabled | 46 | - |  |
|  |  | $\mathrm{f}_{\text {HCLK }}=262 \mathrm{kHz}$ <br> Flash switched OFF | 46 | - |  |
| ${ }^{\text {twustop }}$ | Wakeup from Stop mode, regulator in Run mode ULP bit $=1$ and FWU bit $=1$ | $\mathrm{f}_{\mathrm{HCLK}}=\mathrm{f}_{\text {MSI }}=4.2 \mathrm{MHz}$ | 8.2 | - |  |
|  | Wakeup from Stop mode, regulator in low-power mode ULP bit $=1$ and FWU bit = 1 | $\mathrm{f}_{\mathrm{HCLK}}=\mathrm{f}_{\mathrm{MSI}}=4.2 \mathrm{MHz}$ Voltage range 1 and 2 | 7.7 | 8.9 |  |
|  |  | $\mathrm{f}_{\mathrm{HCLK}}=\mathrm{f}_{\mathrm{MSI}}=4.2 \mathrm{MHz}$ Voltage range 3 | 8.2 | 13.1 |  |
|  |  | $\mathrm{f}_{\text {HCLK }}=\mathrm{f}_{\mathrm{MSI}}=2.1 \mathrm{MHz}$ | 10.2 | 13.4 |  |
|  |  | $\mathrm{f}_{\text {HCLK }}=\mathrm{f}_{\text {MSI }}=1.05 \mathrm{MHz}$ | 16 | 20 |  |
|  |  | $\mathrm{f}_{\text {HCLK }}=\mathrm{f}_{\mathrm{MSI}}=524 \mathrm{kHz}$ | 31 | 37 |  |
|  |  | $\mathrm{f}_{\text {HCLK }}=\mathrm{f}_{\text {MSI }}=262 \mathrm{kHz}$ | 57 | 66 |  |
|  |  | $\mathrm{f}_{\text {HCLK }}=\mathrm{f}_{\text {MSI }}=131 \mathrm{kHz}$ | 112 | 123 |  |
|  |  | $\mathrm{f}_{\text {HCLK }}=\mathrm{MSI}=65 \mathrm{kHz}$ | 221 | 236 |  |
| ${ }^{\text {twustdib }}$ | Wakeup from Standby mode ULP bit $=1$ and FWU bit = 1 | $\mathrm{f}_{\text {HCLK }}=\mathrm{MSI}=2.1 \mathrm{MHz}$ | 58 | 104 |  |
|  | Wakeup from Standby mode FWU bit $=0$ | $\mathrm{f}_{\mathrm{HCLK}}=\mathrm{MSI}=2.1 \mathrm{MHz}$ | 2.6 | 3.25 | ms |

1. Guaranteed by characterization, unless otherwise specified

### 6.3.6 External clock source characteristics

High-speed external user clock generated from an external source
In bypass mode the HSE oscillator is switched off and the input pin is a standard GPIO.The external clock signal has to respect the I/O characteristics in Section 6.3.13. However, the recommended clock input waveform is shown in Figure 14.

Table 26. High-speed external user clock characteristics ${ }^{(1)}$

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{f}_{\text {HSE_ext }}$ | User external clock source <br> frequency | CSS is on or <br> PLL is used | 1 | 8 | 32 | MHz |
|  |  | CSS is off, PLL <br> not used | 0 | 8 | 32 | MHz |

Table 26. High-speed external user clock characteristics ${ }^{(1)}$ (continued)

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {HSEH }}$ | OSC_IN input pin high level voltage | - | $0.7 \mathrm{~V}_{\mathrm{DD}}$ | - | $\mathrm{V}_{\mathrm{DD}}$ | V |
| $\mathrm{V}_{\text {HSEL }}$ | OSC_IN input pin low level voltage |  | $\mathrm{V}_{\mathrm{SS}}$ | - | $0.3 \mathrm{~V}_{\mathrm{DD}}$ |  |
| $\begin{aligned} & \left.\mathrm{t}_{\mathrm{w}(\mathrm{HSEH})}\right) \\ & \left.\mathrm{t}_{\mathrm{w}(\mathrm{HSEL}}\right) \end{aligned}$ | OSC_IN high or low time |  | 12 | - | - | ns |
| $\begin{aligned} & \mathrm{t}_{\mathrm{r}(\mathrm{HSE})} \\ & \mathrm{t}_{\mathrm{f}(\text { (HSE })} \end{aligned}$ | OSC_IN rise or fall time |  | - | - | 20 |  |
| $\mathrm{C}_{\text {in(HSE) }}$ | OSC_IN input capacitance |  | - | 2.6 | - | pF |

1. Guaranteed by design.

Figure 14. High-speed external clock source AC timing diagram


## Low-speed external user clock generated from an external source

The characteristics given in the following table result from tests performed using a lowspeed external clock source, and under the conditions summarized in Table 13.

Table 27. Low-speed external user clock characteristics ${ }^{(1)}$

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| flSE_ext | User external clock source frequency | - | 1 | 32.768 | 1000 | kHz |
| V LSEH | OSC32_IN input pin high level voltage |  | $0.7 \mathrm{~V}_{\mathrm{DD}}$ | - | $V_{\text {DD }}$ | V |
| $V_{\text {LSEL }}$ | OSC32_IN input pin low level voltage |  | $\mathrm{V}_{\text {SS }}$ | - | $0.3 \mathrm{~V}_{\mathrm{DD}}$ |  |
| $\mathrm{t}_{\mathrm{w} \text { (LSEH) }}$ <br> $\mathrm{t}_{\mathrm{w}(\mathrm{LSEL})}$ | OSC32_IN high or low time |  | 465 | - | - | ns |
| $\begin{aligned} & \left.\mathrm{t}_{\mathrm{r}(\mathrm{LSE})}\right) \\ & \left.\mathrm{t}_{\mathrm{f}(\mathrm{LSE})}\right) \end{aligned}$ | OSC32_IN rise or fall time |  | - | - | 10 |  |
| $\mathrm{C}_{\text {IN(LSE) }}$ | OSC32_IN input capacitance | - | - | 0.6 | - | pF |

1. Guaranteed by design.

Figure 15. Low-speed external clock source AC timing diagram


## High-speed external clock generated from a crystal/ceramic resonator

The high-speed external (HSE) clock can be supplied with a 1 to 24 MHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on characterization results obtained with typical external components specified in Table 28. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Table 28. HSE oscillator characteristics ${ }^{(1)(2)}$

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| fosc_IN | Oscillator frequency | - | 1 | - | 24 | MHz |
| $\mathrm{R}_{\mathrm{F}}$ | Feedback resistor | - | - | 200 | - | $\mathrm{k} \Omega$ |
| C | Recommended load capacitance versus equivalent serial resistance of the crystal $\left(\mathrm{R}_{\mathrm{S}}\right)^{(3)}$ | $\mathrm{R}_{\mathrm{S}}=30 \Omega$ | - | 20 | - | pF |
| $I_{\text {HSE }}$ | HSE driving current | $\begin{gathered} \mathrm{V}_{\mathrm{DD}}=3.3 \mathrm{~V}, \\ \mathrm{~V}_{\mathrm{IN}}=\begin{array}{c} \mathrm{V}_{\mathrm{SS}} \text { with } 30 \mathrm{pF} \\ \text { load } \end{array} \end{gathered}$ | - | - | 3 | mA |
| $I_{\text {DD(HSE) }}$ | HSE oscillator power consumption | $\begin{gathered} \mathrm{C}=20 \mathrm{pF} \\ \mathrm{f}_{\mathrm{OSC}}=16 \mathrm{MHz} \end{gathered}$ | - | - | $\begin{gathered} 2.5 \text { (startup) } \\ 0.7 \text { (stabilized) } \end{gathered}$ | mA |
|  |  | $\begin{gathered} \mathrm{C}=10 \mathrm{pF} \\ \mathrm{f}_{\mathrm{OSC}}=16 \mathrm{MHz} \end{gathered}$ | - | - | $\begin{gathered} 2.5 \text { (startup) } \\ 0.46 \text { (stabilized) } \end{gathered}$ |  |
| $\mathrm{gm}_{\mathrm{m}}$ | Oscillator transconductance | Startup | 3.5 | - | - | $\mathrm{mA} / \mathrm{V}$ |
| $\mathrm{t}_{\text {SU(HSE) }}{ }^{(4)}$ | Startup time | $V_{D D}$ is stabilized | - | 1 | - | ms |

1. Resonator characteristics given by the crystal/ceramic resonator manufacturer.
2. Guaranteed by characterization results.
3. The relatively low value of the RF resistor offers a good protection against issues resulting from use in a humid environment, due to the induced leakage and the bias condition change. However, it is recommended to take this point into account if the MCU is used in tough humidity conditions.
4. $t_{\text {SU(HSE }}$ is the startup time measured from the moment it is enabled (by software) to a stabilized 8 MHz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer.

For $\mathrm{C}_{\mathrm{L} 1}$ and $\mathrm{C}_{\mathrm{L} 2}$, it is recommended to use high-quality external ceramic capacitors in the 5 pF to 25 pF range (typ.), designed for high-frequency applications, and selected to match the requirements of the crystal or resonator (see Figure 16). $\mathrm{C}_{\mathrm{L} 1}$ and $\mathrm{C}_{\mathrm{L} 2}$ are usually the same size. The crystal manufacturer typically specifies a load capacitance which is the series combination of $C_{\mathrm{L} 1}$ and $\mathrm{C}_{\mathrm{L} 2}$. PCB and MCU pin capacitance must be included ( 10 pF can be used as a rough estimate of the combined pin and board capacitance) when sizing $\mathrm{C}_{\mathrm{L} 1}$ and $\mathrm{C}_{\mathrm{L} 2}$. Refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website www.st.com.

Figure 16. HSE oscillator circuit diagram


## Low-speed external clock generated from a crystal/ceramic resonator

The low-speed external (LSE) clock can be supplied with a 32.768 kHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on characterization results obtained with typical external components specified in Table 29. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Table 29. LSE oscillator characteristics $\left(\mathrm{f}_{\text {LSE }}=32.768 \mathrm{kHz}\right.$ ) ${ }^{(\mathbf{1 )}}$

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{f}_{\text {LSE }}$ | Low speed external oscillator frequency | - | - | 32.768 | - | kHz |
| $\mathrm{R}_{\mathrm{F}}$ | Feedback resistor | - | - | 1.2 | - | $\mathrm{M} \Omega$ |
| $C^{(2)}$ | Recommended load capacitance versus equivalent serial resistance of the crystal $\left(\mathrm{R}_{\mathrm{S}}\right)^{(3)}$ | $\mathrm{R}_{\mathrm{S}}=30 \mathrm{k} \Omega$ | - | 8 | - | pF |
| $\mathrm{I}_{\text {LSE }}$ | LSE driving current | $\mathrm{V}_{\mathrm{DD}}=3.3 \mathrm{~V}, \mathrm{~V}_{\text {IN }}=\mathrm{V}_{\mathrm{SS}}$ | - | - | 1.1 | $\mu \mathrm{A}$ |
| $\mathrm{I}_{\mathrm{DD} \text { (LSE) }}$ | LSE oscillator current consumption | $\mathrm{V}_{\mathrm{DD}}=1.8 \mathrm{~V}$ | - | 450 | - | nA |
|  |  | $\mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}$ | - | 600 | - |  |
|  |  | $\mathrm{V}_{\mathrm{DD}}=3.6 \mathrm{~V}$ | - | 750 | - |  |
| $\mathrm{g}_{\mathrm{m}}$ | Oscillator transconductance | - | 3 | - | - | $\mu \mathrm{A} / \mathrm{V}$ |
| $\mathrm{t}_{\text {SU(LSE) }}{ }^{(4)}$ | Startup time | $\mathrm{V}_{\mathrm{DD}}$ is stabilized | - | 1 | - | s |

1. Guaranteed by characterization results.
2. Refer to the note and caution paragraphs below the table, and to the application note AN2867 "Oscillator design guide for ST microcontrollers".
3. The oscillator selection can be optimized in terms of supply current using an high quality resonator with small $R_{S}$ value for example MSIV-TIN 32.768 kHz . Refer to crystal manufacturer for more details.
4. $t_{\text {SU(LSE) }}$ is the startup time measured from the moment it is enabled (by software) to a stabilized 32.768 kHz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer.

Note: $\quad$ For $C_{L 1}$ and $C_{L 2}$, it is recommended to use high-quality ceramic capacitors in the 5 pF to 15 pF range selected to match the requirements of the crystal or resonator (see Figure 17). $C_{L 1}$ and $C_{L 2,}$, are usually the same size. The crystal manufacturer typically specifies a load capacitance which is the series combination of $C_{L 1}$ and $C_{L 2}$.
Load capacitance $C_{L}$ has the following formula: $C_{L}=C_{L 1} \times C_{L 2} /\left(C_{L 1}+C_{L 2}\right)+C_{\text {stray }}$ where $C_{\text {stray }}$ is the pin capacitance and board or trace PCB-related capacitance. Typically, it is between 2 pF and 7 pF .

Caution: To avoid exceeding the maximum value of $\mathrm{C}_{\mathrm{L} 1}$ and $\mathrm{C}_{\mathrm{L} 2}(15 \mathrm{pF})$ it is strongly recommended to use a resonator with a load capacitance $\mathrm{C}_{\mathrm{L}} \leq 7 \mathrm{pF}$. Never use a resonator with a load capacitance of 12.5 pF .
Example: if the user chooses a resonator with a load capacitance of $C_{L}=6 \mathrm{pF}$ and $C_{\text {stray }}=2 \mathrm{pF}$, then $\mathrm{C}_{\mathrm{L} 1}=\mathrm{C}_{\mathrm{L} 2}=8 \mathrm{pF}$.

Figure 17. Typical application with a 32.768 kHz crystal


### 6.3.7 Internal clock source characteristics

The parameters given in Table 30 are derived from tests performed under the conditions summarized in Table 13.

High-speed internal (HSI) RC oscillator
Table 30. HSI oscillator characteristics

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{f}_{\mathrm{HSI}}$ | Frequency | $\mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}$ | - | 16 | - | MHz |
| TRIM ${ }^{(1)(2)}$ | HSI user-trimmed resolution | Trimming code is not a multiple of 16 | - | $\pm 0.4$ | 0.7 | \% |
|  |  | Trimming code is a multiple of 16 | - | - | $\pm 1.5$ | \% |
| $\mathrm{ACC}_{\mathrm{HSI}}{ }^{(2)}$ | Accuracy of the factory-calibrated HSI oscillator | $V_{\text {DDA }}=3.0 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ | $-1^{(3)}$ | - | $1^{(3)}$ | \% |
|  |  | $\mathrm{V}_{\text {DDA }}=3.0 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=0$ to $55^{\circ} \mathrm{C}$ | -1.5 | - | 1.5 | \% |
|  |  | $\mathrm{V}_{\text {DDA }}=3.0 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-10$ to $70{ }^{\circ} \mathrm{C}$ | -2 | - | 2 | \% |
|  |  | $V_{\text {DDA }}=3.0 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-10$ to $85{ }^{\circ} \mathrm{C}$ | -2.5 | - | 2 | \% |
|  |  | $\mathrm{V}_{\text {DDA }}=3.0 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-10$ to $105^{\circ} \mathrm{C}$ | -4 | - | 2 | \% |
|  |  | $\begin{aligned} & \mathrm{V}_{\mathrm{DDA}}=1.65 \mathrm{~V} \text { to } 3.6 \mathrm{~V} \\ & \mathrm{~T}_{\mathrm{A}}=-40 \text { to } 105^{\circ} \mathrm{C} \end{aligned}$ | -4 | - | 3 | \% |
| $\mathrm{t}_{\text {SU(HSI) }}{ }^{(2)}$ | HSI oscillator startup time | - | - | 3.7 | 6 | $\mu \mathrm{s}$ |
| $\mathrm{I}_{\mathrm{DD}(\mathrm{HSI})}{ }^{(2)}$ | HSI oscillator power consumption | - | - | 100 | 140 | $\mu \mathrm{A}$ |

1. The trimming step differs depending on the trimming code. It is usually negative on the codes which are multiples of 16 ( $0 \times 00,0 \times 10,0 \times 20,0 \times 30 \ldots 0 \times E 0$ ).
2. Guaranteed by characterization results.
3. Guaranteed by test in production.

## Low-speed internal (LSI) RC oscillator

Table 31. LSI oscillator characteristics

| Symbol | Parameter | Min | Typ | Max | Unit |
| :---: | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{f}_{\mathrm{LSI}}{ }^{(1)}$ | LSI frequency | 26 | 38 | 56 | kHz |
| $\mathrm{D}_{\mathrm{LSI}}{ }^{(2)}$ | LSI oscillator frequency drift <br> $0^{\circ} \mathrm{C} \leq \mathrm{T}_{\mathrm{A}} \leq 105^{\circ} \mathrm{C}$ | -10 | - | 4 | $\%$ |
| $\mathrm{t}_{\mathrm{su}(\mathrm{LSI})^{(3)}}$ | LSI oscillator startup time | - | - | 200 | $\mu \mathrm{~s}$ |
| $\mathrm{I}_{\mathrm{DD}(\mathrm{LSI})^{(3)}}$ | LSI oscillator power consumption | - | 400 | 510 | nA |

1. Guaranteed by test in production.
2. This is a deviation for an individual part, once the initial frequency has been measured.
3. Guaranteed by design.

## Multi-speed internal (MSI) RC oscillator

Table 32. MSI oscillator characteristics

| Symbol | Parameter | Condition | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{f}_{\mathrm{MSI}}$ | Frequency after factory calibration, done at $V_{D D}=3.3 \mathrm{~V}$ and $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ | MSI range 0 | 65.5 | - | kHz |
|  |  | MSI range 1 | 131 | - |  |
|  |  | MSI range 2 | 262 | - |  |
|  |  | MSI range 3 | 524 | - |  |
|  |  | MSI range 4 | 1.05 | - | MHz |
|  |  | MSI range 5 | 2.1 | - |  |
|  |  | MSI range 6 | 4.2 | - |  |
| $\mathrm{ACC}_{\text {MSI }}$ | Frequency error after factory calibration | - | $\pm 0.5$ | - | \% |
| $\mathrm{D}_{\text {TEMP(MSI) }}{ }^{(1)}$ | MSI oscillator frequency drift $0^{\circ} \mathrm{C} \leq \mathrm{T}_{\mathrm{A}} \leq 105^{\circ} \mathrm{C}$ | - | $\pm 3$ | - | \% |
| $\mathrm{D}_{\text {Volt(MSI) }}{ }^{(1)}$ | MSI oscillator frequency drift $1.65 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DD}} \leq 3.6 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ | - | - | 2.5 | \%/V |
| $\mathrm{I}_{\mathrm{DD}(\mathrm{MSI})}{ }^{(2)}$ | MSI oscillator power consumption | MSI range 0 | 0.75 | - | $\mu \mathrm{A}$ |
|  |  | MSI range 1 | 1 | - |  |
|  |  | MSI range 2 | 1.5 | - |  |
|  |  | MSI range 3 | 2.5 | - |  |
|  |  | MSI range 4 | 4.5 | - |  |
|  |  | MSI range 5 | 8 | - |  |
|  |  | MSI range 6 | 15 | - |  |
| $\mathrm{t}_{\text {SU(MSI) }}$ | MSI oscillator startup time | MSI range 0 | 30 | - | $\mu \mathrm{s}$ |
|  |  | MSI range 1 | 20 | - |  |
|  |  | MSI range 2 | 15 | - |  |
|  |  | MSI range 3 | 10 | - |  |
|  |  | MSI range 4 | 6 | - |  |
|  |  | MSI range 5 | 5 | - |  |
|  |  | MSI range 6, Voltage range 1 and 2 | 3.5 | - |  |
|  |  | MSI range 6, Voltage range 3 | 5 | - |  |

Table 32. MSI oscillator characteristics (continued)

| Symbol | Parameter | Condition | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\text {STAB(MSI) }}{ }^{(2)}$ | MSI oscillator stabilization time | MSI range 0 | - | 40 | $\mu \mathrm{s}$ |
|  |  | MSI range 1 | - | 20 |  |
|  |  | MSI range 2 | - | 10 |  |
|  |  | MSI range 3 | - | 4 |  |
|  |  | MSI range 4 | - | 2.5 |  |
|  |  | MSI range 5 | - | 2 |  |
|  |  | MSI range 6, Voltage range 1 and 2 | - | 2 |  |
|  |  | MSI range 3, Voltage range 3 | - | 3 |  |
| $\mathrm{f}_{\text {OVER(MSI) }}$ | MSI oscillator frequency overshoot | Any range to range 5 | - | 4 | MHz |
|  |  | Any range to range 6 | - | 6 |  |

1. This is a deviation for an individual part, once the initial frequency has been measured.
2. Guaranteed by characterization results.

### 6.3.8 PLL characteristics

The parameters given in Table 33 are derived from tests performed under the conditions summarized in Table 13.

Table 33. PLL characteristics

| Symbol | Parameter | Value |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Typ | Max ${ }^{(1)}$ |  |
| $\mathrm{f}_{\text {PLL_IN }}$ | PLL input clock ${ }^{(2)}$ | 2 | - | 24 | MHz |
|  | PLL input clock duty cycle | 45 | - | 55 | \% |
| $\mathrm{f}_{\text {PLL_OUT }}$ | PLL output clock | 2 | - | 32 | MHz |
| $t_{\text {LOCK }}$ | PLL lock time PLL input $=16 \mathrm{MHz}$ <br> PLL VCO = 96 MHz | - | 115 | 160 | $\mu \mathrm{s}$ |
| Jitter | Cycle-to-cycle jitter | - | - | $\pm 600$ | ps |
| $\mathrm{I}_{\mathrm{DDA}}(\mathrm{PLL})$ | Current consumption on $\mathrm{V}_{\text {DDA }}$ | - | 220 | 450 | $\mu \mathrm{A}$ |
| $\mathrm{I}_{\mathrm{DD}}$ (PLL) | Current consumption on $\mathrm{V}_{\mathrm{DD}}$ | - | 120 | 150 |  |

1. Guaranteed by characterization results.
2. Take care of using the appropriate multiplier factors so as to have PLL input clock values compatible with the range defined by fPLL_out.

### 6.3.9 Memory characteristics

The characteristics are given at $\mathrm{T}_{\mathrm{A}}=-40$ to $105^{\circ} \mathrm{C}$ unless otherwise specified.

## RAM memory

Table 34. RAM and hardware registers

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| VRM | Data retention mode ${ }^{(1)}$ | STOP mode (or RESET) | 1.65 | - | - | V |

1. Minimum supply voltage without losing data stored in RAM (in Stop mode or under Reset) or in hardware registers (only in Stop mode).

## Flash memory and data EEPROM

Table 35. Flash memory and data EEPROM characteristics

| Symbol | Parameter | Conditions | Min | Typ | Max ${ }^{(1)}$ | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $V_{D D}$ | Operating voltage Read / Write / Erase | - | 1.65 | - | 3.6 | V |
| $\mathrm{t}_{\text {prog }}$ | Programming/ erasing time for byte / word / double word / half-page | Erasing | - | 3.28 | 3.94 | ms |
|  |  | Programming | - | 3.28 | 3.94 |  |
| $\mathrm{I}_{\mathrm{DD}}$ | Average current during the whole programming / erase operation | $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{DD}}=3.6 \mathrm{~V}$ | - | 600 | 900 | $\mu \mathrm{A}$ |
|  | Maximum current (peak) during the whole programming / erase operation |  | - | 1.5 | 2.5 | mA |

1. Guaranteed by design.

Table 36. Flash memory and data EEPROM endurance and retention

| Symbol | Parameter | Conditions | Value |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Min ${ }^{(1)}$ | Typ | Max |  |
| $\mathrm{N}_{\mathrm{CYC}}{ }^{(2)}$ | Cycling (erase / write) Program memory | $\begin{aligned} & \mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C} \text { to } \\ & 105^{\circ} \mathrm{C} \end{aligned}$ | 10 | - | - | kcycles |
|  | Cycling (erase / write) EEPROM data memory |  | 300 | - | - |  |
| $\mathrm{t}_{\mathrm{RET}}{ }^{(2)}$ | Data retention (program memory) after 10 kcycles at $\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ | $\mathrm{T}_{\text {RET }}=+85^{\circ} \mathrm{C}$ | 30 | - | - | years |
|  | Data retention (EEPROM data memory) after 300 kcycles at $\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ |  | 30 | - | - |  |
|  | Data retention (program memory) after 10 kcycles at $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ | $\mathrm{T}_{\text {RET }}=+105^{\circ} \mathrm{C}$ | 10 | - | - |  |
|  | Data retention (EEPROM data memory) after 300 kcycles at $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ |  | 10 | - | - |  |

1. Guaranteed by characterization results.
2. Characterization is done according to JEDEC JESD22-A117.

### 6.3.10 FSMC characteristics

## Asynchronous waveforms and timings

Figure 18 through Figure 21 represent asynchronous waveforms and Table 37 through Table 40 provide the corresponding timings. The results shown in these tables are obtained with the following FSMC configuration:

- AddressSetupTime $=0$ (AddressSetupTime $=1$, for asynchronous multiplexed modes)
- AddressHoldTime = 1
- DataSetupTime = 1

Figure 18. Asynchronous non-multiplexed SRAM/PSRAM/NOR read waveforms


1. Mode 2/B, C and D only. In Mode 1, FSMC_NADV is not used.

Table 37. Asynchronous non-multiplexed SRAM/PSRAM/NOR read timings ${ }^{(1)}$

| Symbol | Parameter | Min | Max | Unit |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\mathrm{w} \text { (NE) }}$ | FSMC_NE low time | THCLK ${ }^{-2}$ | THCLK | ns |
| $\mathrm{t}_{\mathrm{v} \text { ( } \mathrm{NOE} \text { - NE) }}$ | FSMC_NEx low to FSMC_NOE low | 0 | 2 | ns |
| $\mathrm{t}_{\mathrm{w} \text { (NOE) }}$ | FSMC_NOE low time | $\mathrm{T}_{\text {HCLK }}$ | $\mathrm{T}_{\text {HCLK }}$ - 1 | ns |
| $\mathrm{t}_{\mathrm{h}}$ (NE_NOE) | FSMC_NOE high to FSMC_NE high hold time | 0 | - | ns |
| $\mathrm{t}_{\mathrm{V}(\mathrm{A} \text { - } \mathrm{NE})}$ | FSMC_NEx low to FSMC_A valid | - | 4 | ns |
| $\mathrm{t}_{\mathrm{h} \text { (A_NOE) }}$ | Address hold time after FSMC_NOE high | $\mathrm{T}_{\text {HCLK }}+1.5$ | - | ns |
| $\mathrm{t}_{\mathrm{v} \text { (BL_NE) }}$ | FSMC_NEx low to FSMC_BL valid | - | 0.5 | ns |
| $\mathrm{th}_{\text {(BL_NOE) }}$ | FSMC_BL hold time after FSMC_NOE high | 2* ${ }_{\text {HCLK }}-0.5$ | - | ns |
| $\mathrm{t}_{\text {su( }}$ (Data_NE) | Data to FSMC_NEx high setup time | $\mathrm{T}_{\text {HCLK }}$ | - | ns |
| $\mathrm{t}_{\text {su(Data_NOE) }}$ | Data to FSMC_NOEx high setup time | $\mathrm{T}_{\text {HCLK }}$ | - | ns |
| $\mathrm{th}_{\text {(Data_NOE) }}$ | Data hold time after FSMC_NOE high | 0 | - | ns |
| $\mathrm{t}_{\text {h(Data_NE) }}$ | Data hold time after FSMC_NEx high | 0 | - | ns |
| $\mathrm{t}_{\mathrm{v} \text { (NADV_NE) }}$ | FSMC_NEx low to FSMC_NADV low | - | 2 | ns |
| $\mathrm{t}_{\text {w }}$ (NADV) | FSMC_NADV low time | - | $\mathrm{T}_{\text {HCLK }}$ | ns |

1. $C_{L}=30 \mathrm{pF}$.

Figure 19. Asynchronous non-multiplexed SRAM/PSRAM/NOR write waveforms

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1. Mode 2/B, C and D only. In Mode 1, FSMC_NADV is not used.

Table 38. Asynchronous non-multiplexed SRAM/PSRAM/NOR write timings ${ }^{(1)}$

| Symbol | Parameter | Min | Max | Unit |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\mathrm{w} \text { (NE) }}$ | FSMC_NE low time | 2*T ${ }_{\text {HCLK }}-3$ | $2^{\star} \mathrm{T}_{\text {HCLK }}+2$ | ns |
| $\mathrm{t}_{\text {v }}$ (NWE_NE) | FSMC_NEx low to FSMC_NWE low | 0.5 | 1 | ns |
| $\mathrm{t}_{\mathrm{w} \text { (NWE) }}$ | FSMC_NWE low time | $\mathrm{T}_{\text {HCLK }}-2$ | $\mathrm{T}_{\text {HCLK }}+3$ | ns |
| $\mathrm{t}_{\mathrm{h} \text { (NE_NWE) }}$ | FSMC_NWE high to FSMC_NE high hold time | THCLK -2.5 | - | ns |
| $\mathrm{t}_{\mathrm{v}}\left(\mathrm{A} \_\mathrm{NE}\right)$ | FSMC_NEx low to FSMC_A valid | - | 0 | ns |
| $\mathrm{t}_{\text {h(A_NWE) }}$ | Address hold time after FSMC_NWE high | T ${ }_{\text {HCLK }}$ - 2.5 | - | ns |
| $\mathrm{t}_{\mathrm{v} \text { (BL_NE) }}$ | FSMC_NEx low to FSMC_BL valid | - | 0 | ns |
| $\mathrm{t}_{\mathrm{h} \text { (BL_NWE) }}$ | FSMC_BL hold time after FSMC_NWE high | $\mathrm{T}_{\text {HCLK }}-4$ | - | ns |
| $\mathrm{t}_{\mathrm{v} \text { (Data_NE) }}$ | FSMC_NEx low to Data valid | - | $\mathrm{T}_{\text {HCLK }}$ | ns |
| $\mathrm{t}_{\mathrm{h} \text { (Data_NWE) }}$ | Data hold time after FSMC_NWE high | $\mathrm{T}_{\text {HCLK }}-2.5$ | - | ns |

1. $\mathrm{C}_{\mathrm{L}}=30 \mathrm{pF}$.

Figure 20. Asynchronous multiplexed PSRAM/NOR read waveforms


Table 39. Asynchronous multiplexed PSRAM/NOR read timings ${ }^{(1)}$

| Symbol | Parameter | Min | Max | Unit |
| :--- | :--- | :---: | :---: | :---: |
| $\mathrm{t}_{\mathrm{w} \text { (NE) }}$ | FSMC_NE low time | $3^{*} \mathrm{~T}_{\text {HCLK }}-1.5$ | $3^{*} \mathrm{~T}_{\text {HCLK }}+1$ | ns |
| $\mathrm{t}_{\text {v(NOE_NE) }}$ | FSMC_NEx low to FSMC_NOE low | $2^{*} \mathrm{~T}_{\text {HCLK }}-1$ | $2^{*} \mathrm{~T}_{\text {HCLK }}$ | ns |
| $\mathrm{t}_{\mathrm{w} \text { (NOE) }}$ | FSMC_NOE low time | $\mathrm{T}_{\text {HCLK }}-0.5$ | $\mathrm{~T}_{\text {HCLK }}+0.5$ | ns |
| $\mathrm{t}_{\mathrm{h} \text { (NE_NOE) }}$ | FSMC_NOE high to FSMC_NE high hold time | 0 | - | ns |
| $\mathrm{t}_{\text {v(A_NE) }}$ | FSMC_NEx low to FSMC_A valid | - | 5 | ns |
| $\mathrm{t}_{\text {v(NADV_NE) }}$ | FSMC_NEx low to FSMC_NADV low | 1.5 | 2 | ns |
| $\mathrm{t}_{\mathrm{w} \text { (NADV) }}$ | FSMC_NADV low time | $\mathrm{T}_{\text {HCLK }}-0.5$ | $\mathrm{~T}_{\text {HCLK }}$ | ns |
| $\mathrm{t}_{\mathrm{h} \text { (AD_NADV) }}$ | FSMC_AD(address) valid hold time after <br> FSMC_NADV high | $\mathrm{T}_{\text {HCLK }}-6$ | - | ns |
| $\mathrm{t}_{\mathrm{h} \text { (A_NOE) }}$ | Address hold time after FSMC_NOE high | $2^{*} \mathrm{~T}_{\text {HCLK }}-1$ | - | ns |
| $\mathrm{t}_{\mathrm{h} \text { (BL_NOE) }}$ | FSMC_BL time after FSMC_NOE high | 1.5 | - | ns |
| $\mathrm{t}_{\mathrm{v} \text { (BL_NE) }}$ | FSMC_NEx low to FSMC_BL valid | - | 0 | ns |
| $\mathrm{t}_{\text {su(Data_NE) }}$ | Data to FSMC_NEx high setup time | $\mathrm{T}_{\text {HCLK }}$ | - | ns |
| $\mathrm{t}_{\text {su(Data_NOE) }}$ | Data to FSMC_NOE high setup time | $\mathrm{T}_{\text {HCLK }}$ | - | ns |
| $\mathrm{t}_{\mathrm{h} \text { (Data_NE) }}$ | Data hold time after FSMC_NEx high | 0 | - | ns |
| $\mathrm{t}_{\mathrm{h} \text { (Data_NOE) }}$ | Data hold time after FSMC_NOE high | 0 | - | ns |

1. $C_{L}=30 \mathrm{pF}$.

Figure 21. Asynchronous multiplexed PSRAM/NOR write waveforms

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Table 40. Asynchronous multiplexed PSRAM/NOR write timings ${ }^{(1)}$

| Symbol | Parameter | Min | Max | Unit |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\mathrm{w} \text { (NE) }}$ | FSMC_NE low time | $4^{*} \mathrm{~T}_{\text {HCLK }}-3$ | $4^{*} \mathrm{~T}_{\text {HCLK }}+2$ | ns |
| $\mathrm{t}_{\mathrm{v} \text { (NWE_NE) }}$ | FSMC_NEx low to FSMC_NWE low | $\mathrm{T}_{\text {HCLK }}$ | $\mathrm{T}_{\text {HCLK }}+1$ | ns |
| $\mathrm{t}_{\mathrm{w} \text { (NWE) }}$ | FSMC_NWE low time | $2^{*} \mathrm{~T}_{\text {HCLK }}-2$ | $2^{*} \mathrm{~T}_{\text {HCLK }}+4$ | ns |
| $t_{\text {h(NE_NWE) }}$ | FSMC_NWE high to FSMC_NE high hold time | T ${ }_{\text {HCLK }}-2.5$ | - | ns |
| $\mathrm{t}_{\mathrm{v} \text { (A_NE) }}$ | FSMC_NEx low to FSMC_A valid | - | 6 | ns |
| $t_{\text {v(NADV_NE) }}$ | FSMC_NEx low to FSMC_NADV low | 1.5 | 2 | ns |
| $\mathrm{t}_{\mathrm{w} \text { (NADV) }}$ | FSMC_NADV low time | $\mathrm{T}_{\text {HCLK }}-4$ | $\mathrm{T}_{\text {HCLK }}+4$ | ns |
| $\mathrm{t}_{\text {( }}^{\text {(AD_NADV })}$ | FSMC_AD (address) valid hold time after FSMC_NADV high | $\mathrm{T}_{\text {HCLK }}{ }^{-5}$ | - | ns |
| $\mathrm{t}_{\text {h(A_NWE) }}$ | Address hold time after FSMC_NWE high | $\mathrm{T}_{\text {HCLK }}-2.5$ | - | ns |
| $t_{\text {h (BL_NWE) }}$ | FSMC_BL hold time after FSMC_NWE high | $\mathrm{T}_{\text {HCLK }}$ - 3 | - | ns |
| $\mathrm{t}_{\mathrm{v} \text { (BL_NE) }}$ | FSMC_NEx low to FSMC_BL valid | - | 0.5 | ns |
| $\mathrm{t}_{\mathrm{v} \text { (Data_NADV) }}$ | FSMC_NADV high to Data valid | - | $\mathrm{T}_{\text {HCLK }}+6$ | ns |
| $t_{\text {h( }}$ (Data_NWE) | Data hold time after FSMC_NWE high | T ${ }_{\text {HCLK }}-2.5$ | - | ns |

1. $\mathrm{C}_{\mathrm{L}}=30 \mathrm{pF}$.

## Synchronous waveforms and timings

Figure 22 through Figure 25 represent synchronous waveforms and Table 42 through Table 44 provide the corresponding timings. The results shown in these tables are obtained with the following FSMC configuration:

- BurstAccessMode = FSMC_BurstAccessMode_Enable;
- MemoryType = FSMC_MemoryType_CRAM;
- WriteBurst = FSMC_WriteBurst_Enable;
- CLKDivision = 1;
- DataLatency = 1 for NOR Flash; DataLatency $=0$ for PSRAM

Figure 22. Synchronous multiplexed NOR/PSRAM read timings


Table 41. Synchronous multiplexed NOR/PSRAM read timings ${ }^{(1)}$

| Symbol | Parameter | Min | Max | Unit |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\mathrm{w} \text { (CLK) }}$ | FSMC_CLK period | $\begin{gathered} 2^{\star} \mathrm{T}_{\text {HCLK }}- \\ 0.5 \end{gathered}$ | - | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-NExL) }}$ | FSMC_CLK low to FSMC_NEx low (x = 0...2) | - | 0 | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-NExH) }}$ | FSMC_CLK low to FSMC_NEx high ( $\mathrm{x}=0 . . .2$ ) | $\underset{1.5}{\mathrm{~T}_{\mathrm{HCLK}}+}$ | - | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-NADVL) }}$ | FSMC_CLK low to FSMC_NADV low | - | 3 | ns |
| $\mathrm{t}_{\mathrm{d}(\text { CLKL-NADVH) }}$ | FSMC_CLK low to FSMC_NADV high | 3.5 | - | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-AV) }}$ | FSMC_CLK low to FSMC_Ax valid ( $\mathrm{x}=16 . . .25$ ) | - | 0 | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-AIV) }}$ | FSMC_CLK low to FSMC_Ax invalid ( $x=16 . .25$ ) | 0 | - | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-NOEL) }}$ | FSMC_CLK low to FSMC_NOE low | - | $\mathrm{T}_{\text {HCLK }}$ - 1 | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-NOEH) }}$ | FSMC_CLK low to FSMC_NOE high | 2.5 | - | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-ADV) }}$ | FSMC_CLK low to FSMC_AD[15:0] valid | - | 4 | ns |
| $\mathrm{t}_{\text {(} \text { (CLKL-ADIV) }}$ | FSMC_CLK low to FSMC_AD[15:0] invalid | 0 | - | ns |
| $\mathrm{t}_{\text {su(ADV-CLKH) }}$ | FSMC_A/D[15:0] valid data before FSMC_CLK high | 6 | - | ns |
| $\mathrm{th}_{\text {( }}$ CLKH-ADV) | FSMC_A/D[15:0] valid data after FSMC_CLK high | 4 | - | ns |
| $\mathrm{t}_{\text {su(NWAITV-CLKH) }}$ | FSMC_NWAIT valid before FSMC_CLK high | 6 | - | ns |
| $\mathrm{t}_{\mathrm{h} \text { (CLKH-NWAITV) }}$ | FSMC_NWAIT valid after FSMC_CLK high | 0 | - | ns |

1. $\mathrm{C}_{\mathrm{L}}=30 \mathrm{pF}$.

Figure 23. Synchronous multiplexed PSRAM write timings


Table 42. Synchronous multiplexed PSRAM write timings ${ }^{(1)}$

| Symbol | Parameter | Min | Max | Unit |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\mathrm{w} \text { (CLK) }}$ | FSMC_CLK period | 2*T ${ }_{\text {HCLK }}$ | - | ns |
| $\mathrm{t}_{\text {d(CLKL-NExL) }}$ | FSMC_CLK low to FSMC_NEx low (x = 0...2) | - | 0 | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-NExH) }}$ | FSMC_CLK low to FSMC_NEx high ( $\mathrm{x}=0 . . .2$ ) | 0 | - | ns |
| $\mathrm{t}_{\mathrm{d}(\text { CLKL-NADVL) }}$ | FSMC_CLK low to FSMC_NADV low | - | 0 | ns |
| $\mathrm{t}_{\mathrm{d}(\mathrm{CLKL}}$ (NADVH) | FSMC_CLK low to FSMC_NADV high | 0 | - | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-AV) }}$ | FSMC_CLK low to FSMC_Ax valid ( $\mathrm{x}=16 \ldots 25$ ) | - | 0 | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-AIV) }}$ | FSMC_CLK low to FSMC_Ax invalid (x = 16...25) | $\mathrm{T}_{\text {HCLK }}+4$ | - | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-NWEL) }}$ | FSMC_CLK low to FSMC_NWE low | - | 0 | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-NWEH) }}$ | FSMC_CLK low to FSMC_NWE high | 1 | - | ns |
| $\mathrm{t}_{\text {(CLKL-ADIV) }}$ | FSMC_CLK low to FSMC_AD[15:0] invalid | 5 | - | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-DATA) }}$ | FSMC_A/D[15:0] valid after FSMC_CLK low | - | 6 | ns |
| $\mathrm{t}_{\text {su(NWAITV-CLKH) }}$ | FSMC_NWAIT valid before FSMC_CLK high | 6 | - | ns |
| $\mathrm{t}_{\mathrm{h} \text { (CLKH-NWAITV) }}$ | FSMC_NWAIT valid after FSMC_CLK high | 0 | - | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-NBLH) }}$ | FSMC_CLK low to FSMC_NBL high | 1 | - | ns |

1. $C_{L}=30 \mathrm{pF}$.

Figure 24. Synchronous non-multiplexed NOR/PSRAM read timings


Table 43. Synchronous non-multiplexed NOR/PSRAM read timings ${ }^{(1)}$

| Symbol | Parameter | Min | Max | Unit |
| :---: | :---: | :---: | :---: | :---: |
| $t_{\text {w (CLK }}$ | FSMC_CLK period | $\begin{gathered} 2^{*} \mathrm{~T}_{\text {HCLK }} \\ 0.5 \end{gathered}$ | - | ns |
| $\mathrm{t}_{\mathrm{d}(\text { CLKL-NExL) }}$ | FSMC_CLK low to FSMC_NEx low (x = 0...2) | - | 0 | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-NExH) }}$ | FSMC_CLK low to FSMC_NEx high ( $\mathrm{x}=0 . . .2$ ) | 0 | - | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-NADVL) }}$ | FSMC_CLK low to FSMC_NADV low | - | 3 | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-NADVH) }}$ | FSMC_CLK low to FSMC_NADV high | 3.5 | - | ns |
| $\mathrm{t}_{\mathrm{d}(\mathrm{CLKL}-\mathrm{AV})}$ | FSMC_CLK low to FSMC_Ax valid ( $\mathrm{x}=16 . .25$ ) | - | 0 | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-AIV) }}$ | FSMC_CLK low to FSMC_Ax invalid (x = 16...25) | 0 | - | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-NOEL) }}$ | FSMC_CLK low to FSMC_NOE low | - | $\mathrm{T}_{\text {HCLK }}+1$ | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-NOEH) }}$ | FSMC_CLK low to FSMC_NOE high | 2.5 | - | ns |
| $\mathrm{t}_{\text {su( }}$ (DV-CLKH) | FSMC_D[15:0] valid data before FSMC_CLK high | 4 | - | ns |
| $\mathrm{t}_{\mathrm{h}}$ (CLKH-DV) | FSMC_D[15:0] valid data after FSMC_CLK high | 4 | - | ns |

Table 43. Synchronous non-multiplexed NOR/PSRAM read timings ${ }^{(1)}$ (continued)

| Symbol | Parameter | Min | Max | Unit |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\text {su(NWAITV-CLKH) }}$ | FSMC_NWAIT valid before FSMC_CLK high | 6 | - | ns |
| $\mathrm{t}_{\mathrm{h} \text { (CLKH-NWAITV) }}$ | FSMC_NWAIT valid after FSMC_CLK high | 0 | - | ns |

1. $\mathrm{C}_{\mathrm{L}}=30 \mathrm{pF}$.

Figure 25. Synchronous non-multiplexed PSRAM write timings


Table 44. Synchronous non-multiplexed PSRAM write timings ${ }^{(1)}$

| Symbol | Parameter | Min | Max | Unit |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\mathrm{w} \text { (CLK) }}$ | FSMC_CLK period | $2^{*} \mathrm{~T}_{\text {HCLK }}{ }^{-3}$ | - | ns |
| $\mathrm{t}_{\text {d(CLKL-NExL) }}$ | FSMC_CLK low to FSMC_NEx low (x = 0...2) | - | 0 | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-NExH) }}$ | FSMC_CLK low to FSMC_NEx high ( $\mathrm{x}=0 . . .2$ ) | 1 | - | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-NADVL) }}$ | FSMC_CLK low to FSMC_NADV low | - | 5 | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-NADVH) }}$ | FSMC_CLK low to FSMC_NADV high | 7 | - | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-AV) }}$ | FSMC_CLK low to FSMC_Ax valid ( $\mathrm{x}=16 . . .25$ ) | - | 0 | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-AIV) }}$ | FSMC_CLK low to FSMC_Ax invalid ( $\mathrm{x}=16 . . .25$ ) | $\mathrm{T}_{\text {HCLK }}+4$ | - | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-NWEL) }}$ | FSMC_CLK low to FSMC_NWE low | - | 2 | ns |

Table 44. Synchronous non-multiplexed PSRAM write timings ${ }^{(1)}$ (continued)

| Symbol | Parameter | Min | Max | Unit |
| :--- | :--- | :---: | :---: | :--- |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-NWEH) }}$ | FSMC_CLK low to FSMC_NWE high | 5 | - | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-DATA) }}$ | FSMC_D[15:0] valid data after FSMC_CLK low | - | 7 | ns |
| $\mathrm{t}_{\mathrm{d} \text { (CLKL-NBLH) }}$ | FSMC_CLK low to FSMC_NBL high | 3 | - | ns |
| $\mathrm{t}_{\text {su(NWAITV-CLKH) }}$ | FSMC_NWAIT valid before FSMC_CLK high | 6 | - | ns |
| $\mathrm{t}_{\text {h(CLKH-NWAITV) }}$ | FSMC_NWAIT valid after FSMC_CLK high | 0 | - | ns |

1. $\mathrm{C}_{\mathrm{L}}=30 \mathrm{pF}$.

### 6.3.11 EMC characteristics

Susceptibility tests are performed on a sample basis during device characterization.

## Functional EMS (electromagnetic susceptibility)

While a simple application is executed on the device (toggling 2 LEDs through I/O ports). the device is stressed by two electromagnetic events until a failure occurs. The failure is indicated by the LEDs:

- Electrostatic discharge (ESD) (positive and negative) is applied to all device pins until a functional disturbance occurs. This test is compliant with the IEC 61000-4-2 standard.
- FTB: A Burst of Fast Transient voltage (positive and negative) is applied to $V_{D D}$ and $\mathrm{V}_{\mathrm{SS}}$ through a 100 pF capacitor, until a functional disturbance occurs. This test is compliant with the IEC 61000-4-4 standard.

A device reset allows normal operations to be resumed.
The test results are given in Table 45. They are based on the EMS levels and classes defined in application note AN1709.

Table 45. EMS characteristics

| Symbol | Parameter | Conditions | Level/ <br> Class |
| :--- | :--- | :--- | :---: |
| $\mathrm{V}_{\text {FESD }}$ | Voltage limits to be applied on any $\mathrm{I} / \mathrm{O}$ pin to <br> induce a functional disturbance | $\mathrm{V}_{\mathrm{DD}}=3.3 \mathrm{~V}, \mathrm{LQFP} 100, \mathrm{~T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, <br> $\mathrm{f}_{\mathrm{HCLK}}=32 \mathrm{MHz}$ <br> conforms to IEC 61000-4-2 | 2 B |
| $\mathrm{~V}_{\text {EFTB }}$ | Fast transient voltage burst limits to be <br> applied through 100 pF on $\mathrm{V}_{\mathrm{DD}}$ and $\mathrm{V}_{\mathrm{SS}}$ <br> pins to induce a functional disturbance | $\mathrm{V}_{\mathrm{DD}}=3.3 \mathrm{~V}, \mathrm{LQFP} 100, \mathrm{~T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, <br> $\mathrm{f}_{\mathrm{HCLK}}=32 \mathrm{MHz}$ <br> conforms to IEC 61000-4-4 | 4 A |

## Designing hardened software to avoid noise problems

EMC characterization and optimization are performed at component level with a typical application environment and simplified MCU software. It must be noted that good EMC performance is highly dependent on the user application and the software in particular.
Therefore it is recommended that the user applies EMC software optimization and prequalification tests in relation with the EMC level requested for his application.

## Software recommendations

The software flowchart must include the management of runaway conditions such as:

- Corrupted program counter
- Unexpected reset
- Critical data corruption (control registers...)

Prequalification trials
Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the NRST pin or the oscillator pins for 1 second

To complete these trials, ESD stress can be applied directly on the device, over the range of specification values. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors occurring (see application note AN1015).

## Electromagnetic Interference (EMI)

The electromagnetic field emitted by the device are monitored while a simple application is executed (toggling 2 LEDs through the I/O ports). This emission test is compliant with IEC 61967-2 standard which specifies the test board and the pin loading.

Table 46. EMI characteristics

| Symbol | Parameter | Conditions | Monitored frequency band | Max vs. frequency range |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | 4 MHz <br> voltage <br> range 3 | 16 MHz voltage range 2 | 32 MHz voltage range 1 |  |
| $\mathrm{S}_{\text {EMI }}$ | Peak level | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=3.3 \mathrm{~V}, \\ & \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \end{aligned}$ <br> LQFP100 package compliant with IEC 61967-2 | 0.1 to 30 MHz | 3 | -6 | -5 | $\mathrm{dB} \mu \mathrm{V}$ |
|  |  |  | 30 to 130 MHz | 18 | 4 | -7 |  |
|  |  |  | 130 MHz to 1GHz | 15 | 5 | -7 |  |
|  |  |  | SAE EMI Level | 2.5 | 2 | 1 | - |

### 6.3.12 Electrical sensitivity characteristics

Based on three different tests (ESD, LU) using specific measurement methods, the device is stressed in order to determine its performance in terms of electrical sensitivity.

## Electrostatic discharge (ESD)

Electrostatic discharges (a positive then a negative pulse separated by 1 second) are applied to the pins of each sample according to each pin combination. The sample size depends on the number of supply pins in the device ( 3 parts $\times(n+1$ ) supply pins). This test conforms to the JESD22-A114/C101 standard.

Table 47. ESD absolute maximum ratings

| Symbol | Ratings | Conditions | Class | Maximum <br> value | Unit |
| :---: | :--- | :--- | :---: | :---: | :---: |
| $\mathrm{V}_{\text {ESD(HBM) }}$ | Electrostatic <br> discharge voltage <br> (human body model) | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, conforming <br> to JESD22-A114 | 2 | 2000 | V |
| $\mathrm{~V}_{\text {ESD(CDM) }}$ | Electrostatic <br> discharge voltage <br> (charge device model) | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, conforming <br> to JESD22-C101 | III | 500 | V |

1. Guaranteed by characterization results.

## Static latch-up

Two complementary static tests are required on six parts to assess the latch-up performance:

- A supply overvoltage is applied to each power supply pin
- A current injection is applied to each input, output and configurable I/O pin

These tests are compliant with EIA/JESD 78A IC latch-up standard.
Table 48. Electrical sensitivities

| Symbol | Parameter | Conditions | Class |
| :---: | :---: | :---: | :---: |
| LU | Static latch-up class | $\mathrm{T}_{\mathrm{A}}=+105^{\circ} \mathrm{C}$ conforming to JESD78A | II level A |

### 6.3.13 I/O current injection characteristics

As a general rule, current injection to the I/O pins, due to external voltage below $\mathrm{V}_{\mathrm{SS}}$ or above $\mathrm{V}_{\mathrm{DD}}$ (for standard pins) must be avoided during normal product operation. However, in order to give an indication of the robustness of the microcontroller in cases when abnormal injection accidentally happens, susceptibility tests are performed on a sample basis during device characterization.

## Functional susceptibility to I/O current injection

While a simple application is executed on the device, the device is stressed by injecting current into the I/O pins programmed in floating input mode. While current is injected into the I/O pin, one at a time, the device is checked for functional failures.

The failure is indicated by an out of range parameter: ADC error above a certain limit (higher than 5 LSB TUE), out of conventional limits of induced leakage current on adjacent pins (out of $-5 \mu \mathrm{~A} /+0 \mu \mathrm{~A}$ range), or other functional failure (for example reset occurrence oscillator frequency deviation, LCD levels).
The test results are given in the Table 49.
Table 49. I/O current injection susceptibility

| Symbol | Description | Functional susceptibility |  | Unit |
| :--- | :--- | :---: | :---: | :---: |
|  |  | Negative <br> injection | Positive <br> injection |  |
|  | Injected current on all 5 V tolerant (FT) pins | $-5^{(1)}$ | $\mathrm{NA}^{(2)}$ |  |
|  | Injected current on BOOT0 | -0 | $\mathrm{NA} \mathrm{A}^{(2)}$ | +5 |
|  | Injected current on any other pin | $-5^{(1)}$ | +5 |  |

1. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative currents.
2. Injection is not possible.

### 6.3.14 I/O port characteristics

## General input/output characteristics

Unless otherwise specified, the parameters given in Table 56 are derived from tests performed under the conditions summarized in Table 13. All I/Os are CMOS and TTL compliant.

Table 50. I/O static characteristics

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {IL }}$ | Input low level voltage | - | - | - | $\underset{(1)}{0.3 V_{\mathrm{DD}}}$ | V |
| $\mathrm{V}_{\mathrm{IH}}$ | Input high level voltage | Standard I/O | $0.7 \mathrm{~V}_{\mathrm{DD}}$ | - | - |  |
|  |  | FT I/O |  | - | - |  |
|  |  | BOOTO I/O |  | - | - |  |
| $V_{\text {hys }}$ | I/O Schmitt trigger voltage hysteresis ${ }^{(2)}$ | Standard I/O | - | $\begin{gathered} 10 \% \\ V_{D D}^{(3)} \end{gathered}$ | - |  |
| $\mathrm{I}_{\mathrm{lkg}}$ | Input leakage current ${ }^{(4)}$ | $\begin{aligned} & \mathrm{V}_{S S} \leq \mathrm{V}_{\mathrm{IN}} \leq \mathrm{V}_{\mathrm{DD}} \\ & \mathrm{I} / \mathrm{Os} \text { with } \mathrm{LCD} \end{aligned}$ | - | - | $\pm 50$ | nA |
|  |  | $V_{S S} \leq V_{I N} \leq V_{D D}$ $\mathrm{I} / \mathrm{Os}$ with analog switches | - | - | $\pm 50$ |  |
|  |  | $\mathrm{V}_{\text {SS }} \leq \mathrm{V}_{\text {IN }} \leq \mathrm{V}_{\mathrm{DD}}$ I/Os with analog switches and LCD | - | - | $\pm 50$ |  |
|  |  | $\begin{aligned} & V_{S S} \leq V_{I N} \leq V_{D D} \\ & \text { I/Os with USB } \end{aligned}$ | - | - | $\pm 250$ |  |
|  |  | $\begin{aligned} & \mathrm{V}_{\mathrm{SS}} \leq \mathrm{V}_{\mathrm{IN}} \leq \mathrm{V}_{\mathrm{DD}} \\ & \text { Standard I/Os } \end{aligned}$ | - | - | $\pm 50$ |  |
|  |  | $\begin{gathered} \mathrm{FT} \mathrm{I} / \mathrm{O} \\ \mathrm{~V}_{\mathrm{DD}} \leq \mathrm{V}_{\mathrm{IN}} \leq 5 \mathrm{~V} \end{gathered}$ | - | - | $\pm 10$ | uA |
| RPU | Weak pull-up equivalent resistor ${ }^{(1)(5)}$ | $\mathrm{V}_{\mathrm{IN}}=\mathrm{V}_{\text {SS }}$ | 30 | 45 | 60 | $\mathrm{k} \Omega$ |
| $\mathrm{R}_{\mathrm{PD}}$ | Weak pull-down equivalent resistor ${ }^{(5)}$ | $\mathrm{V}_{\mathrm{IN}}=\mathrm{V}_{\mathrm{DD}}$ | 30 | 45 | 60 | $\mathrm{k} \Omega$ |
| $\mathrm{C}_{10}$ | I/O pin capacitance | - | - | 5 | - | pF |

1. Guaranteed by test in production
2. Hysteresis voltage between Schmitt trigger switching levels. Guaranteed by characterization results.
3. With a minimum of 200 mV . Guaranteed by characterization results.
4. The max. value may be exceeded if negative current is injected on adjacent pins.
5. Pull-up and pull-down resistors are designed with a true resistance in series with a switchable PMOS/NMOS.

## Output driving current

The GPIOs (general purpose input/outputs) can sink or source up to $\pm 8 \mathrm{~mA}$, and sink or source up to $\pm 20 \mathrm{~mA}$ with the non-standard $\mathrm{V}_{\mathrm{OL}} / \mathrm{V}_{\mathrm{OH}}$ specifications given in Table 51 .
In the user application, the number of I/O pins which can drive current must be limited to respect the absolute maximum rating specified in Section 6.2:

- The sum of the currents sourced by all the $\mathrm{I} / \mathrm{Os}$ on $\mathrm{V}_{\mathrm{DD}}$, plus the maximum Run consumption of the MCU sourced on $V_{D D}$, cannot exceed the absolute maximum rating $I_{\operatorname{VDD}(\Sigma)}$ (see Table 11).
- The sum of the currents sunk by all the $I / O s$ on $V_{S S}$ plus the maximum Run consumption of the MCU sunk on $\mathrm{V}_{\mathrm{SS}}$ cannot exceed the absolute maximum rating IVss(£) (see Table 11).


## Output voltage levels

Unless otherwise specified, the parameters given in Table 51 are derived from tests performed under the conditions summarized in Table 13. All I/Os are CMOS and TTL compliant.

Table 51. Output voltage characteristics

| Symbol | Parameter | Conditions | Min | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {OL }}{ }^{(1)(2)}$ | Output low level voltage for an I/O pin | $\begin{gathered} \mathrm{I}_{\mathrm{IO}}=8 \mathrm{~mA} \\ 2.7 \mathrm{~V}<\mathrm{V}_{\mathrm{DD}}<3.6 \mathrm{~V} \end{gathered}$ | - | 0.4 | V |
| $\mathrm{V}_{\mathrm{OH}}{ }^{(2)(3)}$ | Output high level voltage for an I/O pin |  | $\mathrm{V}_{\mathrm{DD}}-0.4$ | - |  |
| $\mathrm{V}_{\mathrm{OL}}{ }^{(3)(4)}$ | Output low level voltage for an I/O pin | $\begin{gathered} \mathrm{l}_{\mathrm{IO}}=4 \mathrm{~mA} \\ 1.65 \mathrm{~V}<\mathrm{V}_{\mathrm{DD}}<3.6 \mathrm{~V} \end{gathered}$ | - | 0.45 |  |
| $\mathrm{V}_{\mathrm{OH}}{ }^{(3)(4)}$ | Output high level voltage for an I/O pin |  | $\mathrm{V}_{\mathrm{DD}}-0.45$ | - |  |
| $\mathrm{V}_{\mathrm{OL}}{ }^{(1)(4)}$ | Output low level voltage for an I/O pin | $\begin{gathered} \mathrm{I}_{\mathrm{IO}}=20 \mathrm{~mA} \\ 2.7 \mathrm{~V}<\mathrm{V}_{\mathrm{DD}}<3.6 \mathrm{~V} \end{gathered}$ | - | 1.3 |  |
| $\mathrm{V}_{\mathrm{OH}}{ }^{(3)(4)}$ | Output high level voltage for an I/O pin |  | $\mathrm{V}_{\mathrm{DD}}-1.3$ | - |  |

1. The $\mathrm{I}_{\mathrm{IO}}$ current sunk by the device must always respect the absolute maximum rating specified in Table 11 and the sum of $I_{I O}$ (I/O ports and control pins) must not exceed $I_{\text {VSs }}$.
2. Guaranteed by test in production.
3. The $I_{10}$ current sourced by the device must always respect the absolute maximum rating specified in Table 11 and the sum of $\mathrm{I}_{\mathrm{IO}}$ (I/O ports and control pins) must not exceed IVDD.
4. Guaranteed by characterization results.

## Input/output AC characteristics

The definition and values of input/output AC characteristics are given in Figure 26 and Table 52, respectively.
Unless otherwise specified, the parameters given in Table 52 are derived from tests performed under the conditions summarized in Table 13.

Table 52. I/O AC characteristics ${ }^{(1)}$

| $\begin{aligned} & \text { OSPEEDRX } \\ & {[1: 0] \text { bit }} \\ & \text { value }^{(1)} \end{aligned}$ | Symbol | Parameter | Conditions | Min | Max ${ }^{(2)}$ | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 00 | $\mathrm{f}_{\max (10) \mathrm{out}}$ | Maximum frequency ${ }^{(3)}$ | $\mathrm{C}_{\mathrm{L}}=50 \mathrm{pF}, \mathrm{V}_{\mathrm{DD}}=2.7 \mathrm{~V}$ to 3.6 V | - | 400 | kHz |
|  |  |  | $\mathrm{C}_{\mathrm{L}}=50 \mathrm{pF}, \mathrm{V}_{\mathrm{DD}}=1.65 \mathrm{~V}$ to 2.7 V | - | 400 |  |
|  | $\mathrm{t}_{\text {f(IO) out }}$ <br> $\mathrm{t}_{\mathrm{r}(\mathrm{IO}) \text { out }}$ | Output rise and fall time | $\mathrm{C}_{\mathrm{L}}=50 \mathrm{pF}, \mathrm{V}_{\mathrm{DD}}=2.7 \mathrm{~V}$ to 3.6 V | - | 625 | ns |
|  |  |  | $\mathrm{C}_{\mathrm{L}}=50 \mathrm{pF}, \mathrm{V}_{\mathrm{DD}}=1.65 \mathrm{~V}$ to 2.7 V | - | 625 |  |
| 01 | $\mathrm{f}_{\max (10) \text { out }}$ | Maximum frequency ${ }^{(3)}$ | $\mathrm{C}_{\mathrm{L}}=50 \mathrm{pF}, \mathrm{V}_{\mathrm{DD}}=2.7 \mathrm{~V}$ to 3.6 V | - | 2 | MHz |
|  |  |  | $\mathrm{C}_{\mathrm{L}}=50 \mathrm{pF}, \mathrm{V}_{\mathrm{DD}}=1.65 \mathrm{~V}$ to 2.7 V | - | 1 |  |
|  | $\mathrm{t}_{\mathrm{f}}(\mathrm{O})$ out <br> $\mathrm{t}_{\mathrm{r}(\mathrm{IO}) \text { out }}$ | Output rise and fall time | $\mathrm{C}_{\mathrm{L}}=50 \mathrm{pF}, \mathrm{V}_{\mathrm{DD}}=2.7 \mathrm{~V}$ to 3.6 V | - | 125 | ns |
|  |  |  | $\mathrm{C}_{\mathrm{L}}=50 \mathrm{pF}, \mathrm{V}_{\mathrm{DD}}=1.65 \mathrm{~V}$ to 2.7 V | - | 250 |  |
| 10 | $F_{\text {max (IO)out }}$ | Maximum frequency ${ }^{(3)}$ | $\mathrm{C}_{\mathrm{L}}=50 \mathrm{pF}, \mathrm{V}_{\mathrm{DD}}=2.7 \mathrm{~V}$ to 3.6 V | - | 10 | MHz |
|  |  |  | $\mathrm{C}_{\mathrm{L}}=50 \mathrm{pF}, \mathrm{V}_{\mathrm{DD}}=1.65 \mathrm{~V}$ to 2.7 V | - | 2 |  |
|  | $\mathrm{t}_{\text {f(IO) out }}$ <br> $\mathrm{t}_{\mathrm{r}(\mathrm{IO}) \text { out }}$ | Output rise and fall time | $\mathrm{C}_{\mathrm{L}}=50 \mathrm{pF}, \mathrm{V}_{\mathrm{DD}}=2.7 \mathrm{~V}$ to 3.6 V | - | 25 | ns |
|  |  |  | $\mathrm{C}_{\mathrm{L}}=50 \mathrm{pF}, \mathrm{V}_{\mathrm{DD}}=1.65 \mathrm{~V}$ to 2.7 V | - | 125 |  |
| 11 | $F_{\text {max (10)out }}$ | Maximum frequency ${ }^{(3)}$ | $\mathrm{C}_{\mathrm{L}}=30 \mathrm{pF}, \mathrm{V}_{\mathrm{DD}}=2.7 \mathrm{~V}$ to 3.6 V | - | 50 | MHz |
|  |  |  | $\mathrm{C}_{\mathrm{L}}=50 \mathrm{pF}, \mathrm{V}_{\mathrm{DD}}=1.65 \mathrm{~V}$ to 2.7 V | - | 8 |  |
|  | $\mathrm{t}_{\text {f(IO)out }}$ <br> $\mathrm{t}_{\mathrm{r}(\mathrm{IO}) \text { out }}$ | Output rise and fall time | $\mathrm{C}_{\mathrm{L}}=30 \mathrm{pF}, \mathrm{V}_{\mathrm{DD}}=2.7 \mathrm{~V}$ to 3.6 V | - | 5 | ns |
|  |  |  | $\mathrm{C}_{\mathrm{L}}=50 \mathrm{pF}, \mathrm{V}_{\mathrm{DD}}=1.65 \mathrm{~V}$ to 2.7 V | - | 30 |  |
| - | $t_{\text {EXTIpw }}$ | Pulse width of external signals detected by the EXTI controller | - | 8 | - |  |

1. The I/O speed is configured using the OSPEEDRx[1:0] bits. Refer to the STM32L151xx, STM32L152xx and STM32L162xx reference manual for a description of GPIO Port configuration register.
2. Guaranteed by design.
3. The maximum frequency is defined in Figure 26.

Figure 26. I/O AC characteristics definition


### 6.3.15 NRST pin characteristics

The NRST pin input driver uses CMOS technology. It is connected to a permanent pull-up resistor, R RU (see Table 53)

Unless otherwise specified, the parameters given in Table 53 are derived from tests performed under the conditions summarized in Table 13.

Table 53. NRST pin characteristics

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {IL(NRST) }}{ }^{(1)}$ | NRST input low level voltage | - | - | - | 0.3 V DD | V |
| $\mathrm{V}_{\mathrm{IH}(\mathrm{NRST})}{ }^{(1)}$ | NRST input high level voltage | - | $0.7 \mathrm{~V}_{\mathrm{DD}}$ | - | - |  |
| $\mathrm{V}_{\text {OL(NRST) }}{ }^{(1)}$ | NRST output low level voltage | $\begin{gathered} \mathrm{I}_{\mathrm{OL}}=2 \mathrm{~mA} \\ 2.7 \mathrm{~V}<\mathrm{V}_{\mathrm{DD}}<3.6 \mathrm{~V} \end{gathered}$ | - | - | 0.4 |  |
|  |  | $\begin{gathered} \mathrm{I}_{\mathrm{OL}}=1.5 \mathrm{~mA} \\ 1.65 \mathrm{~V}<\mathrm{V}_{\mathrm{DD}}<2.7 \mathrm{~V} \end{gathered}$ | - | - |  |  |
| $\mathrm{V}_{\text {hys(NRST) }}{ }^{(1)}$ | NRST Schmitt trigger voltage hysteresis | - | - | $10 \% \mathrm{~V}_{\mathrm{DD}}{ }^{(2)}$ | - | mV |
| $\mathrm{R}_{\mathrm{PU}}$ | Weak pull-up equivalent resistor ${ }^{(3)}$ | $\mathrm{V}_{\mathrm{IN}}=\mathrm{V}_{\mathrm{SS}}$ | 30 | 45 | 60 | k $\Omega$ |
| $\mathrm{V}_{\mathrm{F}(\mathrm{NRST})}{ }^{(1)}$ | NRST input filtered pulse | - | - | - | 50 | ns |
| $\mathrm{V}_{\mathrm{NF}(\mathrm{NRST})^{(3)}}{ }^{(3)}$ | NRST input not filtered pulse | - | 350 | - | - | ns |

1. Guaranteed by design.
2. With a minimum of 200 mV .
3. The pull-up is designed with a true resistance in series with a switchable PMOS. This PMOS contribution to the series resistance is around $10 \%$.

Figure 27. Recommended NRST pin protection


1. The reset network protects the device against parasitic resets. 0.1 uF capacitor must be placed as close as possible to the chip.
2. The user must ensure that the level on the NRST pin can go below the $\mathrm{V}_{\text {IL(NRST) }}$ max level specified in Table 53. Otherwise the reset is not taken into account by the device.

### 6.3.16 TIM timer characteristics

The parameters given in the Table 54 are guaranteed by design.
Refer to Section 6.3.14: I/O port characteristics for details on the input/output ction characteristics (output compare, input capture, external clock, PWM output).

Table 54. TIMx ${ }^{(1)}$ characteristics

| Symbol | Parameter | Conditions | Min | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\text {res(TIM) }}$ | Timer resolution time | - | 1 | - | $\mathrm{t}_{\text {TIMxCLK }}$ |
|  |  | $\mathrm{f}_{\text {TIMxCLK }}=32 \mathrm{MHz}$ | 31.25 | - | ns |
| $\mathrm{f}_{\mathrm{EXT}}$ | Timer external clock frequency on CH 1 to CH 4 | - | 0 | $\mathrm{f}_{\text {TIM×CLK }} / 2$ | MHz |
|  |  | $\mathrm{f}_{\text {TIMxCLK }}=32 \mathrm{MHz}$ | 0 | 16 | MHz |
| $\mathrm{Res}_{\text {TIM }}$ | Timer resolution | - | - | 16 | bit |
| $\mathrm{t}_{\text {COUNTER }}$ | 16-bit counter clock period when internal clock is selected (timer's prescaler disabled) | - | 1 | 65536 | ${ }^{\text {t }}$ IMx ${ }^{\text {CLK }}$ |
|  |  | $\mathrm{f}_{\text {TIMxCLK }}=32 \mathrm{MHz}$ | 0.0312 | 2048 | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\text {MAX_COUNT }}$ | Maximum possible count | - | - | $65536 \times 65536$ | ${ }_{\text {timx }}$ |
|  |  | $\mathrm{f}_{\text {TIMxCLK }}=32 \mathrm{MHz}$ | - | 134.2 | s |

1. TIMx is used as a general term to refer to the TIM2, TIM3 and TIM4 timers.

### 6.3.17 Communications interfaces

## $1^{2} \mathrm{C}$ interface characteristics

The device $\mathrm{I}^{2} \mathrm{C}$ interface meets the requirements of the standard $\mathrm{I}^{2} \mathrm{C}$ communication protocol with the following restrictions: SDA and SCL are not "true" open-drain I/O pins. When configured as open-drain, the PMOS connected between the I/O pin and $\mathrm{V}_{\mathrm{DD}}$ is disabled, but is still present.
The $I^{2} \mathrm{C}$ characteristics are described in Table 55. Refer also to Section 6.3.14: I/O port characteristics for more details on the input/output ction characteristics (SDA and SCL).

Table 55. $I^{2} \mathrm{C}$ characteristics

| Symbol | Parameter | Standard mode $\mathbf{I}^{2} \mathbf{C}^{(1)(2)}$ |  | Fast mode $\mathrm{I}^{2} \mathrm{C}^{(1)(2)}$ |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Max | Min | Max |  |
| $\mathrm{t}_{\text {w (SCLL) }}$ | SCL clock low time | 4.7 | - | 1.3 | - | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\text {w(SCLH) }}$ | SCL clock high time | 4.0 | - | 0.6 | - |  |
| $\mathrm{t}_{\text {su(SDA) }}$ | SDA setup time | 250 | - | 100 | - | ns |
| $\mathrm{t}_{\mathrm{h} \text { (SDA) }}$ | SDA data hold time | - | $3450{ }^{(3)}$ | - | 900 ${ }^{(3)}$ |  |
| $\begin{aligned} & \left.\mathrm{t}_{\mathrm{r}(\mathrm{SDA})}\right) \\ & \mathrm{t}_{\mathrm{r}(\mathrm{SCL})} \\ & \hline \end{aligned}$ | SDA and SCL rise time | - | 1000 | - | 300 |  |
| $\begin{aligned} & \mathrm{t}_{\mathrm{f}(\mathrm{SDA})} \\ & \mathrm{t}_{\mathrm{f}(\mathrm{SCL})} \\ & \hline \end{aligned}$ | SDA and SCL fall time | - | 300 | - | 300 |  |
| $\mathrm{t}_{\text {h(STA) }}$ | Start condition hold time | 4.0 | - | 0.6 | - | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\text {su(STA) }}$ | Repeated Start condition setup time | 4.7 | - | 0.6 | - |  |
| $\mathrm{t}_{\text {su(STO) }}$ | Stop condition setup time | 4.0 | - | 0.6 | - | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\text {w(STO:STA) }}$ | Stop to Start condition time (bus free) | 4.7 | - | 1.3 | - | $\mu \mathrm{s}$ |
| $\mathrm{C}_{\mathrm{b}}$ | Capacitive load for each bus line | - | 400 | - | 400 | pF |
| $\mathrm{t}_{\text {SP }}$ | Pulse width of spikes that are suppressed by the analog filter | 0 | $50^{(4)}$ | 0 | $50^{(4)}$ | ns |

1. Guaranteed by design.
2. $f_{\text {PCLK } 1}$ must be at least 2 MHz to achieve standard mode $\mathrm{I}^{2} \mathrm{C}$ frequencies. It must be at least 4 MHz to achieve fast mode $\mathrm{l}^{2} \mathrm{C}$ frequencies. It must be a multiple of 10 MHz to reach the 400 kHz maximum $\mathrm{I}^{2} \mathrm{C}$ fast mode clock.
3. The maximum Data hold time has only to be met if the interface does not stretch the low period of SCL signal.
4. The minimum width of the spikes filtered by the analog filter is above $\mathrm{t}_{\mathrm{SP}(\max )}$.

Figure 28. $I^{2} \mathrm{C}$ bus AC waveforms and measurement circuit


1. $R_{S}=$ series protection resistor.
2. $R_{P}=$ external pull-up resistor.
3. $V_{D D \_I 2 C}$ is the I2C bus power supply.
4. Measurement points are done at CMOS levels: $0.3 \mathrm{~V}_{\mathrm{DD}}$ and $0.7 \mathrm{~V}_{\mathrm{DD}}$.

Table 56. SCL frequency ( $\left.\mathrm{f}_{\text {PCLK1 }}=32 \mathrm{MHz}, \mathrm{V}_{\mathrm{DD}}=\mathrm{V}_{\mathrm{DD} \_12 \mathrm{C}}=3.3 \mathrm{~V}\right)^{(1)(2)}$

| $\mathbf{f}_{\text {SCL }}(\mathbf{k H z})$ | I2C_CCR value |
| :---: | :---: |
|  | $\mathbf{R}_{\mathbf{P}}=\mathbf{4 . 7} \mathbf{~ k} \Omega$ |
| 400 | $0 \times 801 \mathrm{~B}$ |
| 300 | $0 \times 8024$ |
| 200 | $0 \times 8035$ |
| 100 | $0 \times 00 \mathrm{A0}$ |
| 50 | $0 \times 0140$ |
| 20 | $0 \times 0320$ |

1. $R_{P}=$ External pull-up resistance, $f_{S C L}=I^{2} C$ speed.
2. For speeds around 200 kHz , the tolerance on the achieved speed is of $\pm 5 \%$. For other speed ranges, the tolerance on the achieved speed is $\pm 2 \%$. These variations depend on the accuracy of the external components used to design the application.

## SPI characteristics

Unless otherwise specified, the parameters given in the following table are derived from tests performed under the conditions summarized in Table 13.

Refer to Section 6.3.13: I/O current injection characteristics for more details on the input/output alternate function characteristics (NSS, SCK, MOSI, MISO).

Table 57. SPI characteristics ${ }^{(1)}$

| Symbol | Parameter | Conditions | Min | Max ${ }^{(2)}$ | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{gathered} \mathrm{f}_{\mathrm{SCK}} \\ 1 / \mathrm{t}_{\mathrm{c}(\mathrm{SCK})} \end{gathered}$ | SPI clock frequency | Master mode | - | 16 | MHz |
|  |  | Slave mode | - | 16 |  |
|  |  | Slave transmitter | - | $12^{(3)}$ |  |
| $\begin{aligned} & \mathrm{t}_{\mathrm{r}_{(S C K)}{ }^{(2)}}^{\mathrm{t}_{\left.\mathrm{f}_{(S C K}\right)}^{(2)}} \end{aligned}$ | SPI clock rise and fall time | Capacitive load: $\mathrm{C}=30 \mathrm{pF}$ | - | 6 | ns |
| DuCy(SCK) | SPI slave input clock duty cycle | Slave mode | 30 | 70 | \% |
| $\mathrm{t}_{\text {su(NSS }}$ | NSS setup time | Slave mode | $4 \mathrm{t}_{\text {HCLK }}$ | - | ns |
| $\mathrm{t}_{\mathrm{h} \text { (NSS) }}$ | NSS hold time | Slave mode | $2 \mathrm{t}_{\text {HCLK }}$ | - |  |
| $\begin{aligned} & \mathrm{t}_{\mathrm{w}(\mathrm{SCKH})}{ }^{(2)} \\ & \mathrm{t}_{\mathrm{w}(\mathrm{SCKL})}{ }^{(2)} \end{aligned}$ | SCK high and low time | Master mode | $t_{\text {SCK }} / 2-5$ | $t_{\text {SCK }} / 2+3$ |  |
| $\mathrm{t}_{\text {su(MI) }}{ }^{(2)}$ | Data input setup time | Master mode | 5 | - |  |
| $\mathrm{t}_{\mathrm{su}(\mathrm{SI})^{(2)}}$ |  | Slave mode | 6 | - |  |
| $\mathrm{th}_{\mathrm{h}(\mathrm{MI})^{(2)}}$ | Data input hold time | Master mode | 5 | - |  |
| $\mathrm{th}_{\mathrm{h}(\mathrm{SI})}{ }^{(2)}$ |  | Slave mode | 5 | - |  |
| $\mathrm{ta}_{\text {(SO) }}{ }^{(4)}$ | Data output access time | Slave mode | 0 | $3 \mathrm{t}_{\text {HCLK }}$ |  |
| $\mathrm{t}_{\mathrm{V} \text { (SO) }}{ }^{(2)}$ | Data output valid time | Slave mode | - | 33 |  |
| $\mathrm{t}_{\mathrm{V}(\mathrm{MO})^{(2)}}$ | Data output valid time | Master mode | - | 6.5 |  |
| $\mathrm{t}_{\mathrm{h}(\mathrm{SO})^{(2)}}$ | Data output hold time | Slave mode | 17 | - |  |
| $\left.\mathrm{t}_{\mathrm{h}} \mathrm{MO}\right)^{(2)}$ |  | Master mode | 0.5 | - |  |

1. The characteristics above are given for voltage range 1.
2. Guaranteed by characterization results.
3. The maximum SPI clock frequency in slave transmitter mode is given for an SPI slave input clock duty cycle (DuCy(SCK)) ranging between 40 to 60\%.
4. Min time is for the minimum time to drive the output and max time is for the maximum time to validate the data.

Figure 29. SPI timing diagram - slave mode and CPHA = 0


Figure 30. SPI timing diagram - slave mode and CPHA = $1^{(1)}$


1. Measurement points are done at CMOS levels: $0.3 \mathrm{~V}_{\mathrm{DD}}$ and $0.7 \mathrm{~V}_{\mathrm{DD}}$.

Figure 31. SPI timing diagram - master mode ${ }^{(1)}$


1. Measurement points are done at $C M O S$ levels: $0.3 \mathrm{~V}_{\mathrm{DD}}$ and $0.7 \mathrm{~V}_{\mathrm{DD}}$.

## USB characteristics

The USB interface is USB-IF certified (full speed).
Table 58. USB startup time

| Symbol | Parameter | Max | Unit |
| :---: | :---: | :---: | :---: |
| tSTARTUP $^{(1)}$ | USB transceiver startup time | 1 | $\mu \mathrm{~s}$ |

1. Guaranteed by design.

Table 59. USB DC electrical characteristics

| Symbol | Parameter | Conditions | Min. ${ }^{(1)}$ | Max. ${ }^{(1)}$ | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Input levels |  |  |  |  |  |
| $V_{D D}$ | USB operating voltage | - | 3.0 | 3.6 | V |
| $\mathrm{V}_{D I}{ }^{(2)}$ | Differential input sensitivity | I(USB_DP, USB_DM) | 0.2 | - | V |
| $\mathrm{V}_{\mathrm{CM}}{ }^{(2)}$ | Differential common mode range | Includes $V_{\text {DI }}$ range | 0.8 | 2.5 |  |
| $\mathrm{V}_{S E}{ }^{(2)}$ | Single ended receiver threshold | - | 1.3 | 2.0 |  |
| Output levels |  |  |  |  |  |
| $\mathrm{V}_{\mathrm{OL}}{ }^{(3)}$ | Static output level low | $\mathrm{R}_{\mathrm{L}}$ of $1.5 \mathrm{k} \Omega$ to $3.6 \mathrm{~V}^{(4)}$ | - | 0.3 | V |
| $\mathrm{V}_{\mathrm{OH}}{ }^{(3)}$ | Static output level high | $\mathrm{R}_{\mathrm{L}}$ of $15 \mathrm{k} \Omega$ to $\mathrm{V}_{\mathrm{SS}}{ }^{(4)}$ | 2.8 | 3.6 |  |

1. All the voltages are measured from the local ground potential.
2. Guaranteed by characterization results.
3. Guaranteed by test in production.
4. $R_{L}$ is the load connected on the USB drivers.

Figure 32. USB timings: definition of data signal rise and fall time


Table 60. USB: full speed electrical characteristics

| Driver characteristics $^{(1)}$ |  |  |  |  |  |
| :---: | :--- | :---: | :---: | :---: | :---: |
| Symbol | Parameter | Conditions | Min | Max | Unit |
| $\mathrm{t}_{\mathrm{r}}$ | Rise time $^{(2)}$ | $\mathrm{C}_{\mathrm{L}}=50 \mathrm{pF}$ | 4 | 20 | ns |
| $\mathrm{t}_{\mathrm{f}}$ | Fall Time $^{(2)}$ | $\mathrm{C}_{\mathrm{L}}=50 \mathrm{pF}$ | 4 | 20 | ns |
| $\mathrm{t}_{\mathrm{rfm}}$ | Rise/ fall time matching | $\mathrm{t}_{\mathrm{r}} / \mathrm{t}_{\mathrm{f}}$ | 90 | 110 | $\%$ |
| $\mathrm{~V}_{\mathrm{CRS}}$ | Output signal crossover voltage | - | 1.3 | 2.0 | V |

1. Guaranteed by design.
2. Measured from $10 \%$ to $90 \%$ of the data signal. For more detailed informations, refer to USB Specification Chapter 7 (version 2.0).

## I2S characteristics

Table 61. I2S characteristics

| Symbol | Parameter | Conditions | Min | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{f}_{\text {MCK }}$ | I2S Main Clock Output | - | $256 \times 8 \mathrm{~K}$ | $256 \mathrm{xFs}{ }^{(1)}$ | MHz |
| $\mathrm{f}_{\mathrm{CK}}$ | 12S clock frequency | Master data: 32 bits | - | $64 x F s$ | MHz |
|  |  | Slave data: 32 bits | - | 64xFs |  |
| $\mathrm{D}_{\mathrm{CK}}$ | I2S clock frequency duty cycle | Slave receiver, 48 KHz | 30 | 70 | \% |
| $\mathrm{t}_{\text {( }}$ (CK) | I2S clock rise time | Capacitive load CL=30pF | - | 8 | ns |
| $\mathrm{t}_{\mathrm{f}}(\mathrm{CK})$ | I2S clock fall time |  |  | 8 |  |
| $\mathrm{t}_{\mathrm{v} \text { (WS) }}$ | WS valid time | Master mode | 4 | 24 |  |
| $\mathrm{t}_{\mathrm{h}}(\mathrm{WS})$ | WS hold time | Master mode | 0 | - |  |
| $\mathrm{t}_{\text {su( }}$ (WS) | WS setup time | Slave mode | 15 | - |  |
| $\mathrm{t}_{\mathrm{h} \text { (WS) }}$ | WS hold time | Slave mode | 0 | - |  |
| $\mathrm{t}_{\text {su(SD_MR) }}$ | Data input setup time | Master receiver | 8 | - |  |
| $\mathrm{t}_{\text {su(SD_SR) }}$ | Data input setup time | Slave receiver | 9 | - |  |
| $\mathrm{th}_{\text {(SD_MR) }}$ | Data input hold time | Master receiver | 5 | - |  |
| $\mathrm{t}_{\mathrm{h} \text { (SD_SR) }}$ |  | Slave receiver | 4 | - |  |
| $\mathrm{t}_{\mathrm{v} \text { (SD_ST) }}$ | Data output valid time | Slave transmitter (after enable edge) | - | 64 |  |
| $\mathrm{th}_{\text {(SD_ST) }}$ | Data output hold time | Slave transmitter (after enable edge) | 22 | - |  |
| $\mathrm{t}_{\mathrm{v} \text { (SD_MT) }}$ | Data output valid time | Master transmitter (after enable edge) | - | 12 |  |
| $\mathrm{t}_{\text {h(SD_MT) }}$ | Data output hold time | Master transmitter (after enable edge) | 8 | - |  |

1. The maximum for $256 x F s$ is 8 MHz

Note: $\quad$ Refer to the I2S section of the product reference manual for more details about the sampling frequency (Fs), $f_{M C K}, f_{C K}$ and $D_{C K}$ values. These values reflect only the digital peripheral behavior, source clock precision might slightly change them. DCK depends mainly on the ODD bit value, digital contribution leads to a min of (I2SDIV/(2*I2SDIV+ODD) and a max of $(I 2 S D I V+O D D) /\left(2^{*} I 2 S D I V+O D D\right)$. Fs max is supported for each mode/condition.

Figure 33. $1^{2}$ S slave timing diagram (Philips protocol) ${ }^{(1)}$


1. Measurement points are done at CMOS levels: $0.3 \times V_{D D}$ and $0.7 \times V_{D D}$.
2. LSB transmit/receive of the previously transmitted byte. No LSB transmit/receive is sent before the first byte.

Figure 34. $\mathrm{I}^{\mathbf{2}} \mathrm{S}$ master timing diagram (Philips protocol) ${ }^{(1)}$


1. Guaranteed by characterization results.
2. LSB transmit/receive of the previously transmitted byte. No LSB transmit/receive is sent before the first byte.

### 6.3.18 SDIO characteristics

Table 62. SDIO characteristics ${ }^{(1)}$

| Symbol | Parameter | Conditions | Min | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{f}_{\mathrm{PP}}$ | Clock frequency in data transfer mode | $\mathrm{CL} \leq 30 \mathrm{pF}$ | 0 | 24 | MHz |
| $\mathrm{t}_{\text {W(CKL) }}$ | Clock low time, $\mathrm{f}_{\text {PP }}=24 \mathrm{MHz}$ | $\mathrm{CL} \leq 30 \mathrm{pF}$ | $20^{(2)}$ | - |  |
| $\mathrm{t}_{\text {W(CKH) }}$ | Clock high time, $\mathrm{f}_{\mathrm{PP}}=24 \mathrm{MHz}$ | $\mathrm{CL} \leq 30 \mathrm{pF}$ | $18^{(2)}$ | - |  |
| $\mathrm{t}_{\mathrm{r}}$ | Clock rise time, $\mathrm{f}_{\mathrm{PP}}=24 \mathrm{MHz}$ | $\mathrm{CL} \leq 30 \mathrm{pF}$ | - | 5 | ns |
| $\mathrm{t}_{\mathrm{f}}$ | Clock fall time, $\mathrm{f}_{\mathrm{PP}}=24 \mathrm{MHz}$ | $\mathrm{CL} \leq 30 \mathrm{pF}$ | - | 5 |  |
| CMD, D inputs (referenced to CK) in SD default mode |  |  |  |  |  |
| - |  |  | From 2.8 to 3.6 V | - | - |
| tisu | Input setup time, $\mathrm{f}_{\mathrm{PP}}=24 \mathrm{MHz}$ | $\mathrm{CL} \leq 30 \mathrm{pF}$ | 2 | - |  |
| $\mathrm{t}_{\mathrm{H}}$ | Input hold time, $\mathrm{f}_{\mathrm{PP}}=24 \mathrm{MHz}$ | $\mathrm{CL} \leq 30 \mathrm{pF}$ | 1.6 | - |  |
| CMD, D outputs (referenced to CK) in SD default mode |  |  |  |  |  |
| tovo | Output valid default time, $\mathrm{f}_{\mathrm{PP}}=24 \mathrm{MHz}$ | $\mathrm{CL} \leq 30 \mathrm{pF}$ | 0 | 14 |  |
| $\mathrm{t}_{\text {OHD }}$ | Output hold default time, $\mathrm{f}_{\mathrm{PP}}=24 \mathrm{MHz}$ | $\mathrm{CL} \leq 30 \mathrm{pF}$ | 0 | - |  |

1. Guaranteed by characterization results.
2. Values measured with a threshold level equal to $\mathrm{V}_{\mathrm{DD}} / 2$.

Figure 35. SDIO timings


### 6.3.19 12-bit ADC characteristics

Unless otherwise specified, the parameters given in Table 64 are guaranteed by design.
Table 63. ADC clock frequency

| Symbol | Parameter | Conditions |  |  | Min | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{f}_{\text {ADC }}$ | ADC clock frequency | Voltage range 1 \& 2 | $2.4 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DDA}} \leq 3.6 \mathrm{~V}$ | $\mathrm{V}_{\text {REF }+}=\mathrm{V}_{\text {DDA }}$ | 0.480 | 16 | MHz |
|  |  |  |  | $\begin{aligned} & \mathrm{V}_{\text {REF }+}<\mathrm{V}_{\mathrm{DDA}} \\ & \mathrm{~V}_{\mathrm{REF}+}>2.4 \mathrm{~V} \end{aligned}$ |  | 8 |  |
|  |  |  |  | $\begin{aligned} & \mathrm{V}_{\mathrm{REF}+}<\mathrm{V}_{\mathrm{DDA}} \\ & \mathrm{~V}_{\mathrm{REF}+} \leq 2.4 \mathrm{~V} \end{aligned}$ |  | 4 |  |
|  |  |  | $1.8 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DDA}} \leq 2.4 \mathrm{~V}$ | $\mathrm{V}_{\text {REF }+}=\mathrm{V}_{\text {DDA }}$ |  | 8 |  |
|  |  |  |  | $\mathrm{V}_{\text {REF+ }}<\mathrm{V}_{\text {DDA }}$ |  | 4 |  |
|  |  | Voltage range 3 |  |  |  | 4 |  |

Table 64. ADC characteristics

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {DDA }}$ | Power supply | - | 1.8 | - | 3.6 | V |
| $\mathrm{V}_{\text {REF+ }}$ | Positive reference voltage | - | $1.8{ }^{(1)}$ | - | $\mathrm{V}_{\text {DDA }}$ |  |
| $\mathrm{V}_{\text {REF- }}$ | Negative reference voltage | - | - | $\mathrm{V}_{\text {SSA }}$ | - |  |
| $\mathrm{I}_{\text {VDDA }}$ | Current on the $\mathrm{V}_{\text {DDA }}$ input pin | - | - | 1000 | 1450 | $\mu \mathrm{A}$ |
| $\mathrm{IVREF}^{(2)}$ | Current on the $\mathrm{V}_{\text {REF }}$ input pin | Peak | - | 400 | 700 |  |
|  |  | Average | -- |  | 450 |  |
| $\mathrm{V}_{\text {AIN }}$ | Conversion voltage range ${ }^{(3)}$ | - | $0^{(4)}$ | - | $\mathrm{V}_{\text {REF }+}$ | V |
| $\mathrm{f}_{S}$ | 12-bit sampling rate | Direct channels | - | - | 1 | Msps |
|  |  | Multiplexed channels | - | - | 0.76 |  |
|  | 10-bit sampling rate | Direct channels | - | - | 1.07 | Msps |
|  |  | Multiplexed channels | - | - | 0.8 |  |
|  | 8-bit sampling rate | Direct channels | - | - | 1.23 | Msps |
|  |  | Multiplexed channels | - | - | 0.89 |  |
|  | 6-bit sampling rate | Direct channels | - | - | 1.45 | Msps |
|  |  | Multiplexed channels | - | - | 1 |  |

Table 64. ADC characteristics (continued)

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $t_{s}{ }^{(5)}$ | Sampling time | Direct channels $2.4 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DDA}} \leq 3.6 \mathrm{~V}$ | 0.25 | - | - | $\mu \mathrm{s}$ |
|  |  | Multiplexed channels $2.4 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DDA}} \leq 3.6 \mathrm{~V}$ | 0.56 | - | - |  |
|  |  | Direct channels $1.8 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DDA}} \leq 2.4 \mathrm{~V}$ | 0.56 | - | - |  |
|  |  | Multiplexed channels $1.8 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DDA}} \leq 2.4 \mathrm{~V}$ | 1 | - | - |  |
|  |  | - | 4 | - | 384 | $1 / f_{\text {ADC }}$ |
| ${ }^{\text {t }}$ CONV | Total conversion time (including sampling time) | $\mathrm{f}_{\mathrm{ADC}}=16 \mathrm{MHz}$ | 1 | - | 24.75 | $\mu \mathrm{s}$ |
|  |  | - | 4 to 384 (sampling phase) +12 (successive approximation) |  |  | 1/f ${ }_{\text {ADC }}$ |
| $\mathrm{C}_{\text {ADC }}$ | Internal sample and hold capacitor | Direct channels | - | 16 | - | pF |
|  |  | Multiplexed channels | - |  | - |  |
| $\mathrm{f}_{\text {TRIG }}$ | External trigger frequency Regular sequencer | 12-bit conversions | - | - | Tconv+1 | $1 / f_{\text {ADC }}$ |
|  |  | 6/8/10-bit conversions | - | - | Tconv | $1 / f_{\text {ADC }}$ |
| $\mathrm{f}_{\text {TRIG }}$ | External trigger frequency Injected sequencer | 12-bit conversions | - | - | Tconv+2 | $1 / f_{\text {ADC }}$ |
|  |  | 6/8/10-bit conversions | - | - | Tconv+1 | $1 / f_{\text {ADC }}$ |
| $\mathrm{R}_{\text {AIN }}{ }^{(6)}$ | Signal source impedance | - | - | - | 50 | $\mathrm{k} \Omega$ |
| $t_{\text {lat }}$ | Injection trigger conversion latency | $\mathrm{f}_{\text {ADC }}=16 \mathrm{MHz}$ | 219 | - | 281 | ns |
|  |  | - | 3.5 | - | 4.5 | $1 / f_{\text {ADC }}$ |
| $t_{\text {latr }}$ | Regular trigger conversion latency | $\mathrm{f}_{\mathrm{ADC}}=16 \mathrm{MHz}$ | 156 | - | 219 | ns |
|  |  | - | 2.5 | - | 3.5 | $1 / f_{\text {ADC }}$ |
| $\mathrm{t}_{\text {Stab }}$ | Power-up time | - | - | - | 3.5 | $\mu \mathrm{s}$ |

1. The Vref+ input can be grounded if neither the ADC nor the DAC are used (this allows to shut down an external voltage
reference).
2. The current consumption through VREF is composed of two parameters:

- one constant ( $\max 300 \mu \mathrm{~A}$ )
- one variable ( $\max 400 \mu \mathrm{~A}$ ), only during sampling time +2 first conversion pulses

So, peak consumption is $300+400=700 \mu \mathrm{~A}$ and average consumption is $300+[(4$ sampling +2$) / 16] \times 400=450 \mu \mathrm{~A}$ at 1Msps
3. $\mathrm{V}_{\text {REF+ }}$ can be internally connected to $\mathrm{V}_{\text {DDA }}$ and $\mathrm{V}_{\text {REF- }}$ can be internally connected to $\mathrm{V}_{\mathrm{SSA}}$, depending on the package. Refer to Section 4: Pin descriptions for further details.
4. $V_{S S A}$ or $V_{\text {REF- }}$ must be tied to ground.
5. Minimum sampling time is reached for an external input impedance limited to a value as defined in Table 66: Maximum source impedance $R_{\text {AIN }}$ max.
6. External impedance has another high value limitation when using short sampling time as defined in Table 66: Maximum source impedance $R_{\text {AIN }}$ max.

Table 65. ADC accuracy ${ }^{(1)(2)}$

| Symbol | Parameter | Test conditions | $\operatorname{Min}^{(3)}$ | Typ | Max ${ }^{(3)}$ | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ET | Total unadjusted error | $\begin{aligned} & 2.4 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DDA}} \leq 3.6 \mathrm{~V} \\ & 2.4 \mathrm{~V} \leq \mathrm{V}_{\mathrm{REF}} \leq 3.6 \mathrm{~V} \\ & \mathrm{f}_{\mathrm{ADC}}=8 \mathrm{MHz}, \mathrm{R}_{\text {AIN }}=50 \Omega \\ & \mathrm{~T}_{\mathrm{A}}=-40 \text { to } 105{ }^{\circ} \mathrm{C} \end{aligned}$ | - | 2 | 4 | LSB |
| EO | Offset error |  | - | 1 | 2 |  |
| EG | Gain error |  | - | 1.5 | 3.5 |  |
| ED | Differential linearity error |  | - | 1 | 2 |  |
| EL | Integral linearity error |  | - | 1.7 | 3 |  |
| ENOB | Effective number of bits | $\begin{aligned} & 2.4 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DDA}} \leq 3.6 \mathrm{~V} \\ & \mathrm{~V}_{\mathrm{DDA}}=\mathrm{V}_{\mathrm{REF}} \\ & \mathrm{f}_{\mathrm{ADC}}=16 \mathrm{MHz}, \mathrm{R}_{\mathrm{AIN}}=50 \Omega \\ & \mathrm{~T}_{\mathrm{A}}=-40 \text { to } 105^{\circ} \mathrm{C} \\ & \mathrm{~F}_{\text {input }}=10 \mathrm{kHz} \end{aligned}$ | 9.2 | 10 | - | bits |
| SINAD | Signal-to-noise and distortion ratio |  | 57.5 | 62 | - | dB |
| SNR | Signal-to-noise ratio |  | 57.5 | 62 | - |  |
| THD | Total harmonic distortion |  | - | -70 | -65 |  |
| ENOB | Effective number of bits | $\begin{aligned} & 1.8 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DDA}} \leq 2.4 \mathrm{~V} \\ & \mathrm{~V}_{\mathrm{DDA}}=\mathrm{V}_{\mathrm{REF}} \\ & \mathrm{f}_{\mathrm{ADC}}=8 \mathrm{MHz} \text { or } 4 \mathrm{MHz}, \mathrm{R}_{\text {AIN }}=50 \Omega \\ & \mathrm{~T}_{\mathrm{A}}=-40 \text { to } 105^{\circ} \mathrm{C} \\ & \mathrm{~F}_{\text {input }}=10 \mathrm{kHz} \end{aligned}$ | 9.2 | 10 | - | bits |
| SINAD | Signal-to-noise and distortion ratio |  | 57.5 | 62 | - | dB |
| SNR | Signal-to-noise ratio |  | 57.5 | 62 | - |  |
| THD | Total harmonic distortion |  | - | -70 | -65 |  |
| ET | Total unadjusted error | $\begin{aligned} & 2.4 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DDA}} \leq 3.6 \mathrm{~V} \\ & 1.8 \mathrm{~V} \leq \mathrm{V}_{\mathrm{REF}} \leq 2.4 \mathrm{~V} \\ & \mathrm{f}_{\mathrm{ADC}}=4 \mathrm{MHz}, \mathrm{R}_{\text {AIN }}=50 \Omega \\ & \mathrm{~T}_{\mathrm{A}}=-40 \text { to } 105^{\circ} \mathrm{C} \end{aligned}$ | - | 4 | 6.5 | LSB |
| EO | Offset error |  | - | 2 | 4 |  |
| EG | Gain error |  | - | 4 | 6 |  |
| ED | Differential linearity error |  | - | 1 | 2 |  |
| EL | Integral linearity error |  | - | 1.5 | 3 |  |
| ET | Total unadjusted error | $\begin{aligned} & 1.8 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DDA}} \leq 2.4 \mathrm{~V} \\ & 1.8 \mathrm{~V} \leq \mathrm{V}_{\mathrm{REF}} \leq 2.4 \mathrm{~V} \\ & \mathrm{f}_{\mathrm{ADC}}=4 \mathrm{MHz}, \mathrm{R}_{\text {AIN }}=50 \Omega \\ & \mathrm{~T}_{\mathrm{A}}=-40 \text { to } 105^{\circ} \mathrm{C} \end{aligned}$ | - | 2 | 3 | LSB |
| EO | Offset error |  | - | 1 | 1.5 |  |
| EG | Gain error |  | - | 1.5 | 2 |  |
| ED | Differential linearity error |  | - | 1 | 2 |  |
| EL | Integral linearity error |  | - | 1 | 1.5 |  |

1. ADC DC accuracy values are measured after internal calibration.
2. ADC accuracy vs. negative injection current: Injecting a negative current on any analog input pins must be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative currents. Any positive injection current within the limits specified for $I_{I N J(P I N)}$ and $\Sigma I_{\text {INJ(PIN) }}$ in Section 6.3.13 does not affect the ADC accuracy.
3. Guaranteed by characterization results.

Figure 36. ADC accuracy characteristics


Figure 37. Typical connection diagram using the ADC


1. Refer to Table 66: Maximum source impedance $R_{\text {AIN }}$ max for the value of $R_{\text {AIN }}$ and Table 64: ADC characteristics for the value of $\mathrm{C}_{A D C}$.
2. $C_{\text {parasitic }}$ represents the capacitance of the PCB (dependent on soldering and PCB layout quality) plus the pad capacitance (roughly 7 pF ). A high $\mathrm{C}_{\text {parasitic }}$ value downgrades conversion accuracy. To remedy this, $\mathrm{f}_{\mathrm{ADC}}$ must be reduced.

Figure 38. Maximum dynamic current consumption on $\mathrm{V}_{\text {REF+ }}$ supply pin during ADC conversion


Table 66. Maximum source impedance RAIN $_{\text {max }}{ }^{(1)}$

| Ts ( $\mu \mathrm{s}$ ) | $\mathrm{R}_{\text {AIN }} \max (\mathrm{k} \Omega)$ |  |  |  | Ts (cycles)$\mathrm{f}_{\mathrm{ADC}}=16 \mathrm{MHz}^{(2)}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | Multiplexed channels |  | Direct channels |  |  |
|  | $2.4 \mathrm{~V}<\mathrm{V}_{\text {DDA }}<3.6 \mathrm{~V}$ | $1.8 \mathrm{~V}<\mathrm{V}_{\text {DDA }}<2.4 \mathrm{~V}$ | $2.4 \mathrm{~V}<\mathrm{V}_{\text {DDA }}<3.6 \mathrm{~V}$ | $1.8 \mathrm{~V}<\mathrm{V}_{\text {DDA }}<2.4 \mathrm{~V}$ |  |
| 0.25 | Not allowed | Not allowed | 0.7 | Not allowed | 4 |
| 0.5625 | 0.8 | Not allowed | 2.0 | 1.0 | 9 |
| 1 | 2.0 | 0.8 | 4.0 | 3.0 | 16 |
| 1.5 | 3.0 | 1.8 | 6.0 | 4.5 | 24 |
| 3 | 6.8 | 4.0 | 15.0 | 10.0 | 48 |
| 6 | 15.0 | 10.0 | 30.0 | 20.0 | 96 |
| 12 | 32.0 | 25.0 | 50.0 | 40.0 | 192 |
| 24 | 50.0 | 50.0 | 50.0 | 50.0 | 384 |

1. Guaranteed by design.
2. Number of samples calculated for $f_{A D C}=16 \mathrm{MHz}$. For $\mathrm{f}_{\mathrm{ADC}}=8$ and 4 MHz the number of sampling cycles can be reduced with respect to the minimum sampling time Ts ( $\mu \mathrm{s}$ ),

## General PCB design guidelines

Power supply decoupling must be performed as shown in Figure 11. The applicable procedure depends on whether $\mathrm{V}_{\text {REF+ }}$ is connected to $\mathrm{V}_{\mathrm{DDA}}$ or not. The 100 nF capacitors must be ceramic (good quality). They must be placed as close as possible to the chip.

### 6.3.20 DAC electrical specifications

Data guaranteed by design, unless otherwise specified.
Table 67. DAC characteristics

| Symbol | Parameter | Conditions |  | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {DDA }}$ | Analog supply voltage |  | - | 1.8 | - | 3.6 | V |
| $\mathrm{V}_{\text {REF }+}$ | Reference supply voltage | $\mathrm{V}_{\text {REF+ }}$ must always be below <br> $V_{\text {DDA }}$ |  | 1.8 | - | 3.6 |  |
| $\mathrm{V}_{\text {REF- }}$ | Lower reference voltage |  | - |  |  |  |  |
| $\mathrm{I}_{\text {DVVREF+ }}{ }^{(1)}$ | Current consumption on <br> $\mathrm{V}_{\text {REF+ }}$ supply <br> $\mathrm{V}_{\text {REF }+}=3.3 \mathrm{~V}$ | No load, middle code (0x800) |  | - | 130 | 220 |  |
|  |  | No load, worst code (0x000) |  | - | 220 | 350 |  |
| $\mathrm{I}_{\text {DDA }}{ }^{(1)}$ | Current consumption on <br> $V_{\text {DDA }}$ supply <br> $V_{D D A}=3.3 \mathrm{~V}$ | No load, middle code (0x800) |  | - | 210 | 320 |  |
|  |  | No load, worst code (0xF1C) |  | - | 320 | 520 |  |
| $\mathrm{R}_{\mathrm{L}}$ | Resistive load | DAC output buffer ON | Connected to $\mathrm{V}_{\text {SSA }}$ | 5 | - | - | k $\Omega$ |
|  |  |  | Conected to $V_{\text {DDA }}$ | 25 | - | - |  |
| $\mathrm{C}_{\mathrm{L}}{ }^{(2)}$ | Capacitive load | DAC output | buffer ON | - | - | 50 | pF |
| $\mathrm{R}_{\mathrm{O}}$ | Output impedance | DAC output | buffer OFF | 12 | 16 | 20 | k $\Omega$ |
| V DAC_OUT | Voltage on DAC_OUT output | DAC output buffer ON |  | 0.2 | - | $V_{\text {DDA }}-0.2$ | V |
|  |  | DAC output buffer OFF |  | 0.5 | - | $\begin{gathered} \mathrm{V}_{\mathrm{REF}+}- \\ 1 \mathrm{LSB} \end{gathered}$ | mV |
| DNL ${ }^{(1)}$ | Differential non linearity ${ }^{(3)}$ | $\mathrm{C}_{\mathrm{L}} \leq 50 \mathrm{pF}, \mathrm{R}_{\mathrm{L}} \geq 5 \mathrm{k} \Omega$ <br> DAC output buffer ON |  | - | 1.5 | 3 | LSB |
|  |  | No $R_{L}, C_{L} \leq 50 \mathrm{pF}$ DAC output buffer OFF |  | - | 1.5 | 3 |  |
| $\mathrm{INL}{ }^{(1)}$ | Integral non linearity ${ }^{(4)}$ | $\mathrm{C}_{\mathrm{L}} \leq 50 \mathrm{pF},$ DAC output | $\begin{aligned} & \mathrm{R}_{\mathrm{L}} \geq 5 \mathrm{k} \Omega \\ & \text { buffer } \mathrm{ON} \end{aligned}$ | - | 2 | 4 |  |
|  |  | No $R_{L}, C_{L} \leq$ DAC output | 50 pF uffer OFF | - | 2 | 4 |  |
| Offset ${ }^{(1)}$ | Offset error at code$0 \times 800^{(5)}$ | $\mathrm{C}_{\mathrm{L}} \leq 50 \mathrm{pF}, \mathrm{R}_{\mathrm{L}} \geq 5 \mathrm{k} \Omega$ <br> DAC output buffer ON |  | - | $\pm 10$ | $\pm 25$ |  |
|  |  | No $R_{L}, C_{L} \leq 50 \mathrm{pF}$ DAC output buffer OFF |  | - | $\pm 5$ | $\pm 8$ |  |
| Offset $1^{(1)}$ | Offset error at code $0 \times 001^{(6)}$ | No $R_{L}, C_{L} \leq 50 \mathrm{pF}$ DAC output buffer OFF |  | - | $\pm 1.5$ | $\pm 5$ |  |

Table 67. DAC characteristics (continued)

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| dOffset/dT ${ }^{(1)}$ | Offset error temperature coefficient (code 0x800) | $\begin{aligned} & \hline \mathrm{V}_{\mathrm{DDA}}=3.3 \mathrm{~V} \\ & \mathrm{~V}_{\mathrm{REF}+}=3.0 \mathrm{~V} \\ & \mathrm{~T}_{\mathrm{A}}=0 \text { to } 50^{\circ} \mathrm{C} \\ & \text { DAC output buffer OFF } \end{aligned}$ | -20 | -10 | 0 | $\mu \mathrm{V} /{ }^{\circ} \mathrm{C}$ |
|  |  | $\begin{aligned} & \mathrm{V}_{\text {DDA }}=3.3 \mathrm{~V} \\ & \mathrm{~V}_{\text {REF }}=3.0 \mathrm{~V} \\ & \mathrm{~T}_{\mathrm{A}}=0 \text { to } 50^{\circ} \mathrm{C} \\ & \mathrm{DAC} \text { output buffer } \mathrm{ON} \end{aligned}$ | 0 | 20 | 50 |  |
| Gain ${ }^{(1)}$ | Gain error ${ }^{(7)}$ | $\mathrm{C}_{\mathrm{L}} \leq 50 \mathrm{pF}, \mathrm{R}_{\mathrm{L}} \geq 5 \mathrm{k} \Omega$ <br> DAC output buffer ON | - | +0.1/-0.2\% | +0.2 / -0.5\% | \% |
|  |  | No $R_{L}, C_{L} \leq 50 \mathrm{pF}$ DAC output buffer OFF | - | +0 / -0.2\% | +0 / -0.4\% |  |
| dGain/dT ${ }^{(1)}$ | Gain error temperature coefficient | $\begin{aligned} & \mathrm{V}_{\mathrm{DDA}}=3.3 \mathrm{~V} \\ & \mathrm{~V}_{\mathrm{REF}+}=3.0 \mathrm{~V} \\ & \mathrm{~T}_{\mathrm{A}}=0 \text { to } 50^{\circ} \mathrm{C} \\ & \text { DAC output buffer OFF } \end{aligned}$ | -10 | -2 | 0 | $\mu \mathrm{V} /{ }^{\circ} \mathrm{C}$ |
|  |  | $\begin{aligned} & \mathrm{V}_{\mathrm{DDA}}=3.3 \mathrm{~V} \\ & \mathrm{~V}_{\text {REF }}=3.0 \mathrm{~V} \\ & \mathrm{~T}_{\mathrm{A}}=0 \text { to } 50^{\circ} \mathrm{C} \end{aligned}$ <br> DAC output buffer ON | -40 | -8 | 0 |  |
| TUE ${ }^{(1)}$ | Total unadjusted error | $\mathrm{C}_{\mathrm{L}} \leq 50 \mathrm{pF}, \mathrm{R}_{\mathrm{L}} \geq 5 \mathrm{k} \Omega$ <br> DAC output buffer ON | - | 12 | 30 | LSB |
|  |  | No $R_{L}, C_{L} \leq 50 \mathrm{pF}$ DAC output buffer OFF | - | 8 | 12 |  |
| $\mathrm{t}_{\text {SETTLING }}$ | Settling time (full scale: for a 12-bit code transition between the lowest and the highest input codes till DAC_OUT reaches final value $\pm 1$ LSB | $\mathrm{C}_{\mathrm{L}} \leq 50 \mathrm{pF}, \mathrm{R}_{\mathrm{L}} \geq 5 \mathrm{k} \Omega$ | - | 7 | 12 | $\mu \mathrm{s}$ |
| Update rate | Max frequency for a correct DAC_OUT change (95\% of final value) with 1 LSB variation in the input code | $\mathrm{C}_{\mathrm{L}} \leq 50 \mathrm{pF}, \mathrm{R}_{\mathrm{L}} \geq 5 \mathrm{k} \Omega$ | - | - | 1 | Msps |
| $\mathrm{t}_{\text {WAKEUP }}$ | Wakeup time from off state (setting the ENx bit in the DAC Control register) ${ }^{(8)}$ | $\mathrm{C}_{\mathrm{L}} \leq 50 \mathrm{pF}, \mathrm{R}_{\mathrm{L}} \geq 5 \mathrm{k} \Omega$ | - | 9 | 15 | $\mu \mathrm{s}$ |
| PSRR+ | $V_{\text {DDA }}$ supply rejection ratio (static DC measurement) | $\mathrm{C}_{\mathrm{L}} \leq 50 \mathrm{pF}, \mathrm{R}_{\mathrm{L}} \geq 5 \mathrm{k} \Omega$ | - | -60 | -35 | dB |

1. Data based on characterization results.
2. Connected between DAC_OUT and VSSA.
3. Difference between two consecutive codes - 1 LSB.
4. Difference between measured value at Code i and the value at Code i on a line drawn between Code 0 and last Code 4095.
5. Difference between the value measured at $\operatorname{Code}(0 \times 800)$ and the ideal value $=V_{R E F+} / 2$.
6. Difference between the value measured at Code (0x001) and the ideal value.
7. Difference between ideal slope of the transfer function and measured slope computed from code $0 \times 000$ and $0 x F F F$ when buffer is OFF, and from code giving 0.2 V and $\left(\mathrm{V}_{\mathrm{DDA}}-0.2\right) \mathrm{V}$ when buffer is ON .
8. In buffered mode, the output can overshoot above the final value for low input code (starting from min value).

Figure 39. 12-bit buffered /non-buffered DAC


1. The DAC integrates an output buffer that can be used to reduce the output impedance and to drive external loads directly without the use of an external operational amplifier. The buffer can be bypassed by configuring the BOFFx bit in the DAC_CR register.

### 6.3.21 Operational amplifier characteristics

Table 68. Operational amplifier characteristics

| Symbol | Parameter |  | Condition ${ }^{(1)}$ | Min ${ }^{(2)}$ | Typ | Max ${ }^{(2)}$ | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CMIR | Common mode input range |  | - | 0 | - | $V_{D D}$ | - |
| $\mathrm{VI}_{\text {OFFSET }}$ | Input offset voltage | Maximum calibration range | - | - | - | $\pm 15$ |  |
|  |  | After offset calibration | - | - | - | $\pm 1.5$ |  |
| $\Delta \mathrm{VI}_{\text {OFFSET }}$ | Input offset voltage drift | Normal mode | - | - | - | $\pm 40$ | $\mu \mathrm{V} /{ }^{\circ} \mathrm{C}$ |
|  |  | Low-power mode | - | - | - | $\pm 80$ | - |
| $\mathrm{I}_{\mathrm{B}}$ | Input current bias | Dedicated input | $75^{\circ} \mathrm{C}$ | - | - | 1 | nA |
|  |  | General purpose input |  | - | - | 10 |  |
| I Load | Drive current | Normal mode | - | - | - | 500 | $\mu \mathrm{A}$ |
|  |  | Low-power mode | - | - | - | 100 |  |
| $\mathrm{I}_{\mathrm{DD}}$ | Consumption | Normal mode | No load, quiescent mode | - | 100 | 220 | $\mu \mathrm{A}$ |
|  |  | Low-power mode |  | - | 30 | 60 |  |
| CMRR | Common mode rejection ration | Normal mode | - | - | -85 | - | dB |
|  |  | Low-power mode | - | - | -90 | - |  |

Table 68. Operational amplifier characteristics (continued)

| Symbol | Parameter |  | Condition ${ }^{(1)}$ | $\mathbf{M i n}{ }^{(2)}$ | Typ | Max ${ }^{(2)}$ | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PSRR | Power supply rejection ratio | Normal mode | DC | - | -85 | - | dB |
|  |  | Low-power mode |  | - | -90 | - |  |
| GBW | Bandwidth | Normal mode | $\mathrm{V}_{\mathrm{DD}}>2.4 \mathrm{~V}$ | 400 | 1000 | 3000 | kHZ |
|  |  | Low-power mode |  | 150 | 300 | 800 |  |
|  |  | Normal mode | $\mathrm{V}_{\mathrm{DD}}<2.4 \mathrm{~V}$ | 200 | 500 | 2200 |  |
|  |  | Low-power mode |  | 70 | 150 | 800 |  |
| SR | Slew rate | Normal mode | $\mathrm{V}_{\mathrm{DD}}>2.4 \mathrm{~V}$ <br> (between 0.1 V and $\left.V_{D D}-0.1 \mathrm{~V}\right)$ | - | 700 | - | $\mathrm{V} / \mathrm{ms}$ |
|  |  | Low-power mode | $\mathrm{V}_{\mathrm{DD}}>2.4 \mathrm{~V}$ | - | 100 | - |  |
|  |  | Normal mode | $\mathrm{V}_{\mathrm{DD}}<2.4 \mathrm{~V}$ | - | 300 | - |  |
|  |  | Low-power mode |  | - | 50 | - |  |
| AO | Open loop gain | Normal mode | - | 55 | 100 | - | dB |
|  |  | Low-power mode | - | 65 | 110 | - |  |
| $\mathrm{R}_{\mathrm{L}}$ | Resistive load | Normal mode | $\mathrm{V}_{\mathrm{DD}}<2.4 \mathrm{~V}$ | 4 | - | - | k $\Omega$ |
|  |  | Low-power mode |  | 20 | - | - |  |
| $\mathrm{C}_{\mathrm{L}}$ | Capacitive load |  | - | - | - | 50 | pF |
| $\mathrm{VOH}_{\text {SAT }}$ | High saturation voltage | Normal mode | $\begin{aligned} & \mathrm{I}_{\mathrm{LOAD}}=\max \text { or } \\ & R_{\mathrm{L}}=\min \end{aligned}$ | $\begin{aligned} & V_{D D}- \\ & 100 \end{aligned}$ | - | - | mV |
|  |  | Low-power mode |  | $\mathrm{V}_{\mathrm{DD}}-50$ | - | - |  |
| $\mathrm{VOL}_{\text {SAT }}$ | Low saturation voltage | Normal mode |  | - | - | 100 |  |
|  |  | Low-power mode |  | - | - | 50 |  |
| $\varphi m$ | Phase margin |  | - | - | 60 | - | - |
| GM | Gain margin |  | - | - | -12 | - | dB |
| $\mathrm{t}_{\text {OFFTRIM }}$ | Offset trim time: during calibration, minimum time needed between two steps to have 1 mV accuracy |  | - | - | 1 | - | ms |
| ${ }^{\text {t WAKEUP }}$ | Wakeup time | Normal mode | $\begin{aligned} & \mathrm{C}_{\mathrm{L}} \leq 50 \mathrm{pf}, \\ & \mathrm{R}_{\mathrm{L}} \geq 4 \mathrm{k} \Omega \end{aligned}$ | - | 10 | - | $\mu \mathrm{s}$ |
|  |  | Low-power mode | $\begin{aligned} & \mathrm{C}_{\mathrm{L}} \leq 50 \mathrm{pf}, \\ & \mathrm{R}_{\mathrm{L}} \geq 20 \mathrm{k} \Omega \end{aligned}$ | - | 30 | - |  |

1. Operating conditions are limited to junction temperature $\left(0^{\circ} \mathrm{C}\right.$ to $\left.105^{\circ} \mathrm{C}\right)$ when $\mathrm{V}_{\mathrm{DD}}$ is below 2 V . Otherwise to the full ambient temperature range $\left(-40^{\circ} \mathrm{C}\right.$ to $85^{\circ} \mathrm{C},-40^{\circ} \mathrm{C}$ to $\left.105^{\circ} \mathrm{C}\right)$.
2. Guaranteed by characterization results.

### 6.3.22 Temperature sensor characteristics

Table 69. Temperature sensor calibration values

| Calibration value name | Description | Memory address |
| :--- | :--- | :--- |
| TS_CAL1 | TS ADC raw data acquired at <br> temperature of $30^{\circ} \mathrm{C} \pm 5^{\circ} \mathrm{C}$ <br> $\mathrm{V}_{\text {DDA }}=3 \mathrm{~V} \pm 10 \mathrm{mV}$ | $0 \times 1 \mathrm{FF} 800 \mathrm{FA}-0 \times 1 \mathrm{FF} 800 \mathrm{FB}$ |
| TS_CAL2 | TS ADC raw data acquired at <br> temperature of $110^{\circ} \mathrm{C} \pm 5^{\circ} \mathrm{C}$ <br> $\mathrm{V}_{\text {DDA }}=3 \mathrm{~V} \pm 10 \mathrm{mV}$ | $0 \times 1 \mathrm{FF} 800 \mathrm{FE}-0 \times 1 \mathrm{FF} 800 \mathrm{FF}$ |

Table 70. Temperature sensor characteristics

| Symbol | Parameter | Min | Typ | Max | Unit |
| :--- | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{T}_{\mathrm{L}}{ }^{(1)}$ | $\mathrm{V}_{\text {SENSE }}$ linearity with temperature | - | $\pm 1$ | $\pm 2$ | ${ }^{\circ} \mathrm{C}$ |
| Avg_Slope $^{(1)}$ | Average slope | 1.48 | 1.61 | 1.75 | $\mathrm{mV} /{ }^{\circ} \mathrm{C}$ |
| $\mathrm{V}_{110}$ | Voltage at $110^{\circ} \mathrm{C} \pm 5^{\circ} \mathrm{C}^{(2)}$ | 612 | 626.8 | 641.5 | mV |
| $\mathrm{I}_{\text {DDA(TEMP }}{ }^{(3)}$ | Current consumption | - | 3.4 | 6 | $\mu \mathrm{~A}$ |
| $\mathrm{t}_{\text {START }}{ }^{(3)}$ | Startup time | - | - | 10 | $\mathrm{\mu s}$ |
| $\mathrm{~T}_{\text {S_temp }}{ }^{(3)}$ | ADC sampling time when reading the <br> temperature | 4 | - | - |  |

1. Guaranteed by characterization results.
2. Measured at $\mathrm{V}_{\mathrm{DD}}=3 \mathrm{~V} \pm 10 \mathrm{mV}$. V 110 ADC conversion result is stored in the TS_CAL2 byte.
3. Guaranteed by design.

### 6.3.23 Comparator

Table 71. Comparator 1 characteristics

| Symbol | Parameter | Conditions | Min ${ }^{(1)}$ | Typ | Max ${ }^{(1)}$ | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $V_{\text {DDA }}$ | Analog supply voltage | - | 1.65 | - | 3.6 | V |
| $\mathrm{R}_{400 \mathrm{~K}}$ | $\mathrm{R}_{400 \mathrm{~K}}$ value | - | - | 400 | - | $\mathrm{k} \Omega$ |
| $\mathrm{R}_{10 \mathrm{~K}}$ | $\mathrm{R}_{10 \mathrm{~K}}$ value | - | - | 10 | - |  |
| $\mathrm{V}_{\text {IN }}$ | Comparator 1 input voltage range | - | 0.6 | - | $V_{\text {DDA }}$ | V |
| $\mathrm{t}_{\text {START }}$ | Comparator startup time | - | - | 7 | 10 | $\mu \mathrm{s}$ |
| td | Propagation delay ${ }^{(2)}$ | - | - | 3 | 10 |  |
| Voffset | Comparator offset | - | - | $\pm 3$ | $\pm 10$ | mV |
| $\mathrm{d}_{\text {Voffset }} / \mathrm{dt}$ | Comparator offset variation in worst voltage stress conditions | $\begin{aligned} & \mathrm{V}_{\mathrm{DDA}}=3.6 \mathrm{~V} \\ & \mathrm{~V}_{\text {IN }+}=0 \mathrm{~V} \\ & \mathrm{~V}_{\text {IN }-}=\mathrm{V}_{\text {REFINT }} \\ & \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C} \end{aligned}$ | 0 | 1.5 | 10 | $\mathrm{mV} / 1000 \mathrm{~h}$ |
| $\mathrm{I}_{\text {COMP1 }}$ | Current consumption ${ }^{(3)}$ | - | - | 160 | 260 | nA |

1. Guaranteed by characterization results.
2. The delay is characterized for 100 mV input step with 10 mV overdrive on the inverting input, the noninverting input set to the reference.
3. Comparator consumption only. Internal reference voltage not included.

Table 72. Comparator 2 characteristics

| Symbol | Parameter | Conditions | Min | Typ | Max ${ }^{(1)}$ | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {DDA }}$ | Analog supply voltage | - | 1.65 | - | 3.6 | V |
| $\mathrm{V}_{\text {IN }}$ | Comparator 2 input voltage range | - | 0 | - | $V_{\text {DDA }}$ | V |
| $\mathrm{t}_{\text {START }}$ | Comparator startup time | Fast mode | - | 15 | 20 | $\mu \mathrm{s}$ |
|  |  | Slow mode | - | 20 | 25 |  |
| $\mathrm{t}_{\text {d slow }}$ | Propagation delay ${ }^{(2)}$ in slow mode | $1.65 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DDA}} \leq 2.7 \mathrm{~V}$ | - | 1.8 | 3.5 |  |
|  |  | $2.7 \mathrm{~V} \leq \mathrm{V}_{\text {DDA }} \leq 3.6 \mathrm{~V}$ | - | 2.5 | 6 |  |
| $\mathrm{t}_{\mathrm{d} \text { fast }}$ | Propagation delay ${ }^{(2)}$ in fast mode | $1.65 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DDA}} \leq 2.7 \mathrm{~V}$ | - | 0.8 | 2 |  |
|  |  | $2.7 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DDA}} \leq 3.6 \mathrm{~V}$ | - | 1.2 | 4 |  |
| $\mathrm{V}_{\text {offset }}$ | Comparator offset error | - | - | $\pm 4$ | $\pm 20$ | mV |
| dThreshold/ dt | Threshold voltage temperature coefficient | $\begin{aligned} & \mathrm{V}_{\text {DDA }}=3.3 \mathrm{~V} \\ & \mathrm{~T}_{\mathrm{A}}=0 \text { to } 50{ }^{\circ} \mathrm{C} \\ & \mathrm{~V}_{-}=\mathrm{V}_{\text {REFINT }}, \\ & 3 / 4 \text { V }_{\text {REFINT }}, \\ & 1 / 2 \mathrm{~V}_{\text {REFINT }}, \\ & 1 / 4 \text { V REFINT. }^{2} \end{aligned}$ | - | 15 | 100 | $\underset{1{ }^{\circ} \mathrm{C}}{\mathrm{ppm}}$ |
| $\mathrm{I}_{\text {COMP2 }}$ | Current consumption ${ }^{(3)}$ | Fast mode | - | 3.5 | 5 | $\mu \mathrm{A}$ |
|  |  | Slow mode | - | 0.5 | 2 |  |

1. Guaranteed by characterization results.
2. The delay is characterized for 100 mV input step with 10 mV overdrive on the inverting input, the noninverting input set to the reference.
3. Comparator consumption only. Internal reference voltage (necessary for comparator operation) is not included.

### 6.3.24 LCD controller

The device embeds a built-in step-up converter to provide a constant LCD reference voltage independently from the $V_{D D}$ voltage. An external capacitor $C_{e x t}$ must be connected to the $\mathrm{V}_{\mathrm{LCD}}$ pin to decouple this converter.

Table 73. LCD controller characteristics

| Symbol | Parameter | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {LCD }}$ | LCD external voltage | - | - | 3.6 | V |
| $V_{\text {LCDO }}$ | LCD internal reference voltage 0 | - | 2.6 | - |  |
| $V_{\text {LCD1 }}$ | LCD internal reference voltage 1 | - | 2.73 | - |  |
| $V_{\text {LCD2 }}$ | LCD internal reference voltage 2 | - | 2.86 | - |  |
| V LCD3 | LCD internal reference voltage 3 | - | 2.98 | - |  |
| V ${ }_{\text {LCD4 }}$ | LCD internal reference voltage 4 | - | 3.12 | - |  |
| $V_{\text {LCD5 }}$ | LCD internal reference voltage 5 | - | 3.26 | - |  |
| $V_{\text {LCD6 }}$ | LCD internal reference voltage 6 | - | 3.4 | - |  |
| V LCD7 | LCD internal reference voltage 7 | - | 3.55 | - |  |
| $\mathrm{C}_{\text {ext }}$ | $\mathrm{V}_{\text {LCD }}$ external capacitance | 0.1 | - | 2 | $\mu \mathrm{F}$ |
| $\mathrm{ILCD}^{(1)}$ | Supply current at $\mathrm{V}_{\mathrm{DD}}=2.2 \mathrm{~V}$ | - | 3.3 | - | $\mu \mathrm{A}$ |
|  | Supply current at $\mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}$ | - | 3.1 | - |  |
| $\mathrm{R}_{\text {Htot }}{ }^{(2)}$ | Low drive resistive network overall value | 5.28 | 6.6 | 7.92 | $\mathrm{M} \Omega$ |
| $\mathrm{R}_{\mathrm{L}}{ }^{(2)}$ | High drive resistive network total value | 192 | 240 | 288 | $\mathrm{k} \Omega$ |
| $\mathrm{V}_{44}$ | Segment/Common highest level voltage | - | - | $V_{\text {LCD }}$ | V |
| $V_{34}$ | Segment/Common 3/4 level voltage | - | $3 / 4 \mathrm{~V}_{\text {LCD }}$ | - | V |
| $\mathrm{V}_{23}$ | Segment/Common $2 / 3$ level voltage | - | 2/3 $\mathrm{V}_{\text {LCD }}$ | - |  |
| $\mathrm{V}_{12}$ | Segment/Common 1/2 level voltage | - | $1 / 2 \mathrm{~V}_{\text {LCD }}$ | - |  |
| $\mathrm{V}_{13}$ | Segment/Common 1/3 level voltage | - | $1 / 3 \mathrm{~V}_{\text {LCD }}$ | - |  |
| $\mathrm{V}_{14}$ | Segment/Common 1/4 level voltage | - | $1 / 4 \mathrm{~V}_{\text {LCD }}$ | - |  |
| $\mathrm{V}_{0}$ | Segment/Common lowest level voltage | 0 | - | - |  |
| $\Delta \mathrm{Vxx}{ }^{(3)}$ | Segment/Common level voltage error $\mathrm{T}_{\mathrm{A}}=-40 \text { to } 105^{\circ} \mathrm{C}$ | - | - | $\pm 50$ | mV |

1. LCD enabled with 3 V internal step-up active, $1 / 8$ duty, $1 / 4$ bias, division ratio $=64$, all pixels active, no LCD connected.
2. Guaranteed by design.
3. Guaranteed by characterization results.

## 7 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK packages, depending on their level of environmental compliance. ECOPACK specifications, grade definitions and product status are available at: www.st.com. ECOPACK is an ST trademark.

### 7.1 LQFP64 package information

LQFP64 is a 14-pin, $10 \times 10 \mathrm{~mm}$, low-profile quad flat package information.
Figure 40. LQFP64 outline


1. Drawing is not to scale.

Table 74. LQFP64 mechanical data

| Symbol | millimeters |  |  | inches $^{(1)}$ |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Min | Typ | Max | Min | Typ | Max |
| A | - | - | 1.600 | - | - | 0.0630 |
| A1 | 0.050 | - | 0.150 | 0.0020 | - | 0.0059 |
| A2 | 1.350 | 1.400 | 1.450 | 0.0531 | 0.0551 | 0.0571 |
| b | 0.170 | 0.220 | 0.270 | 0.0067 | 0.0087 | 0.0106 |

Table 74. LQFP64 mechanical data (continued)

| Symbol | millimeters |  |  | inches $^{(1)}$ |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Min | Typ | Max | Min | Typ | Max |
| c | 0.090 | - | 0.200 | 0.0035 | - | 0.0079 |
| D | - | 12.000 | - | - | 0.4724 | - |
| D1 | - | 10.000 | - | - | 0.3937 | - |
| D3 | - | 7.500 | - | - | 0.2953 | - |
| E | - | 12.000 | - | - | 0.4724 | - |
| E1 | - | 10.000 | - | - | 0.3937 | - |
| E3 | - | 7.500 | - | - | 0.2953 | - |
| e | - | 0.500 | - | - | 0.0197 | - |
| K | $0^{\circ}$ | $3.5^{\circ}$ | $7^{\circ}$ | $0^{\circ}$ | $3.5^{\circ}$ | $7^{\circ}$ |
| L | 0.450 | 0.600 | 0.750 | 0.0177 | 0.0236 | 0.0295 |
| L1 | - | 1.000 | - | - | 0.0394 | - |
| Ccc | - | - | 0.080 | - | - | 0.0031 |

1. Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 41. LQFP64 recommended footprint


1. Dimensions are in millimeters.

## LQFP64 device marking

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

The printed markings may differ depending on the supply chain.
Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

Figure 42. LQFP64 top view example


1. Parts marked as ES or E or accompanied by an engineering sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

### 7.2 WLCSP64 package information

WLCSP64 is a 64-ball, 0.4 mm pitch wafer level chip scale package.
Figure 43. WLCSP64 outline


1. Drawing is not to scale.

Table 75. WLCSP64 mechanical data

| Symbol | millimeters |  |  | inches $^{(1)}$ |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Min | Typ | Max | Min | Typ | Max |
| A | 0.540 | 0.570 | 0.600 | 0.0205 | 0.0224 | 0.0236 |
| A1 | - | 0.190 | - | - | 0.0075 | - |
| A2 | - | 0.380 | - | - | 0.0150 | - |
| $\mathrm{b}^{(2)}$ | 0.240 | 0.270 | 0.300 | 0.0094 | 0.0106 | 0.0118 |

Table 75. WLCSP64 mechanical data (continued)

| Symbol | millimeters |  |  | inches $^{(1)}$ |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Min | Typ | Max | Min | Typ | Max |
| D | 4.504 | 4.539 | 4.574 | 0.1773 | 0.1787 | 0.1801 |
| E | 4.876 | 4.911 | 4.946 | 0.1920 | 0.1933 | 0.1947 |
| e | - | 0.400 | - | - | 0.0157 | - |
| e1 | - | 2.800 | - | - | 0.1102 | - |
| F | - | 0.870 | - | - | 0.0343 | - |
| G | - | 1.056 | - | - | 0.0416 | - |
| aaa | - | - | 0.100 | - | - | 0.0039 |
| bbb | - | - | 0.100 | - | - | 0.0039 |
| ccc | - | - | 0.100 | - | - | 0.0039 |
| ddd | - | - | 0.050 | - | - | 0.0020 |
| eee | - | - | 0.050 | - | - | 0.0020 |

1. Values in inches are converted from mm and rounded to 4 decimal digits.
2. Dimension is measured at the maximum bump diameter parallel to primary datum Z .

Figure 44. WLCSP64 recommended footprint


Table 76. WLCSP64 recommended PCB design rules

| Dimension | Recommended values |
| :--- | :--- |
| Pitch | 0.4 |
| Dpad | $260 \mu \mathrm{~m}$ max. (circular) |
|  | $220 \mu \mathrm{~m}$ recommended |
| Dsm | $300 \mu \mathrm{~m}$ min. (for $260 \mu \mathrm{~m}$ diameter pad) |
| PCB pad design | Non-solder mask defined via underbump allowed. |

## WLCSP64 device marking

The following figure gives an example of topside marking orientation versus ball A1 identifier location.

The printed markings may differ depending on the supply chain.
Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

Figure 45. WLCSP64 top view example


1. Parts marked as ES or E or accompanied by an engineering sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

### 7.3 LQFP100 package information

LQFP100 is a 100 -pin, $14 \times 14 \mathrm{~mm}$, low-profile quad flat package.
Figure 46. LQFP100 outline


1. Drawing is not to scale.

Table 77. LQPF100 mechanical data

| Symbol | millimeters |  |  | inches $^{(1)}$ |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Min | Typ | Max | Min | Typ | Max |
| A | - | - | 1.600 | - | - | 0.0630 |
| A1 | 0.050 | - | 0.150 | 0.0020 | - | 0.0059 |
| A2 | 1.350 | 1.400 | 1.450 | 0.0531 | 0.0551 | 0.0571 |
| b | 0.170 | 0.220 | 0.270 | 0.0067 | 0.0087 | 0.0106 |
| c | 0.090 | - | 0.200 | 0.0035 | - | 0.0079 |
| D | 15.800 | 16.000 | 16.200 | 0.6220 | 0.6299 | 0.6378 |
| D1 | 13.800 | 14.000 | 14.200 | 0.5433 | 0.5512 | 0.5591 |
| D3 | - | 12.000 | - | - | 0.4724 | - |
| E | 15.800 | 16.000 | 16.200 | 0.6220 | 0.6299 | 0.6378 |
| E1 | 13.800 | 14.000 | 14.200 | 0.5433 | 0.5512 | 0.5591 |
| E3 | - | 12.000 | - | - | 0.4724 | - |
| e | - | 0.500 | - | - | 0.0197 | - |
| L | 0.450 | 0.600 | 0.750 | 0.0177 | 0.0236 | 0.0295 |
| L1 | - | 1.000 | - | - | 0.0394 | - |

Table 77. LQPF100 mechanical data (continued)

| Symbol | millimeters |  |  | inches $^{(1)}$ |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Min | Typ | Max | Min | Typ | Max |
| k | $0.0^{\circ}$ | $3.5^{\circ}$ | $7.0^{\circ}$ | $0.0^{\circ}$ | $3.5^{\circ}$ | $7.0^{\circ}$ |
| ccc | - | - | 0.080 | - | - | 0.0031 |

1. Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 47. LQFP100 recommended footprint


1. Dimensions are in millimeters

## LQFP100 device marking

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

The printed markings may differ depending on the supply chain.
Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

Figure 48. LQFP100 top view example


1. Parts marked as ES or E or accompanied by an engineering sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

### 7.4 UFBGA132 package information

UFBGA132 is a 132 -ball, $7 \times 7 \mathrm{~mm}$, ultra thin, fine-pitch ball grid array package.
Figure 49. UFBGA132 outline


1. Drawing is not to scale.

Table 78. UFBGA132 mechanical data

| Symbol | millimeters |  |  | inches $^{(1)}$ |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Min | Typ | Max | Min | Typ | Max |
| A | 0.460 | 0.530 | 0.600 | 0.0181 | 0.0209 | 0.0236 |
| A1 | 0.050 | 0.080 | 0.110 | 0.0020 | 0.0031 | 0.0043 |
| A2 | 0.400 | 0.450 | 0.500 | 0.0157 | 0.0177 | 0.0197 |
| A3 | - | 0.130 | - | - | 0.0051 | - |
| A4 | 0.270 | 0.320 | 0.370 | 0.0106 | 0.0126 | 0.0146 |
| b | 0.170 | 0.280 | 0.330 | 0.0067 | 0.0110 | 0.0130 |
| D | 6.950 | 7.000 | 7.050 | 0.2736 | 0.2756 | 0.2776 |
| E | 6.950 | 7.000 | 7.050 | 0.2736 | 0.2756 | 0.2776 |
| e | - | 0.500 | - | - | 0.0197 | - |
| Z | 0.700 | 0.750 | 0.800 | 0.0276 | 0.0295 | 0.0315 |

Table 78. UFBGA132 mechanical data (continued)

| Symbol | millimeters |  |  | inches $^{(1)}$ |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Min | Typ | Max | Min | Typ | Max |
| ddd | - | - | 0.080 | - | - | 0.0031 |
| eee | - | - | 0.150 | - | - | 0.0059 |
| fff | - | - | 0.050 | - | - | 0.0020 |

1. Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 50. UFBGA132 recommended footprint


| Pitch | 0.5 mm |
| :--- | :--- |
| D pad | 0.27 mm |
| Dsm | 0.35 mm typ (depends on <br> the soldermask registration <br> tolerance) |
| Solder paste | 0.27 mm aperture diameter |

## UFBGA132 device marking

The following figure gives an example of topside marking orientation versus ball A1 identifier location.

The printed markings may differ depending on the supply chain.
Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

Figure 51. UFBGA132 top view example


1. Parts marked as ES or E or accompanied by an engineering sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

### 7.5 LQFP144 package information

LQFP144 is a 144-pin $20 \times 20 \mathrm{~mm}$, low-profile quad flat package.
Figure 52. LQFP144 outline


1. Drawing is not to scale.

Table 79. LQFP144 mechanical data

| Symbol | millimeters |  |  | inches $^{(1)}$ |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Min | Typ | Max | Min | Typ | Max |
| A | - | - | 1.600 | - | - | 0.0630 |
| A1 | 0.050 | - | 0.150 | 0.0020 | - | 0.0059 |
| A2 | 1.350 | 1.400 | 1.450 | 0.0531 | 0.0551 | 0.0571 |
| b | 0.170 | 0.220 | 0.270 | 0.0067 | 0.0087 | 0.0106 |
| c | 0.090 | - | 0.200 | 0.0035 | - | 0.0079 |
| D | 21.800 | 22.000 | 22.200 | 0.8583 | 0.8661 | 0.8740 |
| D1 | 19.800 | 20.000 | 20.200 | 0.7795 | 0.7874 | 0.7953 |
| D3 | - | 17.500 | - | - | 0.6890 | - |
| E | 21.800 | 22.000 | 22.200 | 0.8583 | 0.8661 | 0.8740 |
| E1 | 19.800 | 20.000 | 20.200 | 0.7795 | 0.7874 | 0.7953 |
| E3 | - | 17.500 | - | - | 0.6890 | - |
| e | - | 0.500 | - | - | 0.0197 | - |
| L | 0.450 | 0.600 | 0.750 | 0.0177 | 0.0236 | 0.0295 |
| L1 | - | 1.000 | - | - | 0.0394 | - |
| k | $0^{\circ}$ | $3.5^{\circ}$ | $7^{\circ}$ | $0^{\circ}$ | $3.5^{\circ}$ | $7^{\circ}$ |
| ccc | - | - | 0.080 | - | - | 0.0031 |

1. Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 53. LQFP144 recommended footprint


1. Dimensions are in millimeters.

## LQFP144 device marking

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

The printed markings may differ depending on the supply chain.
Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

Figure 54. LQFP144 top view example


1. Parts marked as ES or E or accompanied by an Engineering sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

### 7.6 Thermal characteristics

The maximum chip-junction temperature, $T_{J}$ max, in degrees Celsius, may be calculated using the following equation:
$T_{J} \max =T_{A} \max +\left(P_{D} \max \times \Theta_{J A}\right)$
Where:

- $\quad \mathrm{T}_{\mathrm{A}}$ max is the maximum ambient temperature in ${ }^{\circ} \mathrm{C}$,
- $\quad \Theta_{J A}$ is the package junction-to-ambient thermal resistance, in ${ }^{\circ} \mathrm{C} / \mathrm{W}$,
- $\quad P_{D}$ max is the sum of $P_{I N T} \max$ and $P_{I / O} \max \left(P_{D} \max =P_{I N T} \max +P_{I / O} m a x\right)$,
- $\quad P_{I N T} m a x$ is the product of $I_{D D}$ and $V_{D D}$, expressed in Watts. This is the maximum chip internal power.
$\mathrm{P}_{\mathrm{I} / \mathrm{O}}$ max represents the maximum power dissipation on output pins where:

$$
\mathrm{P}_{\mathrm{I} / \mathrm{O}} \max =\Sigma\left(\mathrm{V}_{\mathrm{OL}} \times \mathrm{l}_{\mathrm{OL}}\right)+\Sigma\left(\left(\mathrm{V}_{\mathrm{DD}}-\mathrm{V}_{\mathrm{OH}}\right) \times \mathrm{I}_{\mathrm{OH}}\right),
$$

taking into account the actual $\mathrm{V}_{\mathrm{OL}} / \mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}} / \mathrm{I}_{\mathrm{OH}}$ of the $\mathrm{I} / \mathrm{Os}$ at low and high level in the application.

Table 80. Thermal characteristics

| Symbol | Parameter | Value | Unit |
| :---: | :---: | :---: | :---: |
| $\Theta_{J A}$ | Thermal resistance junction-ambient LQFP144-20 x $20 \mathrm{~mm} / 0.5 \mathrm{~mm}$ pitch | 40 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
|  | Thermal resistance junction-ambient UFBGA132-7x7mm | 60 |  |
|  | Thermal resistance junction-ambient LQFP100-14 x $14 \mathrm{~mm} / 0.5 \mathrm{~mm}$ pitch | 43 |  |
|  | Thermal resistance junction-ambient LQFP64-10 x $10 \mathrm{~mm} / 0.5 \mathrm{~mm}$ pitch | 46 |  |
|  | Thermal resistance junction-ambient WLCSP64-0.400 mm pitch | 46 |  |

Figure 55. Thermal resistance suffix 6


Figure 56. Thermal resistance suffix 7


### 7.6.1 Reference document

JESD51-2 Integrated Circuits Thermal Test Method Environment Conditions - Natural Convection (Still Air). Available from www.jedec.org.

## 8 Ordering information

Table 81. Ordering information scheme


For a list of available options (speed, package, etc.) or for further information on any aspect of this device, contact the nearest ST sales office.

## $9 \quad$ Revision History

Table 82. Document revision history

| Date | Revision | Changes |
| :---: | :---: | :---: |
| 03-Oct-2011 | 1 | Initial release. |
| 03-Feb-2012 | 2 | Status of the document changed (datasheet instead of preliminary data). <br> Updated low power features on page 1. <br> Removed references to devices with 256 KB of Flash memory. GPIOF replaced with GIOPH. <br> Added SDIO in Table 4: Ultra-low-power STM32L15xxD device features and peripheral counts on page 12 and in Table 19: ction input/output on page 86 (FSMC/SDIO instead of FSMC). <br> Table 4: Ultra-low-power STM32L15xxD device features and peripheral counts: replaced STM32L15xWx with STM32L15xQx. <br> Figure 1: Ultra-low-power STM32L162xC block diagram: updated legend. <br> Modified Section 3.4: Clock management on page 20. <br> Table 4: STM32L15xQD STM32L162QD UFBGA132 ballout: replaced <br> STM32L15xWC/D with STM32L15xQD. <br> Figure 3, Figure 3, Figure 4: updated titles. <br> Table 14: STM32L15xxD pin definitions: updated title, updated pins PF0, PF1, PH2, PF12, PF13, PF14, PF15, PG0, PG1, PG12, PG15, PD0, and PD1. <br> Table 19: ction input/output: Modified ction for PA13 and PA14; removed EVENT OUT for PH2. <br> Figure 5: Memory map: removed the text "APB memory space". <br> Modified Figure 8: Power supply scheme on page 46. <br> Modified Table 2: Functionalities depending on the operating power supply range on page 15. <br> Table 18: Current consumption in Run mode, code with data processing running from RAM: added footnote 3. <br> Table 19: Current consumption in Sleep mode: updated condition for $\mathrm{f}_{\mathrm{HSE}}$; added footnote 3. <br> Table 23: Typical and maximum current consumptions in Standby mode: modified max values. <br> Table 64: USB DC electrical characteristics: removed two footnotes. <br> Modified Table 38: Flash memory and data EEPROM characteristics on page 83. <br> Table 73: Thermal characteristics: updated "TBDs" with values. Modified tables in Section 6.3.4: Supply current characteristics on page 54. |

Table 82. Document revision history (continued)

| Date | Revision | Changes |
| :---: | :---: | :---: |
| 18-Apr-2012 | 3 | Added WLCSP64 package. <br> Section 3: Functional overview: changed ' 128 kHz ' to ' 131 kHz ' in section "Low power run mode". <br> Section 3.17.1: General-purpose timers (TIM2, TIM3, TIM4, TIM5, TIM9, TIM10 and TIM11): changed 'six' to 'seven' synchronizable general-purpose timers. <br> Table 14: STM32L15xxD pin definitions on page 52: updated name of reference manual in footnote 5 . <br> I2C updated: footnote 3. from Table 58 <br> Note about I2C clock updated: footnote 2. from Table 58 modified. Note [non-robust] updated: footnote 2. from Table 68 modified. <br> GPIOs high current capability updated: Section 3.6: GPIOs (generalpurpose inputs/outputs) 'except for analog inputs' was removed. |
| 15-Jun-2012 | 4 | Changed maximum number of touch sensing channels to 34 , and updated Table 4: Ultra-low-power STM32L15xxD device features and peripheral counts. <br> Updated Section 3.10: ADC (analog-to-digital converter) to add <br> Section 3.10.1: Temperature sensor and Section 3.10.2: Internal voltage reference (VREFINT). <br> Removed caution note below Figure 8: Power supply scheme. <br> Added note below Table 4: STM32L15xQD STM32L162QD <br> UFBGA132 ballout. <br> Modified Table 8: STM32L15xRDSTM32L162RD WLCSP64 ballout to match top view. <br> Changed FSMC_LBAR into FSMC_NADV, and I2C1_SMBAI into I2C1_SMBA in Table 14: STM32L15xxD pin definitions. <br> Modified PB10/11/12 for AFIO4 ction, and replaced LBAR by NADV for AFIO12 in Table 19: ction input/output. <br> Updated Table 22: Typical and maximum current consumptions in Stop mode and added Note 6. Updated Table 23: Typical and maximum current consumptions in Standby mode. Updated twustop in Table <br> Updated Table 27: Peripheral current consumption. <br> Updated Table 60: SPI characteristics, added Note 1 and Note 3, and applied Note 2 to $\mathrm{t}_{\mathrm{r}(\mathrm{SCK})}, \mathrm{t}_{\mathrm{f}(\mathrm{SCK})}, \mathrm{t}_{\mathrm{w}(\mathrm{SCKH})}, \mathrm{t}_{\mathrm{w}(\mathrm{SCKL})}, \mathrm{t}_{\mathrm{su}(\mathrm{MI})}, \mathrm{t}_{\mathrm{su}(\mathrm{SI})}, \mathrm{t}_{\mathrm{h}(\mathrm{MI})}$, and $\mathrm{t}_{\mathrm{h}(\mathrm{SI})}$. <br> Updated $\mathrm{I}_{\mathrm{DD}}$ maximum value in Table 38: Flash memory and data EEPROM characteristics. |

Table 82. Document revision history (continued)

| Date | Revision | Changes |
| :---: | :---: | :--- |
|  |  | Updated Features <br> Updated Figure 1: Ultra-low-power STM32L162xC block diagram <br> Added Table 4: Functionalities depending on the working mode (from <br> Run/active down to standby), and Table 3: ange depending on <br> dynamic voltage scaling <br> Updated Figure 3: STM32L162VC LQFP100 pinout <br> Updated Table 14: STM32L15xxD pin definitions |
|  | Added Note 2 in Table 15: Embedded reset and power control block <br> characteristics <br> Replaced TBD values in Table 30: Low-speed external user clock <br> characteristics, Table 38: Flash memory and data EEPROM <br> characteristics and Table 55: I/O AC characteristics |  |
| Added Table 61: I2S characteristics, Figure 29: I2S slave timing |  |  |
| diagram (Philips protocol)(1) and Figure 30: I2S master timing diagram |  |  |
| (Philips protocol)(1) |  |  |
| Added Table 62: SDIO characteristics |  |  |
| Added Figure 31: SDIO timings |  |  |
| Updated Section 6.3.9: FSMC characteristics |  |  |
| Updated Table 72: Temperature sensor characteristics |  |  |
| Added Figure 40: Thermal resistance |  |  |$|$

Table 82. Document revision history (continued)

| Date | Revision | Changes |
| :---: | :---: | :---: |
| 07-Apr-2014 | 7 | Updated current consumption in Section : Features. <br> Updated Section 2.2: Ultra-low-power device continuum. <br> Updated Table 3: Functionalities depending on the operating power supply range.. <br> Added $\mathrm{V}_{\mathrm{DD}}=1.71$ to 1.8 V operating power supply range in Table 5 : Functionalities depending on the working mode (from Run/active down to standby). <br> Updated Section 3.10: LCD (liquid crystal display) to remove $\mathrm{V}_{\mathrm{LCD}}$ rail decoupling. <br> Updated Section 3.16: Touch sensing. <br> Updated Figure 9: Pin loading conditions. <br> Updated Figure 10: Pin input voltage. <br> Updated Figure 11: Power supply scheme. <br> Updated Table 10: Voltage characteristics (added row). <br> Updated Table 11: Current characteristics. <br> Updated Table 13: General operating conditions. Removed figures <br> "Power supply and reference decoupling ( $\mathrm{V}_{\mathrm{REF}}$ not connected to <br> $\mathrm{V}_{\text {DDA }}$ ) and "Power supply and reference decoupling ( $\mathrm{V}_{\text {REF }}$ connected to $V_{\text {DDA }}$ ). <br> Updated Table 15: Embedded internal reference voltage calibration values and moved inside Section 6.3.3: Embedded internal reference voltage. <br> Updated Section 6.3.4: Supply current characteristics. <br> Updated Table 17: Current consumption in Run mode, code with data processing running from Flash, Table 18: Current consumption in Run mode, code with data processing running from RAM, Table 19: Current consumption in Sleep mode, Table 20: Current consumption in Lowpower run mode, Table 21: Current consumption in Low-power sleep mode, Table 22: Typical and maximum current consumptions in Stop mode, and Table 23: Typical and maximum current consumptions in Standby mode. <br> Added Section 6.3.5: Wakeup time from low-power mode. <br> Updated Section 6.3.6: External clock source characteristics. <br> Moved Figure 14: High-speed external clock source AC timing diagram after Table 26: High-speed external user clock characteristics. <br> Updated Figure 17: Typical application with a 32.768 kHz crystal. <br> Updated Table 28: HSE oscillator characteristics. <br> Updated Section 6.3.12: Electrical sensitivity characteristics (title). <br> Updated Section 6.3.13: I/O current injection characteristics. Updated <br> Table 49: I/O current injection susceptibility and added footnote. <br> Updated conditions in Table 51: Output voltage characteristics. <br> Updated Section 6.3.15: NRST pin characteristics.Updated Figure 27: Recommended NRST pin protection. Updated Table 53: NRST pin characteristics. |

Table 82. Document revision history (continued)

| Date | Revision | Changes |
| :---: | :---: | :---: |
| 07-Apr-2014 | $\begin{gathered} 7 \\ \text { (continued) } \end{gathered}$ | Updated Figure 28: $I^{2} C$ bus $A C$ waveforms and measurement circuit. Updated "SDA data hold time" and "SDA and SCL rise time" values and added "Pulse width of spikes that are suppressed by the analog filter" row in Table 55: $I^{2} \mathrm{C}$ characteristics. <br> Updated Table 64: ADC characteristics and Table 65: ADC accuracy. Updated Table 67: DAC characteristics. <br> Updated Table 69: Temperature sensor calibration values and moved inside Section 6.3.22: Temperature sensor characteristics. Removed note 4 in Table 70: Temperature sensor characteristics. <br> Updated Table 78: LQFP64, $10 \times 10 \mathrm{~mm}$ 64-pin low-profile quad flat package mechanical data and Table 79: UFBGA132, $7 \times 7 \mathrm{~mm}$, 132ball ultra thin, fine-pitch ball grid array package mechanical data. Updated Section 8: Ordering information (title). <br> Added Table 50: UFBGA132 recommended footprint.. |
| 23-Oct-2014 | 8 | Updated Section 3.18: Communication interfaces putting I2S characteristics inside. <br> Updated DMIPS features in cover page and Section 2: Description. Updated max temperature at $105^{\circ} \mathrm{C}$ instead of $85^{\circ} \mathrm{C}$ in the whole datasheet. <br> Updated currents consumption in Table 19: Current consumption in Sleep mode. <br> Updated Table 24: Peripheral current consumption with new measured current values. <br> Updated Table 66: Maximum source impedance $R_{\text {AIN }}$ max adding note 2. |
| 04-Feb-2015 | 9 | Updated Section 7: Package information with new package device marking. <br> Updated Figure 8: Memory map. |
| 02-Apr-2015 | 10 | Updated Section 7: Package information with new package paragraph structure (paragraph title and heading level) and adding note for device orientation versus pin 1/ ball A1 identifier. <br> Updated Figure 54: LQFP144 top view example and Figure 48: LQFP100 top view example removing gate mark. <br> Added Figure 44: WLCSP64 recommended footprint and Table 81: WLCSP64, 0.4 mm pitch package recommended $P C B$ design rules. Updated Table 8: STM32L151xD and STM32L152xD pin definitions ADC inputs. <br> Updated Table 16: Embedded internal reference voltage temperature coefficient at $100 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$. <br> and table footnote 3: "guaranteed by design" changed by "guaranteed by characterization results". <br> Updated Table 72: Comparator 2 characteristics new maximum threshold voltage temperature coefficient at $100 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$. |

Table 82. Document revision history (continued)

| Date | Revision | Changes |
| :---: | :---: | :---: |
| 10-Feb-2016 | 11 | Updated cover page putting eight SPIs in the peripheral communication interface list. <br> Updated Table 2: Ultra-low-power STM32L151xD and STM32L152xD device features and peripheral counts SPI and I2S lines. <br> Updated Table 47: ESD absolute maximum ratings CDM class II by class III. <br> Updated all the notes, removing 'not tested in production'. <br> Updated Table 10: Voltage characteristics adding note about $\mathrm{V}_{\text {REF }}$ pin. <br> Updated Table 5: Functionalities depending on the working mode (from Run/active down to standby) LSI and LSE functionalities putting " $Y$ " in Standby mode. |
| 19-Oct-2017 | 12 | Updated Section 7: Package information adding information about other optional marking or inset/upset marks. <br> Updated note 1 below all the package device marking figures. <br> Updated Section 7: Package information replacing "Marking of engineering samples" by "device marking". <br> Updated Nested vectored interrupt controller (NVIC) in Section 3.2: <br> Arm ${ }^{\circledR}$ Cortex ${ }^{\circledR}-M 3$ core with MPU about process state automatically saved. <br> Updated Table 3: Functionalities depending on the operating power supply range removing I/O operation column and adding note about GPIO speed. <br> Updated Table 49: I/O current injection susceptibility note by 'injection is not possible'. <br> Updated Figure 27: Recommended NRST pin protection note about the 0.1 uF capacitor. <br> Updated Table 67: DAC characteristics resistive load. <br> Updated Section 3.1: Low-power modes Low-power run mode (MSI) RC oscillator clock. <br> Updated Table 5: Functionalities depending on the working mode (from Run/active down to standby) disabling I2C functionality in Lowpower Run and Low-power Sleep modes. |
| 20-Sep-2021 | 13 | Updated: <br> - Cover features. <br> - Section 1: Introduction adding reference to errata sheet. <br> - Section 4: Pin descriptions. <br> - Figure 36: ADC accuracy characteristics. <br> - Section 7: Package information. |

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